

Noise and Singularities in Bayesian Calibration Methods for Global 21-cm Cosmology Experiments

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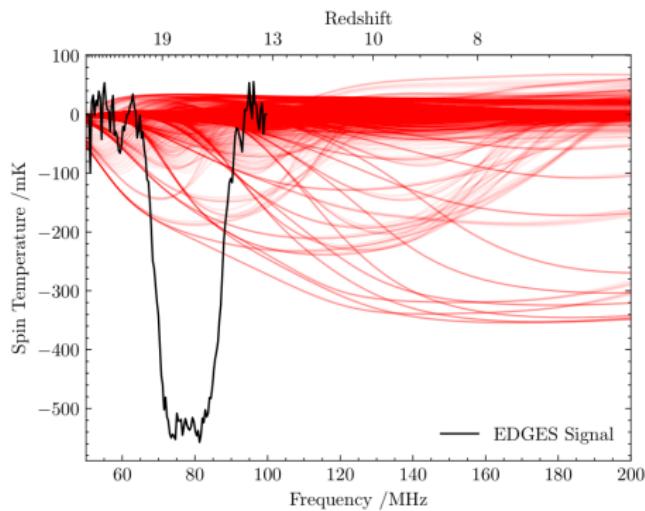
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Global 21-cm Cosmology



EDGES (Experiment to Detect the Global EoR Signature)



- Claimed detection of 21-cm signal at 78 MHz (Bowman et al., 2018)
- Unusually deep signal, requiring exotic physics
- Concerns that there is a residual systematic in the data (Hills et al., 2018; Sims and Pober, 2020)



REACH



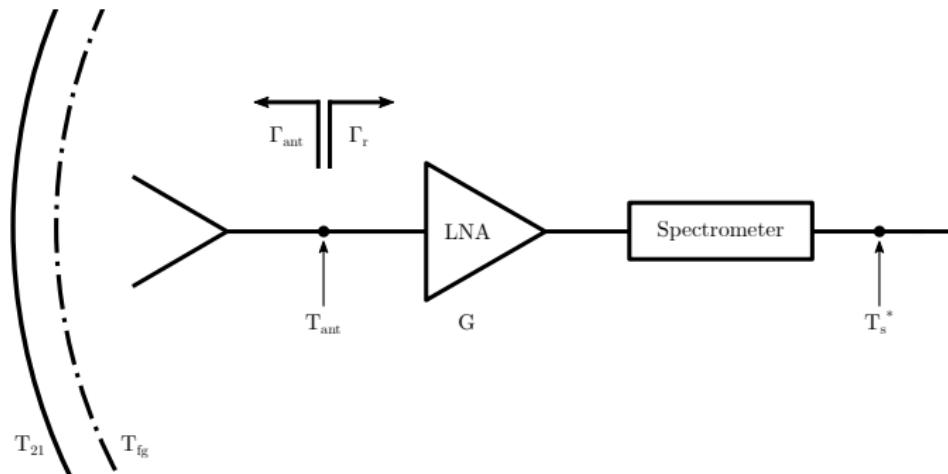
REACH Antenna



REACH Receiver



Global Experiment Receiver System



- Low Noise Amplifier (LNA) with gain G transforms $T_{ant} \rightarrow T_s^*$



What Is Calibration?

- Calibration relates input voltage (measured by a spectrometer as power spectral density, PSD) to temperature
- In general this relation can be written as

$$P_{\text{source}} = kBgM(T_{\text{source}} + T_{\text{rec}}) \quad (1)$$

- g is the source independent gain of the low noise amplifier (LNA) and M is the impedance mismatch factor



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Characterising the LNA with Noise Waves

- When the sources are not impedance matched M and T_{rec} are *source dependent* – simple Dicke switching cannot be used
- One way to characterise the LNA is using ‘noise waves’ (Meys, 1978; Monsalve et al., 2017)

$$\begin{aligned}
 T_{\text{NS}} \left(\frac{P_s - P_L}{P_{\text{NS}} - P_L} \right) + T_L = & T_s \left[\frac{1 - |\Gamma_s|^2}{|1 - \Gamma_s \Gamma_r|^2} \right] + T_{\text{unc}} \left[\frac{|\Gamma_s|^2}{|1 - \Gamma_s \Gamma_r|^2} \right] \\
 & + T_{\cos} \left[\frac{\text{Re} \left(\frac{\Gamma_s}{1 - \Gamma_s \Gamma_r} \right)}{\sqrt{1 - |\Gamma_r|^2}} \right] + T_{\sin} \left[\frac{\text{Im} \left(\frac{\Gamma_s}{1 - \Gamma_s \Gamma_r} \right)}{\sqrt{1 - |\Gamma_r|^2}} \right]
 \end{aligned} \quad (2)$$

- “Calibration equation” - three noise wave parameters: T_{unc} , T_{\cos} , T_{\sin}
- We also fit for T_{NS} , T_L (Roque et al., 2021)



REACH Receiver

The 12 calibrators

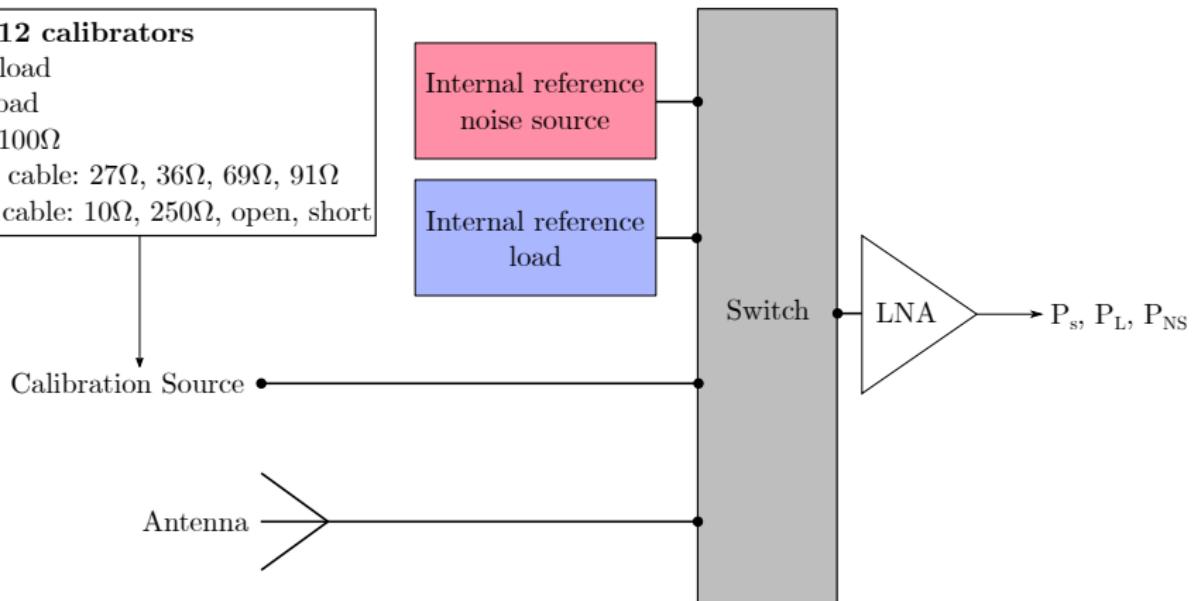
Cold load

Hot load

25Ω , 100Ω

Short cable: 27Ω , 36Ω , 69Ω , 91Ω

Long cable: 10Ω , 250Ω , open, short



Exploiting the Linearity

- All measured quantities - PSDs, reflection coefficients absorbed into X terms to give

$$T_s(\nu) = X_{\text{unc}} T_{\text{unc}} + X_{\cos} T_{\cos} + X_{\sin} T_{\sin} + X_{\text{NS}} T_{\text{NS}} + X_L T_L \quad (3)$$

- Can write this as a linear equation

$$\mathbf{T}_s = \mathbf{X}\boldsymbol{\Theta} + \sigma \quad (4)$$

- *measured* – $\mathbf{X} = (X_{\text{unc}} \quad X_{\cos} \quad X_{\sin} \quad X_{\text{NS}} \quad X_L)$
fitted for – $\boldsymbol{\Theta} = (T_{\text{unc}} \quad T_{\cos} \quad T_{\sin} \quad T_{\text{NS}} \quad T_L)^T$



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Calibration Likelihood

- For all methods in this talk we will use a Gaussian Likelihood

$$\mathcal{L} = \frac{1}{\sqrt{|2\pi\mathbf{C}|}} \exp \left\{ -\frac{1}{2} (\mathbf{T} - \mathbf{X}\Theta)^T \mathbf{C}^{-1} (\mathbf{T} - \mathbf{X}\Theta) \right\} \quad (5)$$

- We fit polynomials to the noise wave parameters, where Θ are the coefficients
- Use this and prior, $\pi(\Theta)$, to sample the posterior distribution, $\mathcal{P} = \mathcal{L}\pi/\mathcal{Z}$
- Calibration solution is the weighted mean of posterior



Conjugate Priors Method

- Method introduced in Roque et al., 2021
- Choose prior to be a Normal Inverse Gamma distribution, $N\text{-}\Gamma(\mu_\Theta, \mathbf{V}_\Theta, a, b)$
- Posterior is then calculated analytically to be the same distribution,
 $N\text{-}\Gamma(\mu^*, \mathbf{V}^*, a^*, b^*)$
- Transformations of $\mu_\Theta \rightarrow \mu^*$ etc. can be calculated analytically with data to quickly evaluate the posterior
- Requires data covariance to be $\mathbf{C} = \frac{1}{\sigma^2} \mathbf{I}$



Estimating the Noise

- Need a figure of merit for calibration
- Assumes PSD noise is Gaussian and S_{11} measurements are noiseless
- Propagate PSD noise through the noise wave parameter equations
- Key points:

$$\sigma_{T_s} \propto 1/(1 - |\Gamma_s|^2) \quad (6)$$

$$\sigma_{T_s} \propto T_{\text{NS}}^{\text{fit}} \quad (7)$$



Drawbacks of Conjugate Priors

- Choice of conjugate prior assumes all source temperatures have same noise, σ
 - As seen $\sigma \propto 1/(1 - |\Gamma_s|^2)$, where $|\Gamma_s| \in (0, 1)$
- Gradient descent can get stuck in local minima



Γ -Weighted Conjugate Priors

- Rewrite the calibration equation as

$$\mathbf{T}'_s = (1 - |\Gamma_s|^2) \mathbf{T}_s = (1 - |\Gamma_s|^2) \mathbf{X} \boldsymbol{\Theta} = \mathbf{X}' \boldsymbol{\Theta} \quad (8)$$

- As a result, $\sigma' \propto 1/(1 - |\Gamma_s|^2)$
- Physical motivation for EDGES' down-weighting of cables in Murray et al., 2022



Marginalised Polynomial Method

- Fit for calibrator noise parameters separately

$$\boldsymbol{\eta} = (\sigma_0 \quad \sigma_1 \quad \dots) \quad (9)$$

- We can also fit for the polynomial orders as parameters

$$\mathbf{n} = (n_{\text{unc}} \quad n_{\text{cos}} \quad n_{\text{sin}} \quad n_{\text{NS}} \quad n_{\text{L}}) \quad (10)$$

- We then marginalise over Θ giving the marginal likelihood

$$\begin{aligned} \log \mathcal{L}(\boldsymbol{\eta}, \mathbf{n}) &= \frac{1}{2} \log \left| \frac{\boldsymbol{\Sigma}_P}{\boldsymbol{\Sigma}_\pi} \right| - \frac{1}{2} (\boldsymbol{\mu}_P - \boldsymbol{\mu}_\pi)^T \boldsymbol{\Sigma}_\pi^{-1} (\boldsymbol{\mu}_P - \boldsymbol{\mu}_\pi) \\ &\quad - \frac{1}{2} \log |2\pi \mathbf{C}| - \frac{1}{2} (\mathbf{T} - \mathbf{X}\boldsymbol{\mu}_P)^T \mathbf{C}^{-1} (\mathbf{T} - \mathbf{X}\boldsymbol{\mu}_P), \end{aligned} \quad (11)$$



Marginalised Polynomial Method (cont.)

- Now we numerically sample over noise parameters and the polynomial orders
- Polynomial coefficients are then sampled from

$$\boldsymbol{\Theta} \sim \mathcal{N}(\boldsymbol{\mu}, \boldsymbol{\Sigma}). \quad (12)$$

- Much more flexible method which allows the use of arbitrary noise models



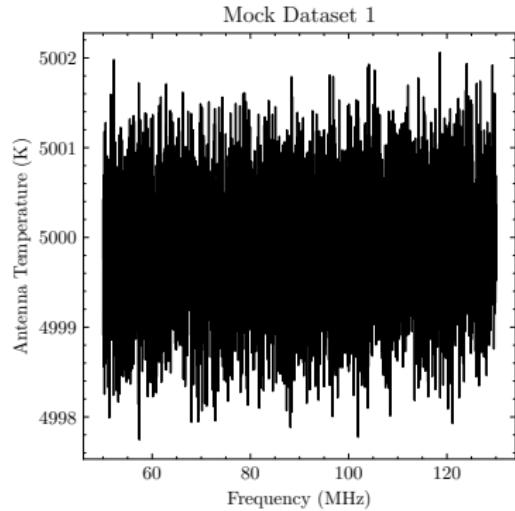
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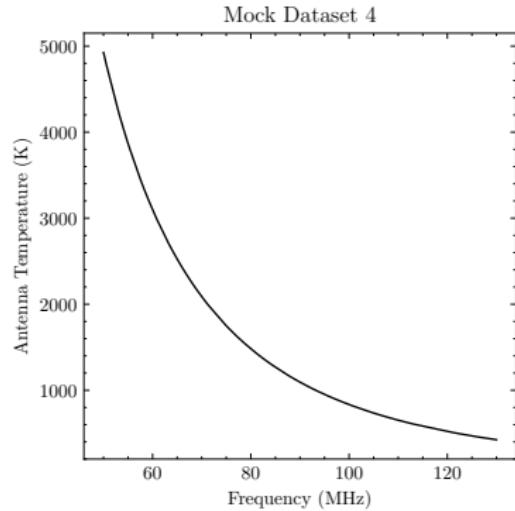


Mock Data

- Full simulation of REACH



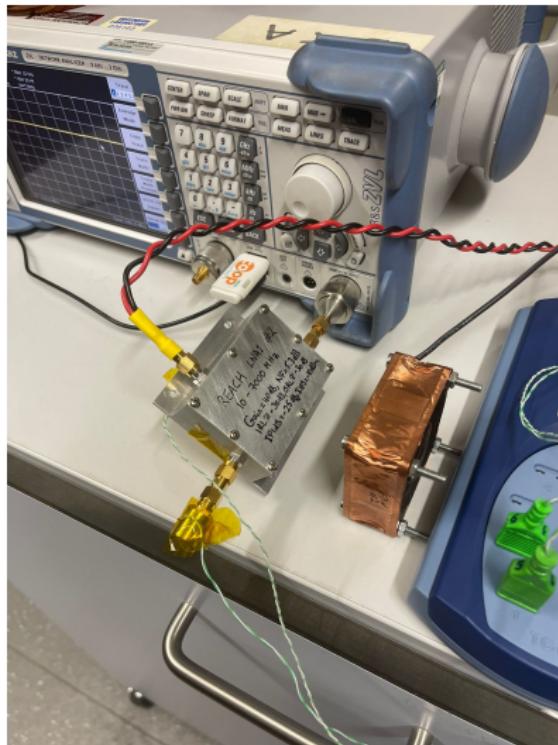
Flat Spectrum



EDGES Foreground

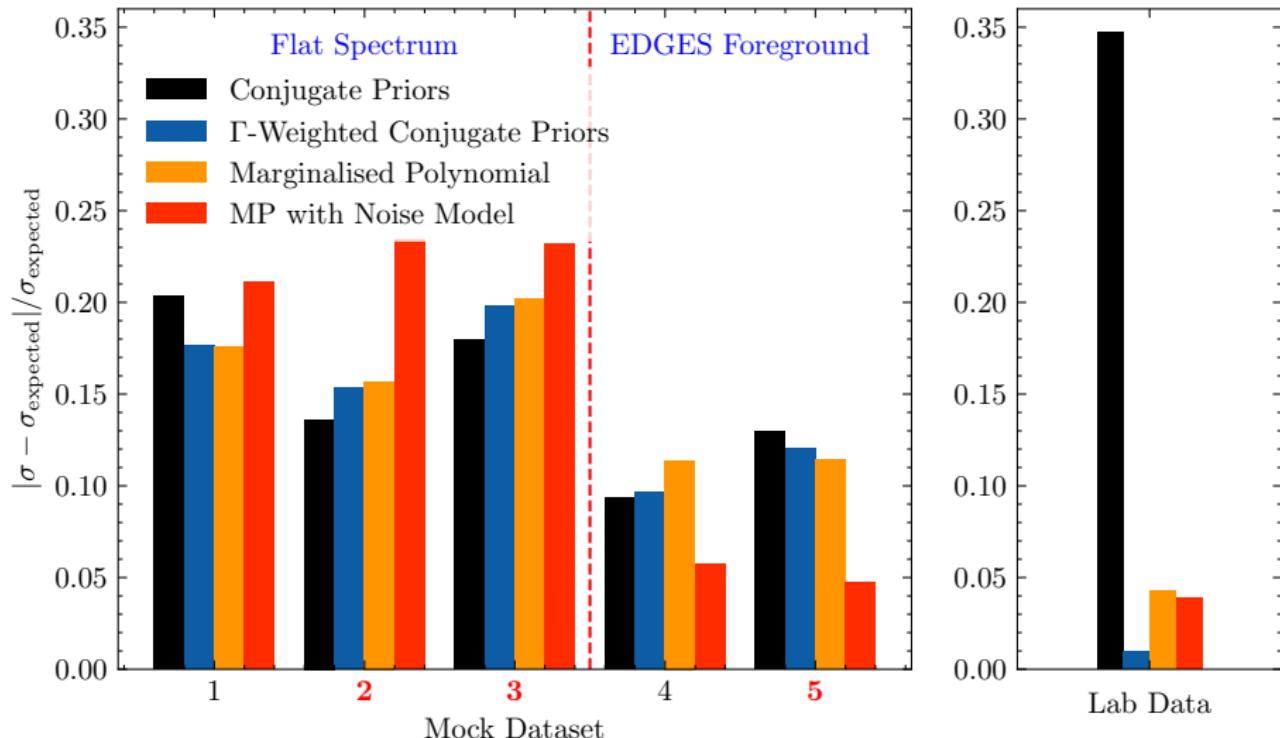


Lab Data

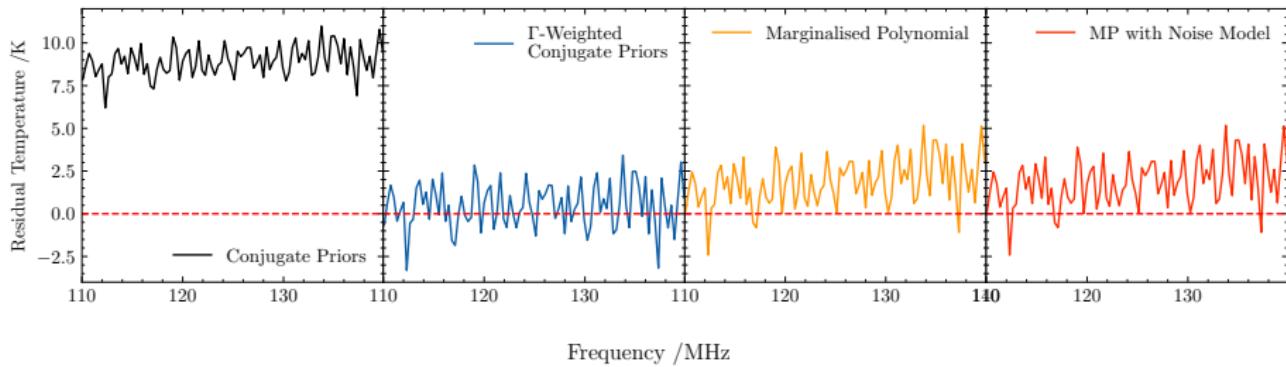


Lab Data Setup

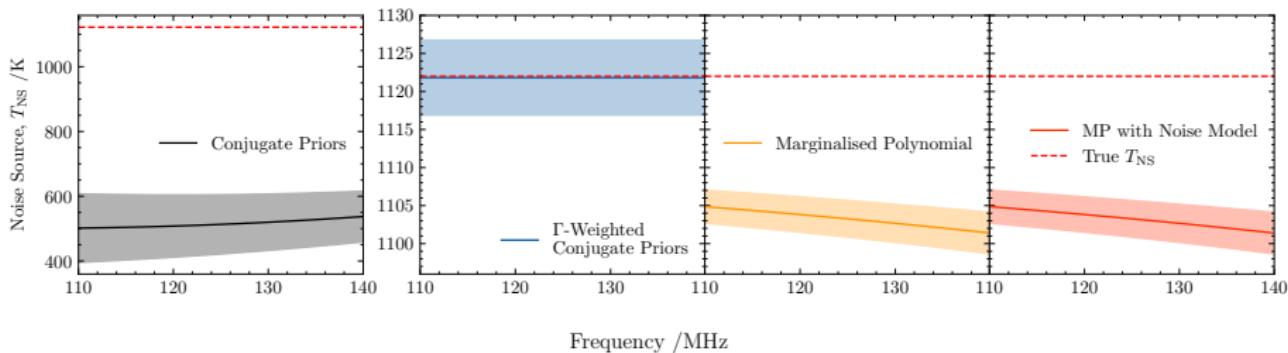
Method Comparison (RMSE)



Method Comparison (Residuals)

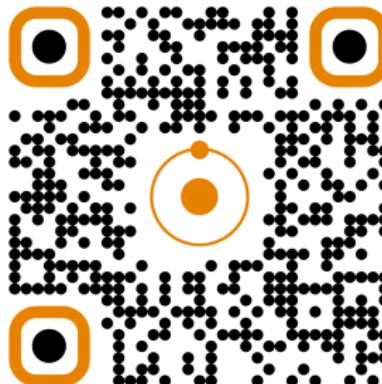


Method Comparison (T_{NS} Recovery)



Summary

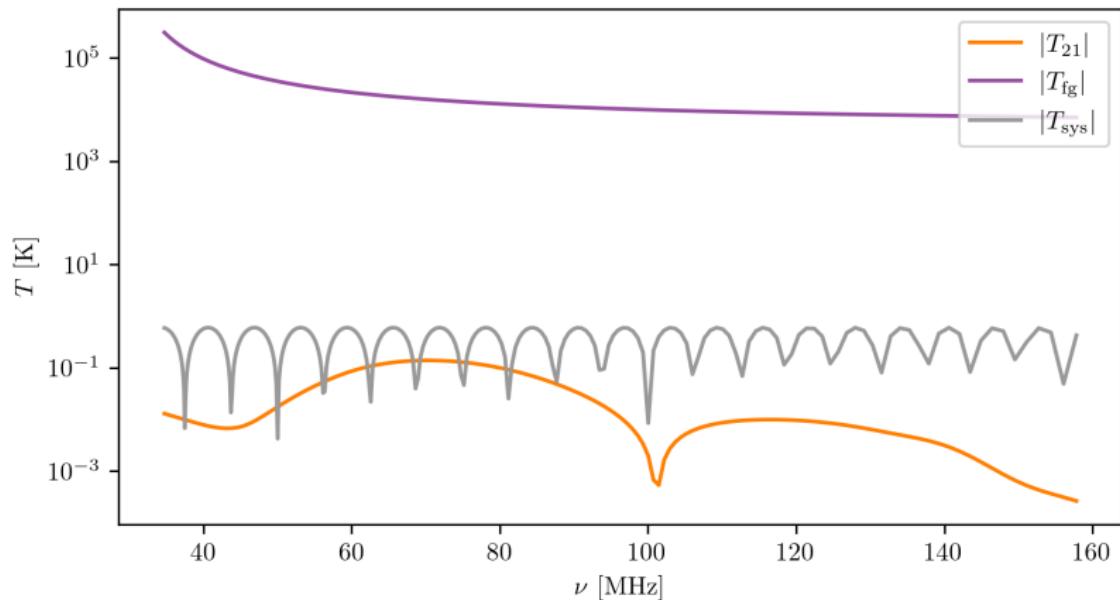
- In order to determine absolute temperatures and remove receiver systematics we must calibrate – noise waves provide a framework that we can use to characterise the LNA
- We can modify the conjugate priors method to mitigate the biases introduced by singular equations and varying noise
- Achieving the lowest RMSE is not the ultimate goal of calibration – the theoretical noise floor must be considered



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Dynamic Range Problem



Credit: Harry Bevins



Polynomial Order Posterior

