

# Noise and Singularities in Bayesian Calibration Methods for Global 21-cm Cosmology Experiments

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- 2 Noise Wave Parameterisation
- 3 Calibration Methods
- 4 Testing the Methods

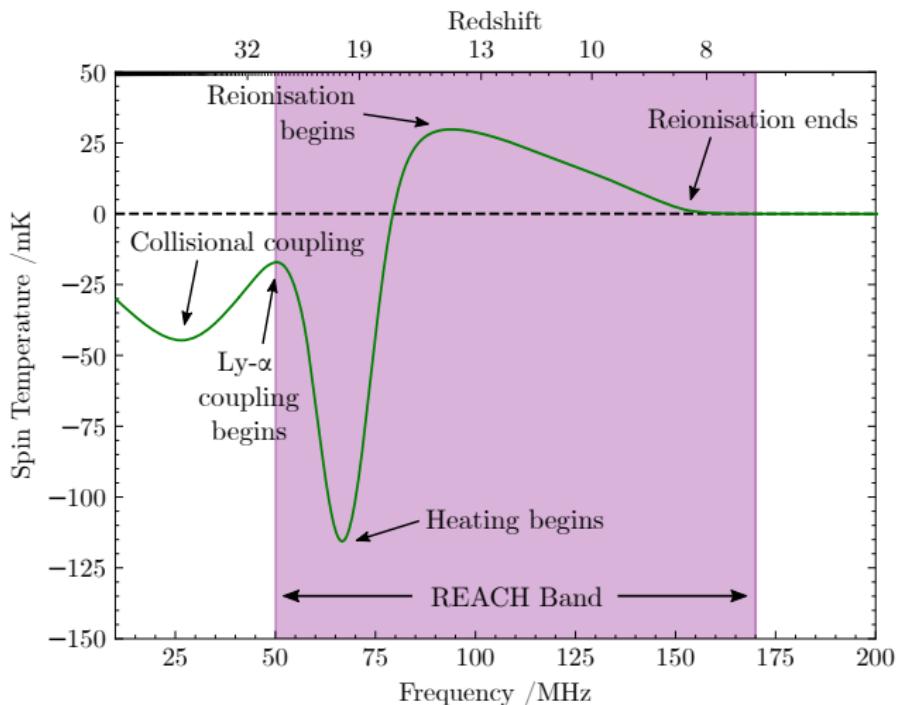


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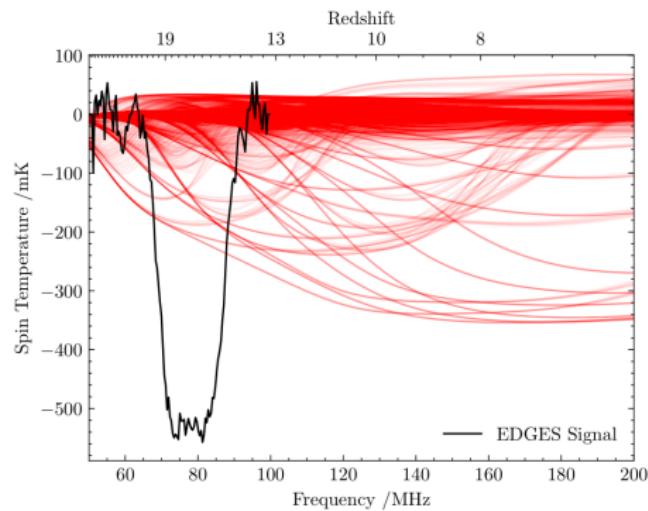
# Global 21-cm Cosmology



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# EDGES (Experiment to Detect the Global EoR Signature)



- Claimed detection of 21-cm signal at 78 MHz (Bowman et al., 2018)
- Unusually deep signal, requiring exotic physics
- Concerns that there is a residual systematic in the data (Hills et al., 2018; Sims and Pober, 2020)



# REACH



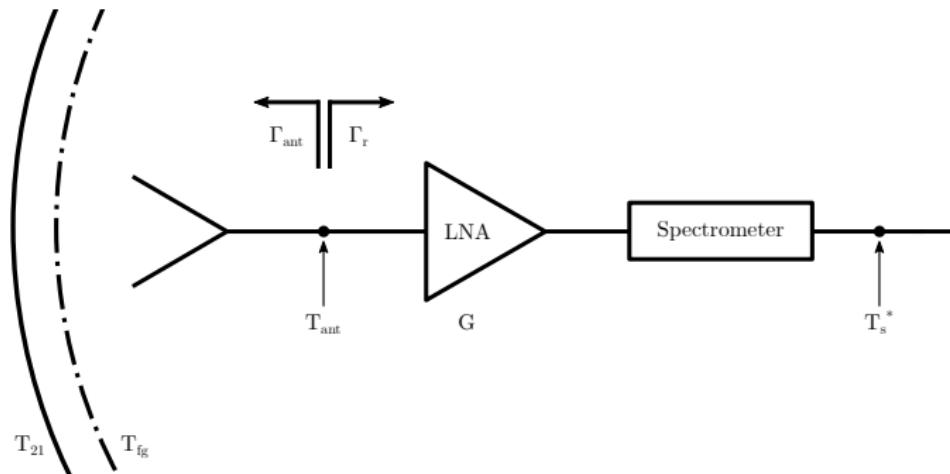
REACH Antenna



REACH Receiver



# Global Experiment Receiver System



- Low Noise Amplifier (LNA) with gain  $G$  transforms  $T_{ant} \rightarrow T_s^*$



# What Is Calibration?

- Calibration relates input voltage (measured by a spectrometer as power spectral density, PSD) to temperature
- In general this relation can be written as

$$P_{\text{source}} = kBgM(T_{\text{source}} + T_{\text{rec}}) \quad (1)$$

- $g$  is the source independent gain of the low noise amplifier (LNA) and  $M$  is the impedance mismatch factor



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# Characterising the LNA with Noise Waves

- When the sources are not impedance matched  $M$  and  $T_{\text{rec}}$  are *source dependent* – simple Dicke switching cannot be used
- One way to characterise the LNA is using ‘noise waves’ (Meys, 1978; Monsalve et al., 2017)

$$\begin{aligned}
 T_{\text{NS}} \left( \frac{P_s - P_L}{P_{\text{NS}} - P_L} \right) + T_L = & T_s \left[ \frac{1 - |\Gamma_s|^2}{|1 - \Gamma_s \Gamma_r|^2} \right] + T_{\text{unc}} \left[ \frac{|\Gamma_s|^2}{|1 - \Gamma_s \Gamma_r|^2} \right] \\
 & + T_{\cos} \left[ \frac{\text{Re} \left( \frac{\Gamma_s}{1 - \Gamma_s \Gamma_r} \right)}{\sqrt{1 - |\Gamma_r|^2}} \right] + T_{\sin} \left[ \frac{\text{Im} \left( \frac{\Gamma_s}{1 - \Gamma_s \Gamma_r} \right)}{\sqrt{1 - |\Gamma_r|^2}} \right]
 \end{aligned} \quad (2)$$

- “Calibration equation” - three noise wave parameters:  $T_{\text{unc}}$ ,  $T_{\cos}$ ,  $T_{\sin}$
- We also fit for  $T_{\text{NS}}$ ,  $T_L$  (Roque et al., 2021)



# REACH Receiver

## The 12 calibrators

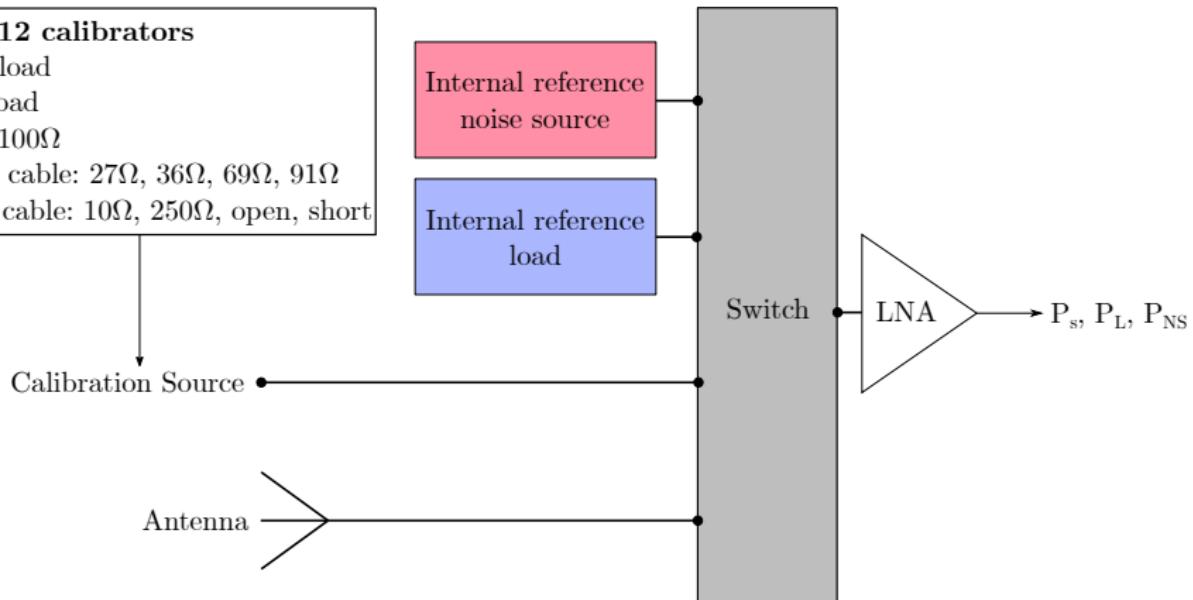
Cold load

Hot load

$25\Omega$ ,  $100\Omega$

Short cable:  $27\Omega$ ,  $36\Omega$ ,  $69\Omega$ ,  $91\Omega$

Long cable:  $10\Omega$ ,  $250\Omega$ , open, short



# Exploiting the Linearity

- All measured quantities - PSDs, reflection coefficients absorbed into  $X$  terms to give

$$T_s(\nu) = X_{\text{unc}} T_{\text{unc}} + X_{\cos} T_{\cos} + X_{\sin} T_{\sin} + X_{\text{NS}} T_{\text{NS}} + X_L T_L \quad (3)$$

- Can write this as a linear equation

$$\mathbf{T}_s = \mathbf{X}\boldsymbol{\Theta} + \sigma \quad (4)$$

- *measured* –  $\mathbf{X} = (X_{\text{unc}} \quad X_{\cos} \quad X_{\sin} \quad X_{\text{NS}} \quad X_L)$   
*fitted for* –  $\boldsymbol{\Theta} = (T_{\text{unc}} \quad T_{\cos} \quad T_{\sin} \quad T_{\text{NS}} \quad T_L)^T$



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# Calibration Likelihood

- For all methods in this talk we will use a Gaussian Likelihood

$$\mathcal{L} = \frac{1}{\sqrt{|2\pi\mathbf{C}|}} \exp \left\{ -\frac{1}{2} (\mathbf{T} - \mathbf{X}\Theta)^T \mathbf{C}^{-1} (\mathbf{T} - \mathbf{X}\Theta) \right\} \quad (5)$$

- We fit polynomials to the noise wave parameters, where  $\Theta$  are the coefficients
- Use this and the prior,  $\pi(\Theta)$ , to sample the posterior distribution,  $\mathcal{P} = \mathcal{L}\pi/\mathcal{Z}$
- Calibration solution is the weighted mean of posterior



# Conjugate Priors Method

- Method introduced in Roque et al., 2021
- Choose prior to be a Normal Inverse Gamma distribution,  $N\text{-}\Gamma(\mu_\Theta, \mathbf{V}_\Theta, a, b)$
- Posterior is then calculated analytically to be the same distribution,  
 $N\text{-}\Gamma(\mu^*, \mathbf{V}^*, a^*, b^*)$
- Transformations of  $\mu_\Theta \rightarrow \mu^*$  etc. can be calculated analytically with data to quickly evaluate the posterior
- Requires data covariance to be  $\mathbf{C} = \frac{1}{\sigma^2} \mathbf{I}$



# Estimating the Noise

- Need a figure of merit for calibration
- Assumes PSD noise is Gaussian and  $S_{11}$  measurements are noiseless
- Propagate PSD noise through the noise wave parameter equations
- Key points:

$$\sigma_{T_s} \propto 1/(1 - |\Gamma_s|^2) \quad (6)$$

$$\sigma_{T_s} \propto T_{\text{NS}}^{\text{fit}} \quad (7)$$



# Drawbacks of Conjugate Priors

- Choice of conjugate prior assumes all source temperatures have same noise,  $\sigma$ 
  - As seen  $\sigma \propto 1/(1 - |\Gamma_s|^2)$ , where  $|\Gamma_s| \in (0, 1)$
- Gradient descent can get stuck in local minima



# $\Gamma$ -Weighted Conjugate Priors

- Rewrite the calibration equation as

$$\mathbf{T}'_s = (1 - |\Gamma_s|^2) \mathbf{T}_s = (1 - |\Gamma_s|^2) \mathbf{X} \boldsymbol{\Theta} = \mathbf{X}' \boldsymbol{\Theta} \quad (8)$$

- As a result,  $\sigma' \propto 1/(1 - |\Gamma_s|^2)$
- Physical motivation for EDGES' down-weighting of cables in Murray et al., 2022



# Marginalised Polynomial Method

- Fit for calibrator noise parameters separately

$$\boldsymbol{\eta} = (\sigma_0 \quad \sigma_1 \quad \dots) \quad (9)$$

- We can also fit for the polynomial orders as parameters

$$\mathbf{n} = (n_{\text{unc}} \quad n_{\cos} \quad n_{\sin} \quad n_{\text{NS}} \quad n_{\text{L}}) \quad (10)$$

- We then marginalise over  $\Theta$  giving the marginal likelihood

$$\begin{aligned} \log \mathcal{L}(\boldsymbol{\eta}, \mathbf{n}) &= \frac{1}{2} \log \left| \frac{\boldsymbol{\Sigma}_P}{\boldsymbol{\Sigma}_\pi} \right| - \frac{1}{2} (\boldsymbol{\mu}_P - \boldsymbol{\mu}_\pi)^T \boldsymbol{\Sigma}_\pi^{-1} (\boldsymbol{\mu}_P - \boldsymbol{\mu}_\pi) \\ &\quad - \frac{1}{2} \log |2\pi \mathbf{C}| - \frac{1}{2} (\mathbf{T} - \mathbf{X}\boldsymbol{\mu}_P)^T \mathbf{C}^{-1} (\mathbf{T} - \mathbf{X}\boldsymbol{\mu}_P), \end{aligned} \quad (11)$$



# Marginalised Polynomial Method (cont.)

- Now we numerically sample over noise parameters and the polynomial orders
- Polynomial coefficients are then sampled from

$$\Theta \sim \mathcal{N}(\mu, \Sigma). \quad (12)$$

- Much more flexible method which allows the use of arbitrary noise models



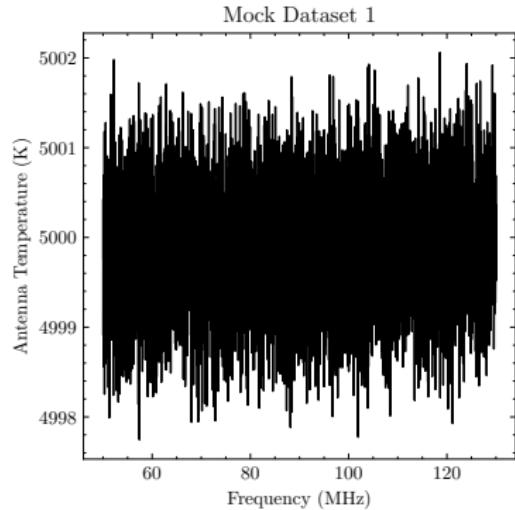
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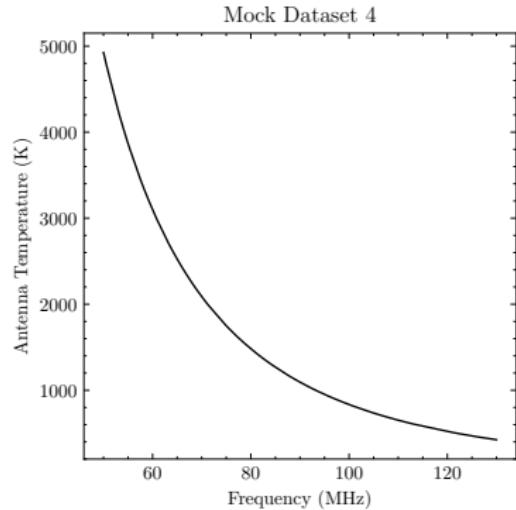


# Mock Data

- Full simulation of REACH



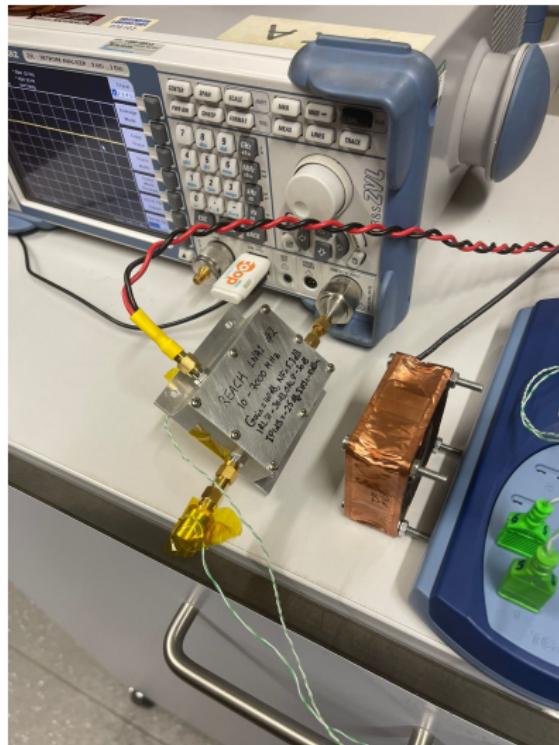
Flat Spectrum



EDGES Foreground



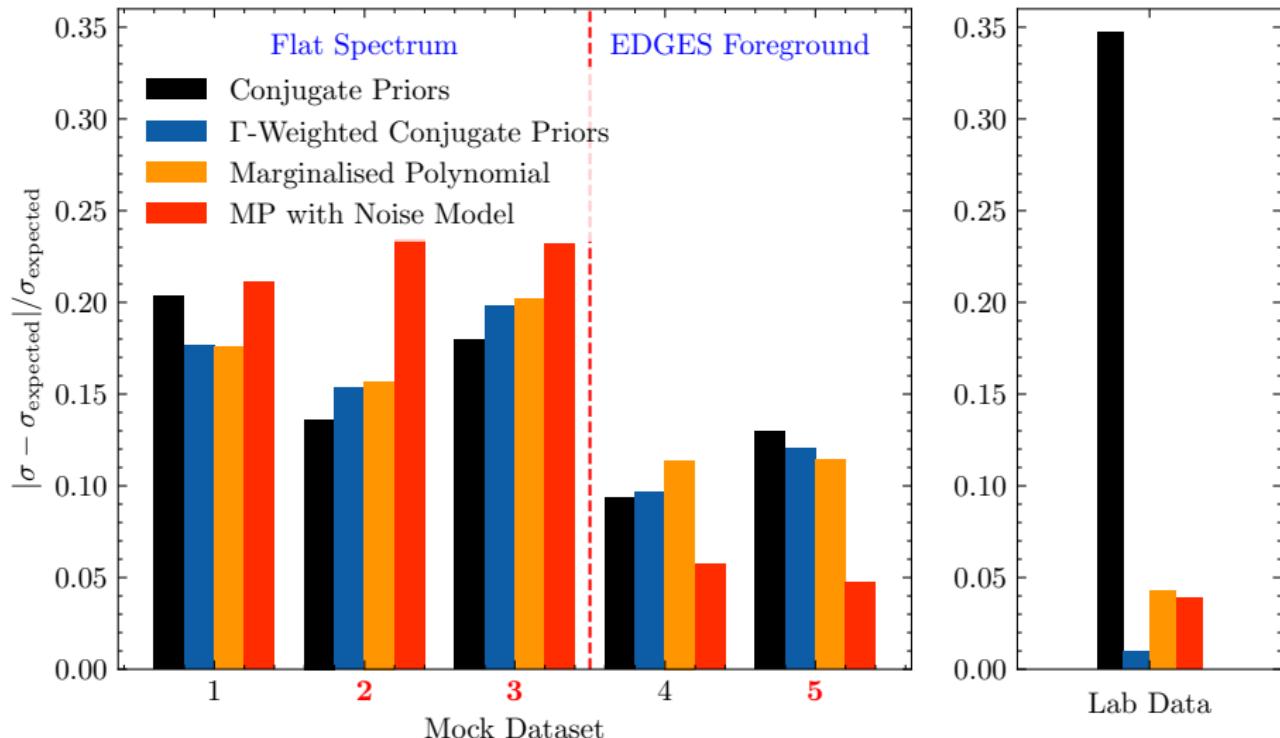
# Lab Data



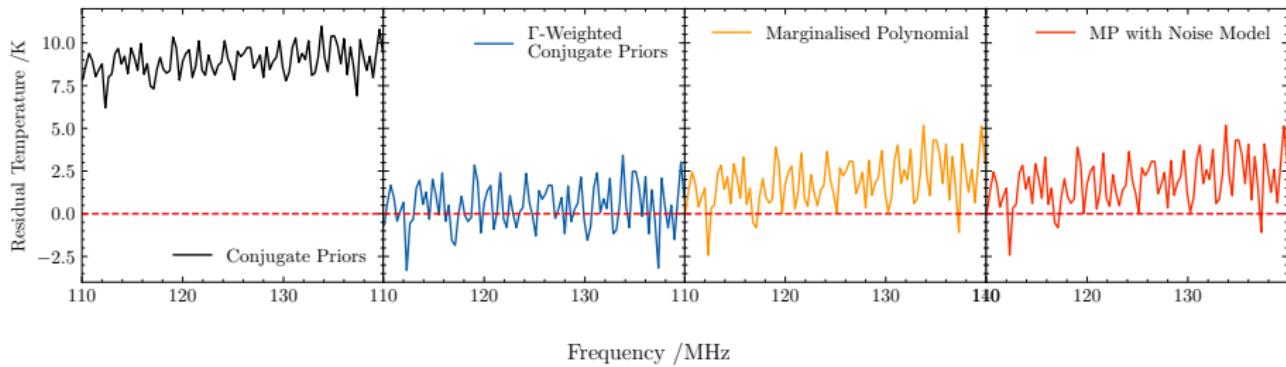
Lab Data Setup



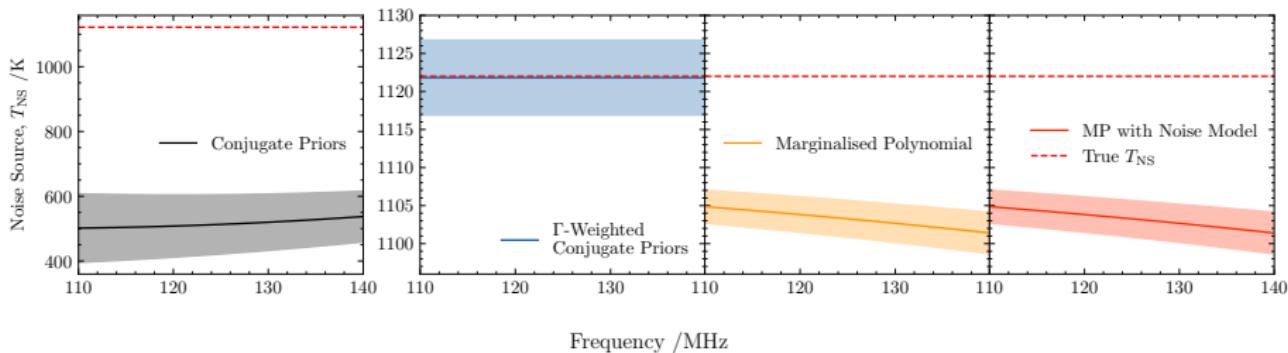
# Method Comparison (RMSE)



# Method Comparison (Residuals)

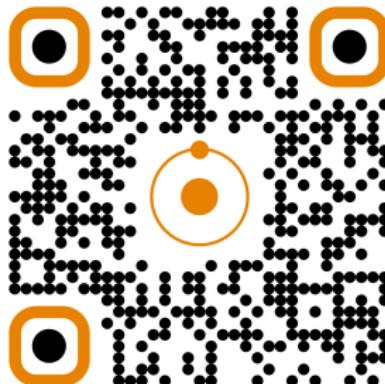


# Method Comparison ( $T_{\text{NS}}$ Recovery)



# Summary

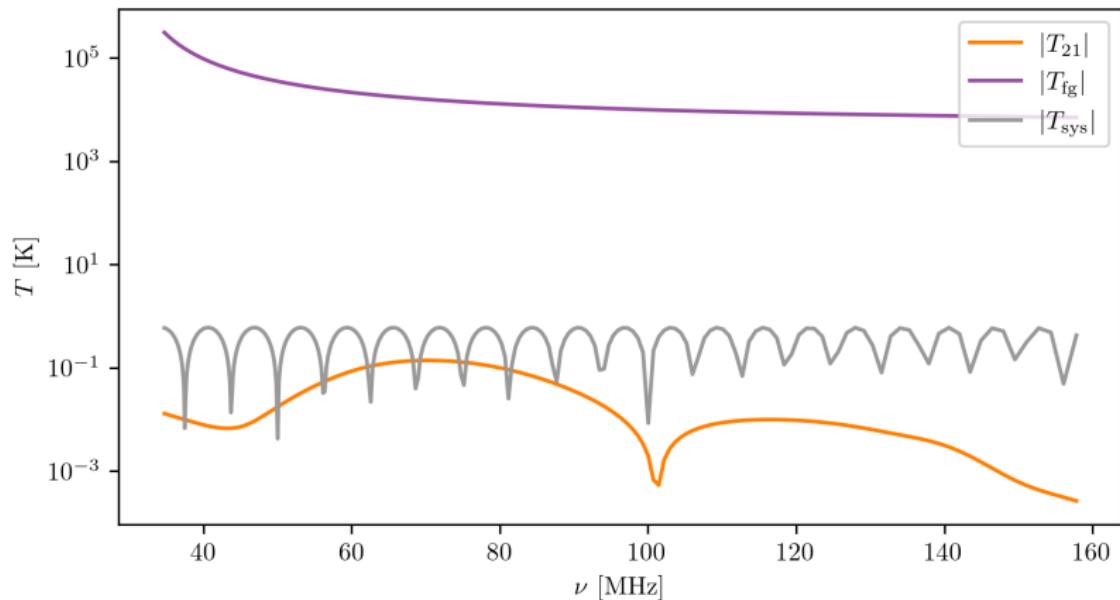
- In order to determine absolute temperatures and remove receiver systematics we must calibrate – noise waves provide a framework that we can use to characterise the LNA
- We can modify the conjugate priors method to mitigate the biases introduced by singular equations and varying noise
- Achieving the lowest RMSE is not the ultimate goal of calibration – the theoretical noise floor must be considered



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# Dynamic Range Problem



Credit: Harry Bevins



# Polynomial Order Posterior

