

The Influence of CSI in Ultra-Reliable Low-Latency Communications with IR-HARQ

Apostolos Avranas, Marios Kountouris, Philippe Ciblat



Supporting New Applications:



Mission-critical control

smart grid (5ms for tx/grid backbone, 50ms for grid backhaul)

Industrial Automation

real time control (20 μ s – 10 ms), time-critical sensing (10ms), over-the-air: 0.25~1 ms



HF Trading & eCommerce

1ms delay advantage = 100M\$/year

A page load slowdown of 1 sec could cost 1.6B\$ in sales each year (Amazon)



Media & Entertainment

4K Video, live streaming in crowded areas (20ms), collaborative gaming (20ms)

AR/VR

MTP 15~20 ms // over-the-air: 1~5 ms

Tactile Internet

1ms e2e latency



Automotive industry

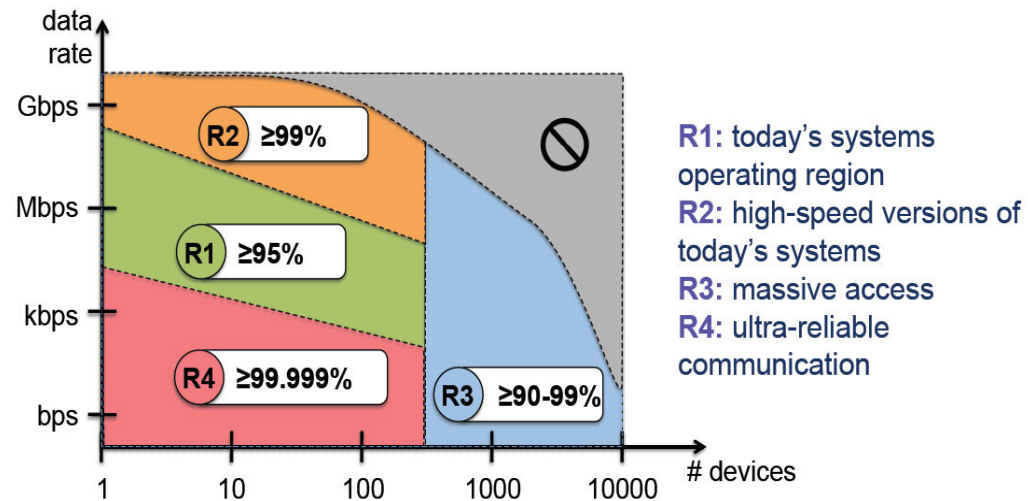
autonomous/cooperative driving, V2X (3-10ms), V2N for remote vehicle operation (10-30ms E2E)



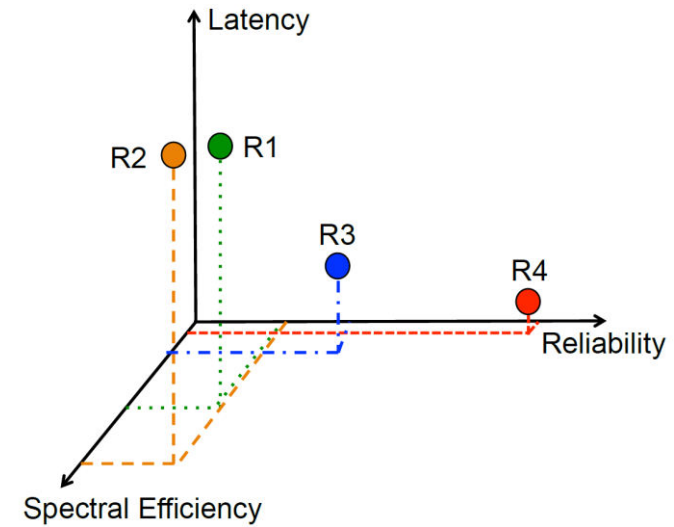
Healthcare

video/AR for remote surgery (100ms), real-time command/control (10-100ms)

New Operation Regimes - URLLC

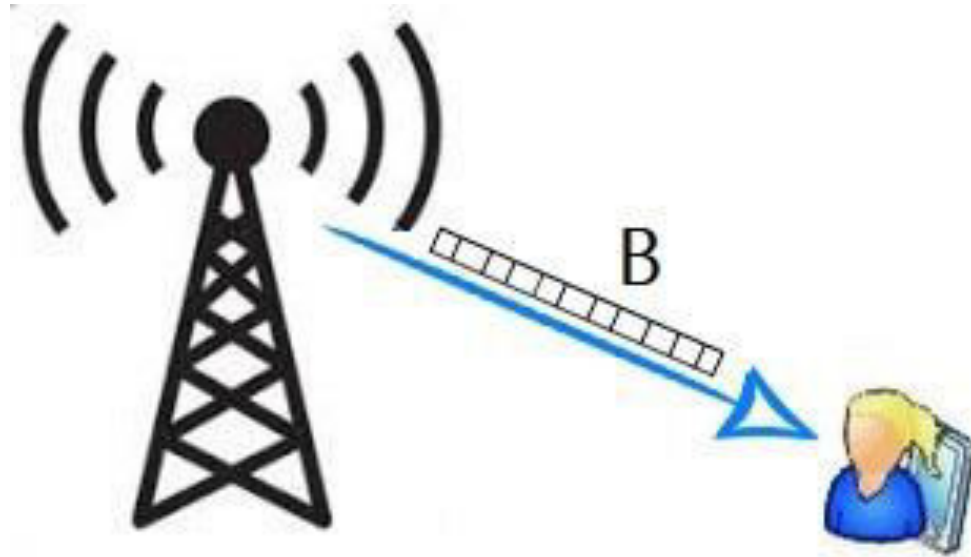


From F. Boccardi et al. "Five Disruptive Technology Directions for 5G"



Use case	Latency (ms)	Reliability (PEP)	Data size (bytes)	Communication range (m)
Industrial automation	0.25 – 10	$10^{-6} - 10^{-9}$	10 – 300	10 – 100
Smart grids	5 – 50	10^{-6}	80 – 1000	Few m to km
Intelligent transport systems	5 – 100	$10^{-3} - 10^{-5}$	500 – 1k	200 – 5000
Telemedicine	1 – 10 (haptics) 20 – 100 (video, audio)	10^{-5}	200 – 4k	< 200 km

System model



- Point-2-Point communication
- Fixed number B of information bits
- Message transmitted using maximum N channel uses
- Transmissions within time coherence interval, i.e. constant channel gain g

Error Probability: ε


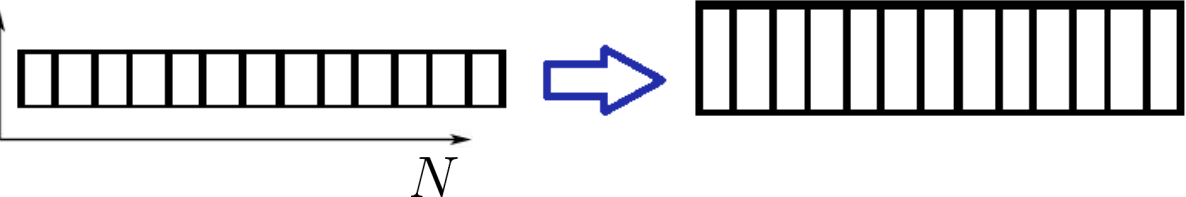
- If block-length $N \rightarrow \infty$ and mutual information above a threshold then:
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- Increase N : 
- Increase P : 

Error Probability: ε

$$\varepsilon \approx Q \left(\frac{N \ln(1 + gP) - B \ln 2 + \frac{\ln N}{2}}{\sqrt{N(1 - \frac{1}{(1 + gP)^2})}} \right)$$

[1] Y. Polyanskiy, “Channel coding: Non-asymptotic fundamental limits”, Ph.D. dissertation, Princeton University, Nov. 2010.

[2] M. Hayashi, “Information spectrum approach to second-order coding rate in channel coding,” IEEE Trans. on Inf. Theory, vol. 55, no. 11, pp. 4947–4966, Nov. 2009

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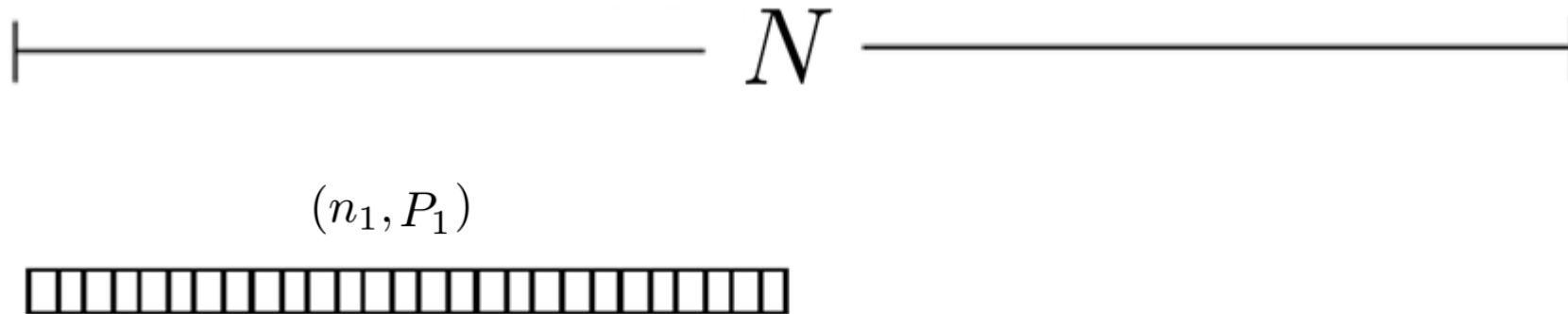
[2] M. Hayashi, “Information spectrum approach to second-order coding rate in channel coding,” IEEE Trans. on Inf. Theory, vol. 55, no. 11, pp. 4947–4966, Nov. 2009

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- Incremental Redundancy Hybrid Automatic Repeat Request (IR-HARQ). For $M = 2$ rounds:



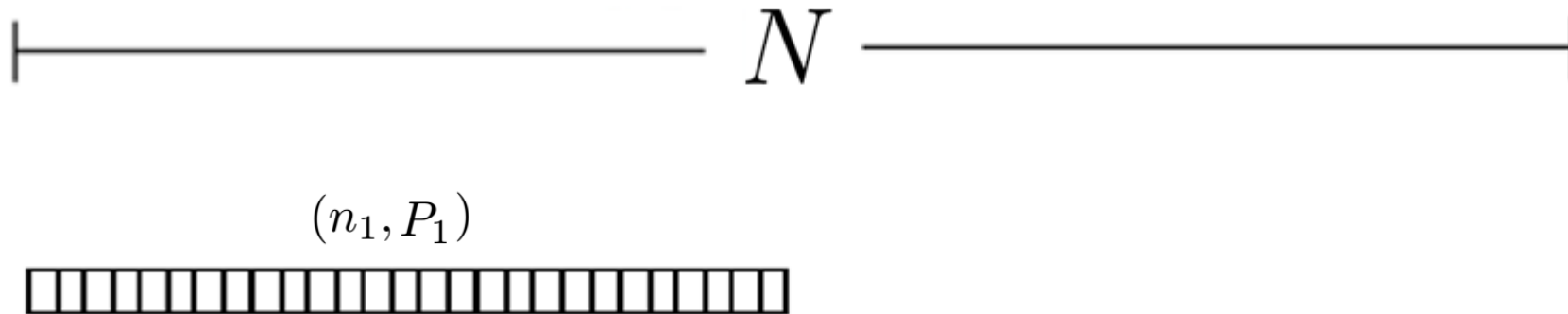
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$$(n_1, P_1)$$

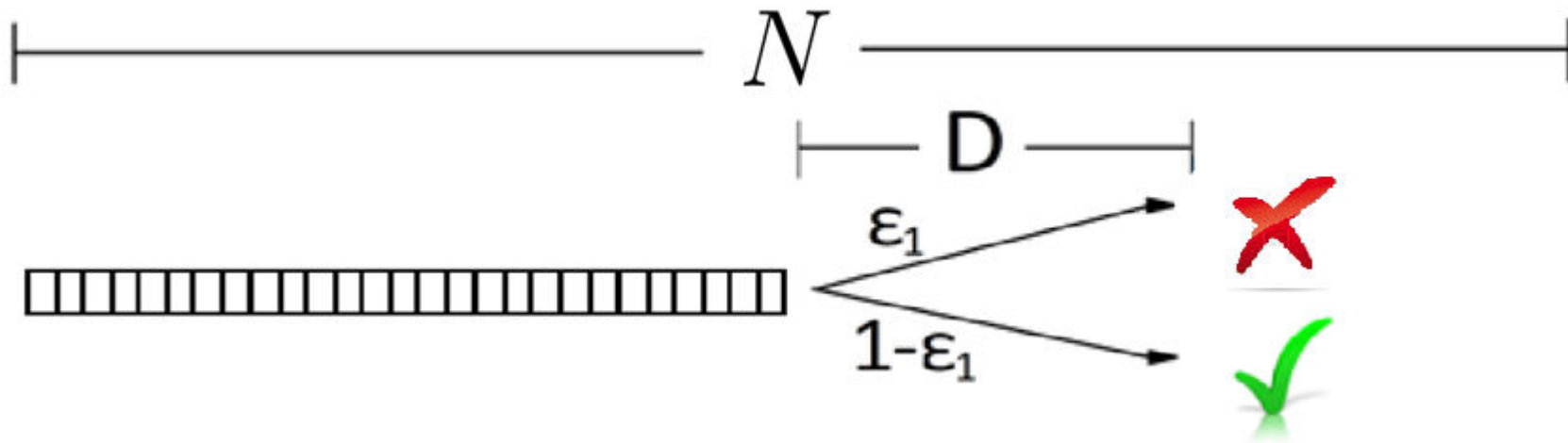
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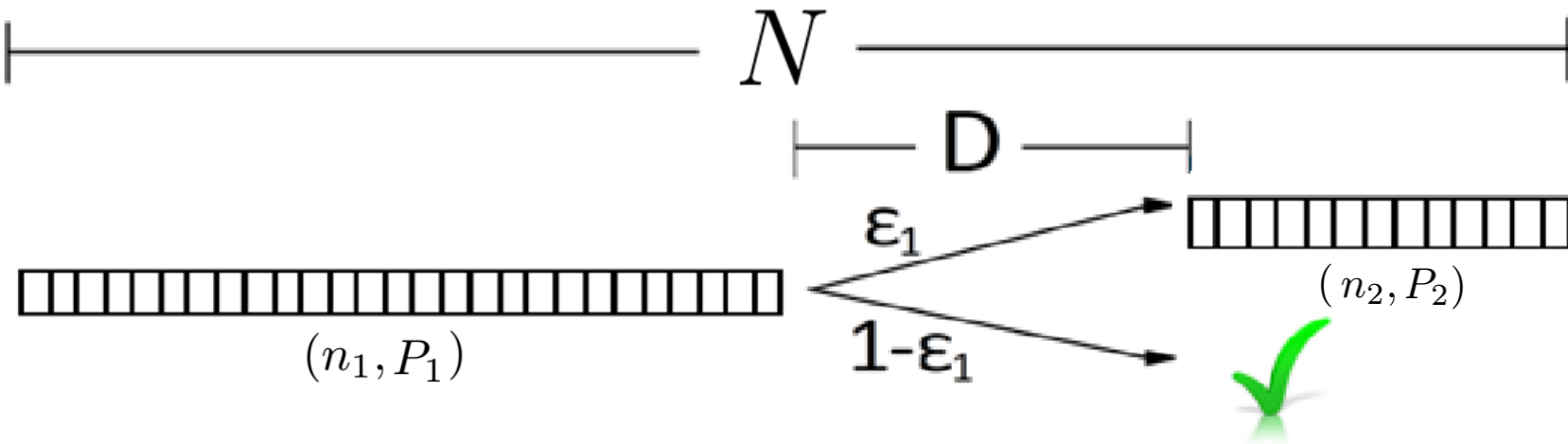
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Error Probability ε_m for IR-HARQ at m round

$$\varepsilon_m \approx Q \left(\frac{\sum_{i=1}^m n_i \ln(1 + gP_i) - B \ln 2}{\sqrt{\sum_{i=1}^m n_i \left(1 - \frac{1}{(1 + gP_i)^2}\right)}} \right)$$

Optimization Problem, URLLC,

Fixing: B information bits, $M = 2$, $D = 0$

$$\min_{n_1, n_2, P_1, P_2} (\alpha-1)\text{Throughput} + \alpha\text{Energy}$$

$s.t.$ Reliability

Latency

Energy & Maximum Power

Optimization Problem, *Full-CSI*

Fixing: B information bits, $M = 2$, $D = 0$

$$\min_{n_1(g), n_2(g), P_1(g), P_2(g)} \mathbb{E}_g \left[-\frac{\mathcal{T}_h(a)}{\mathcal{T}_{h,max}} + \frac{\mathcal{E}(a)}{\mathcal{E}_{min}} \right]$$

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$$\min_{n_1(g), n_2(g), P_1(g), P_2(g)} \mathbb{E}_g \left[\underbrace{\left(1 - a\right) \left(\frac{B(1-\varepsilon_2)}{n_1 + n_2 \varepsilon_1} \right)}_{\mathcal{T}_h(a)} \underbrace{a (n_1 P_1 + n_2 P_2 \varepsilon_1)}_{\mathcal{E}(a)} \right] + \frac{\mathcal{E}(a)}{\mathcal{E}_{min}}$$

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$s.t. \quad n_1(g) + n_2(g) \leq N_\ell, \quad \forall g \quad \text{Latency}$

Optimization Problem, *Full-CSI*

Fixing: B information bits, $M = 2$, $D = 0$

$$\begin{aligned} & \underbrace{\left(1 - a\right)\left(\frac{B(1-\varepsilon_2)}{n_1+n_2\varepsilon_1}\right)}_{\mathcal{T}_h(a)} \underbrace{a(n_1P_1 + n_2P_2\varepsilon_1)}_{\mathcal{E}(a)} \\ & \min_{n_1(g), n_2(g), P_1(g), P_2(g)} \mathbb{E}_g \left[-\frac{\mathcal{T}_h(a)}{\mathcal{T}_{h,max}} + \frac{\mathcal{E}(a)}{\mathcal{E}_{min}} \right] \\ & \text{s.t. } n_1(g) + n_2(g) \leq N_\ell, \quad \forall g \quad \text{Latency} \\ & \mathbb{E}_g[\varepsilon_2(n_1(g), n_2(g), P_1(g), P_2(g), g)] \leq \varepsilon_{rel} \quad \text{Reliability} \end{aligned}$$

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Optimization Problem, *Statistical-CSI*

Fixing: B information bits, $M = 2$, $D = 0$

$$\begin{aligned} & \min_{\cancel{n_1(g)}, \cancel{n_2(g)}, \cancel{P_1(g)}, \cancel{P_2(g)}} \mathbb{E}_g \left[-\frac{\mathcal{T}_h(a)}{\mathcal{T}_{h,max}} + \frac{\mathcal{E}(a)}{\mathcal{E}_{min}} \right] \\ & s.t. \quad \cancel{n_1(g)} + \cancel{n_2(g)} \leq N_\ell, \quad \forall g \quad \text{Latency} \\ & \quad \mathbb{E}_g [\varepsilon_2(\cancel{n_1(g)}, \cancel{n_2(g)}, \cancel{P_1(g)}, \cancel{P_2(g)}, g)] \leq \varepsilon_{rel} \quad \text{Reliability} \\ & \quad \cancel{n_1(g)} \cancel{P_1(g)} + \cancel{n_2(g)} \cancel{P_2(g)} \leq E_b, \quad \forall g \quad \text{Energy} \\ & \quad \cancel{P_i(g)} \leq P_{max}, \quad i \in \{1, 2\} \quad \forall g \quad \text{Maximum Power} \end{aligned}$$

The **full-CSI** problem is an optimization problem with *functions as solution*:



Simplifying Assumption:



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Simplifying Assumption:



Reliability Constraint: $\mathbb{P}(g < g_{th}) + \mathbb{P}(g \geq g_{th})\varepsilon_{on} \leq \varepsilon_{rel}$

Optimization Variables: $g_{th}, (n_1, n_2, P_1, P_2) \quad \forall g \geq g_{th}$

Lemma to find feasibility set

Lemma: The solution of the problem:

$$\min_{n_1, \dots, n_M, P_1, \dots, P_M, M} \varepsilon_M(n_1, \dots, n_m, P_1, \dots, P_m, g)$$

$$\begin{aligned} \text{s.t.} \quad & \sum_{i=1}^M n_i \leq N_\ell \\ & \sum_{i=1}^M n_i P_i \leq E_b \end{aligned}$$

Given hard energy and latency constraint, using all the resources on one packet minimizes the error.*

is $M = 1$ with $(n_1, P_1) = (N_\ell, \frac{E_b}{N_\ell})$.

*In [3] A. Avranas, M. Kountouris, and P. Ciblat, "Energy-latency tradeoff in ultra-reliable low-latency communication with retransmissions," IEEE JSAC., Oct. 2018 we used soft constraints for a similar problem leading to quite different behavior.

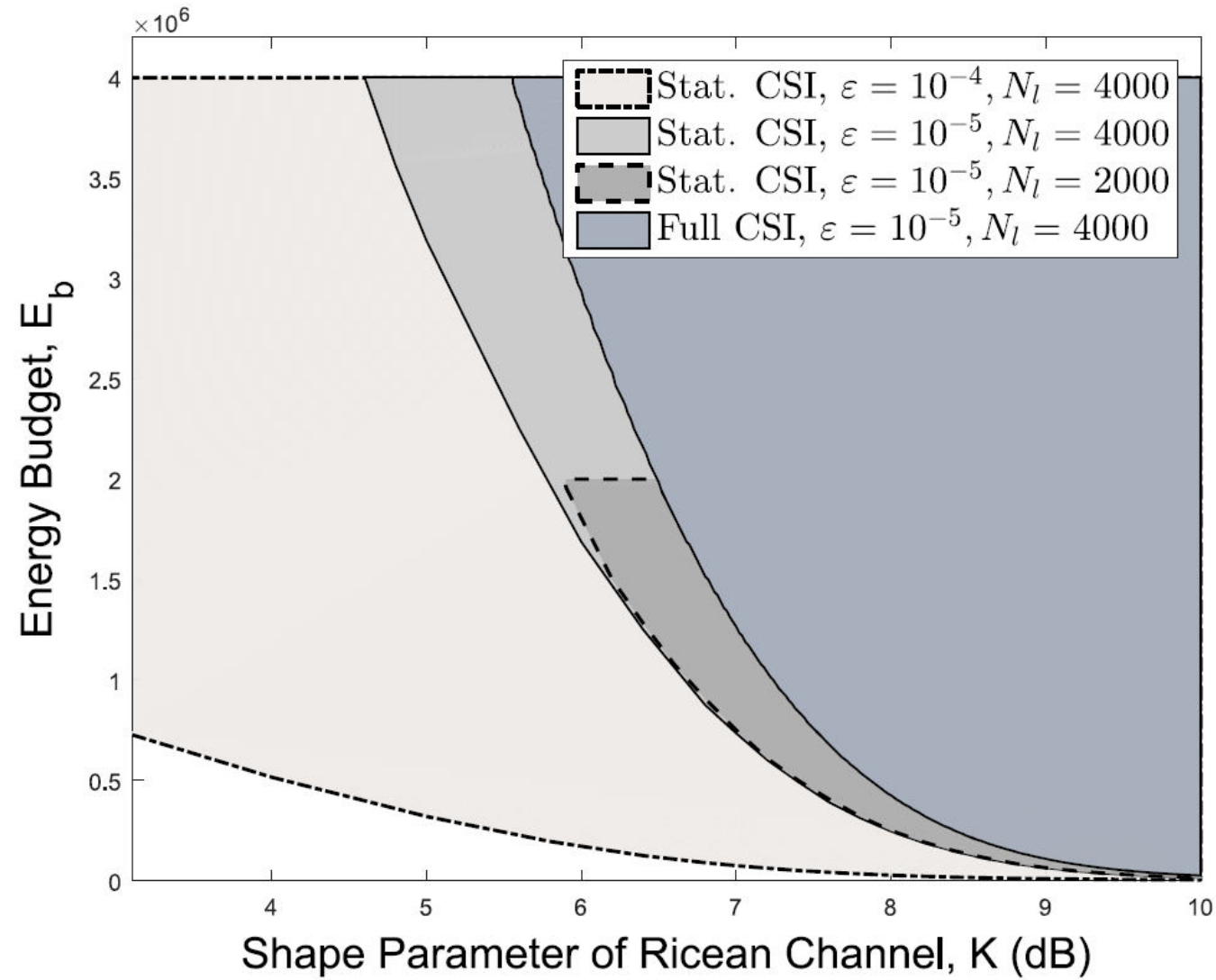


Fig. 1: Feasibility region for different channel, $B = 32\text{Bytes}$, maximum energy budget $E_b = P_{\max}N_\ell$ with $P_{\max} = 30\text{dB}$.

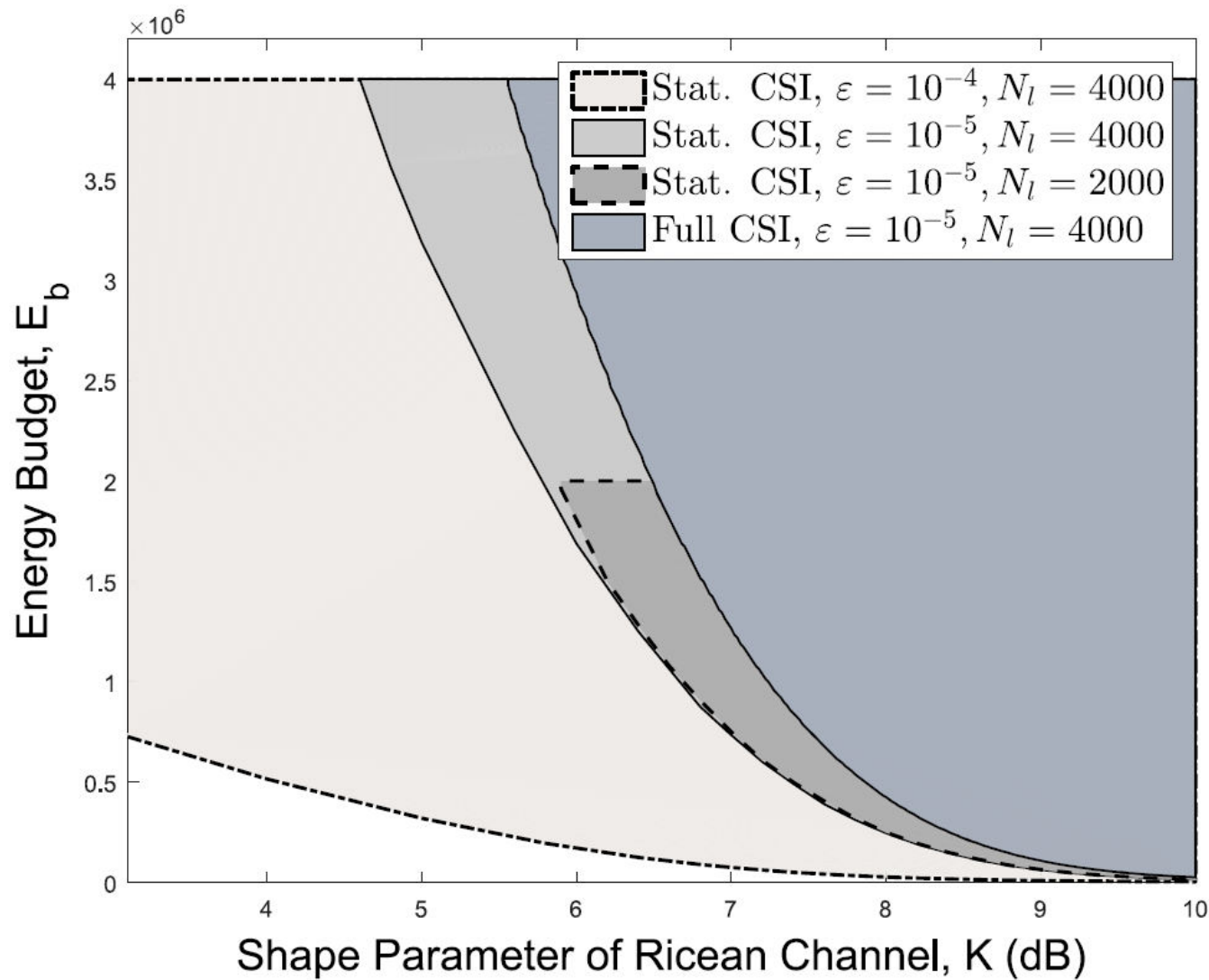


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The trade-off and Pareto frontiers

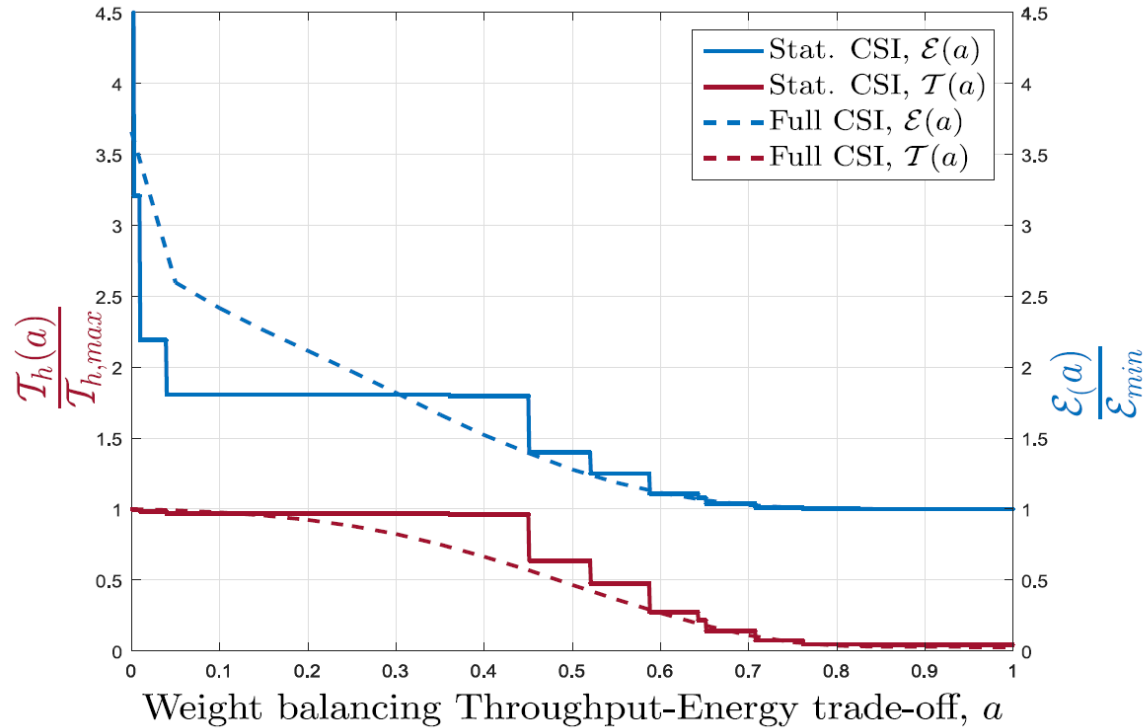


Fig. 2: Throughput and energy relative to their optimal value for Rician channel with $K = 7\text{dB}$, $B = 32\text{Bytes}$, $\varepsilon_{\text{rel}} = 10^{-5}$ and maximum energy $E_b = P_{\text{max}}N_\ell$ with $P_{\text{max}} = 30\text{dB}$ and $N_\ell = 4000$.

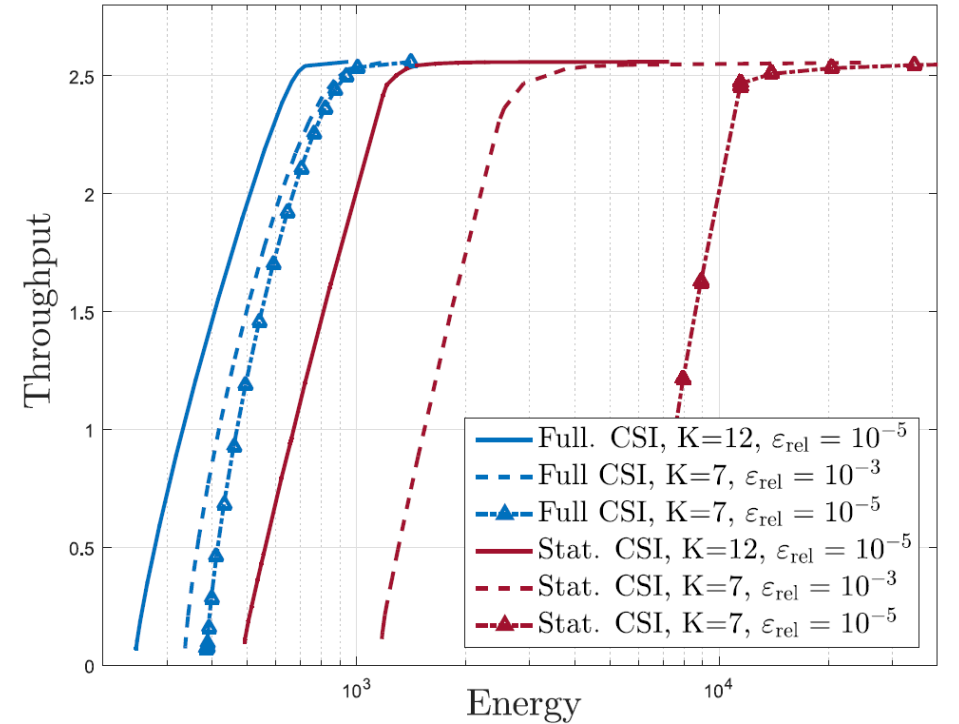


Fig. 3: Pareto frontier for throughput and energy, with $E_b = P_{\text{max}}N_\ell$, $P_{\text{max}} = 30\text{dB}$, and $N_\ell = 4000$.

IR-HARQ vs One-Shot

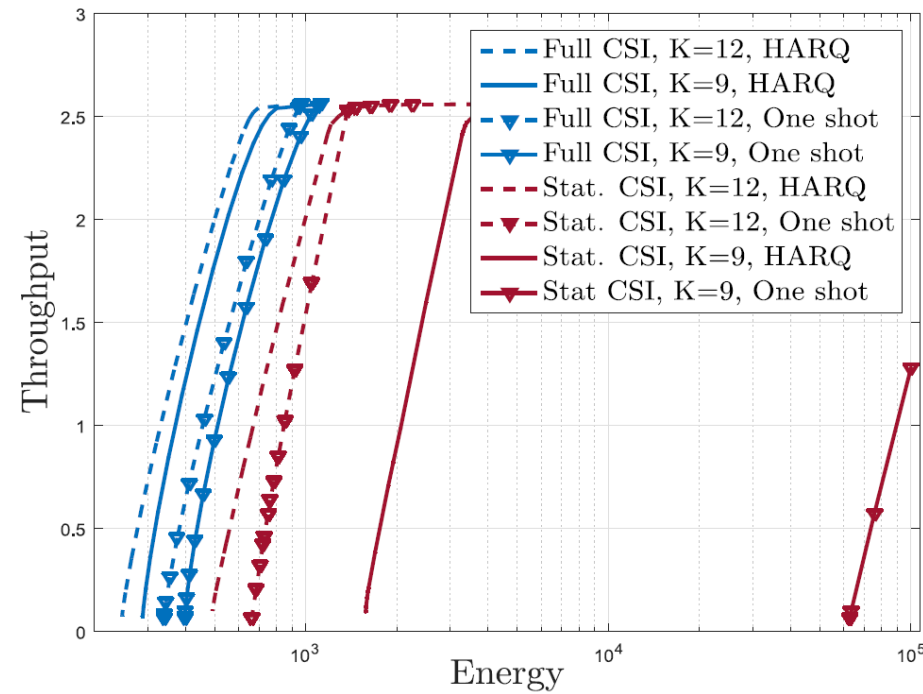


Fig. 4: Pareto frontier for throughput and energy when HARQ or one shot transmission is used, with $E_b = P_{\max} N_\ell$, $P_{\max} = 1000(30dB)$, and $N_\ell = 4000$.

Thank you

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