



Signal Processing for Astronomy

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Narrabri – 25 September 2017
+ some slides from John Tuthill



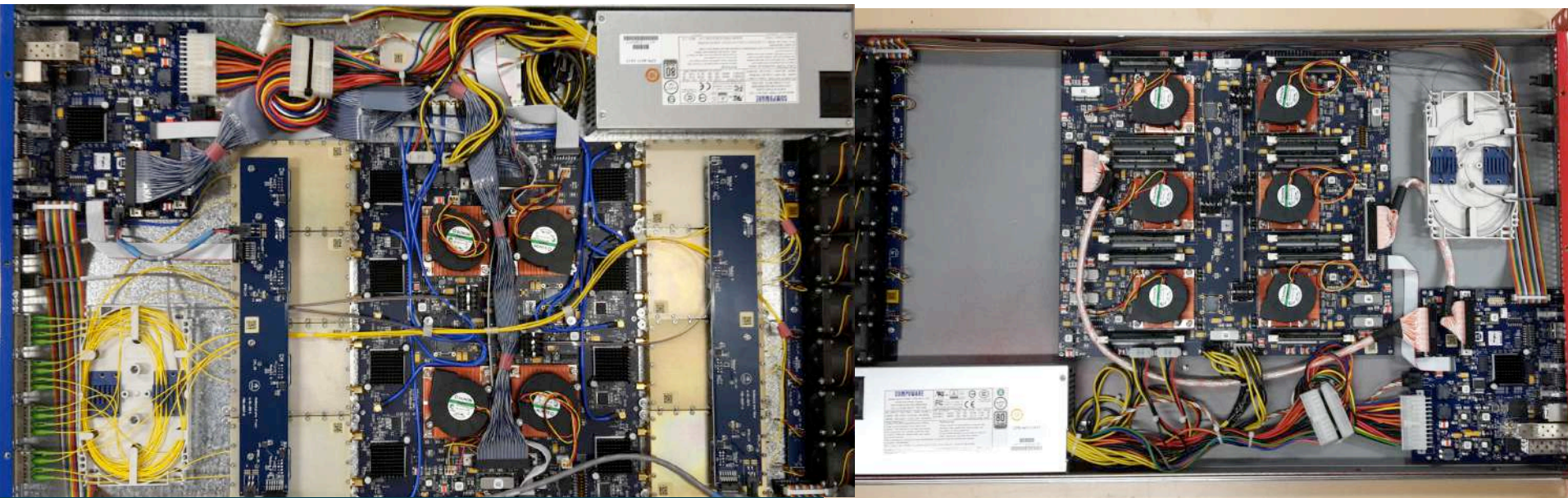
Context

Digital Signal Processing – manipulations of discrete samples approximating some continuous function

- Beamformer
- Tied-array unit
- Fringe rotator
- Correlator

In astronomy:
this is about backends

- well-defined core calculations
- often highly optimised
- often implemented in hardware



Correlator is the brain
of radio-interferometer
(as it computes visibilities)



"black box" for most science work

May need to know details,
if you start pushing the limits



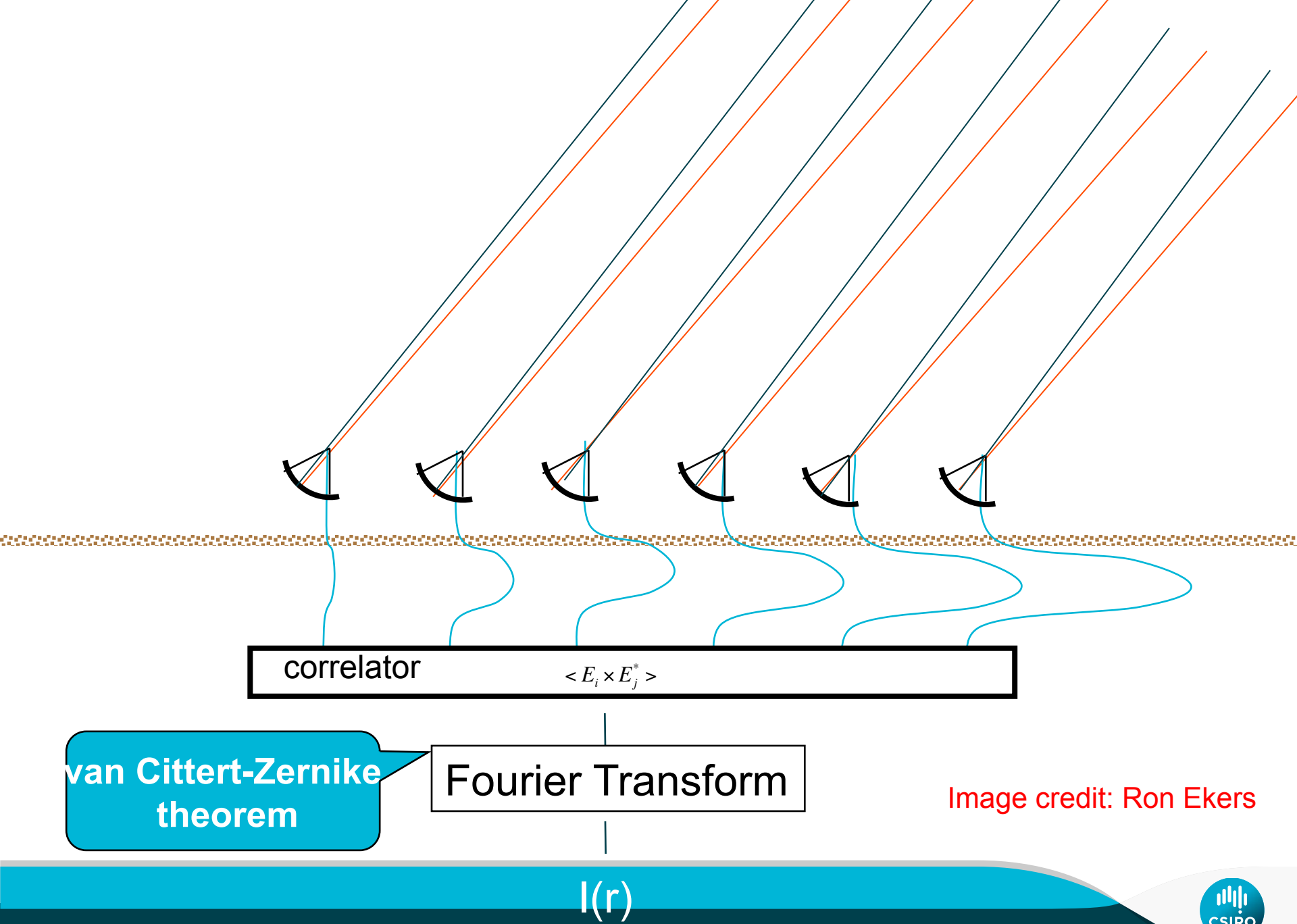
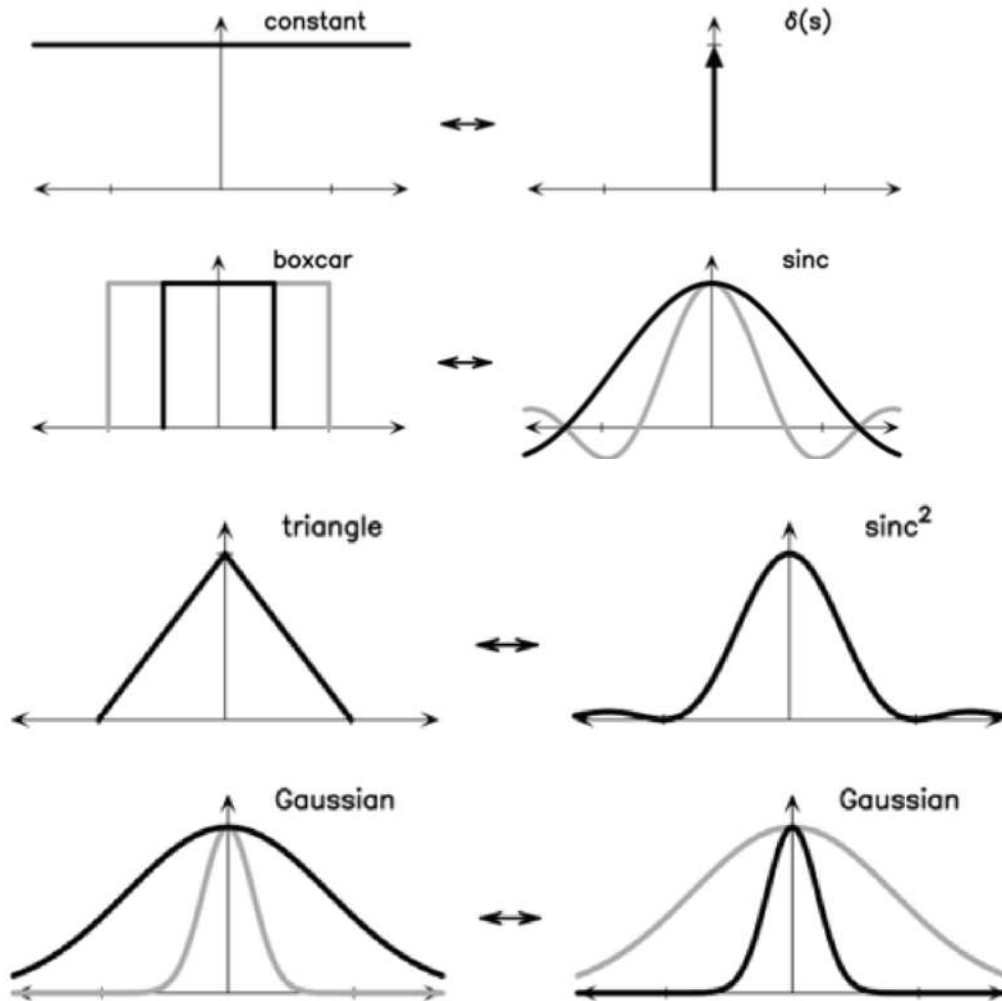


Image credit: Ron Ekers

Fourier transforms – intro



Forward:

$$F(s) \equiv \int_{-\infty}^{\infty} f(x) e^{-2\pi i s x} dx$$

Reverse:

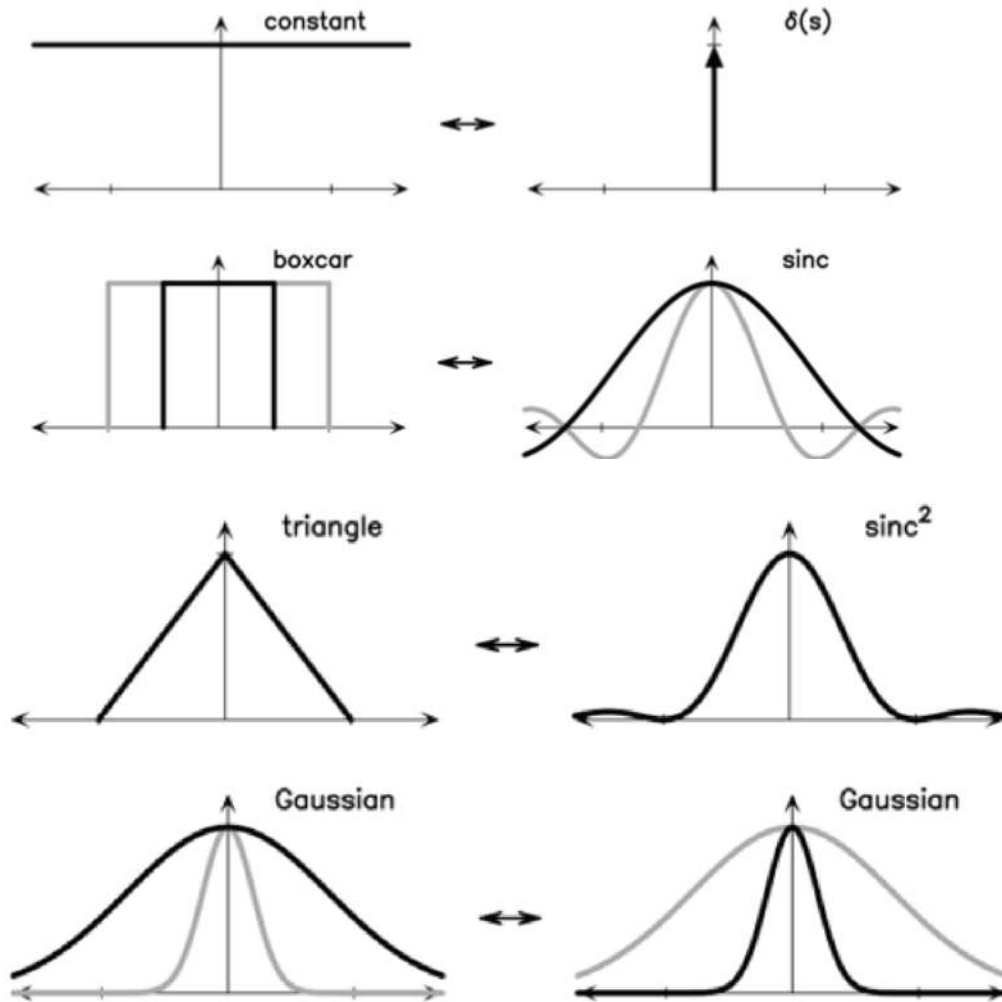
$$f(x) \equiv \int_{-\infty}^{\infty} F(s) e^{2\pi i s x} ds$$

Small \leftrightarrow Large

Convolution \leftrightarrow Product

Fourier transforms – intro

$$j = \sqrt{-1}$$



Forward:

$$F(s) \equiv \int_{-\infty}^{\infty} f(x) e^{-2\pi j s x} dx$$

Reverse:

$$f(x) \equiv \int_{-\infty}^{\infty} F(s) e^{2\pi j s x} ds$$

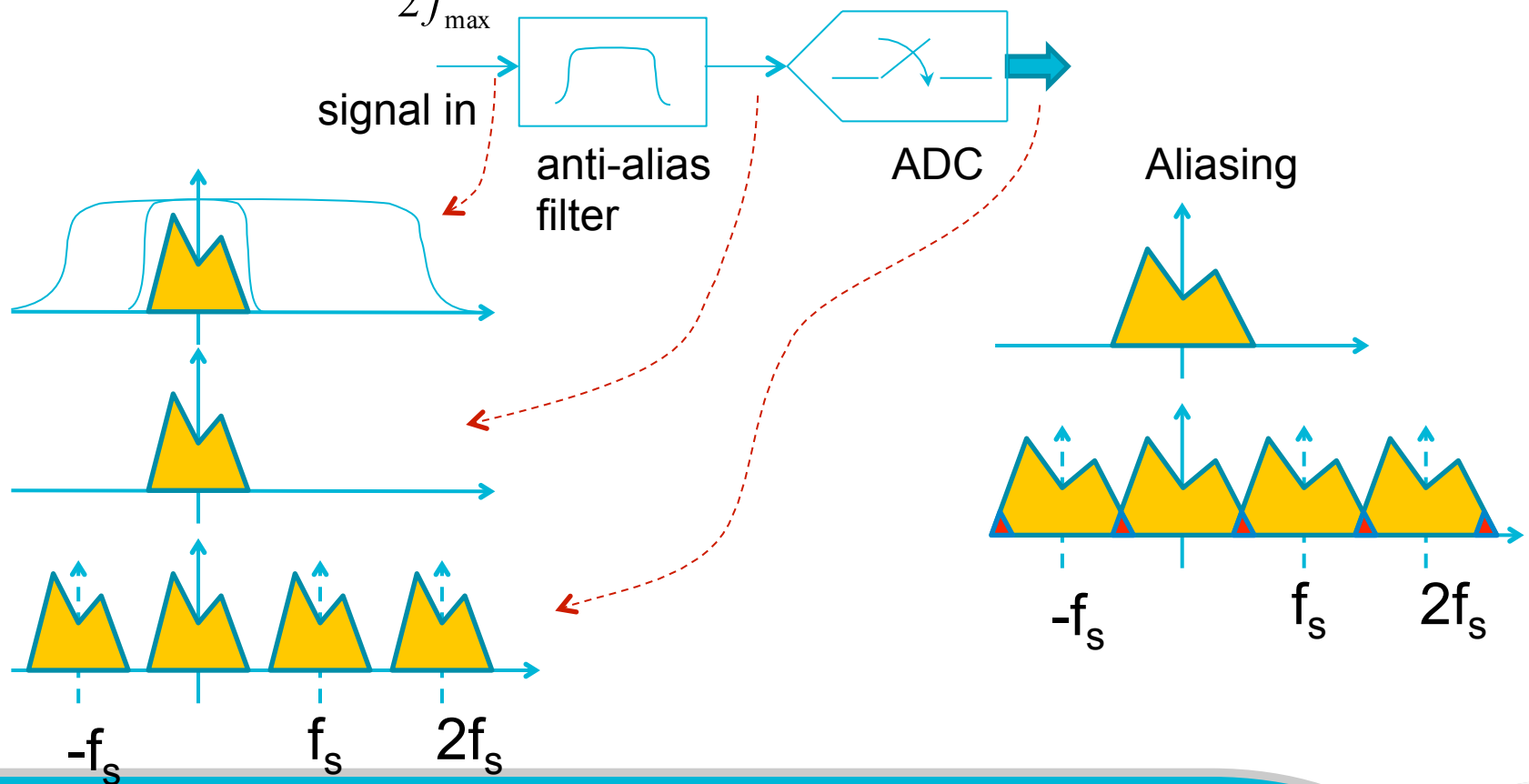
Small ↔ Large

Convolution ↔ Product

Sampling

The Sampling Theorem: A band-limited signal having no frequency components above f_{\max} can be determined uniquely by values sampled at uniform intervals of T_s satisfying:

$$T_s \leq \frac{1}{2f_{\max}}$$



Simplest possible case

Assuming bandwidth is small enough & we don't care about frequency structure

$$V_{ij} = \frac{1}{N} \sum_{k=1}^N E_i(k) \times E_j^*(k - \tau)$$



Compute correlation directly
(e.g. in software)

Code snippet from BETA-3 (3-baseline) software correlator:

```
IndexType offset1 = itsDelay1;  
IndexType offset2 = itsDelay2;  
IndexType offset3 = itsDelay3;  
  
for (; (offset1 < size)&&(offset2 < size)&&(offset3 < size); ++offset1, ++offset2, ++offset3) {  
    itsVis12 += *(stream1 + offset1) * conj(*(stream2+offset2));  
    itsVis13 += *(stream1 + offset1) * conj(*(stream3+offset3));  
    itsVis23 += *(stream2 + offset2) * conj(*(stream3+offset3));  
}
```

Non-monochromatic input signal

More intuitive description via continuous formalism:

$$E(t) = \int s(\nu) \exp\{2\pi j\nu t\} d\nu$$

spectral representation

Correlation between antenna 1 and 2 data streams for a given lag:

$$\gamma(\tau) = \int E_1(t) E_2^*(t - \tau) dt = \int \underbrace{s_1(\nu) s_2^*(\nu)}_{\text{Power (cross-correlation) spectrum}} \exp\{2\pi j\nu \tau\} d\nu$$

Option 1

Correlate streams for a number of lags and Fourier-transform

Option 2

Lag or XF correlator

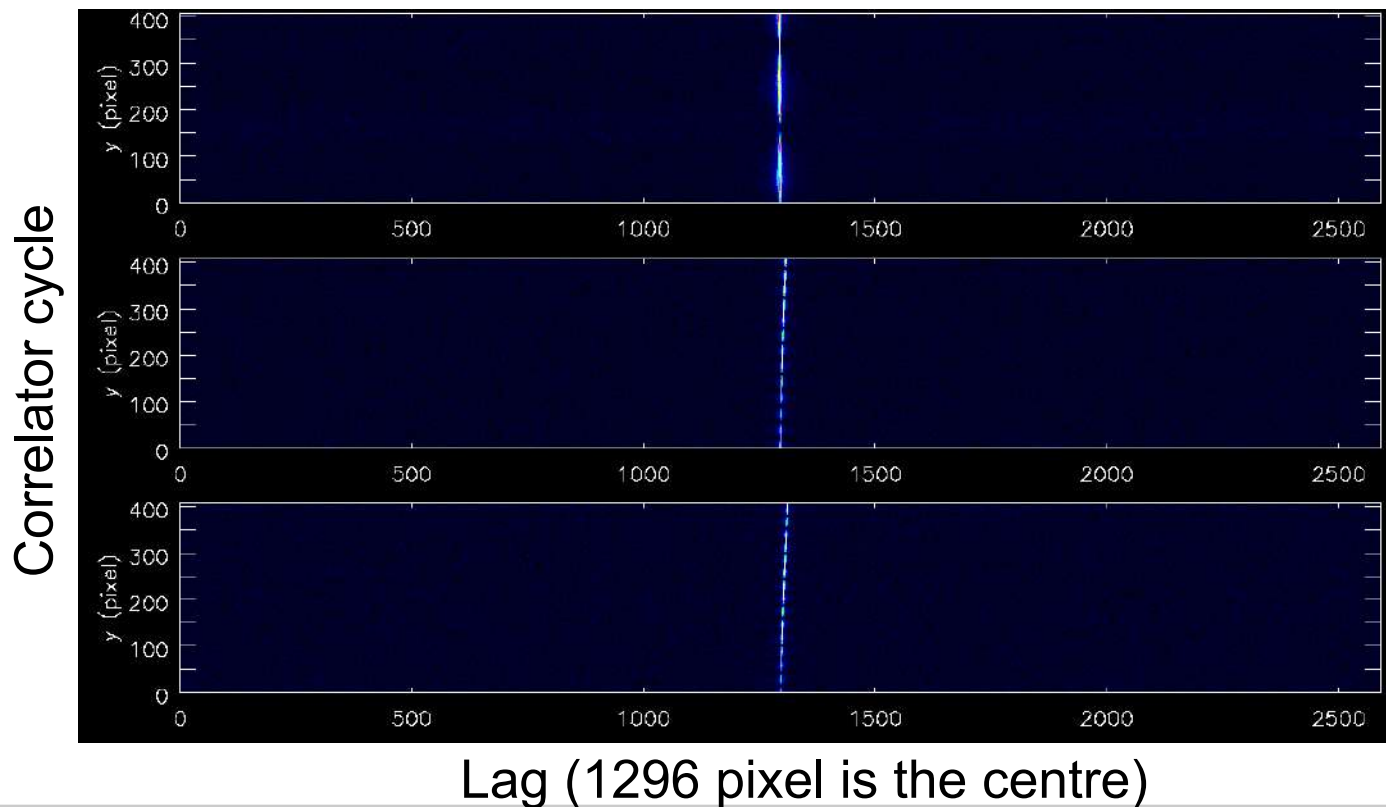
Fourier-transform input streams and cross-multiply

FX-correlator

More on lag domain

Regardless of the correlator architecture, lag spectrum

$$\gamma(\tau) = \int V_{12}(\nu) \exp\{2\pi j\nu\tau\} d\nu \quad \text{is a useful diagnostic tool}$$



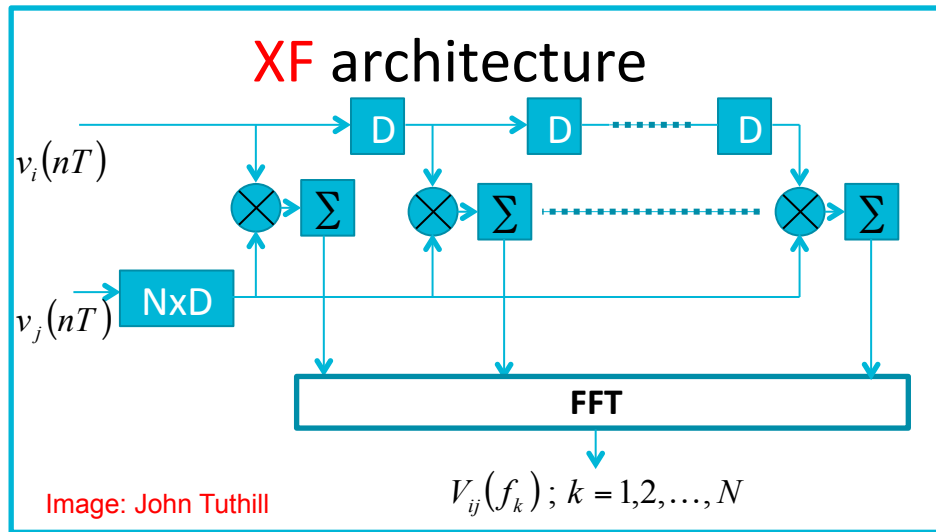
AK04-AK05

AK04-AK12

AK05-AK12

Lag (XF) correlator

Old ATCA correlator (pre-CABB) is a good example



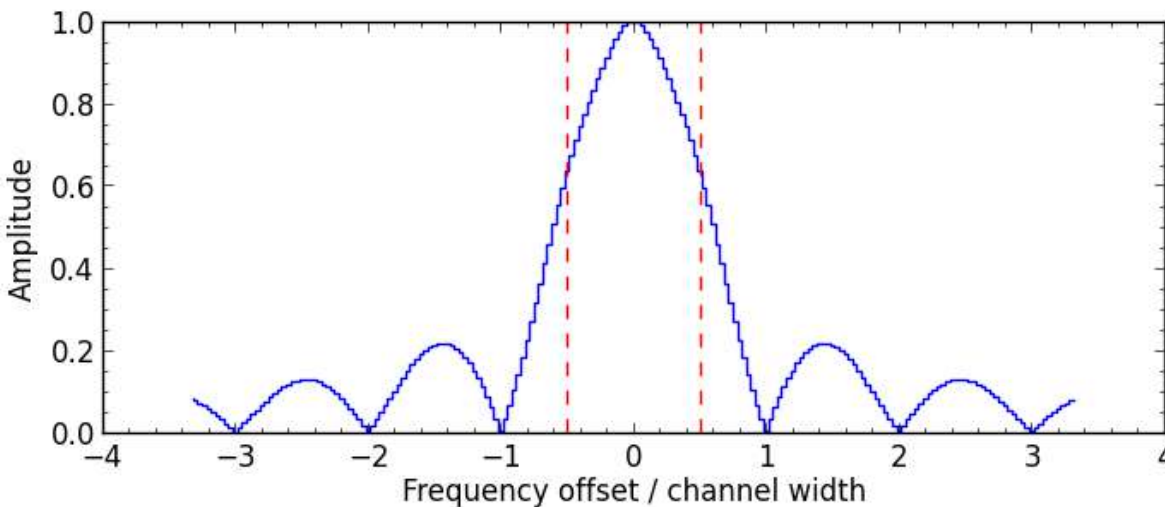
Fractional delays (< 1 sample) need to be corrected in some other way

$$\gamma(\tau) = \int V_{12}(\nu) \exp\{2\pi j\nu\tau\} d\nu$$

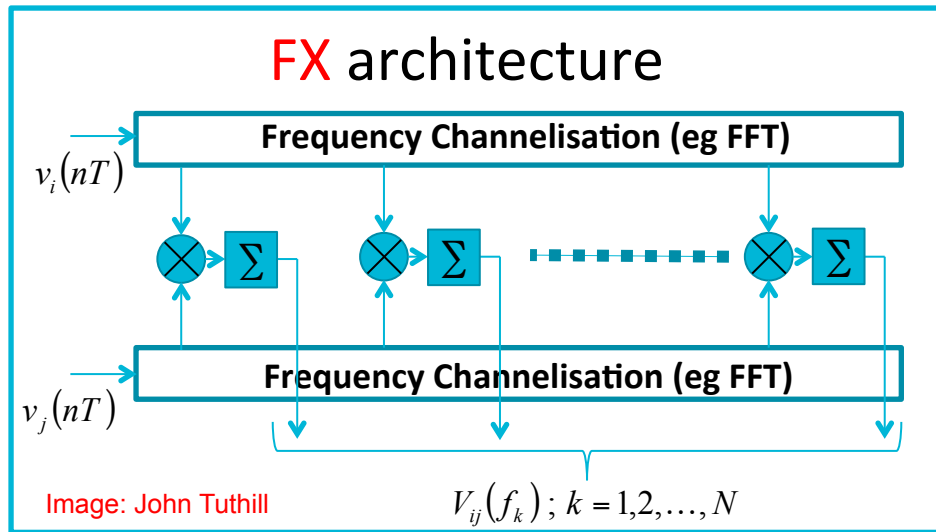
Only finite number of lags can be measured



Convolution with *sinc* in the frequency domain

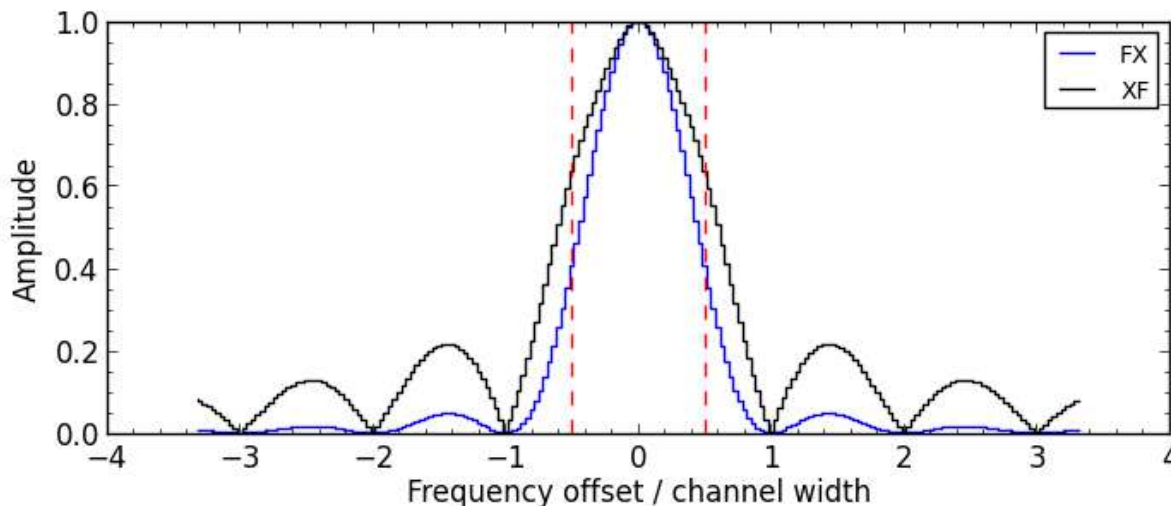


FX correlator



Fractional delays are easy to implement

Output is the product of two Fourier Transforms, each is presented with a finite chunk of data

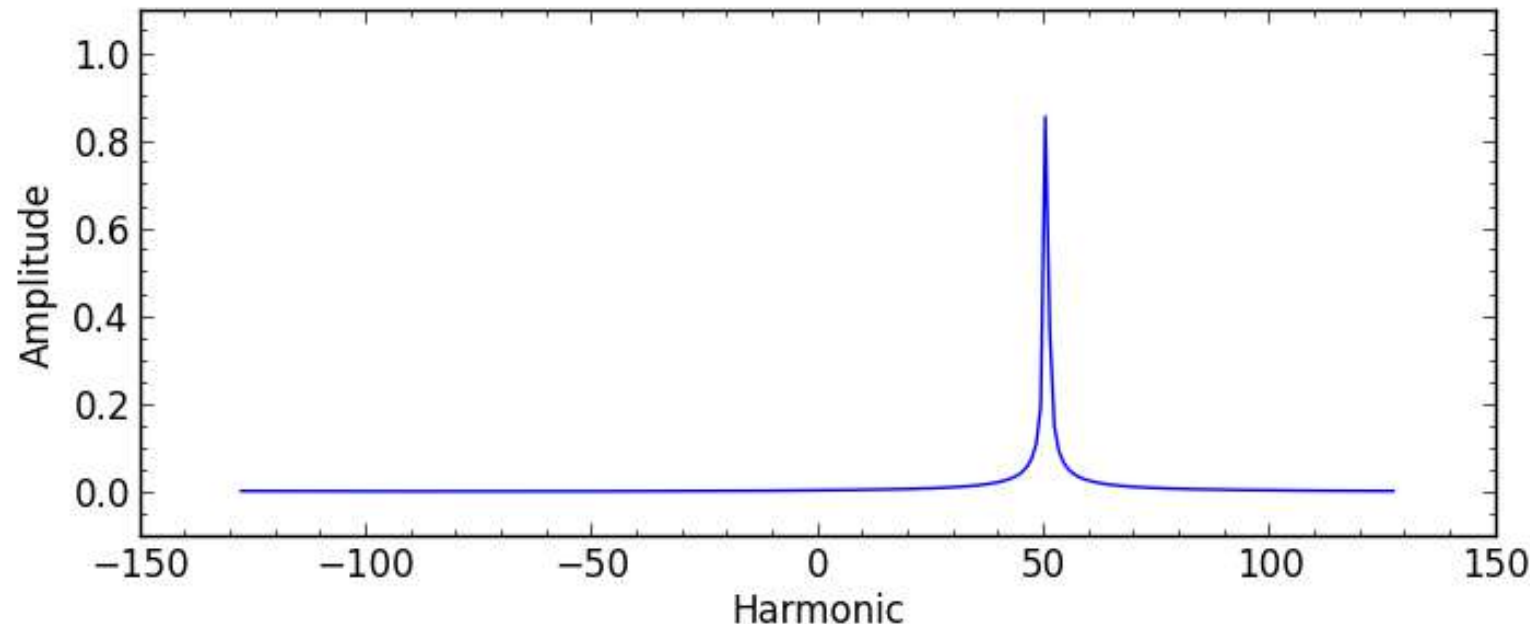


Convolution with sinc^2 in the frequency domain

Can we get a better channel response?

Let's consider FX architecture as a channeliser + cross-multiplier
(think of a simple correlator described earlier)

The channel response is determined by the channeliser performance
and FFT is known to be quite bad



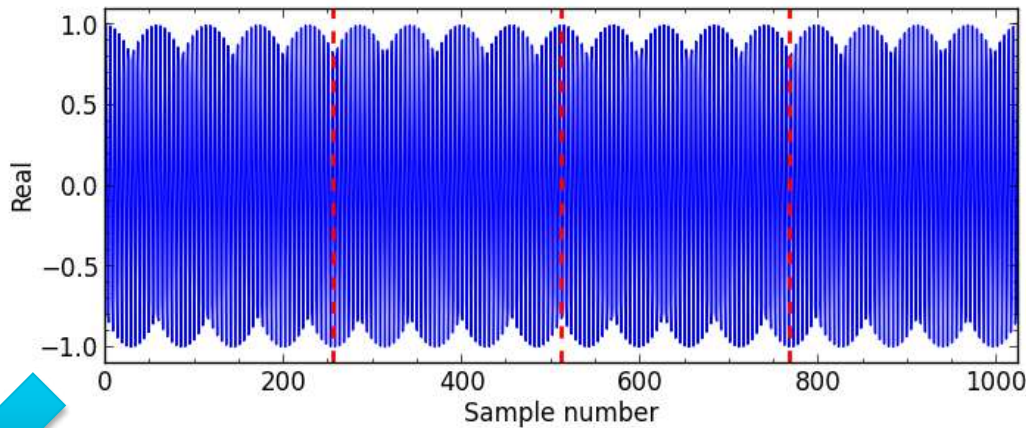
Power leakage
into adjacent
channels

FFT of the sine wave with frequency of $50.3 * f_s / 256$

Filtering

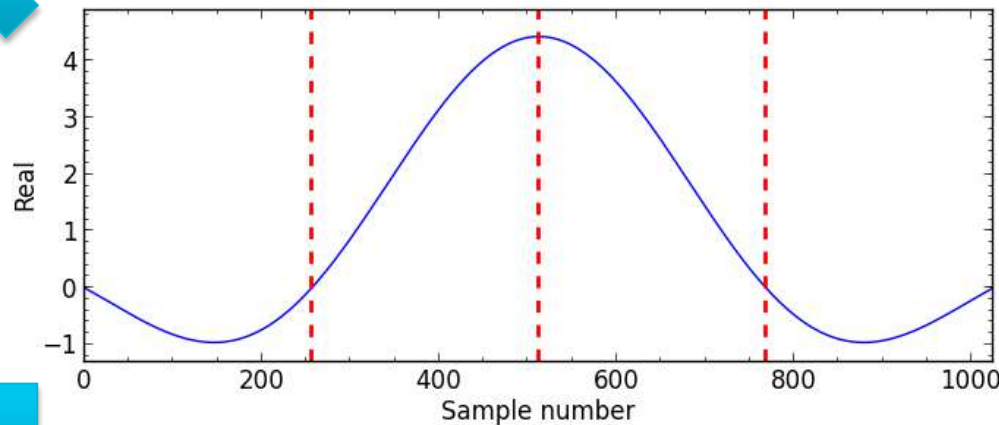
Input data samples

Sine wave with the frequency of $50.3 f_s / 256$

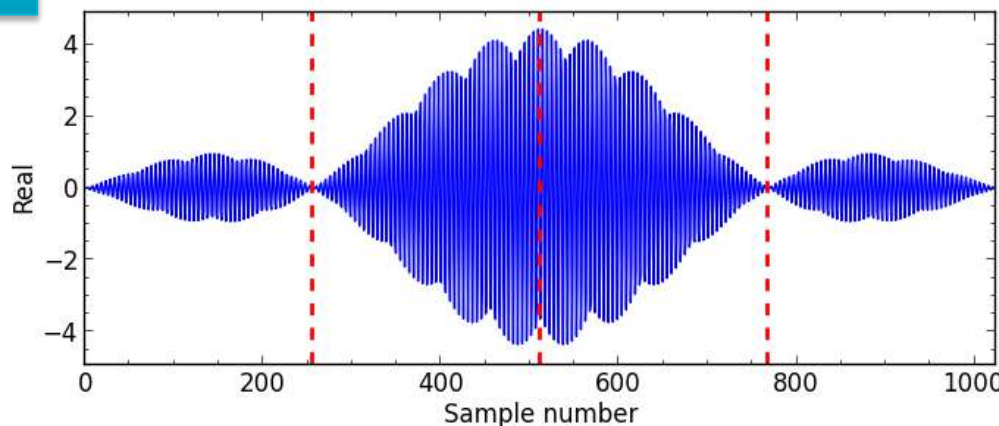


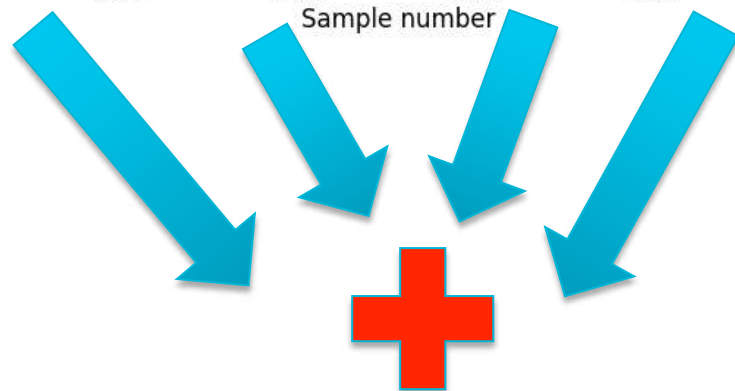
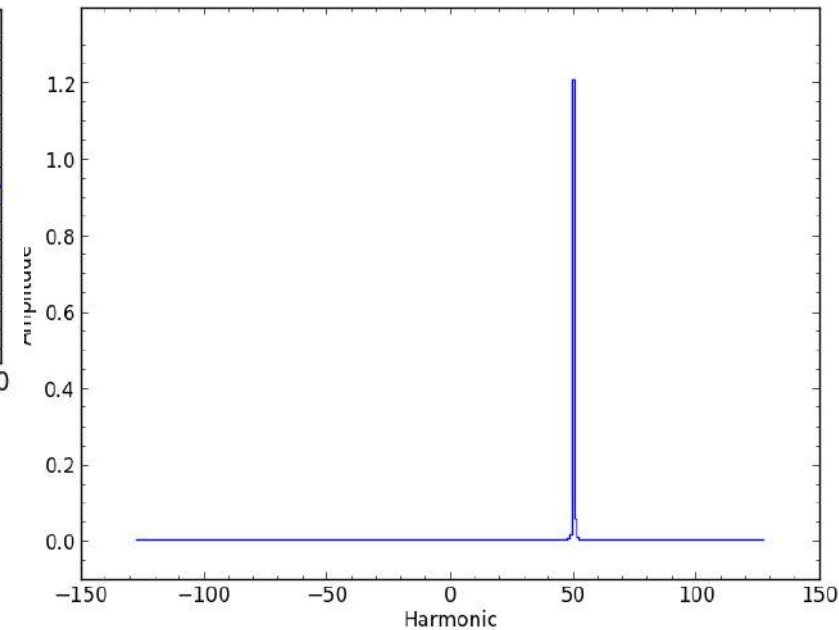
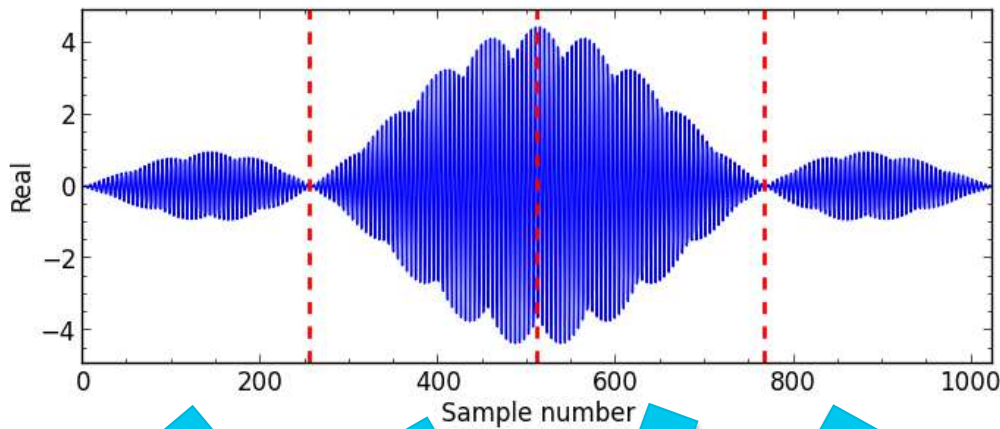
Filter

(Sinc in this particular example)



Filtered data samples





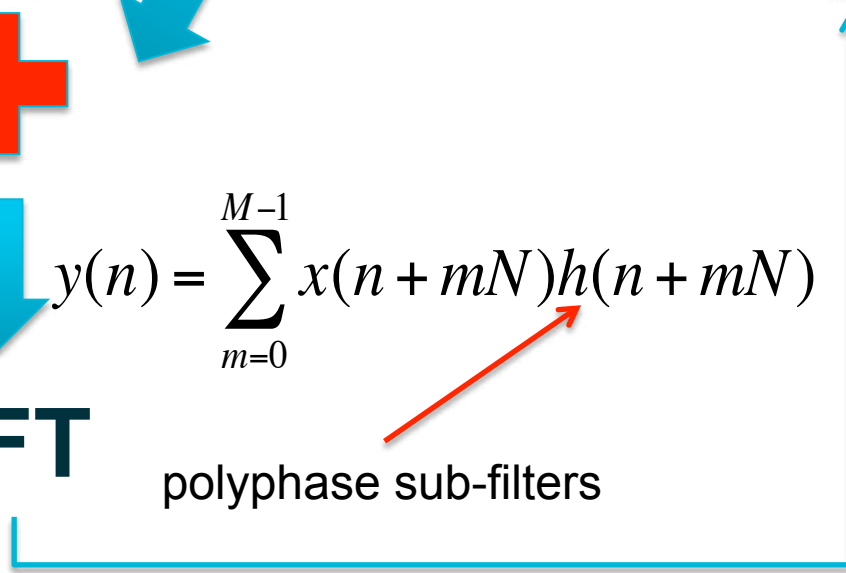
In this example:

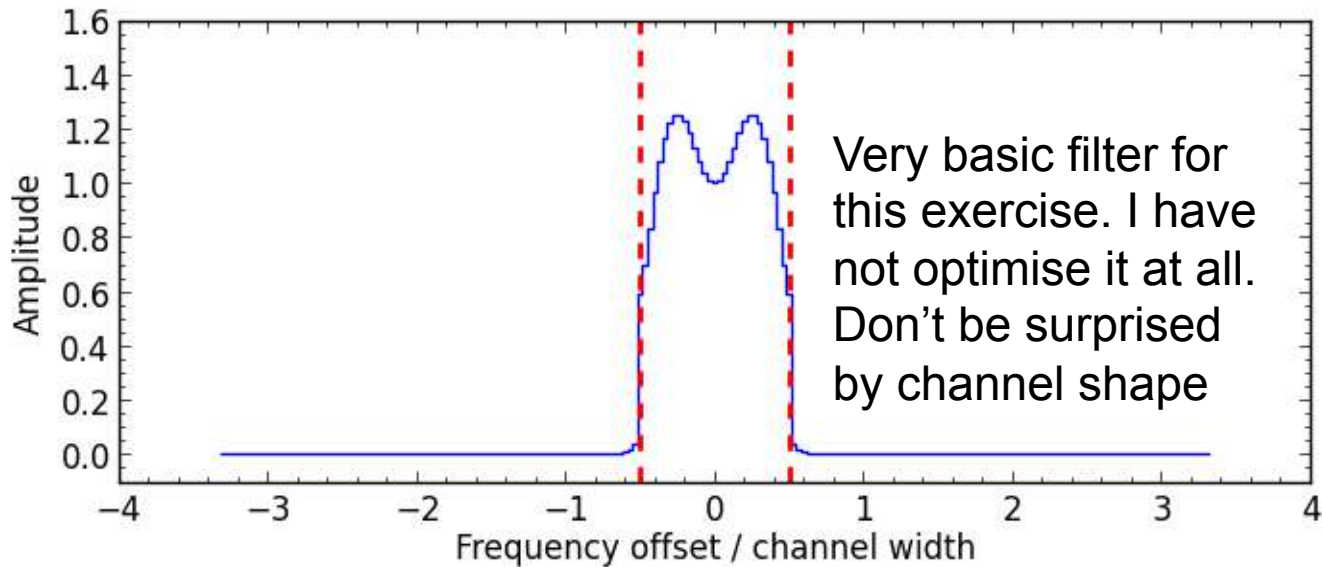
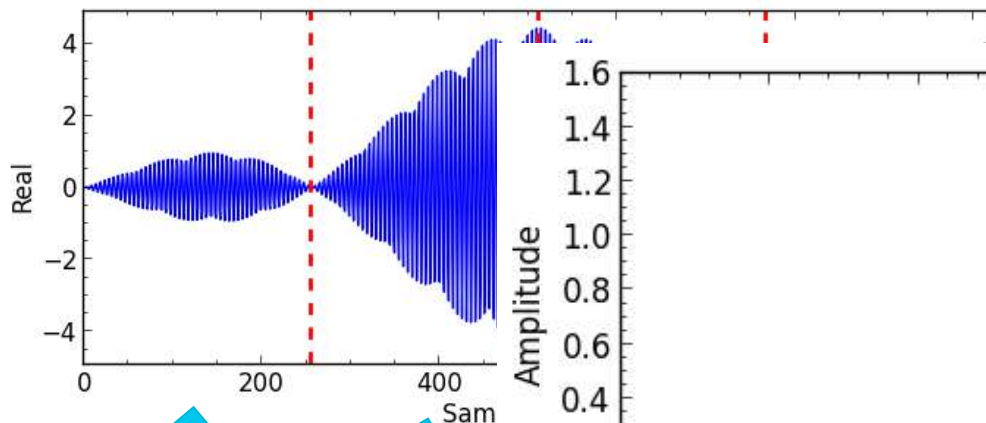
$N = 256$
 $M = 4$

$$y(n) = \sum_{m=0}^{M-1} x(n + mN)h(n + mN)$$

FFT

polyphase sub-filters





In this example:

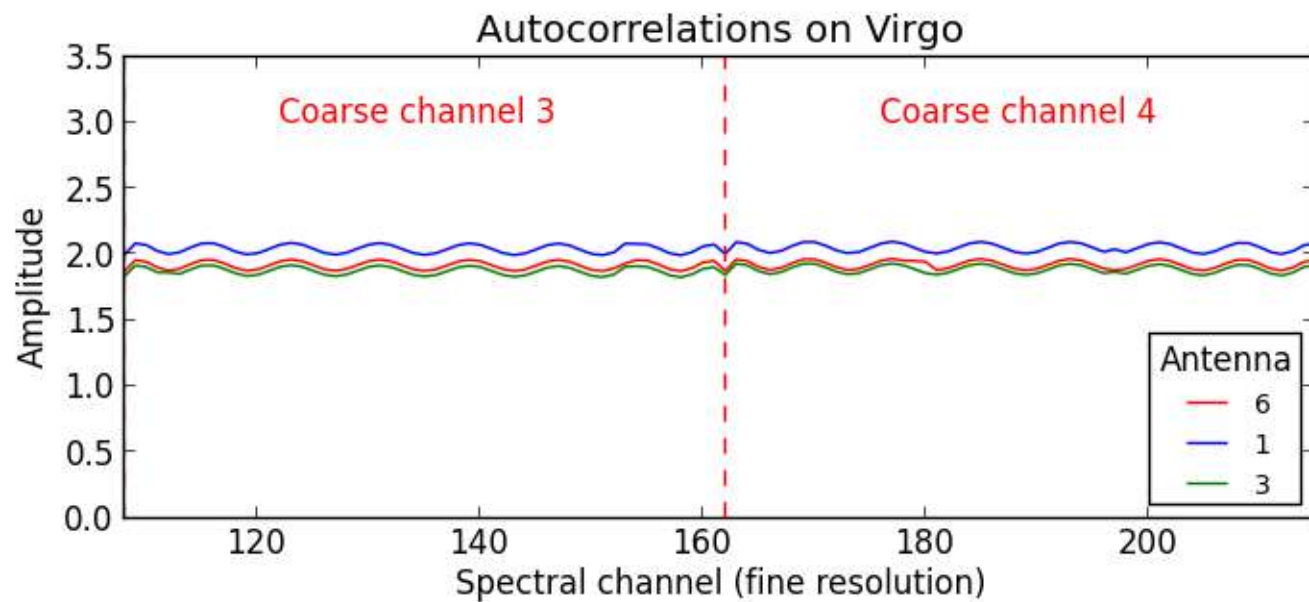
$N = 256$
 $M = 4$



$$y(n) = \sum_{m=0}^{M-1} x(n + mN)h(n + mN)$$

FFT

polyphase sub-filters



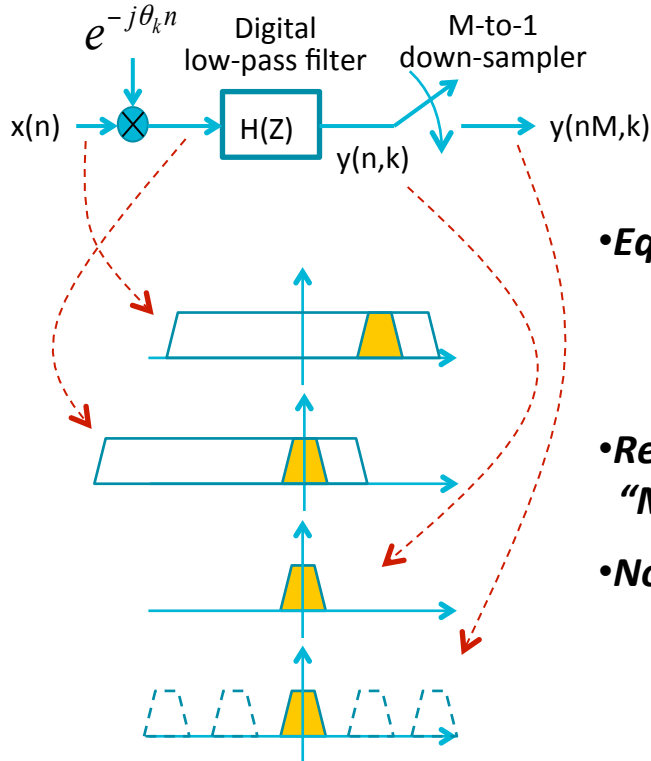
**But filter does
cause a ripple
(can be corrected)**

Early BETA tests



Polyphase decomposition: engineer's view

Standard single-channel down converter



•Equivalency Theorem

- Exchange mixer and low-pass filter with a band-pass filter and a mixer.

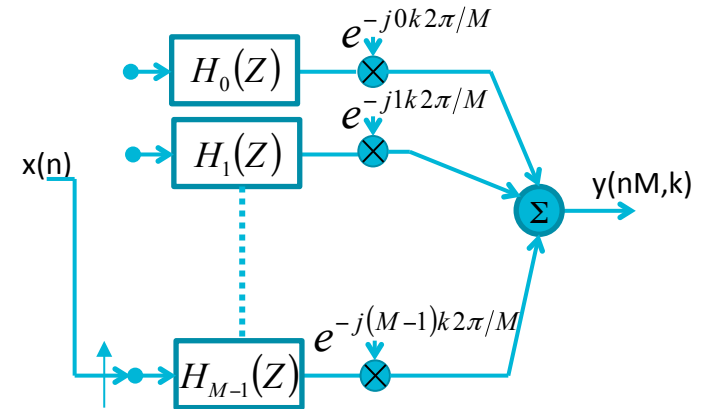
•Re-write the band-pass filter in "M-path form"

•Noble Identity

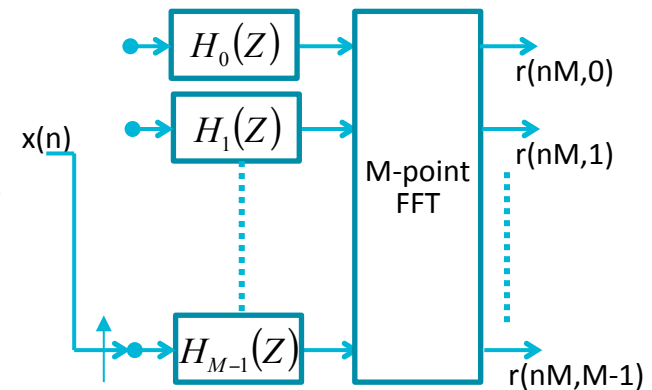
- Move a down-sampler back through a digital filter

$$H(Z^M) \downarrow M = (\downarrow M) H(Z)$$

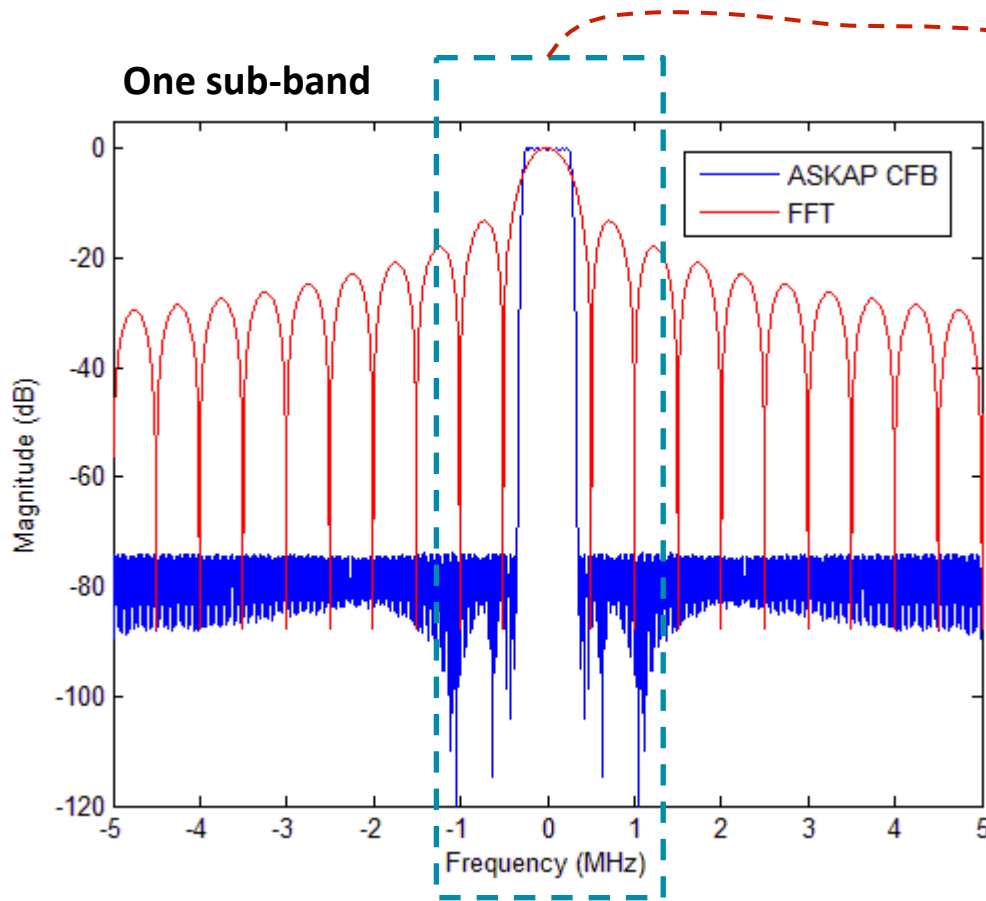
M-path Polyphase down converter



M-path Polyphase channeliser

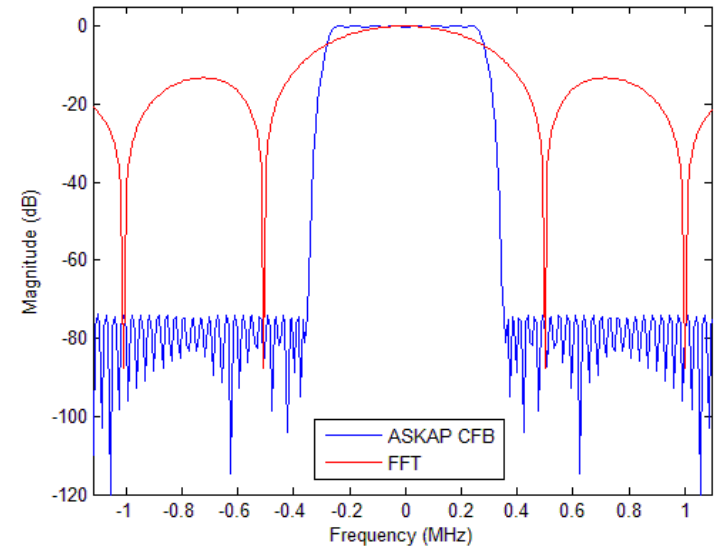


Filterbanks: FFT vs Polyphase Filters

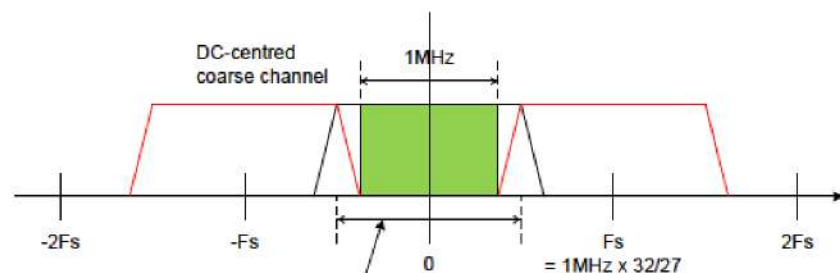


768-point FFT

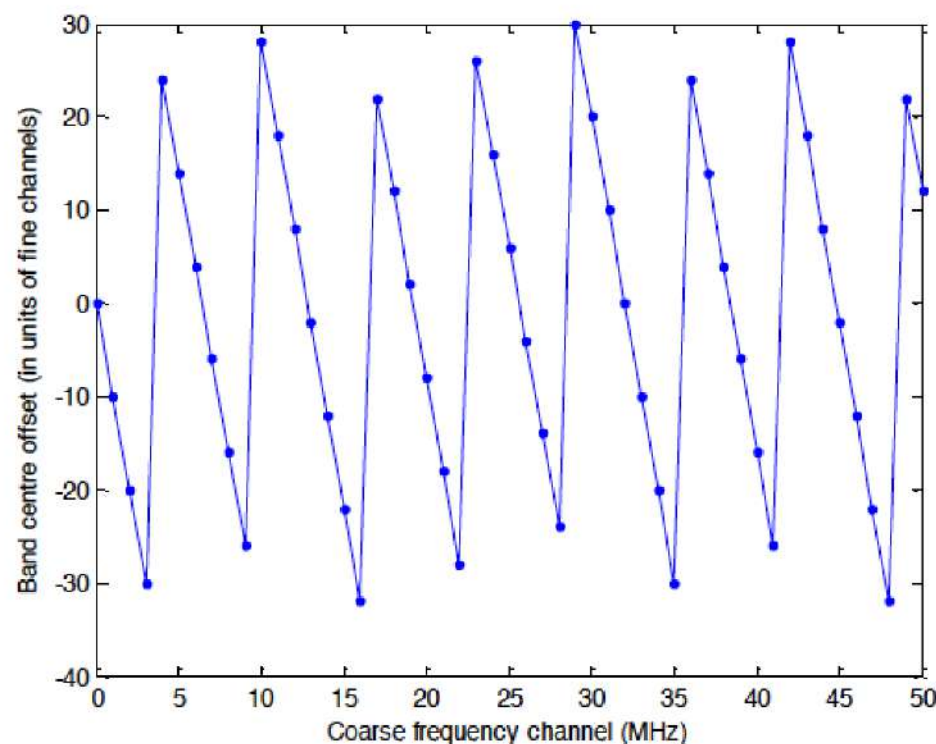
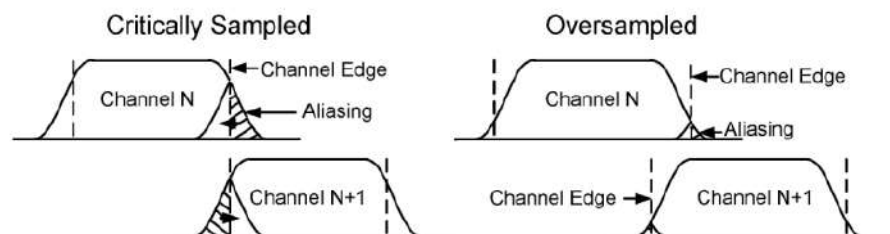
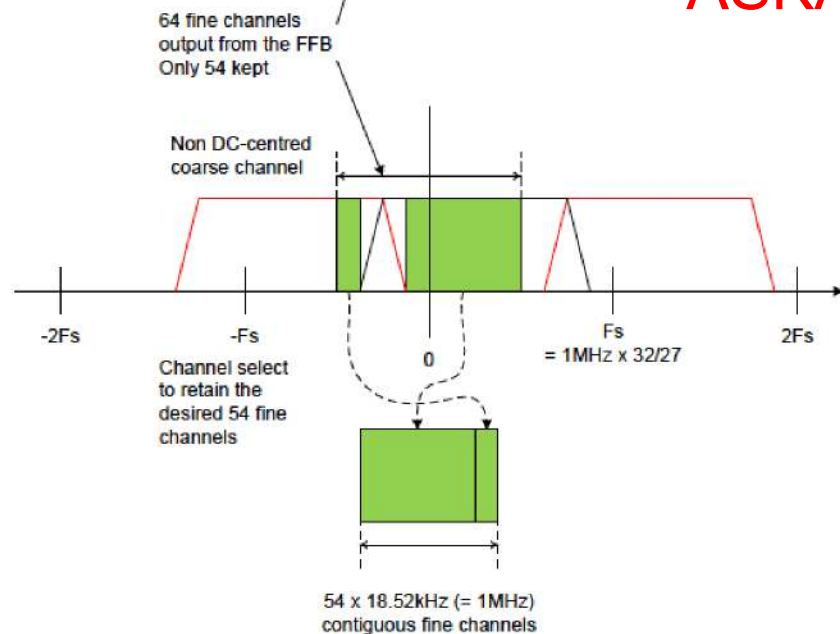
12,288-tap polyphase filter
+ 768-point FFT



A few words on oversampled PFBs



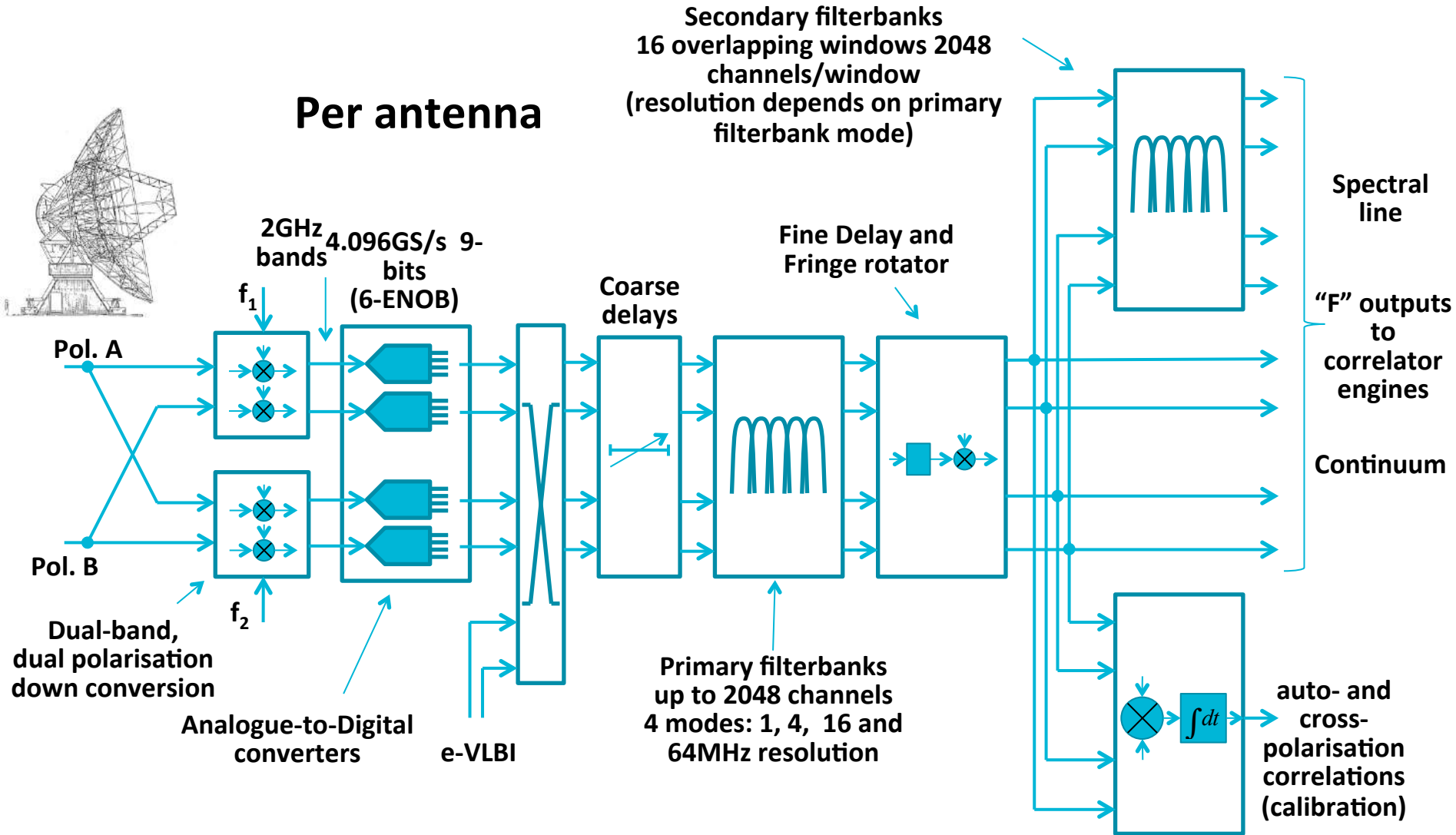
ASKAP



ATCA + CABB see Wilson et al. (2011, MNRAS, 416, 832)



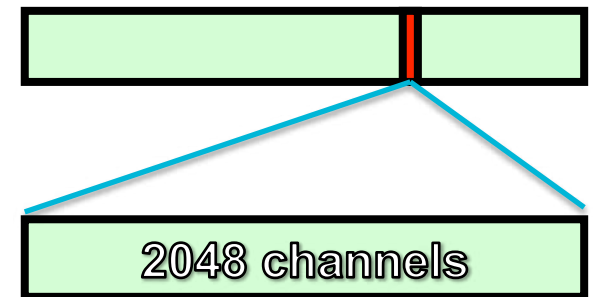
Compact Array Broadband Backend (CABB)



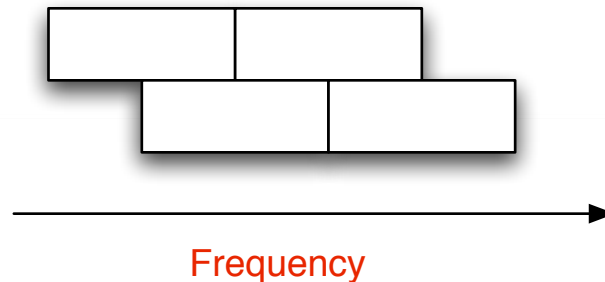
Zoom modes: high spectral resolution



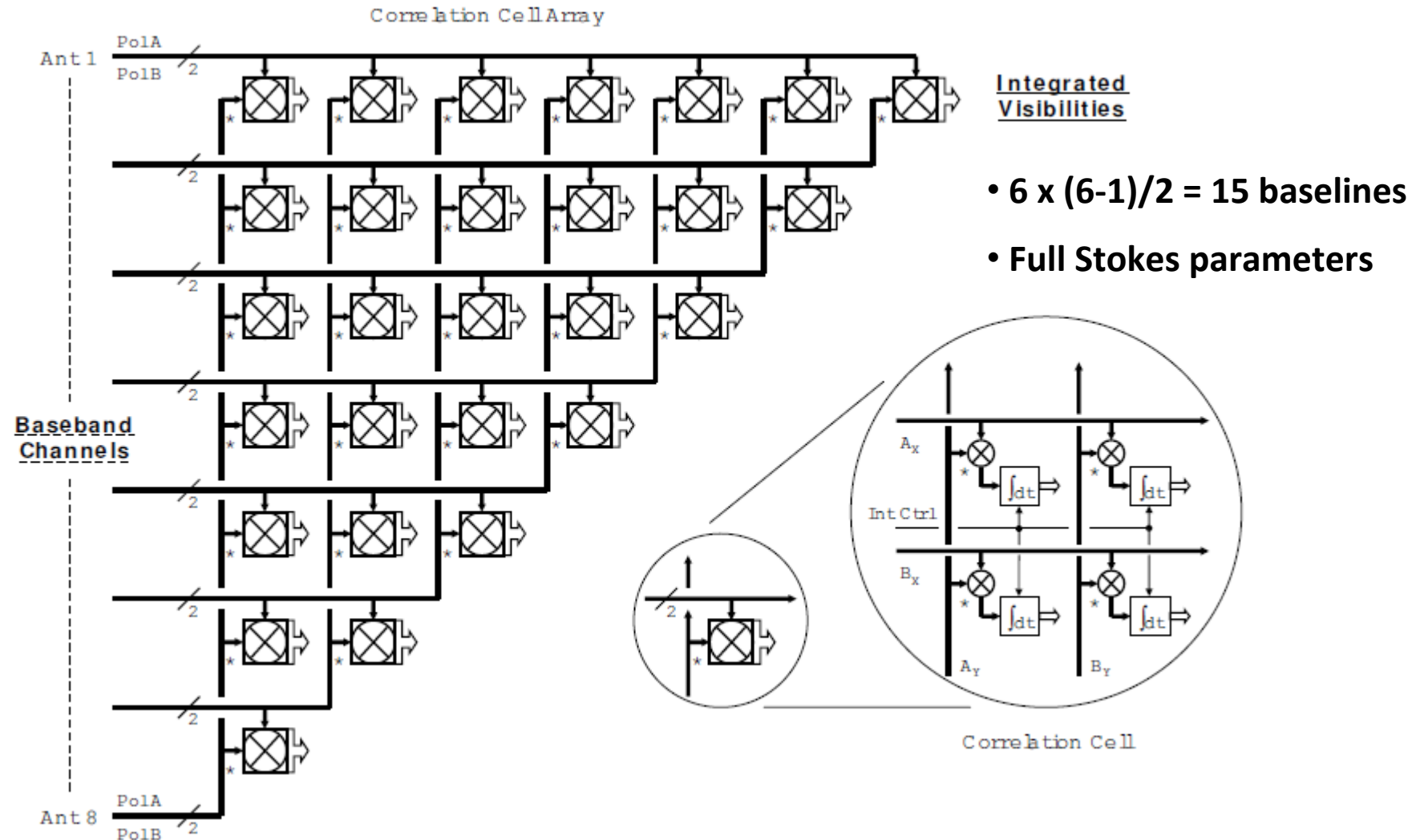
- Up to 16 “zoomed in” channels per each 2 GHz window (user selected)
- Positioned in steps equal to half of the wide-band spectral resolution (i.e. 0.5 MHz or 32 MHz)
- Each zoom window has 2048 spectral channels



Stitching zoom windows:



CABB Correlator – correlation engines

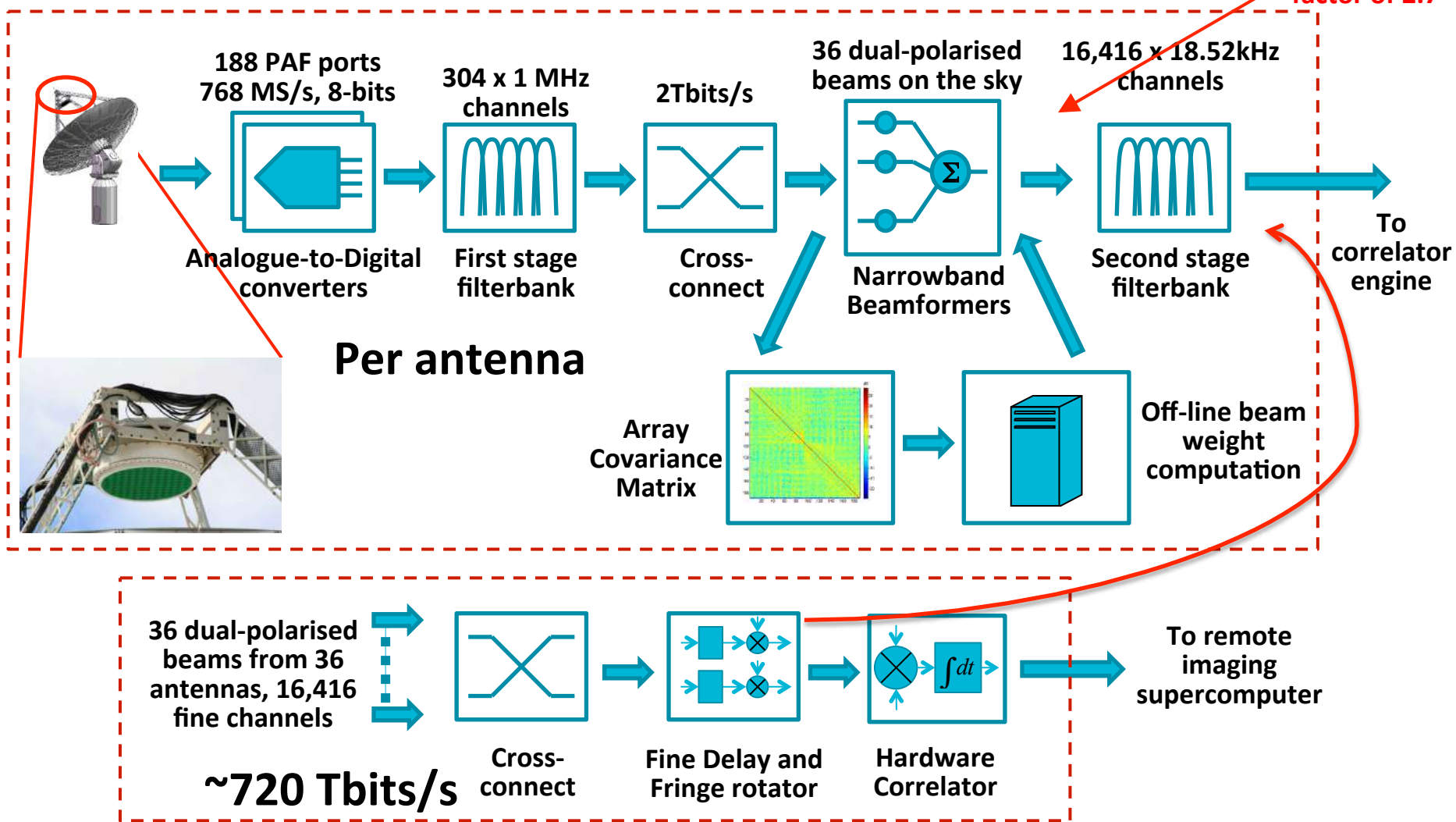


Australian Square Kilometre Array Pathfinder

- Has beamformer before the correlator (needs channelisation)



ASKAP (BETA) digital back-end



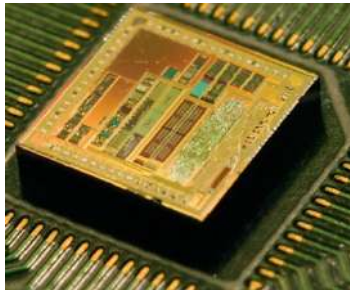
Calculation Engines: so many choices...

Hard-wired logic

Stored (programmed) logic

ASIC's

Application-Specific Integrated Circuit



- EVLA
- ALMA

FPGA's

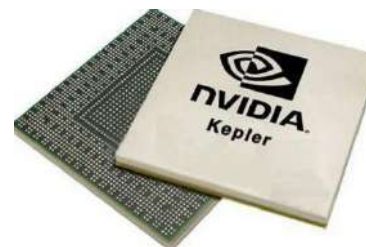
Field Programmable Gate Array



- CABB
- ASKAP

GPU's

Graphics Processing Unit



- MWA
- MeerKAT

CPU's/DSP's

Central Processing Unit/ Digital Signal Processor



- DiFX



- Less flexible
- Lower power/computation
- Higher initial development

- More flexible
- Higher power/computation
- Lower initial development

Further Reading...

- Radio Astronomy:
 - H. C. Ko, "Coherence Theory in Radio-Astronomical Measurements," *IEEE Trans. Antennas & Propagation*, pp. 10-20, Vol. AP-15, No. 1, Jan. 1967.
 - G. B. Taylor, G. L. Carilli and R. A. Perley, *Synthesis Imaging in Radio Astronomy II*, Astron. Soc. Pac. Conf. Series, vol. 180, 2008.
- CABB
 - W. E. Wilson, et. al. "The Australia Telescope Compact Array Broadband Backend (CABB): Description & First Results," *Mon. Not. R. Astron. Soc.*, Feb. 2011
- ASKAP
 - D. R. DeBoer, et.al, "Australian SKA Pathfinder: A High-Dynamic Range Wide-Field of View Survey Telescope," *Proc. IEEE*, 2009.
- Filter Banks
 - R. E. Crochiere and L. R. Rabiner *Multirate Digital Signal Processing*, Prentice Hall, 1983.
 - F. J. Harris, *Multirate Signal Processing for Communication Systems*, Prentice Hall, 2008.
 - P. P. Vaidyanathan, *Multirate Systems And Filter Banks*, Prentice Hall, 1992.
- Beamforming
 - B. D. Van Veen and K. M. Buckley, "Beamforming: A Versatile Approach to Spatial Filtering," *IEEE ASSP Magazine*, April 1988

Thank you

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