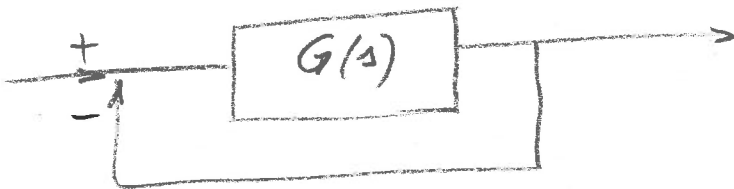


STABILITY ANALYSIS IN FREQ. DOMAIN

Evaluate stability of CL system
by analysing OL system in freq. domain



Two methods:

- Nyquist criterion
- Gain & phase margins

NYQUIST ANALYSIS
OF FEEDBACK STABILITY.

Nyquist circuit

$$G(s) = \frac{1}{(s-p_1)(s-p_2)(s-p_3)}$$

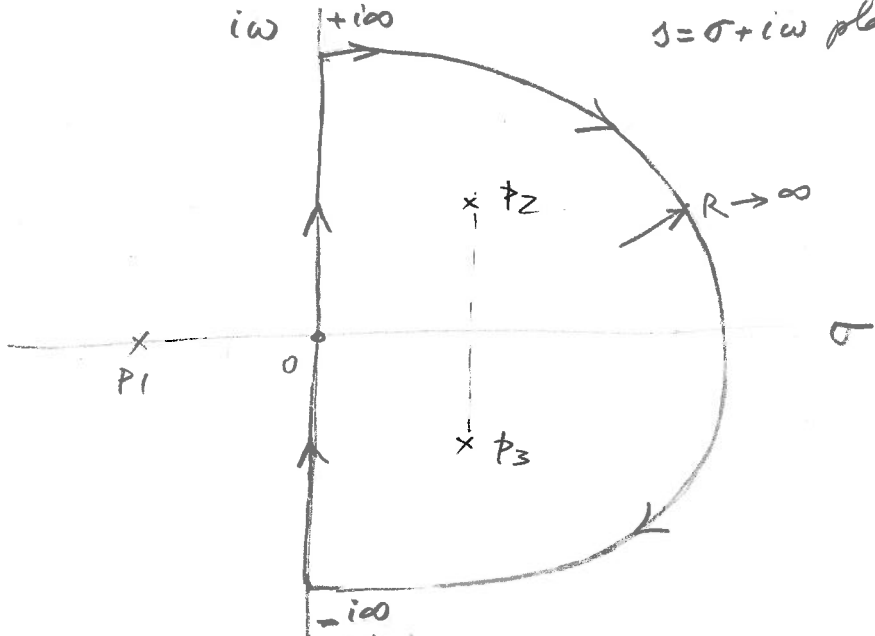
Assume

$$p_1 \in \mathbb{R}, p_1 < 0$$

$$p_2, p_3 \in \mathbb{C} \text{ in R.H.S}$$

$$p_3 = \bar{p}_2$$

$$s = \sigma + i\omega \text{ plane}$$

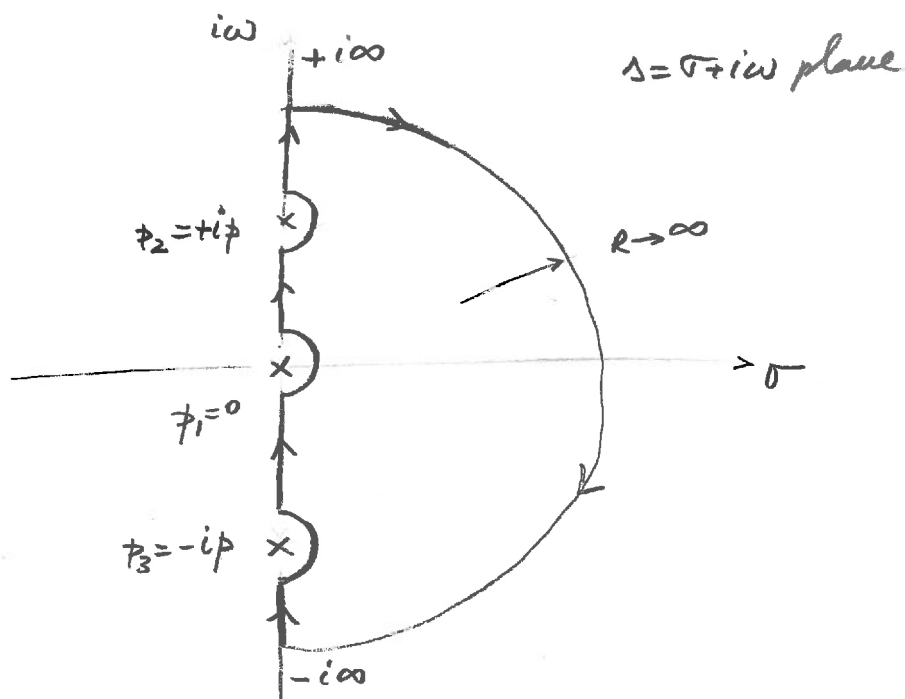


Nyquist circuit is :

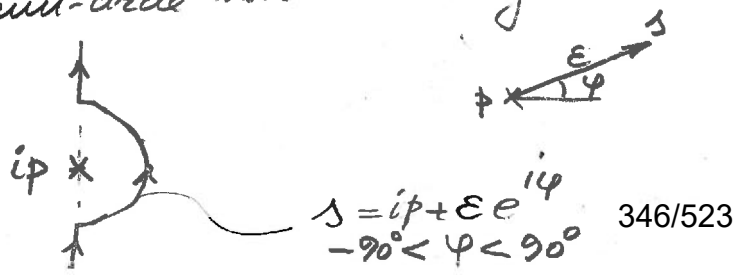
- semi-circle path with $R \rightarrow \infty$
- in the R.H.S of s -plane
- along vertical axis from $-i\infty$ to $+i\infty$
- traveled clockwise (CW)

Nyquist circuit with poles on the vertical axis

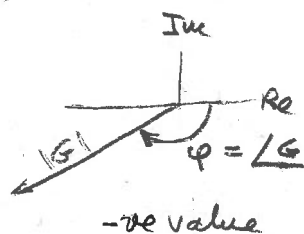
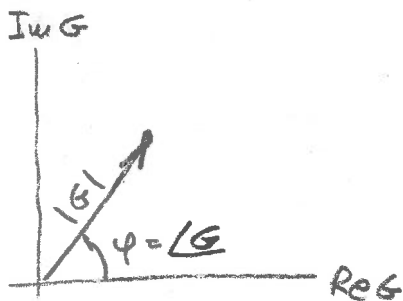
$$G(s) = \frac{1}{s(s-ip)(s+ip)}$$



The poles on the vertical axis must be excluded. We travel around them in the RHS. To do so, we follow a small semi-circle with vanishing radius ϵ



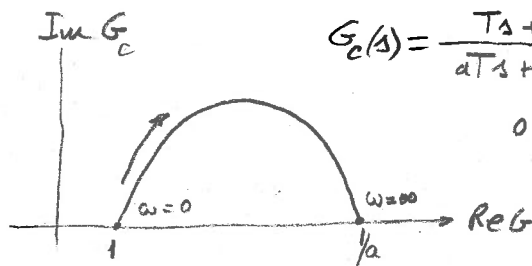
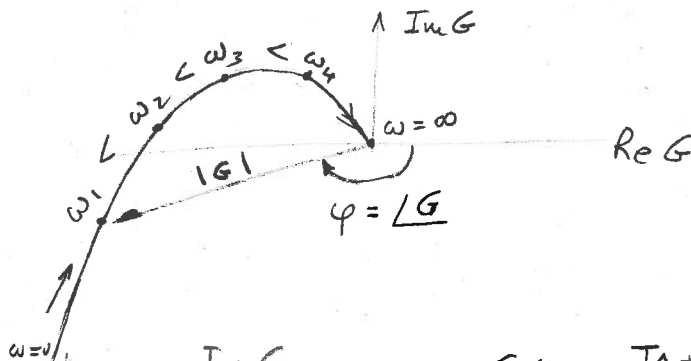
$$G(i\omega) = \text{Re } G(i\omega) + i \text{Im } G(i\omega)$$



$$-\pi < \text{angle}(G) < \pi$$

$$-180^\circ \qquad 180^\circ$$

(see MATLAB Help)



$$G_c(s) = \frac{T_s + 1}{aT_s + 1}$$

$$0 < \alpha < 1$$

Run MATLAB examples.

NPP

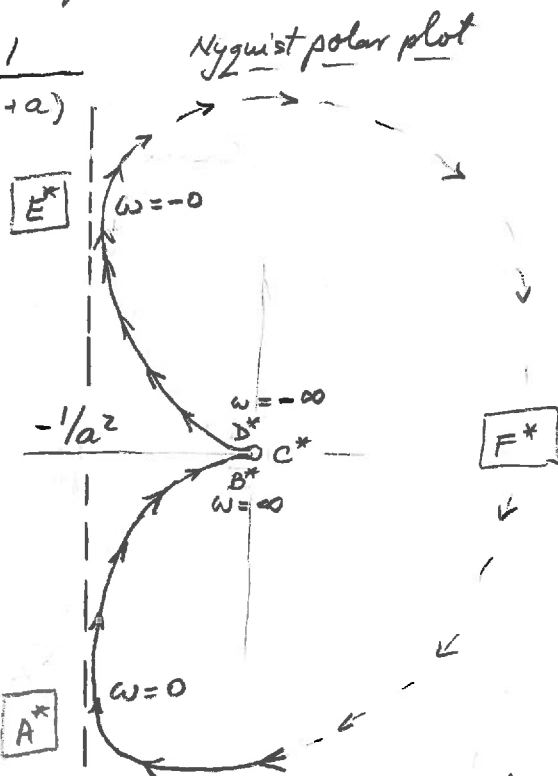
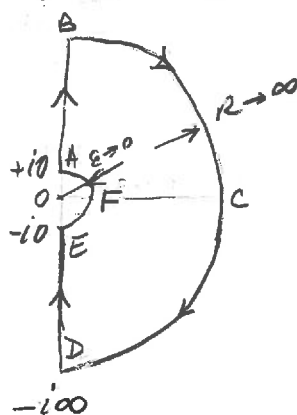
NYQUIST POLAR PLOTS

"s follows N-circuit ; G(s) follows N-polar plot"

Example: $G(s) = \frac{1}{s(s+a)}$

2 poles: $p_1 = 0, p_2 = -a$

Nyquist circuit



- The variable s follows the Nyquist circuit A B C D E F A clock wise in the s-plane
- The function G(s) follows the resulting circuit in the G-plane

In this example, we distinguish 4 segments to be analyzed individually and then assembled in one continuous circuit

2
NPP

Segment $(+i\varepsilon, +iR)$
 $\varepsilon \rightarrow 0 \quad R \rightarrow \infty$



$$G(s) = \frac{1}{s(s+a)} = \frac{1}{as} - \frac{1}{a(s+a)}$$

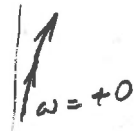
$$\boxed{A} \quad G(i\varepsilon) = \frac{1}{ai\varepsilon} - \frac{1}{a(i\varepsilon+a)} \quad \varepsilon \ll a$$

$$G(i\varepsilon) \approx -\frac{1}{a^2} - i\frac{1}{a\varepsilon}$$

$$G(+i0) = \lim_{\varepsilon \rightarrow 0} G(i\varepsilon) = -\frac{1}{a^2} - i\infty$$

$$|G(+i0)| = \infty$$

$$\angle G(+i0) = \angle -i = -90^\circ$$

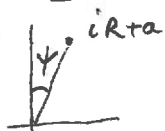


$$\boxed{B} \quad G(iR) = \frac{1}{iR(iR+a)}$$

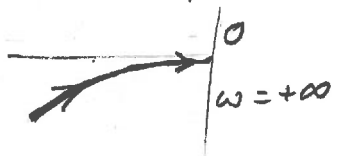
$$|G(iR)| \approx \frac{1}{R^2} \xrightarrow{R \rightarrow \infty} 0$$

$$\angle G(iR) = -[\angle iR + \angle iR+a]$$

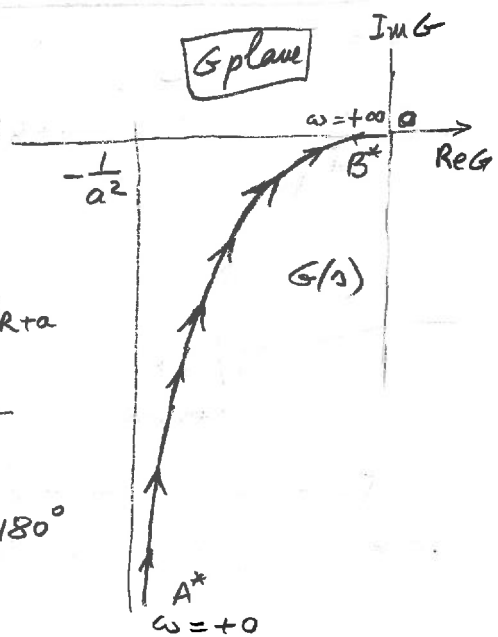
$$\varphi = \tan^{-1} \frac{a}{R}$$



$$\begin{aligned} \angle G(iR) &= -(90^\circ + 90^\circ - \psi) \\ &= -180^\circ + \psi \end{aligned} \quad \begin{matrix} R \rightarrow \infty \\ \psi \rightarrow 0 \end{matrix} \rightarrow -180^\circ$$



G_{plane}



3
NPP Segment $(iR, R, -iR)$ Big circle $R \rightarrow \infty$

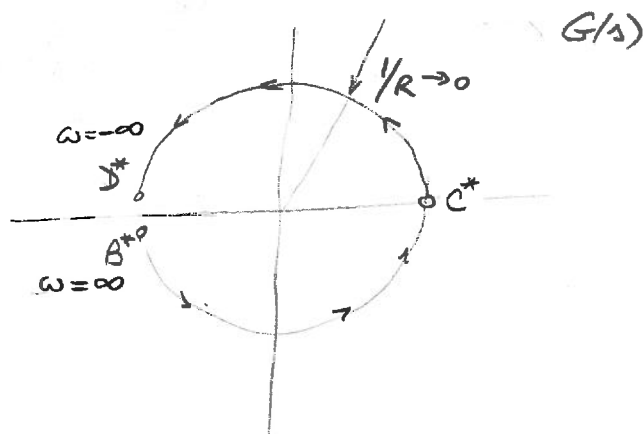
$$s = Re^{i\varphi}, \quad \varphi \in (90^\circ \ 0 \ -90^\circ)$$

$$G(s) = \frac{1}{Re^{i\varphi}(Re^{i\varphi} + a)}$$

$$R \rightarrow \infty \approx \frac{1}{R^2 e^{2i\varphi}} = \frac{1}{R^2} e^{-i2\varphi}$$

$$\angle G(s) = -2\varphi$$

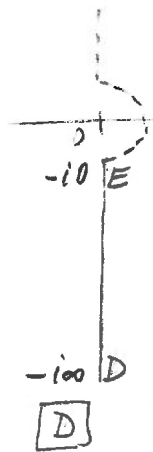
$$\angle G(s) \in \begin{pmatrix} -180^\circ & 0^\circ & 180^\circ \\ B^* & C^* & D^* \end{pmatrix}$$



NP

Segment $(-iR, -i\epsilon)$
 $R \rightarrow \infty \quad \epsilon \rightarrow 0$

s-plane



$$G(s) = \frac{1}{s(s+a)} = \frac{1}{as} - \frac{1}{a(s+a)}$$

$$G(s) \Big|_{s=-iR} = \frac{1}{-iR(-iR+a)}$$

$$|G(s)| = \frac{1}{R^2} \xrightarrow{R \rightarrow \infty} 0$$

$$\angle G(s) = -(\angle -iR + \angle -iR+a)$$

$$= -[-90^\circ + (-90^\circ + \gamma)] = 180^\circ - \gamma \xrightarrow{R \rightarrow \infty, \gamma \rightarrow 0} 180^\circ$$



$$\gamma = \tan^{-1} \frac{a}{R}$$



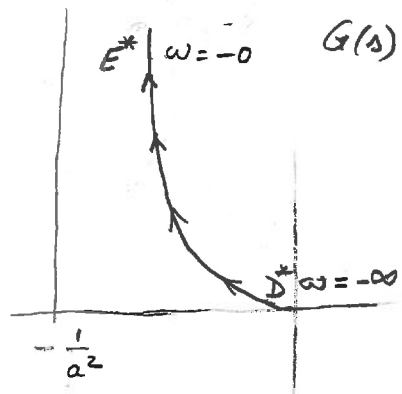
$$\boxed{E} \quad G(-i\epsilon) = -\frac{1}{a(-i\epsilon)} - \frac{1}{a(-i\epsilon+a)} = \frac{i}{a\epsilon} - \frac{1}{a^2}$$

$$G(-i0) = \lim_{\epsilon \rightarrow 0} G(-i\epsilon)$$

$$G(-i0) = -\frac{1}{a^2} + \frac{i}{a\epsilon} \Big|_{\epsilon \rightarrow 0} = -\frac{1}{a^2} + i\infty$$

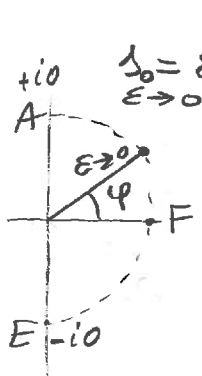
$$|G(-i0)| = \infty$$

$$\angle G(-i0) = 90^\circ$$



S
NPP

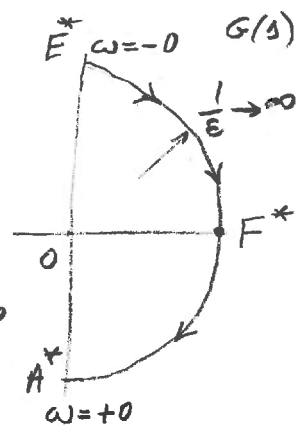
Segment $(-i\varepsilon, \varepsilon, i\varepsilon)$ small circle $\varepsilon \rightarrow 0$



$\Delta_0 = \varepsilon e^{i\varphi}, \varphi \in (-90^\circ, 0, 90^\circ)$
 $\varepsilon \rightarrow 0$

$$G(s) = \frac{1}{\varepsilon e^{i\varphi} (\varepsilon e^{i\varphi} + a)}$$

$\approx \frac{1}{a \varepsilon e^{i\varphi}} = \frac{1}{a \varepsilon} e^{-i\varphi}$
 $\varepsilon \rightarrow 0$



$$|G(\Delta_0)| = \frac{1}{a \varepsilon} \xrightarrow{\varepsilon \rightarrow 0} \infty$$

$\angle G(\Delta_0) = -\varphi \in \begin{pmatrix} 90^\circ & 0^\circ & -90^\circ \\ E^* & F^* & A^* \end{pmatrix}$
 $\varphi = -90^\circ \quad \varphi = 0^\circ \quad \varphi = +90^\circ$

NYQUIST STABILITY CRITERION



Find stability of $G_{CL} = \frac{G}{1+G}$ by analyzing the polar plot of $G(s)$ as s follows the Nyquist circuit.

$$G_{CL}(s) = \frac{G(s)}{1+G(s)}$$

Poles $G_{CL}(s)$

LHS	$j\omega$	RHS
x		

Poles $G_{CL}(s) \Rightarrow$ zeros of $1+G(s)$

stable

unstable!

$Z \leq 0$

$Z > 0$

Nyquist criterion:

$$Z = P + N \leq 0$$

P = number of $G(s)$ poles in RHS

$$G(s) = \frac{B(s)}{A(s)} \rightarrow P_1, P_2, \dots$$

N = number of clockwise encirclements of the $(-1, 0)$ point as s follows the Nyquist path (circuit)

Z = number of zeros of $1+G(s)$ in RHS

STABLE if $Z \leq 0$

Example: aircraft roll model

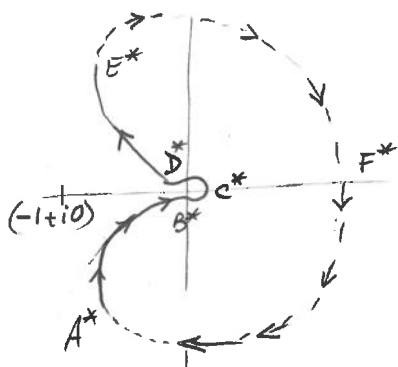
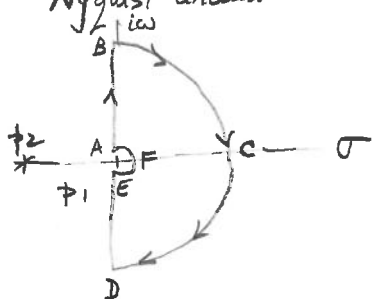
$$G(s) = \frac{114}{10s^2 + 4s} = \frac{114}{s(10s+4)}$$

$$P_1 = 0, P_2 = -2/5 : P = 0$$

Recall $G(s) = \frac{1}{s(s+a)} = \frac{1}{s^2 + sa}$

no poles in RHS

Nyquist circuit



$P = 0$ no poles in RHS

$N = 0$ no encirclements of $(-1+j0)$ point

$$Z = N + P = 0 \quad \text{NO CL poles in RHS}$$

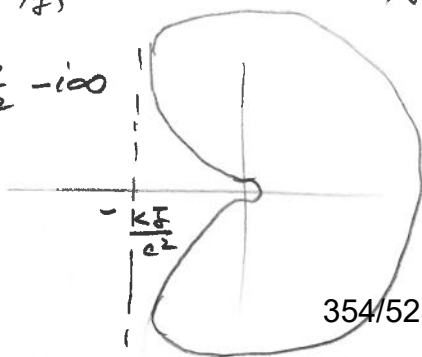
Conclusion: system is unconditionally stable

Note: $G = \frac{K}{Js^2 + cs} = \frac{K}{J} \frac{1}{s(1 + c/J)} = B \frac{1}{s(1 + a/J)} \quad B = \frac{K}{J} \quad a = \frac{c}{J}$

$$G(j\omega) = B \left(-\frac{1}{\omega^2} - j\infty \right) = -\frac{B}{\omega^2} - j\infty$$

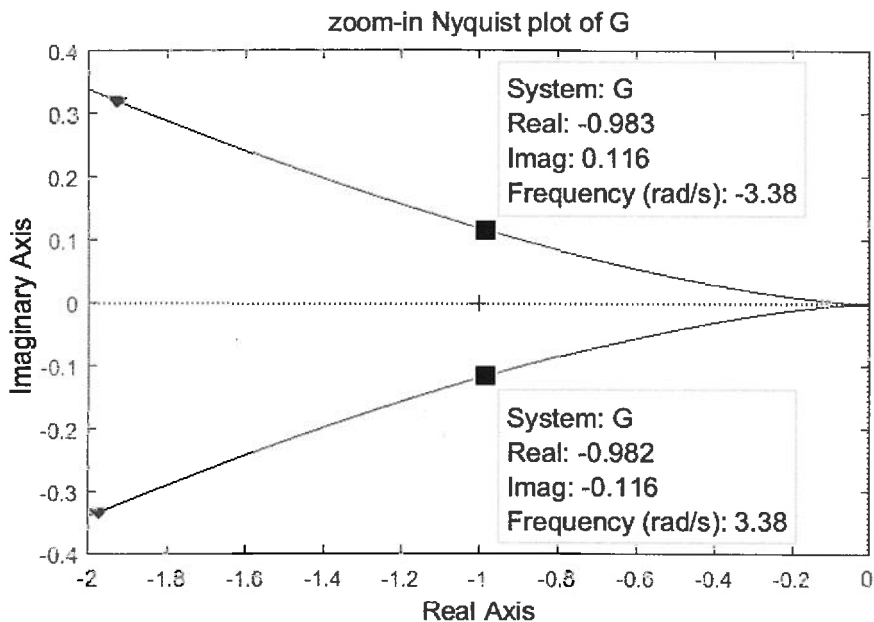
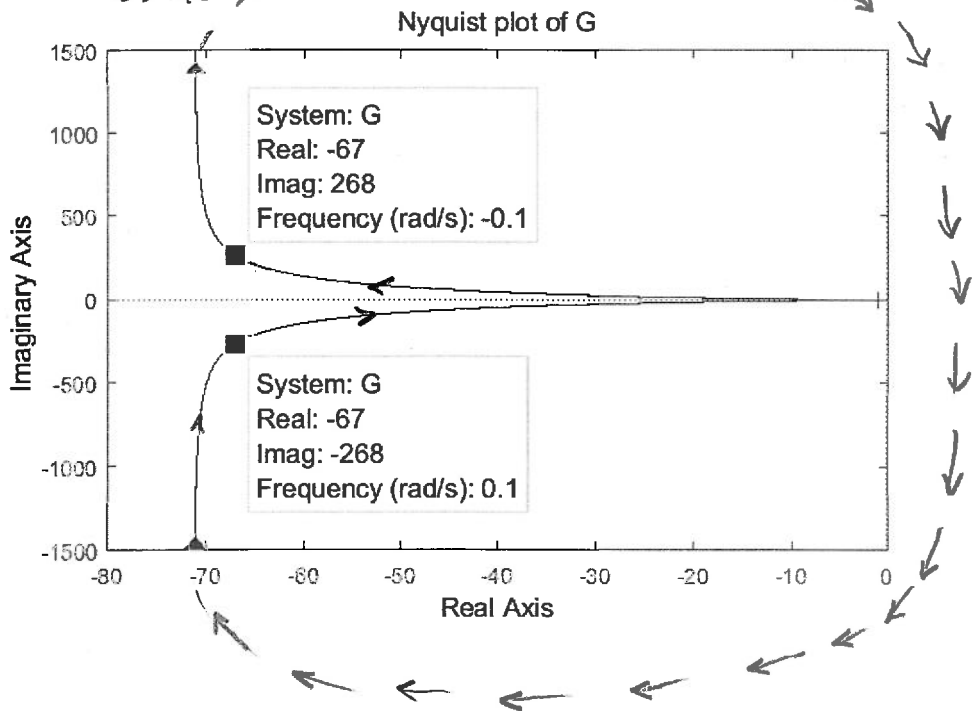
$\omega \rightarrow 0$

$$\frac{B}{\omega^2} = \frac{K}{J} \frac{J^2}{c^2} = \frac{KJ}{c^2}$$



Aircraft

$$G(s) = \frac{114}{10s^2 + 4s}$$



Example AM.8

Given: $G = K \frac{1}{s-1}$



Note: pole in RHS, $s=1$

Find: critical value of K for stability using Nyquist criterion

Solution: do Nyquist plot of $G(s)$

$$G(0) = -\frac{K}{1} = -K$$

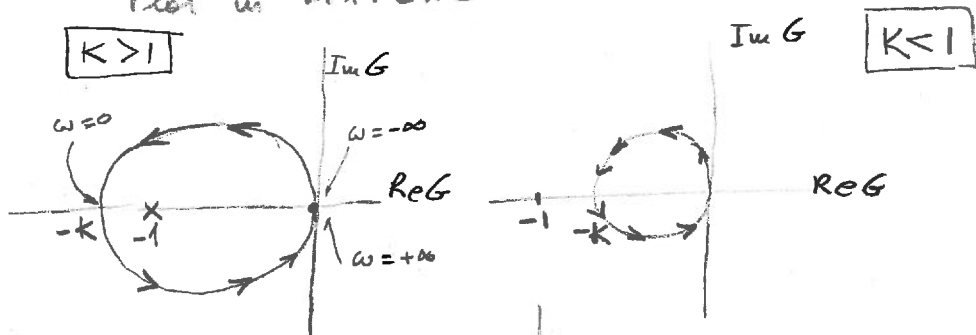
$$G(i\infty) = \lim_{\omega \rightarrow \infty} \frac{K}{i\omega - 1} \approx \frac{K}{i\omega}, \quad |G(i\omega)| = 0$$

$$\angle G(i\omega) = -90^\circ$$

$$G(-i\infty) \rightarrow |G(-i\omega)| = 0$$

$$\angle G(-i\omega) = 90^\circ$$

Plot in MATLAB



$$N = -1 \text{ (counterclockwise)}$$

$$P = 1 \text{ (} s=1 \text{ pole of } G \text{ in RHS)}$$

$$Z = -1 + 1 = 0$$

STABLE for $K > 1$

$$N = 0$$

$$P = 1$$

$$Z = 1$$

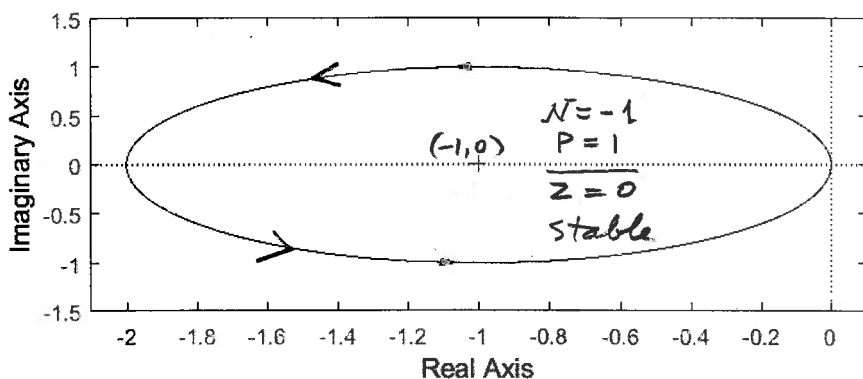
UNSTABLE for $K < 1$

CRITICAL VALUE: $K = 1$

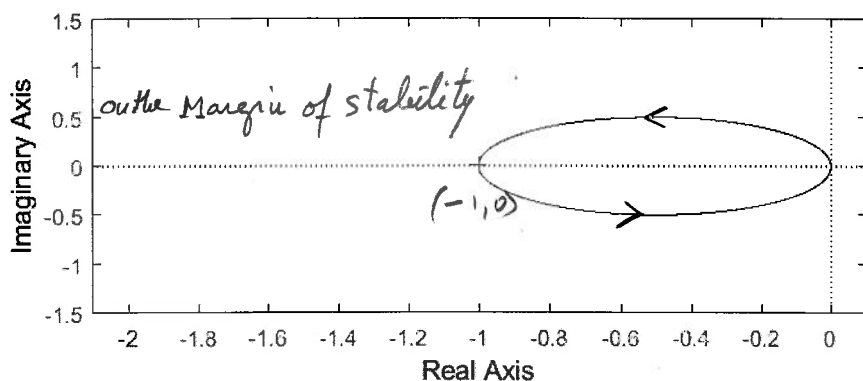
A 11.8

K=2 STABLE

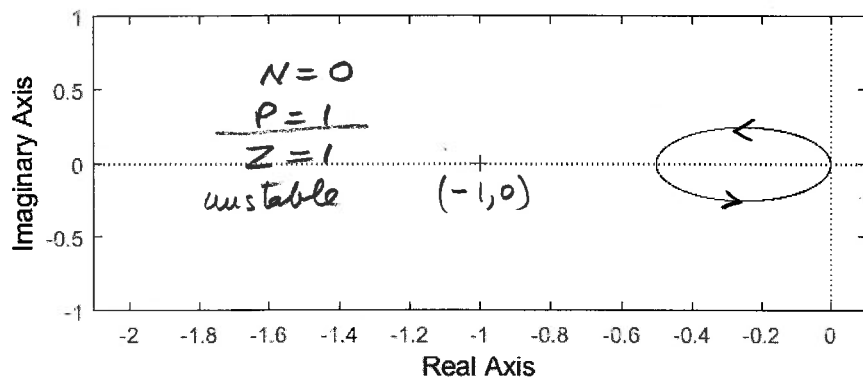
$$G = \frac{2}{s-1}$$



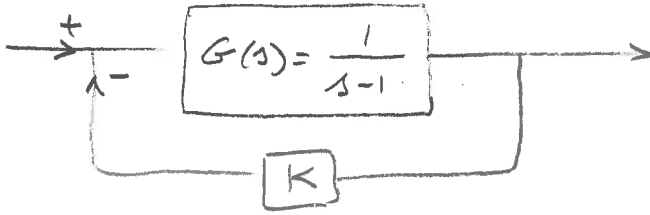
K=1



K=0.5 UNSTABLE; Ex.A11-8



A11.8

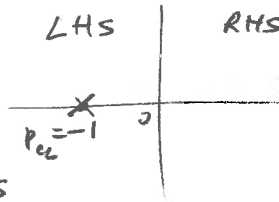


$$G_{CL} = \frac{G}{1+GH} = \frac{1}{s-1+K} = \frac{1}{s-(1-K)}$$

$$p_{CL} = 1-K$$

$$K=2, p_{CL} = -1$$

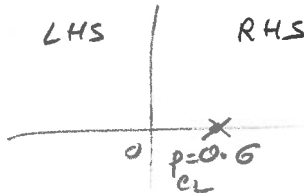
$p_{CL} \in \text{LHS}$



STABLE

$$K=0.4, p_{CL} = 0.6$$

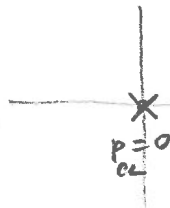
$p_{CL} \in \text{RHS}$



UNSTABLE

$$p_{CL} = 1-K_{cr} = 0$$

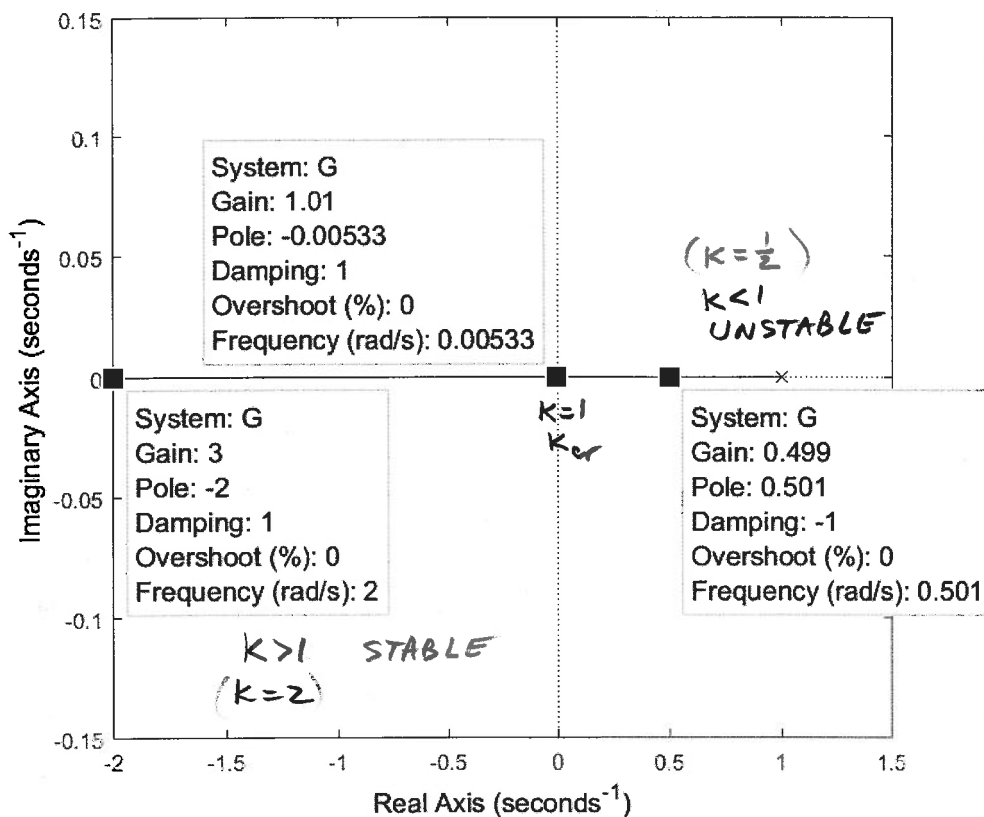
$$K_{cr} = 1$$



critical
K value

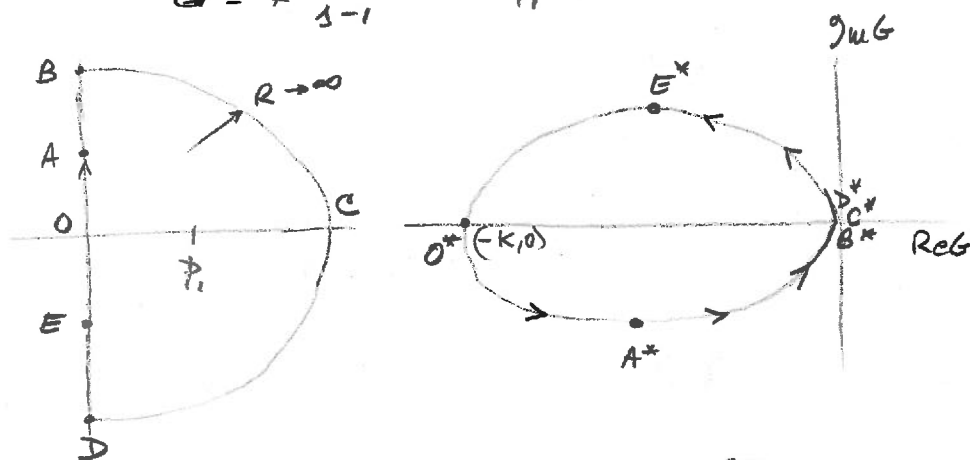
A11.8

Root Locus



Also MANUAL PLOT.

$$G = K \frac{1}{s-1} \quad \tau_1 = 1$$



$$O: \quad G(0) = K \frac{1}{0-1} = -K = K e^{i\pi}$$

$$A = i: \quad G(A) = K \frac{1}{i-1} = -K \frac{1+i}{2} = K \frac{\sqrt{2}}{2} e^{i(\pi + \frac{\pi}{4})}$$

$$\angle G(A) = \frac{5\pi}{4}$$

$$B = iR: \quad G(B) = K \frac{1}{iR-1} = -K \frac{1+iR}{1+R^2} = \frac{K}{\sqrt{1+R^2}} e^{i(\pi + \varphi)}$$

$$\varphi = \tan^{-1} R \xrightarrow{R \rightarrow \infty} \frac{\pi}{2}$$

$$|G(B)| \xrightarrow{R \rightarrow \infty} 0$$

$$\angle G(B) \xrightarrow{R \rightarrow \infty} \frac{3\pi}{2}$$

$$C = R: \quad G(C) = K \frac{1}{R-1} \xrightarrow{R \rightarrow \infty} 0$$

$$D = -iR: \quad G(D) = K \frac{1}{-iR-1} = -K \frac{1-iR}{1+R^2} = \frac{K}{\sqrt{1+R^2}} e^{i(\pi - \varphi)}$$

$$\varphi = \tan^{-1} R \xrightarrow{R \rightarrow \infty} \frac{\pi}{2}$$

$$|G(D)| \xrightarrow{R \rightarrow \infty} 0$$

$$\angle G(D) \xrightarrow{R \rightarrow \infty} \pi - \frac{\pi}{2} = \frac{\pi}{2}$$

A11.86

$$E = -i \quad G(E) = K \frac{1}{-i-1} = -K \frac{1-i}{2} = K \frac{\sqrt{2}}{2} e^{i(\pi - \frac{\pi}{4})}$$

$$\angle G(E) = \frac{3\pi}{4}$$

Nyquist criterion gives $K < 1$ UNSTABLE

$K > 1$ STABLE

To verify, calculate the poles of G_{CL}

$$G_{CL} = \frac{G}{1+G} = \frac{K}{s-1+K} = \frac{K}{s-(1-K)}$$

$$p_{CL} = 1-K$$

for $K < 1$, $p_{CL} > 0$, in RHS, UNSTABLE

$K > 1$, $p_{CL} < 0$, in LHS, STABLE

QED

p634

Example p634

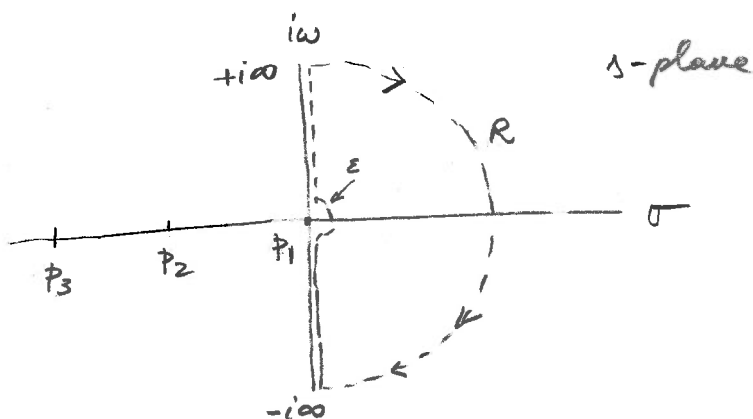
$$G(s) = K \frac{1}{s(T_1 s + 1)(T_2 s + 1)}$$

$$T_2 > T_1$$

Poles: $p_1 = 0$

$$p_2 = -1/T_1$$

$$p_3 = -1/T_2$$



Nyquist circuit will have to go around the origin on a small circle of radius $\epsilon \rightarrow 0$. We study four segments:

- positive $j\omega$ axis $s \in (+i0, +i\infty)$
- big circle $R \rightarrow \infty$
- negative $j\omega$ axis $s \in (-i\infty, -i0)$
- small circle $\epsilon \rightarrow 0$

2
p634

$$+i\omega \text{ axis.}$$

$$\Delta = (+i\epsilon, iR)$$

$$G(+i\epsilon) = -K(T_1 + T_2) - i\frac{K}{\epsilon}$$

$$G(+i\epsilon) \xrightarrow{\epsilon \rightarrow 0} -K(T_1 + T_2) - i\infty$$

$$|G(i\epsilon)| \xrightarrow{\epsilon \rightarrow 0} \infty$$

$$\angle G(i\epsilon) \xrightarrow{\epsilon \rightarrow 0} -90^\circ$$

$$G(iR) = \frac{K}{iR(T_1 iR + 1)(T_2 iR + 1)} \approx \frac{K}{T_1 T_2 R^3 i^3} = \frac{K}{T_1 T_2 R^3} e^{i\frac{\pi}{2}}$$

$$|G(iR)| = \frac{K}{T_1 T_2 R^3} \xrightarrow{R \rightarrow \infty} 0$$

$$\angle G(iR) = \frac{\pi}{2} = 90^\circ$$

$$\epsilon \rightarrow 0$$

$$R \rightarrow \infty$$

Proof

$$\frac{K}{i\epsilon(i\epsilon T_1 + 1)(i\epsilon T_2 + 1)} =$$

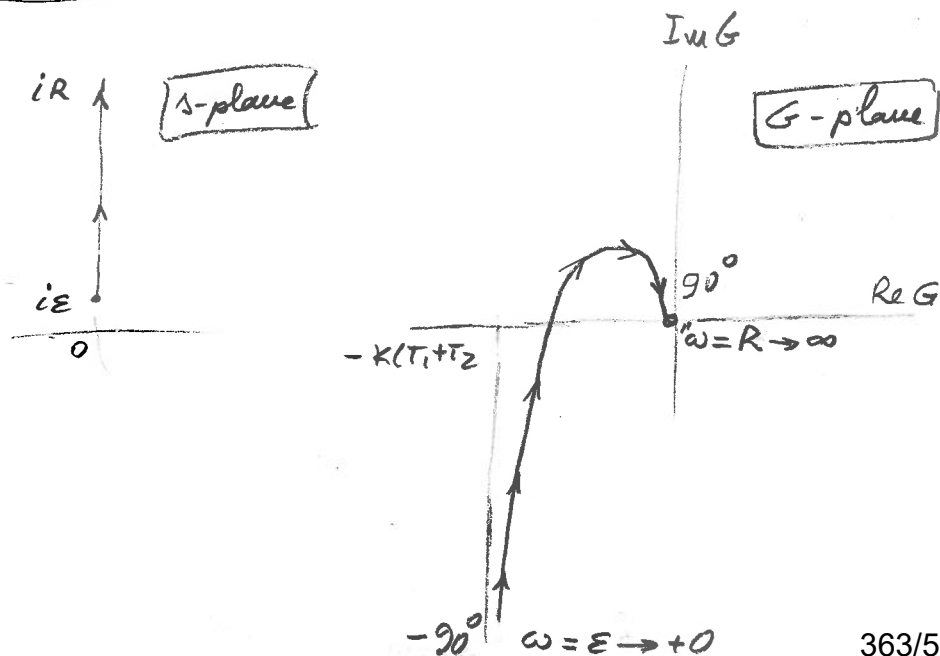
$$\frac{K}{i\epsilon[i^2 \epsilon^2 T_1 T_2 + i\epsilon(T_1 + T_2) + 1]} =$$

$$= \frac{K}{i\epsilon} \frac{1}{i\epsilon(T_1 + T_2) + 1}$$

$$= \frac{K}{i\epsilon} \frac{-i\epsilon(T_1 + T_2) + 1}{\epsilon^2(T_1 + T_2) + 1}$$

$$= -K(T_1 + T_2) - i\frac{K}{\epsilon}$$

$$\frac{1}{i^3} = \frac{i^4}{i^3} = i = e^{i\frac{\pi}{2}}$$



3
P634

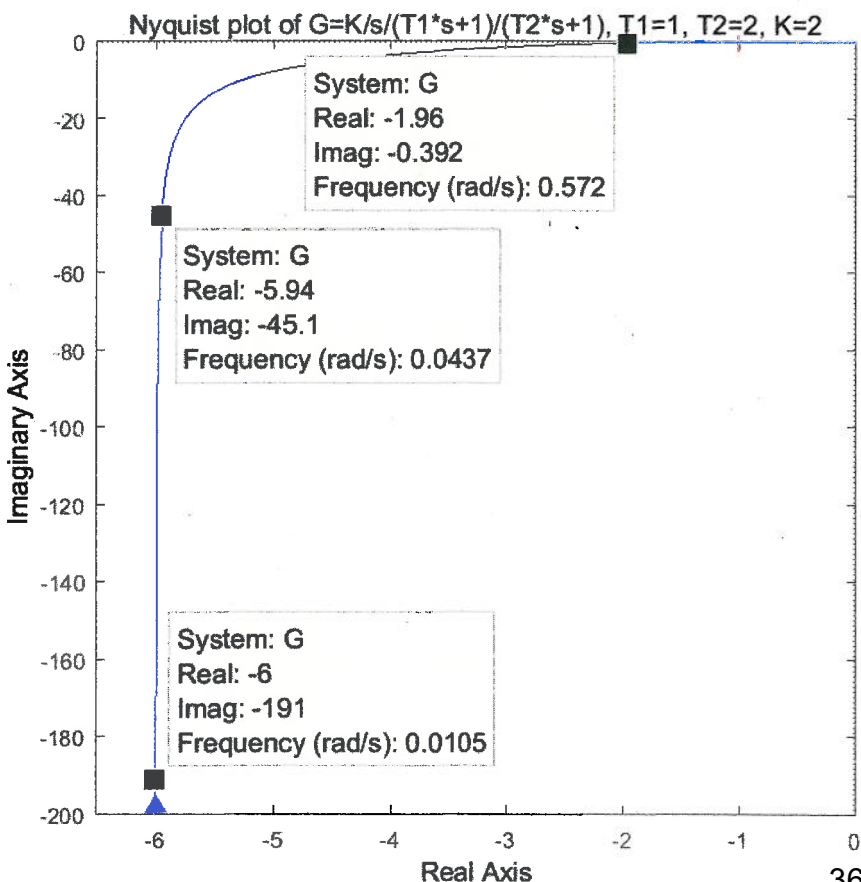
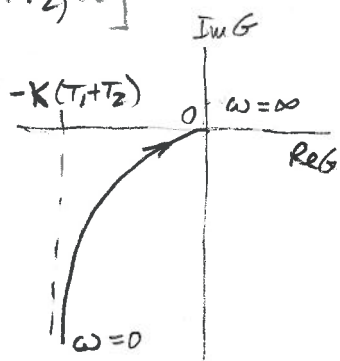
$$G(i\epsilon) = \frac{K}{i\epsilon(i\epsilon T_1 + 1)(i\omega T_2 + 1)}$$

$$= \frac{K}{i\epsilon(\cancel{i\epsilon^2 T_1 T_2} + i\epsilon(T_1 + T_2) + 1)} = \frac{K}{\epsilon} \frac{1}{-\epsilon(T_1 + T_2) + i}$$

$$= \frac{K}{\epsilon} \frac{-\epsilon(T_1 + T_2) - i}{\cancel{\epsilon^2(T_1 + T_2)^2} + 1} = -\frac{K}{\epsilon} [\epsilon(T_1 + T_2) + i]$$

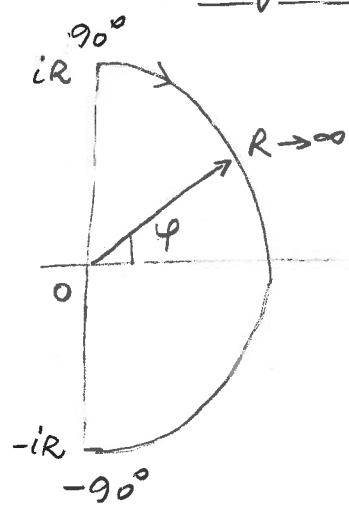
$$G(i\epsilon) = -K(T_1 + T_2) - i\frac{K}{\epsilon}$$

$$\begin{aligned} \xrightarrow{\epsilon \rightarrow 0} & \quad \underbrace{-K(T_1 + T_2)}_{\text{Re } G = \text{Const}} - i\infty \end{aligned}$$

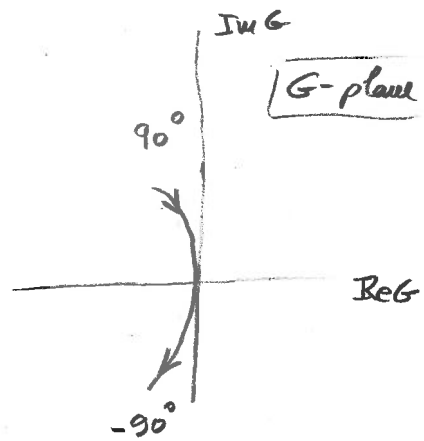


4
p. 634

Big circle



s-plane



G-plane

$$G(s) = \frac{K}{Re^{i\varphi}(T_1 Re^{i\varphi} + 1)(T_2 Re^{i\varphi} + 1)}$$

$$G(s) \underset{R \gg 1}{\approx} \frac{K}{T_1 T_2 R^3 e^{3i\varphi}} = \frac{K}{T_1 T_2 R^3} e^{-3i\varphi}$$

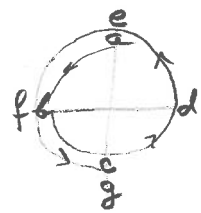
$$|G(s)| = \frac{K}{T_1 T_2 R^3} \xrightarrow{R \rightarrow \infty} 0$$

$$\angle G(s) = -3\varphi = \begin{pmatrix} -270^\circ & +270^\circ \\ \varphi = 90^\circ & -90^\circ \end{pmatrix}$$

$(-270 + 360 = 90^\circ; 270 - 360 = -90^\circ)$

$$\angle G(s) = (90^\circ, -90^\circ)$$

$\varphi = 90^\circ$	60°	30°	0°	-30°	-60°	-90°
$\angle G = -270$	-180	-90	0	90	180	270
90	180	-90	0	90	180	-90
a	b	c	d	e	f	g



$G(s)$ goes round origin 1.5 times

5
p634

$$s = \frac{-i\omega \text{ axis}}{(-iR \quad -i\epsilon)}$$

$$\epsilon \rightarrow 0$$

$$R \rightarrow \infty$$

$$G(-i\epsilon) \approx -K(T_1 + T_2) + i\frac{K}{\epsilon}$$

$$G(-i\epsilon) \xrightarrow{\epsilon \rightarrow 0} -K(T_1 + T_2) + i\infty$$

$$|G(-i\epsilon)| \xrightarrow{\epsilon \rightarrow 0} \infty$$

$$\angle G(-i\epsilon) \xrightarrow{\epsilon \rightarrow 0} 90^\circ$$

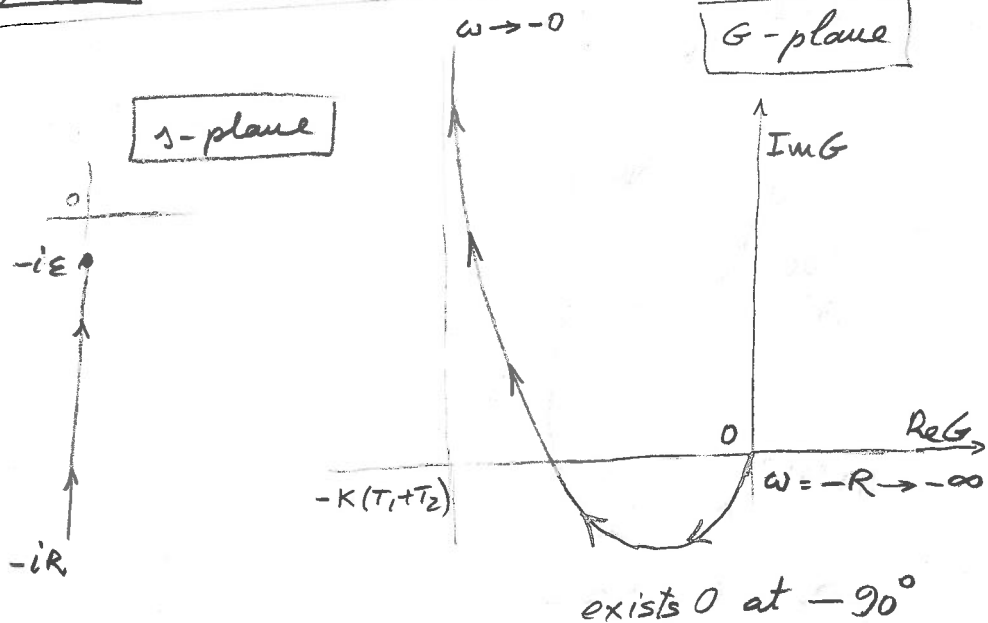
$$G(-iR) = \frac{K}{-iR(-T_1 iR + 1)(-T_2 iR + 1)} \approx \frac{K}{-T_1 T_2 R^3 i^3} = \frac{K}{T_1 T_2 R^3} e^{-i\frac{\pi}{2}}$$

$$|G(-iR)| = \frac{K}{T_1 T_2 R^3} \xrightarrow{R \rightarrow \infty} 0$$

$$\angle G(iR) = -\frac{\pi}{2} = -90^\circ$$

$$\begin{aligned} \frac{\text{Proof}}{K} &= \frac{-i\epsilon(-i\epsilon T_1 + 1)(-i\epsilon T_2 + 1)}{K} \\ &= \frac{-i\epsilon(i^2 \epsilon^2 T_1 T_2 - i\epsilon(T_1 + T_2) + 1)}{K} \\ &= i\frac{K}{\epsilon} \frac{-i\epsilon(T_1 + T_2) + 1}{\epsilon^2(T_1 + T_2)^2 + 1} \\ &= i\frac{K}{\epsilon} [i\epsilon(T_1 + T_2) + 1] \\ &= -K(T_1 + T_2) + i\frac{K}{\epsilon} \end{aligned}$$

$$\frac{-1}{i^3} = \frac{i^2}{i^3} = \frac{1}{i} = -i$$

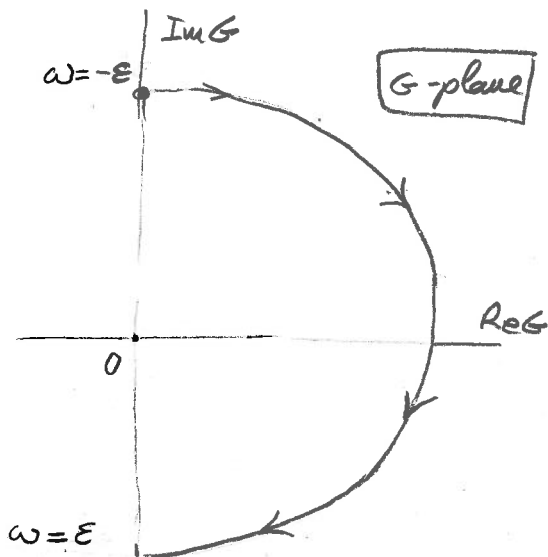
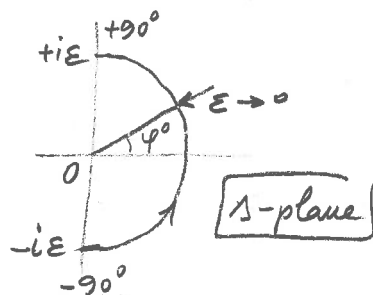


6
p634

Small circle

$$s_0 = \varepsilon e^{i\varphi_0}$$

$$\varphi_0 = (-90^\circ, 90^\circ)$$



$$\varphi_0 = (-90^\circ, 90^\circ)$$

$$G(s_0) = \frac{K}{\varepsilon e^{i\varphi_0} (T_1 \varepsilon e^{i\varphi_0} + 1) (T_2 \varepsilon e^{i\varphi_0} + 1)}$$

$$G(s_0) \approx \frac{K}{\varepsilon e^{i\varphi_0}} = \frac{K}{\varepsilon} e^{-i\varphi_0}$$

$$|G(s_0)| = \frac{K}{\varepsilon} \xrightarrow{\varepsilon \rightarrow 0} \infty$$

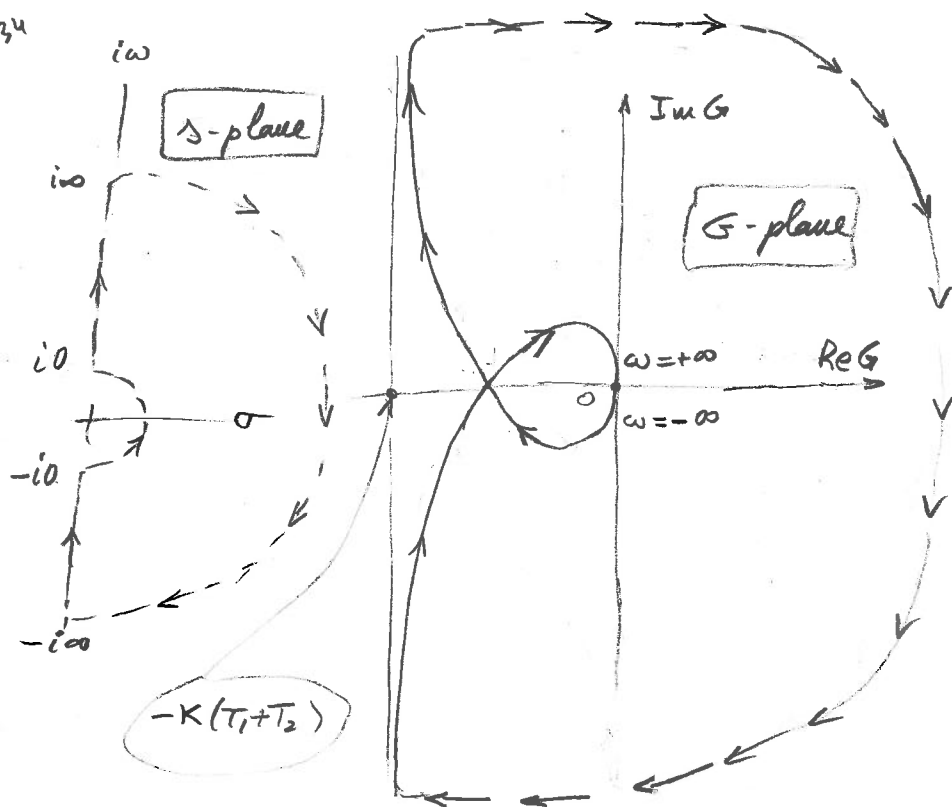
$$\angle G(s_0) = -\varphi_0 = (90^\circ, -90^\circ)$$

$$\varphi_0 = \begin{matrix} -90^\circ & 90^\circ \\ -i\varepsilon & i\varepsilon \end{matrix}$$

$G(s_0)$ travels a big semicircle from 90° to -90° .

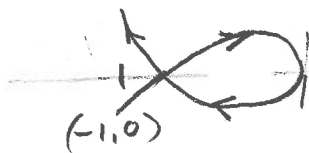
8
P364
7
P634

Finally, assemble the four segments:



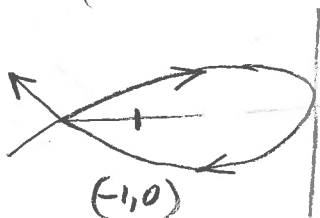
Note that $G(s)$ goes around two times!

Depending on K , it may or may not enclose the point $(-1, 0)$.



STABLE

$$\begin{array}{r} N=0 \\ P=0 \\ \hline Z=0 \end{array}$$



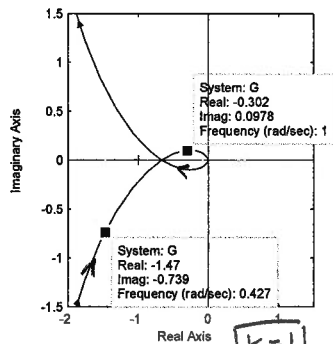
UNSTABLE

$$\begin{array}{r} N=2 \\ P=0 \\ \hline Z=2 \end{array}$$

(see MATLAB plot next page) 368/523

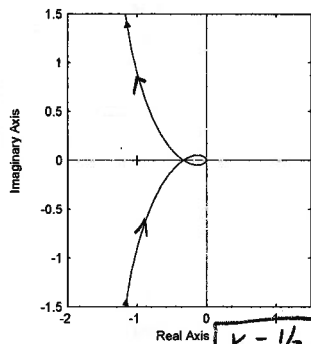
Ex. p634

STABLE
 $N=0$
 $P=0$
 $Z=N+P=0$
 K=1 Nyquist plot; Example pp 634



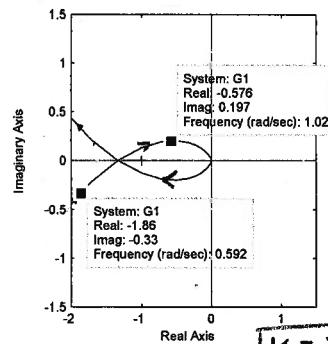
K=1

STABLE
 $N=0$
 $P=0$
 $Z=N+P=0$
 K=0.5 Nyquist plot; Example pp 634



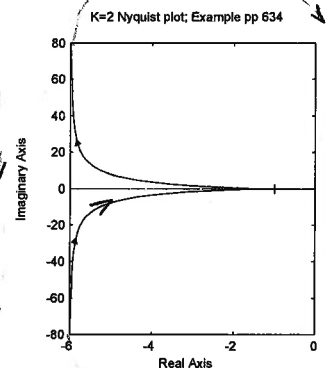
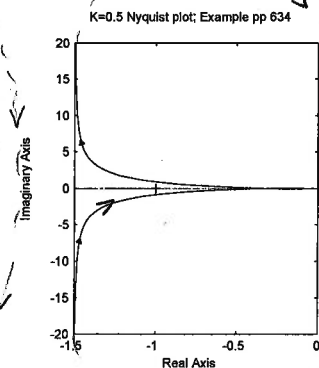
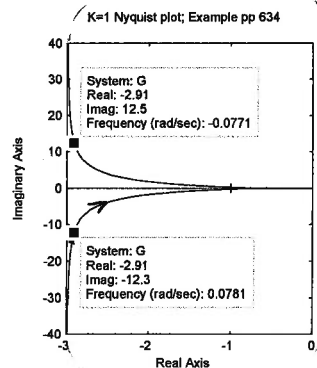
K=1/2

UNSTABLE
 $N=2$
 $P=0$
 $Z=N+P=2$ UNSTABLE
 K=2 Nyquist plot; Example pp 634



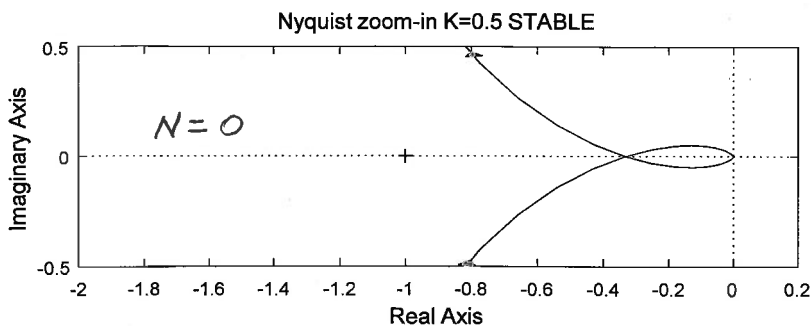
K=2

Zoom
view

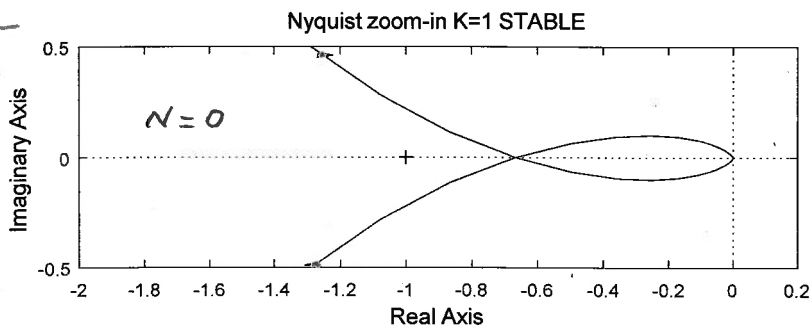


far
view

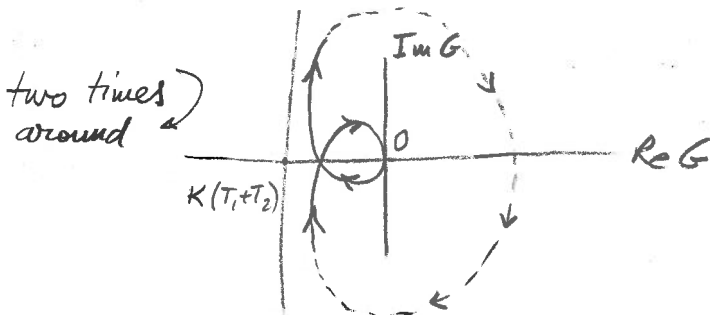
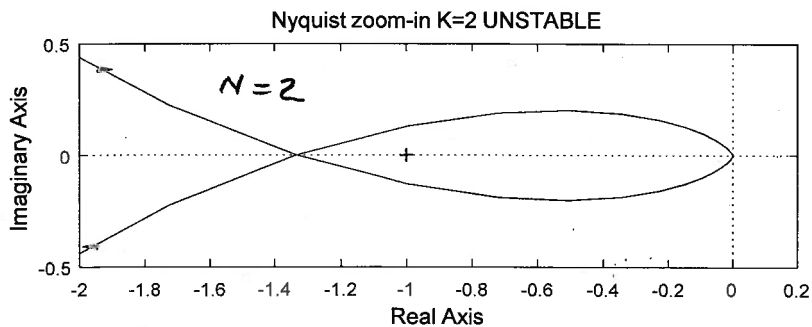
10
P634



$N=0$
 $P=0$
 $Z=0$
stable

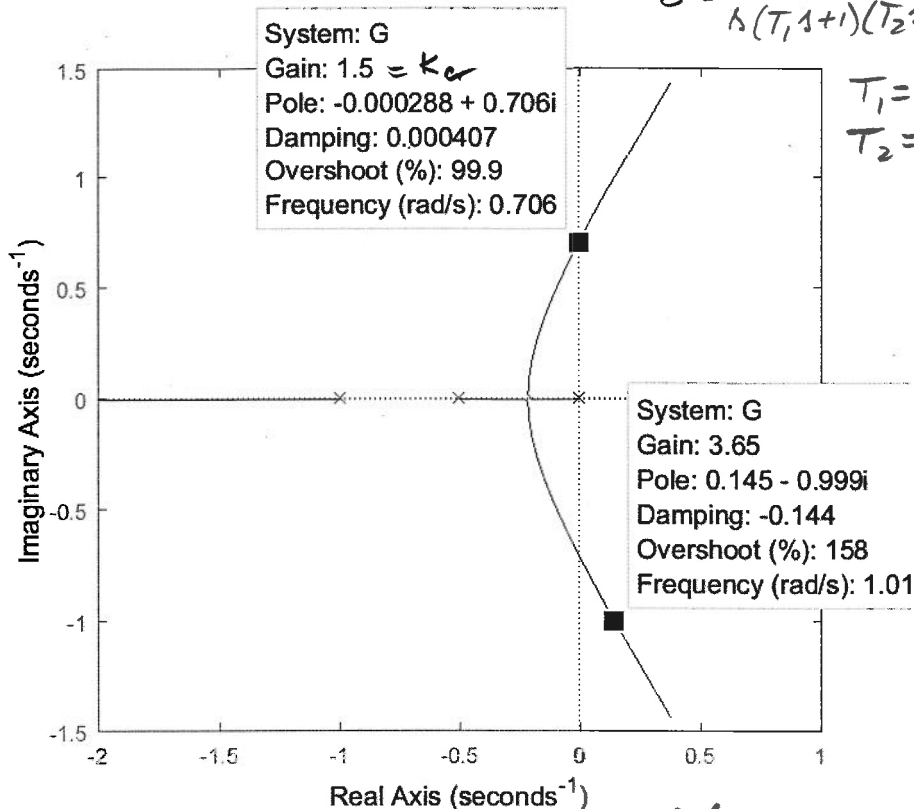


$N=2$
 $P=0$
 $Z=2$
unstable



ROOT LOCUS

$$G = \frac{K}{s(T_1s+1)(T_2s+1)}$$



$K_{cr} = 1.5$ according to root locus plot

Verify using $G_{cl} = \frac{G}{1+G} = \frac{K}{s(T_1s+1)(T_2s+1)+K}$

$$s(T_1s+1)(T_2s+1)+K \Big|_{T_1=1} = s(s+1)(2s+1)+K$$

$$s(2s^2+3s+1)+K \Big|_{T_2=2} = 2s^3+3s^2+s+K$$

For sign change: Routh criterion

$$3-2K=0$$

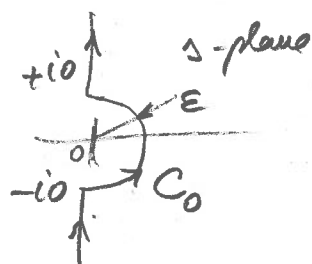
$$K_{cr} = \frac{3}{2} = 1.5$$

QED.

s^3	2	1
s^2	3	K
s^1	$\frac{3-2K}{3}$	
s^0	K	

N3d

Tips on Nyquist Plot



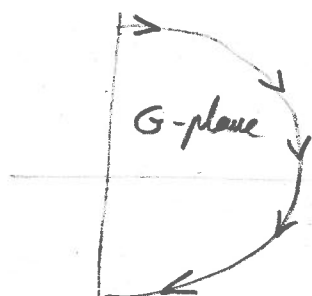
$$s_0 = \epsilon e^{i\varphi}$$

$$-90^\circ < \varphi < 90^\circ$$

$$G = \frac{1}{s} \quad , \quad G(s_0) = \frac{1}{\epsilon e^{i\varphi}} = \frac{1}{\epsilon} e^{-i\varphi}$$

(Type 1 system)

$$+90^\circ > \angle G(s_0) > -90^\circ$$

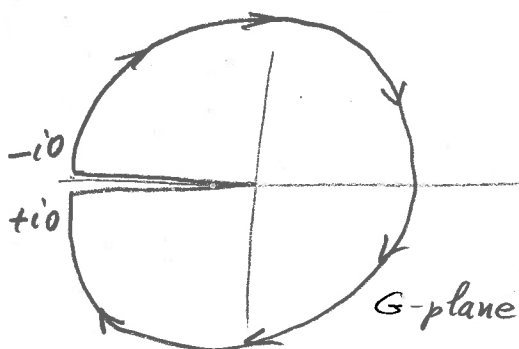


Half-circle, clockwise.

$$G = \frac{1}{s^2} \quad , \quad G(s_0) = \frac{1}{(\epsilon e^{i\varphi})^2} = \frac{1}{\epsilon^2} e^{-i2\varphi}$$

(Type 2 system)

$$+180^\circ > \angle G > -180^\circ$$

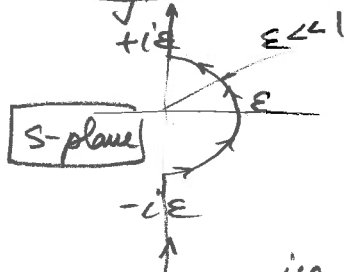


Full circle,
clockwise.

Details on behavior of common functions on the Nyquist circuit

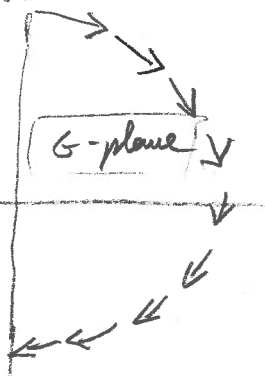
Type 1

$$G(s) = \frac{1}{s}$$



$$s_0 = \epsilon e^{i\varphi}$$

$$-90^\circ < \varphi < 90^\circ$$



$$G(s_0) = \frac{1}{\epsilon e^{i\varphi}}$$

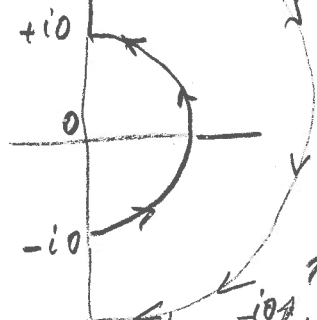
$$= \frac{1}{\epsilon} e^{-i\varphi}$$

$$90^\circ > \angle G(s_0) > -90^\circ$$

Type 2

$$G(s) = \frac{1}{s^2}$$

s-plane

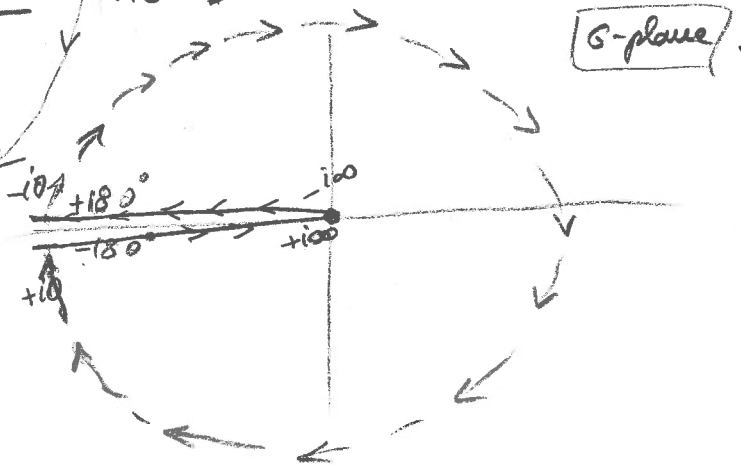


$$G(s_0) = \frac{1}{\epsilon^2} e^{-i2\varphi}, \quad -90^\circ < \varphi < +90^\circ$$

$$+180^\circ > \angle G(s_0) > -180^\circ$$

$$G(-i0) = \frac{1}{(-i0)^2}$$

$$G(+i0) = \frac{1}{(i0)^2}$$



N/a

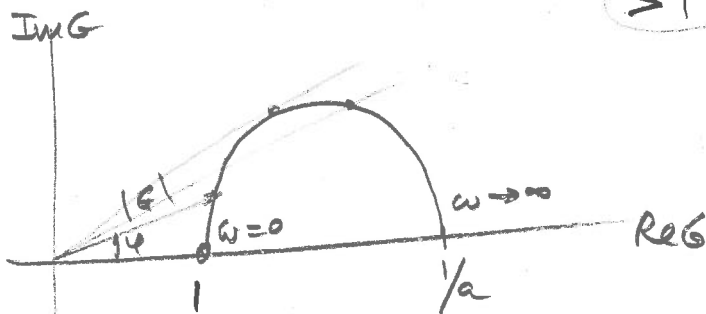
$$G(s) = \frac{Ts+1}{aTs+1} \quad 0 < a < 1$$

$$G(i\omega) = \frac{T \cdot i\omega + 1}{a \cdot T \cdot i\omega + 1}$$

$$\omega \rightarrow 0, \quad G(i0) = 1$$

$$\omega \rightarrow \infty, \quad G(i\infty) = \lim_{\omega \rightarrow \infty} \frac{T \cdot i\omega + 1}{a \cdot T \cdot i\omega + 1} = \frac{T}{aT} = \frac{1}{a}$$

$$> 1$$

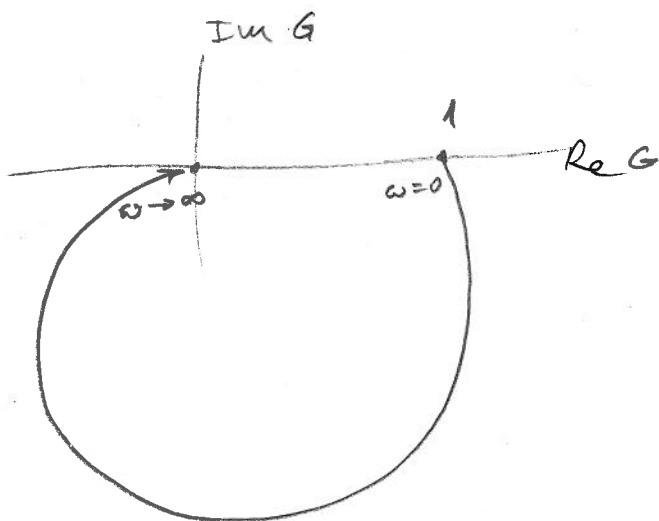
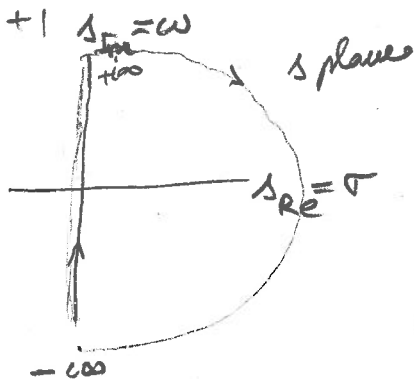


N16. Ex 11.4, p 641

$$G(s) = \frac{1}{s^2 + 0.8s + 1}$$

$G = 1/-$
 nyquist(G)

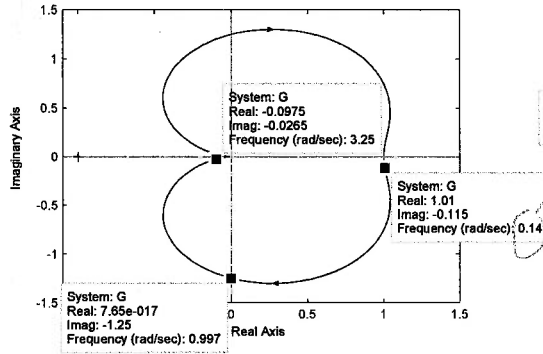
Nyquist circuit for ω



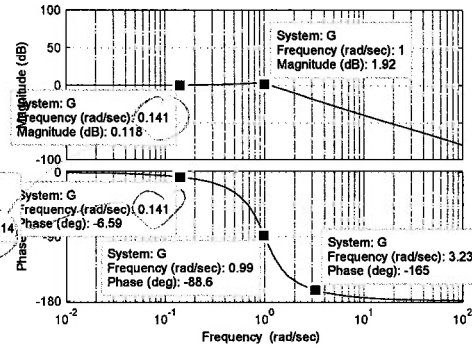
Ex. 11.5 p 642

$$G(s) = \frac{1}{s(s+1)}$$

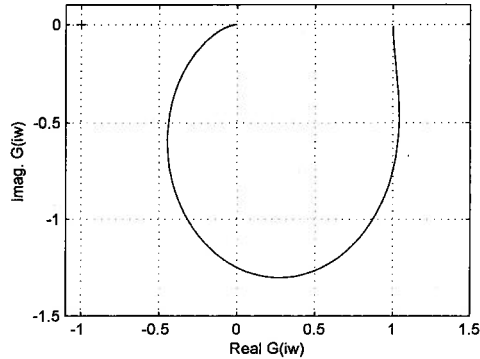
Nyquist plot; Example 11.4 pp 641



Bode plot; Example 11.4 pp 641

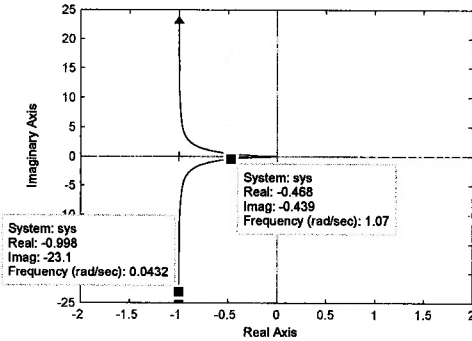


Re-Im plot; Example 11.4 pp 641

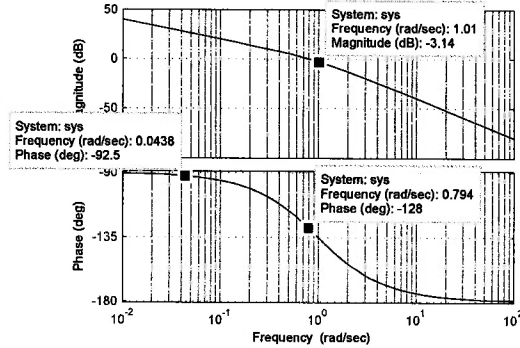


$$G(s) = \frac{1}{s^2 + 0.8s + 1}$$

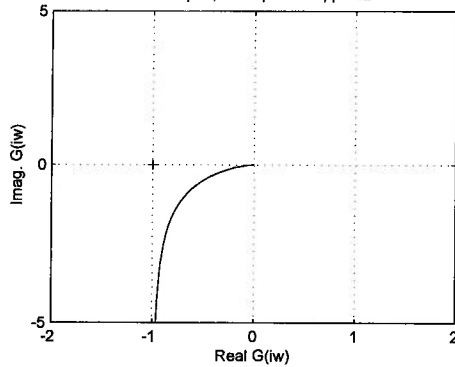
Nyquist plot; Example 11.5 pp 642



Bode plot; Example 11.5 pp 642



Re-Im plot; Example 11.5 pp 642



$$G(s) = \frac{1}{s(s+1)}$$

N2

Nyquist plot trends

$$G(s) = \frac{K (T_a s + 1)(T_b s + 1) \dots}{s^N (T_1 s + 1)(T_2 s + 1) \dots}$$

order m

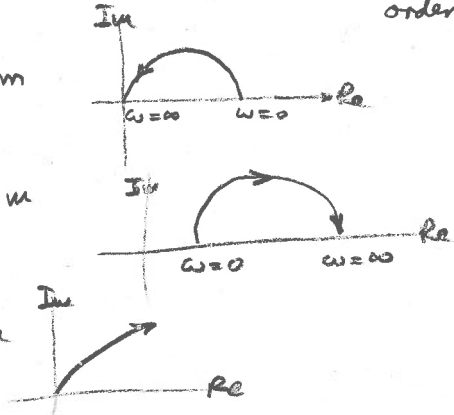
Type 0 systems ($N=0$) , $G(i\omega) = K \frac{(T_a i\omega + 1)(T_b i\omega + 1) \dots}{(T_1 i\omega + 1)(T_2 i\omega + 1) \dots}$

order n

$\omega=0 \quad G(i0) = K$

$\omega \rightarrow \infty$

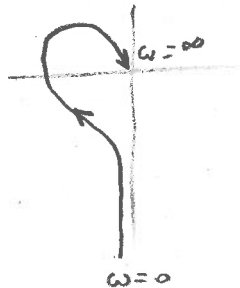
$$G(i\omega) = \begin{cases} 0 & \text{for } n > m \\ \text{const, } n=m \\ \infty & , n < m \end{cases}$$

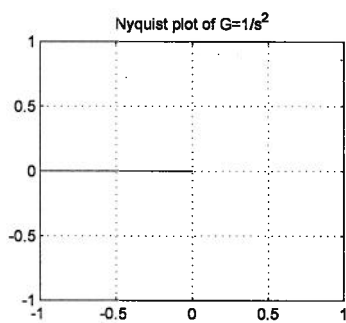
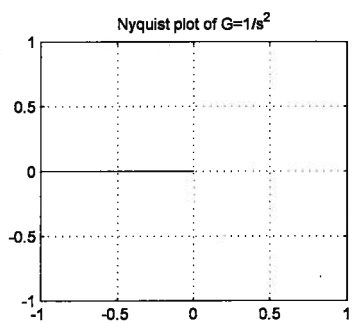
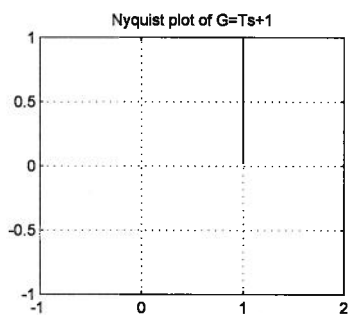
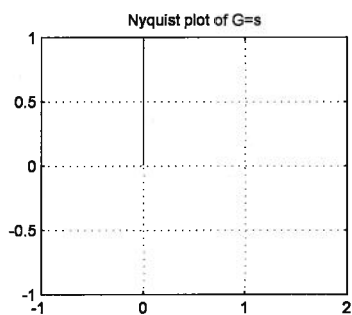
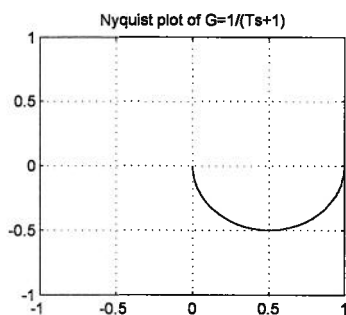
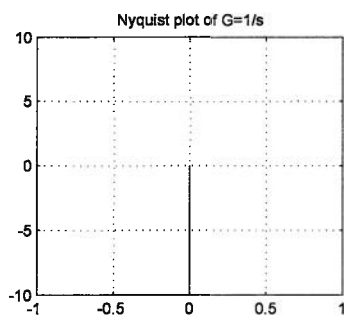


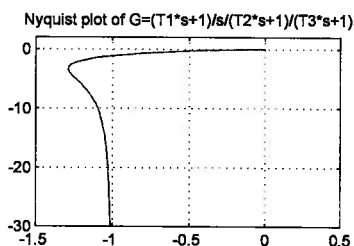
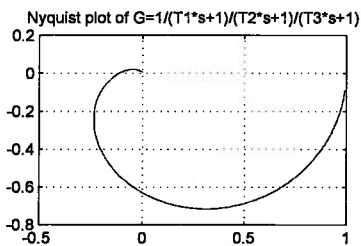
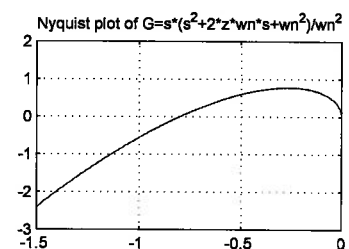
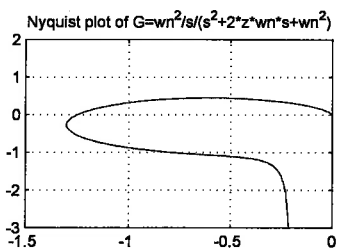
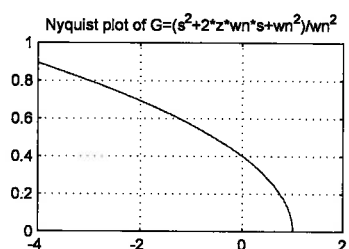
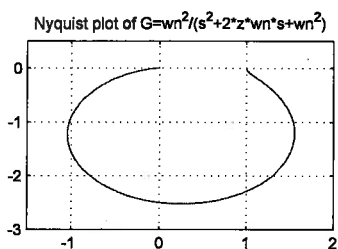
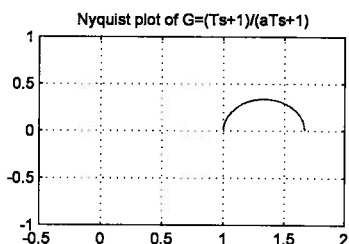
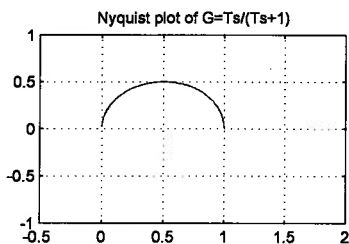
Type 1 systems ($N=1$) , $G(i\omega) = \frac{K}{i\omega} \frac{(\quad)(\quad)}{(\quad)(\quad)}$

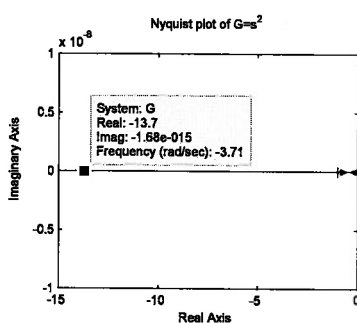
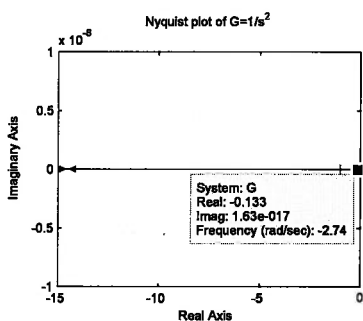
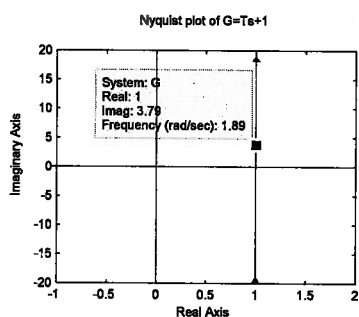
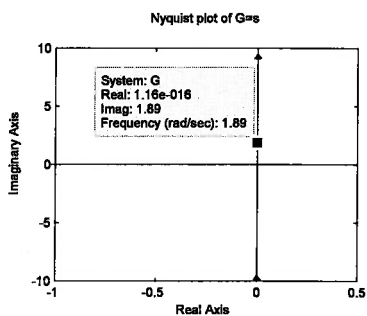
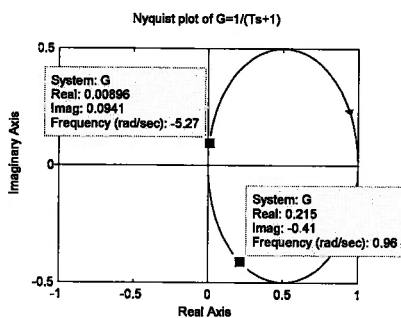
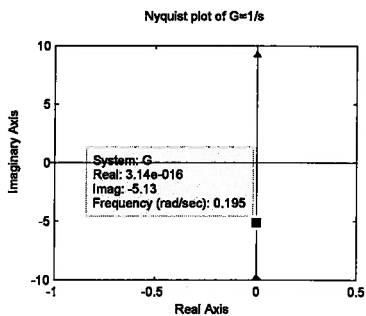
for $n=m$, $G(i\infty) = 0$

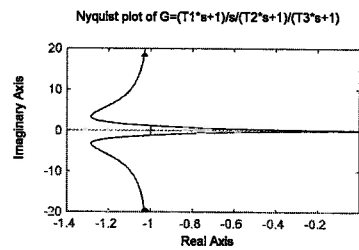
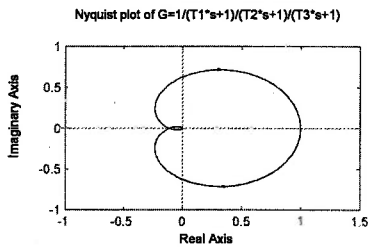
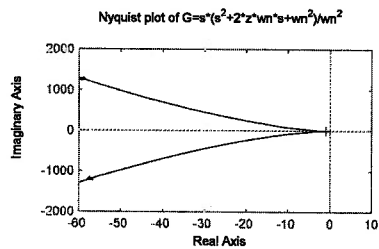
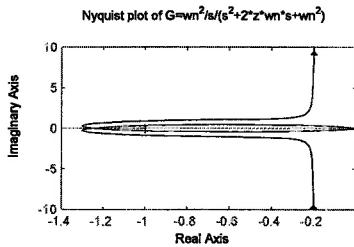
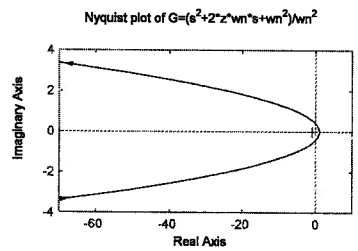
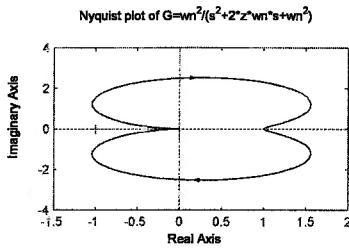
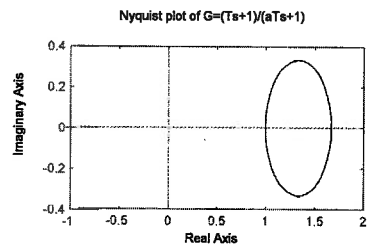
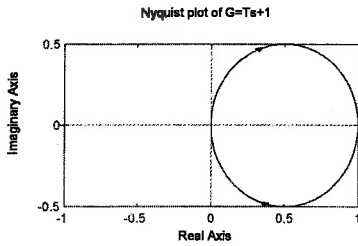
$G(i0) = \frac{1}{i0}$, $\angle G(i0) = -90^\circ$

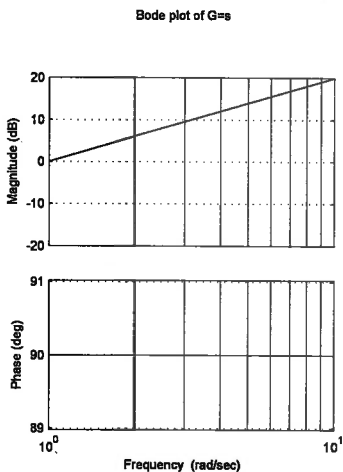
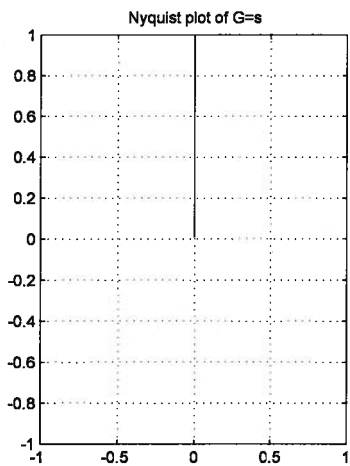
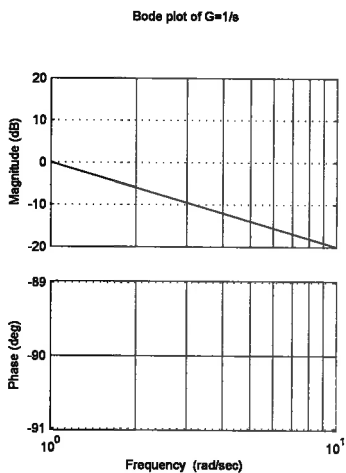
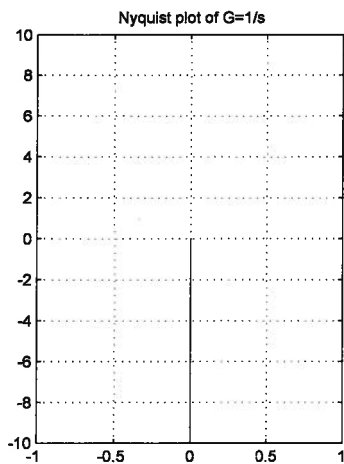


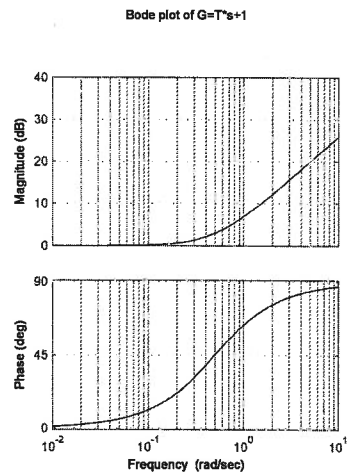
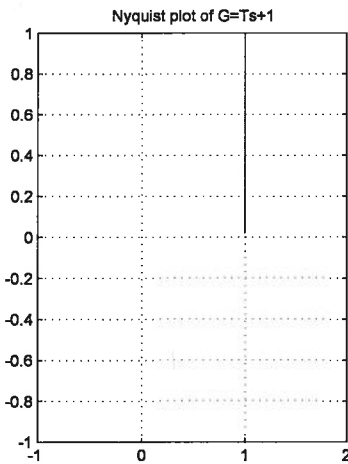
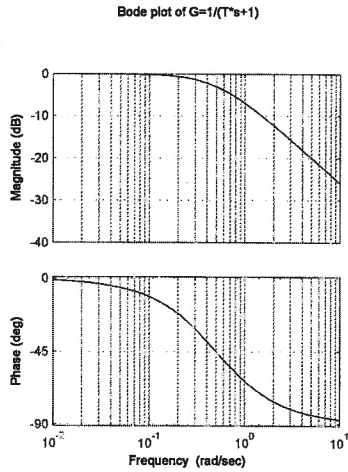
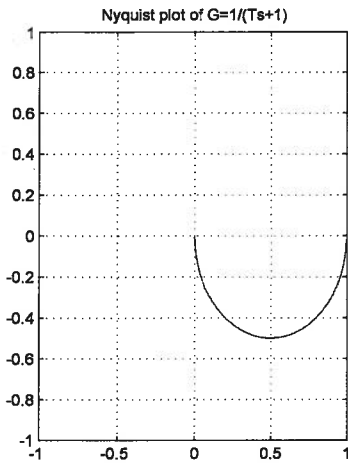


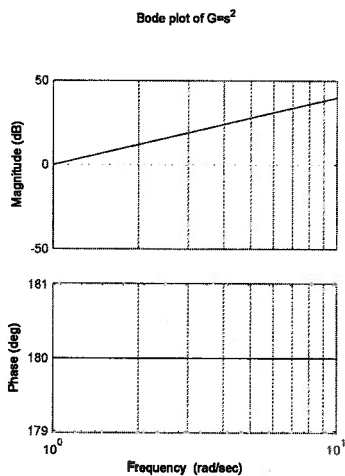
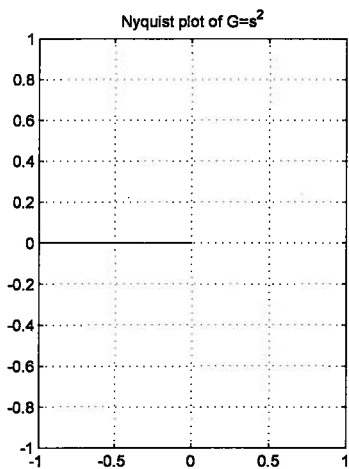
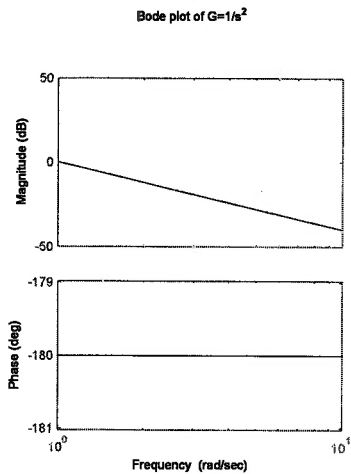
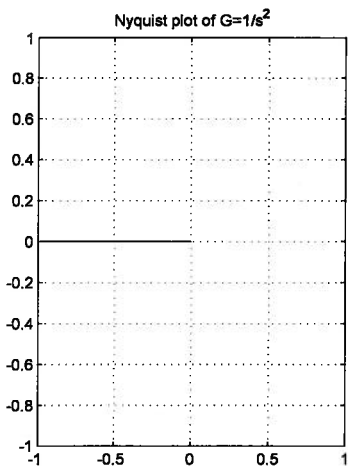


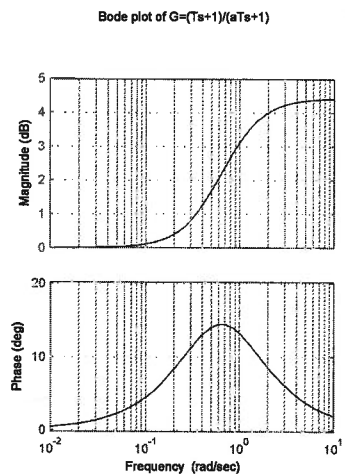
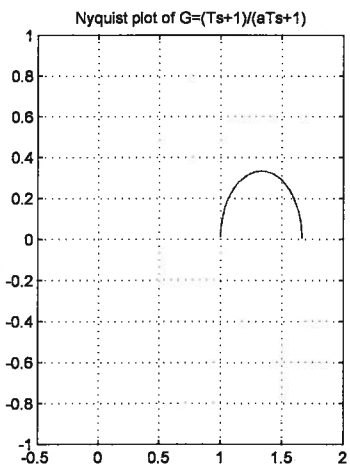
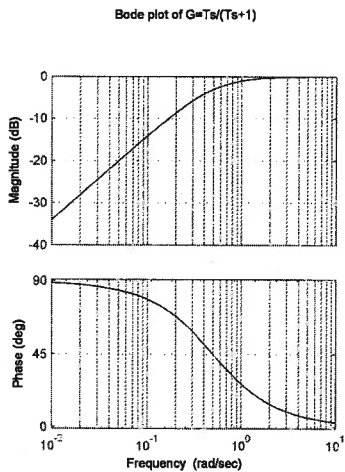
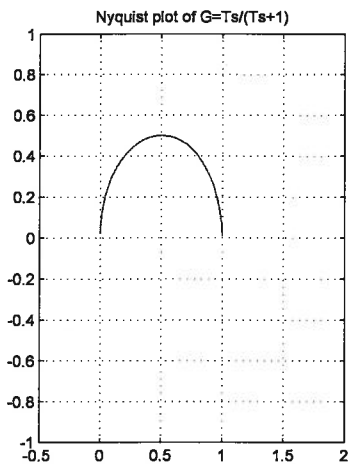


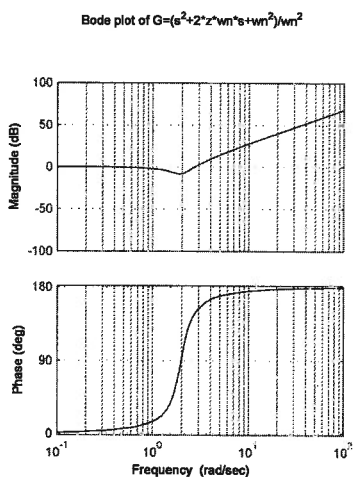
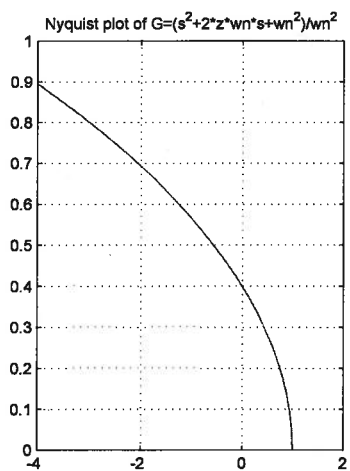
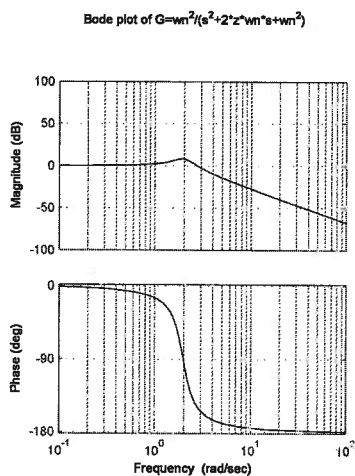
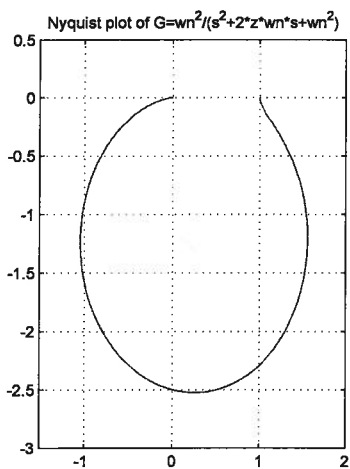


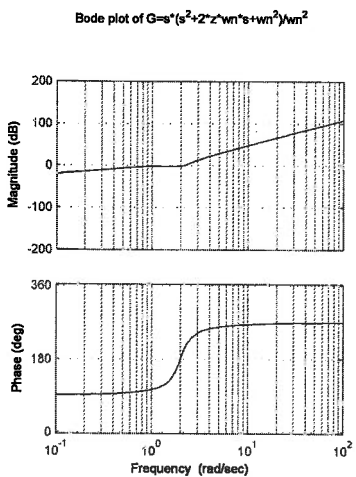
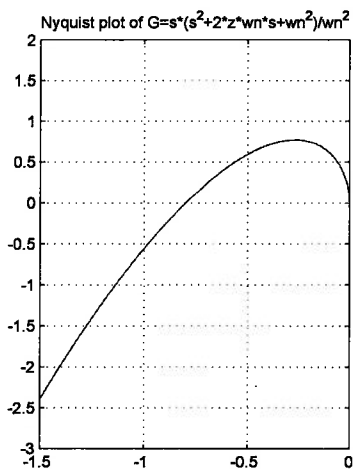
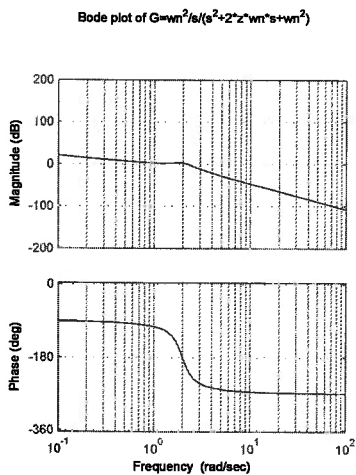
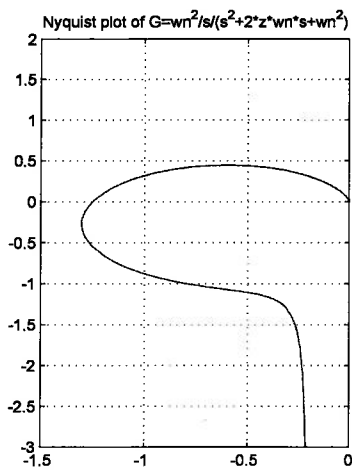




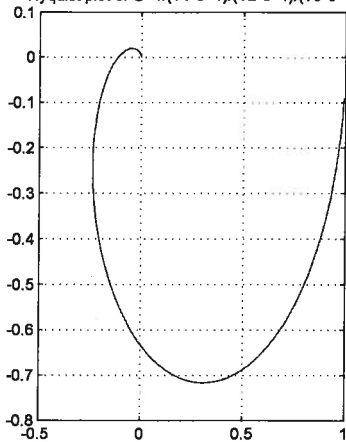




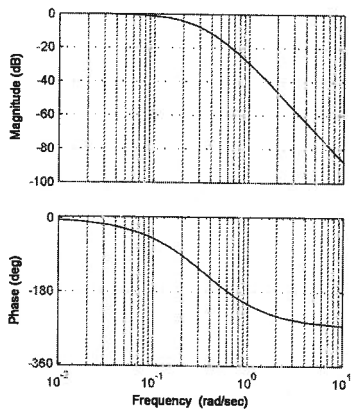




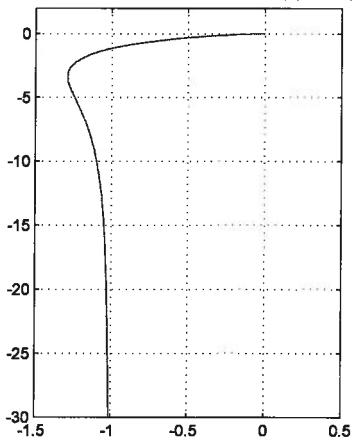
Nyquist plot of $G=1/(T_1s+1)/(T_2s+1)/(T_3s+1)$



Bode plot of $G=1/(T_1s+1)/(T_2s+1)/(T_3s+1)$



Nyquist plot of $G=(T_1s+1)/s/(T_2s+1)/(T_3s+1)$



Bode plot of $G=(T_1s+1)/s/(T_2s+1)/(T_3s+1)$

