

EE16B

Designing Information Devices and Systems II

Lecture 11B

Discrete Signals and Systems

Intro

- Mitterm II yesterday...
- Last time:
 - Sampling Theorem
 - Aliasing
 - Discrete Signals
- Today
 - Discrete systems

Complex Frequencies

- Sinusoids are sums of left and right rotating complex exponentials

$$2 \cos(\omega t) = e^{j\omega t} + e^{-j\omega t}$$

“Positive” and “Negative” frequencies

Discrete frequencies with period N:

$$y[n] = e^{j2\pi n/N}$$

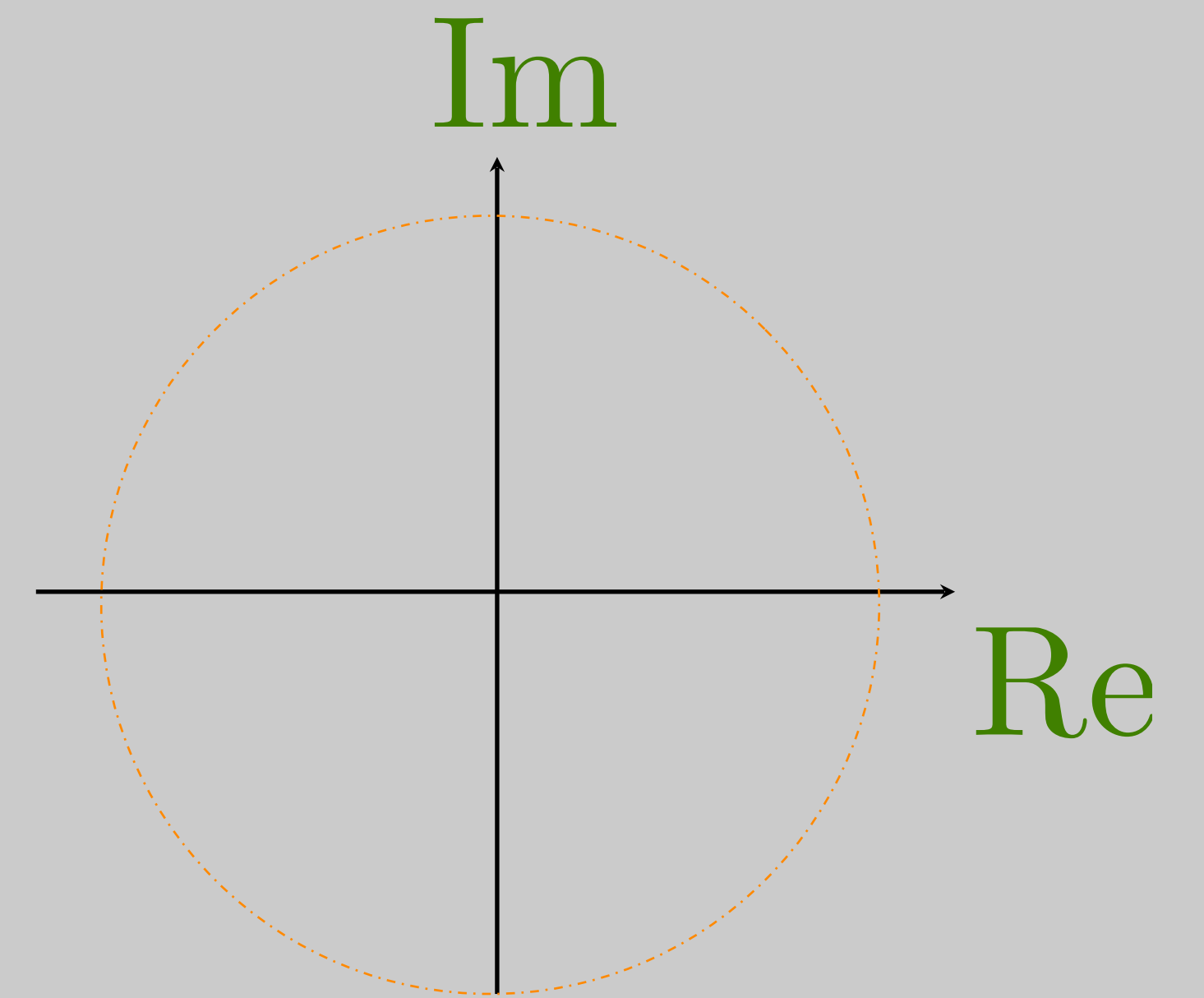
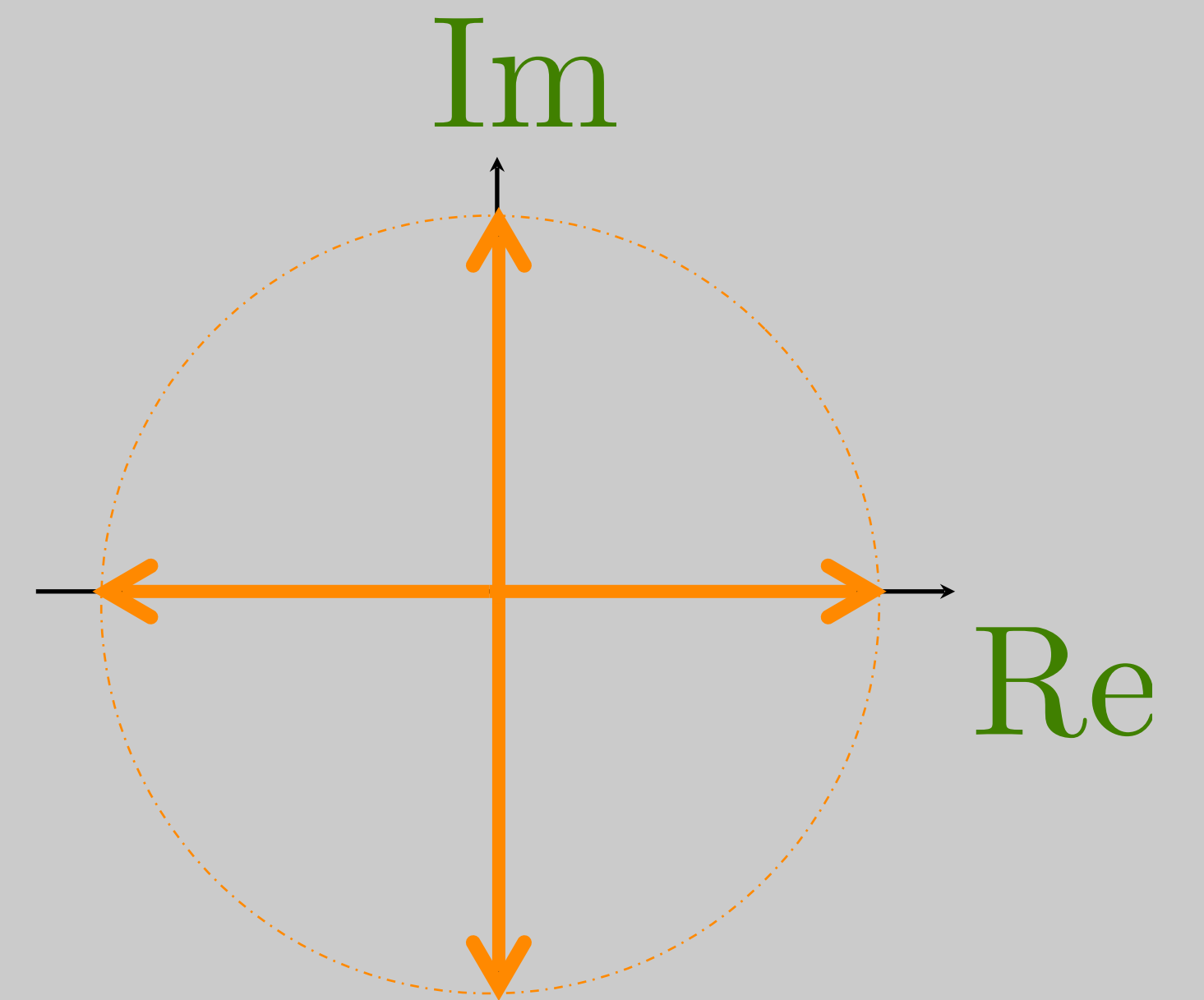
$$W_N \triangleq e^{j2\pi/N} \Rightarrow y[n] = W_N^n$$

Complex Frequencies

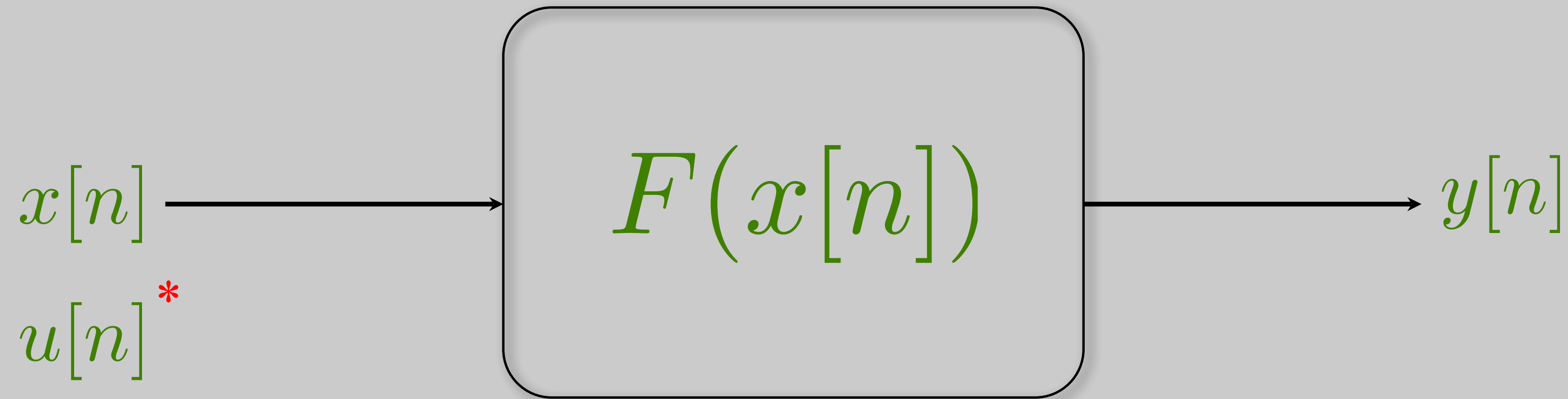
$$W_N \triangleq e^{j2\pi/N} \Rightarrow y[n] = W_N^n$$

• $N = 4$ $y[n] = W_4^n$

• $N = 6$, neg. freq. $y[n] = W_6^{-n}$



Discrete Time Systems



- What Properties?
 - Causality
 - Linearity
 - Stability
 - Time/shift invariance

***WARNING:** Going to interchange $x[n]$ and $u[n]$ as inputs

$\vec{x}[n]$ will be a state, not input

$U[n]$ is unit step, not to be confused with $u[n]$

Properties of D.T. Systems

- Causality:
 - $y[n_0]$ depends only on $x[n]$ for $\infty \leq n \leq n_0$

Causal?

$$\vec{x}[n+1] = A\vec{x}[n] + Bu[n]$$

$$y[n] = C\vec{x}[n]$$

$$\vec{x}[n] = A^n \vec{x}[0] + \sum_{k=0}^{n-1} A^{n-1-k} Bu[k]$$

$$y[n] = CA^n \vec{x}[0] + \sum_{k=0}^{n-1} CA^{n-1-k} Bu[k]$$

Properties of D.T. Systems

$$y[n] = F\{x[n]\}$$

- Linearity
 - Homogeneity: scaling the input, scales the output

$$F\{ax[n]\} = aF\{x[n]\} = ay[n]$$

Properties of D.T. Systems

$$y[n] = F\{x[n]\}$$

- Linearity

- Homogeneity: scaling the input, scales the output

$$F\{ax[n]\} = aF\{x[n]\} = ay[n]$$

- Superposition: sum of inputs \Rightarrow sum of outputs

$$F\{x_1[n] + x_2[n]\} = F\{x_1[n]\} + F\{x_2[n]\} = y_1[n] + y_2[n]$$

Example:

$$\vec{x}[n+1] = A\vec{x}[n] + Bu[n]$$

$$y[n] = C\vec{x}[n]$$

Linear?

$$y[n] = CA^n\vec{x}[0] + \sum_{k=0}^{n-1} CA^{n-1-k}Bu[k]$$

Properties of D.T. Systems

$$y[n] = F\{x[n]\}$$

- BIBO Stability
 - If $x[n]$ is bounded, then $y[n]$ is bounded

$$|x[n]| < M < \infty \quad \forall n \Rightarrow \quad |y[n]| < P < \infty \quad \forall n$$

BIBO stable?

$$y[n] = CA^n \vec{x}[0] + \sum_{k=0}^{n-1} CA^{n-1-k} Bu[k]$$

Properties of D.T. Systems

$$y[n] = F\{x[n]\}$$

- Time Invariance: Shifted input \Rightarrow shifted output

$$y[n - n_0] = F\{x[n - n_0]\}$$

Time Invariant?

$$\vec{x}[n + 1] = A\vec{x}[n] + Bu[n]$$

$$y[n] = C\vec{x}[n]$$

$$y[n] = CA^n\vec{x}[0] + CBu[n - 1] + CABu[n - 2] + \cdots + CA^{n-1}Bu[0]$$

Linear Time Invariant Systems

- Linear Time/Shift Invariant (LTI/LSI) systems are completely characterized by their impulse response $h[n]$



$h[n]$ is the “DNA” of an LTI system

Knowing $h[n]$ is enough to find $y[n]$ for ANY $x[n]$!

Linear Time Invariant Systems



- Decompose $x[n]$:

$$x[n] = \sum_{m=-\infty}^{\infty} x[m] \delta[n - m]$$
$$= \begin{cases} 1 & n = m \\ 0 & n \neq m \end{cases}$$

- Compute output:

$$y[n] = \sum_{m=-\infty}^{\infty} x[m] h[n - m] = x[n] * h[n]$$

Convolution sum

Sum of weighted, delayed impulse responses!

Example:

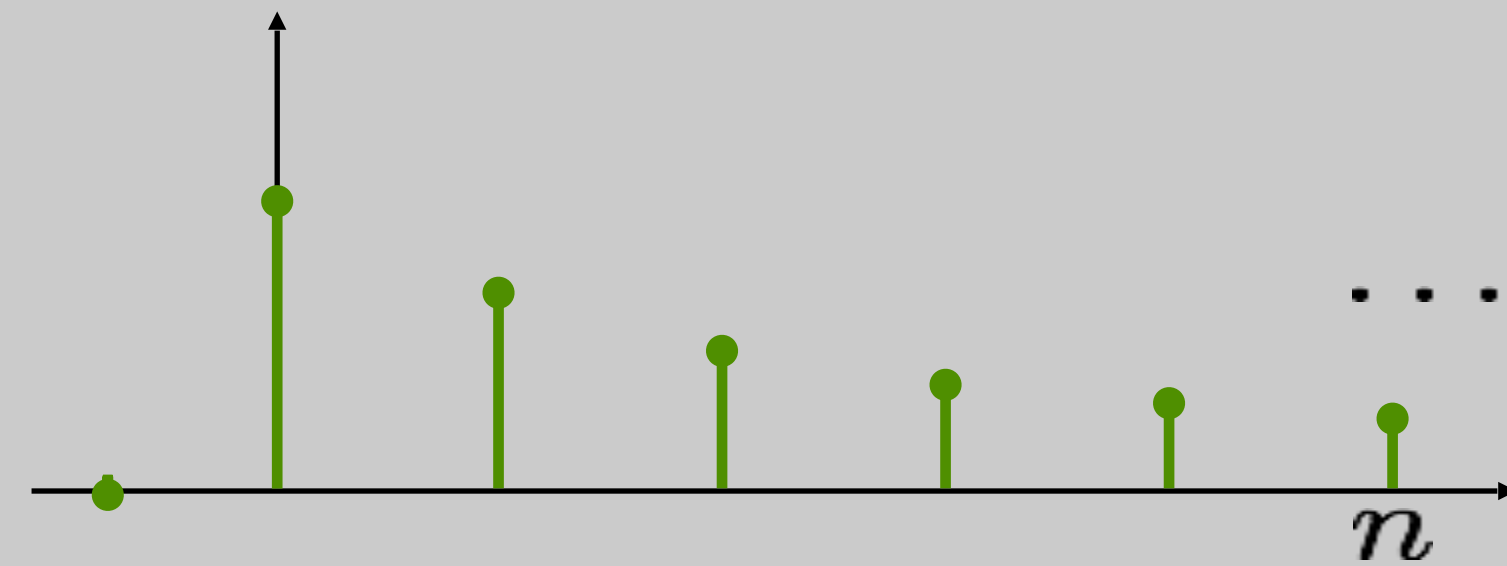
$$y[n] = ay[n-1] + x[n]$$

$$h[n] = \begin{cases} a^n & n \geq 0 \\ 0 & n < 0 \end{cases}$$

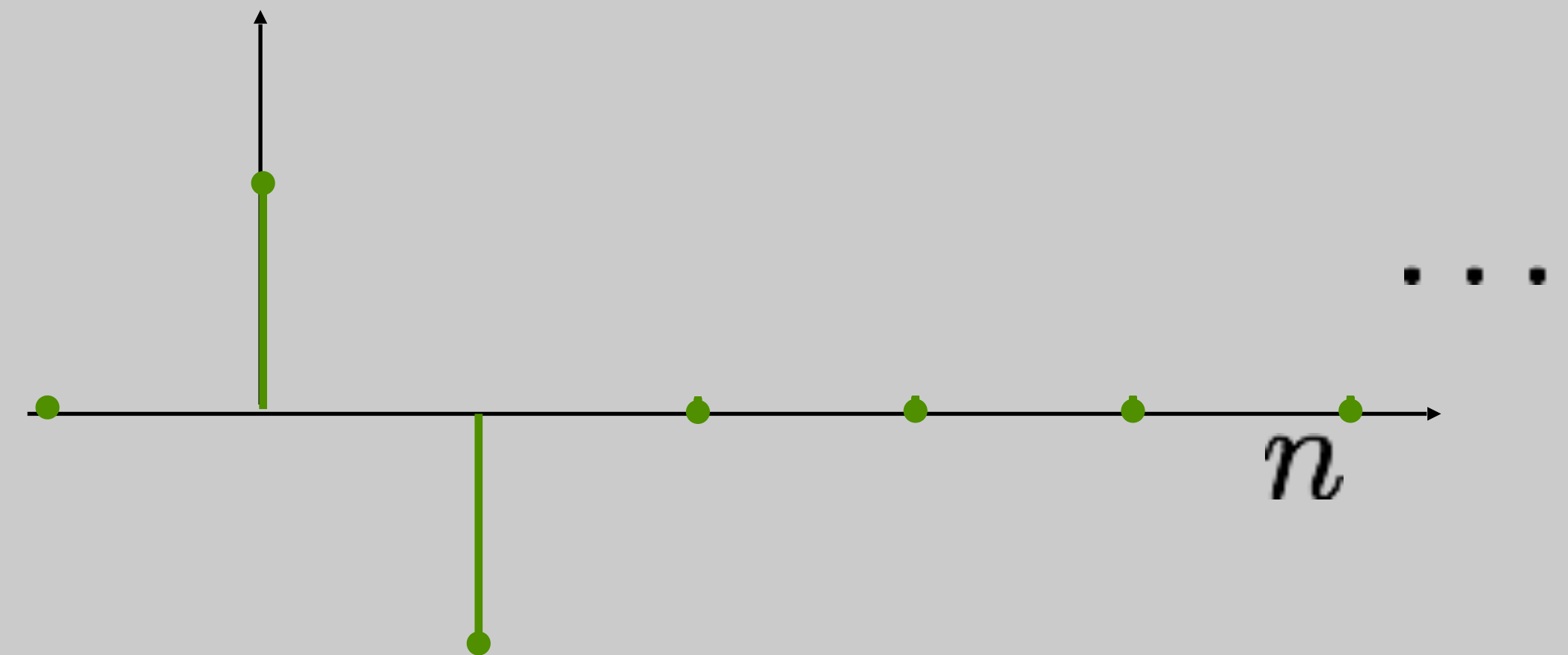
$$y[n] = x[n] - x[n-1]$$

$$h[n] = \delta[n] - \delta[n-1]$$

Infinite impulse response (IIR)



finite impulse response (FIR)

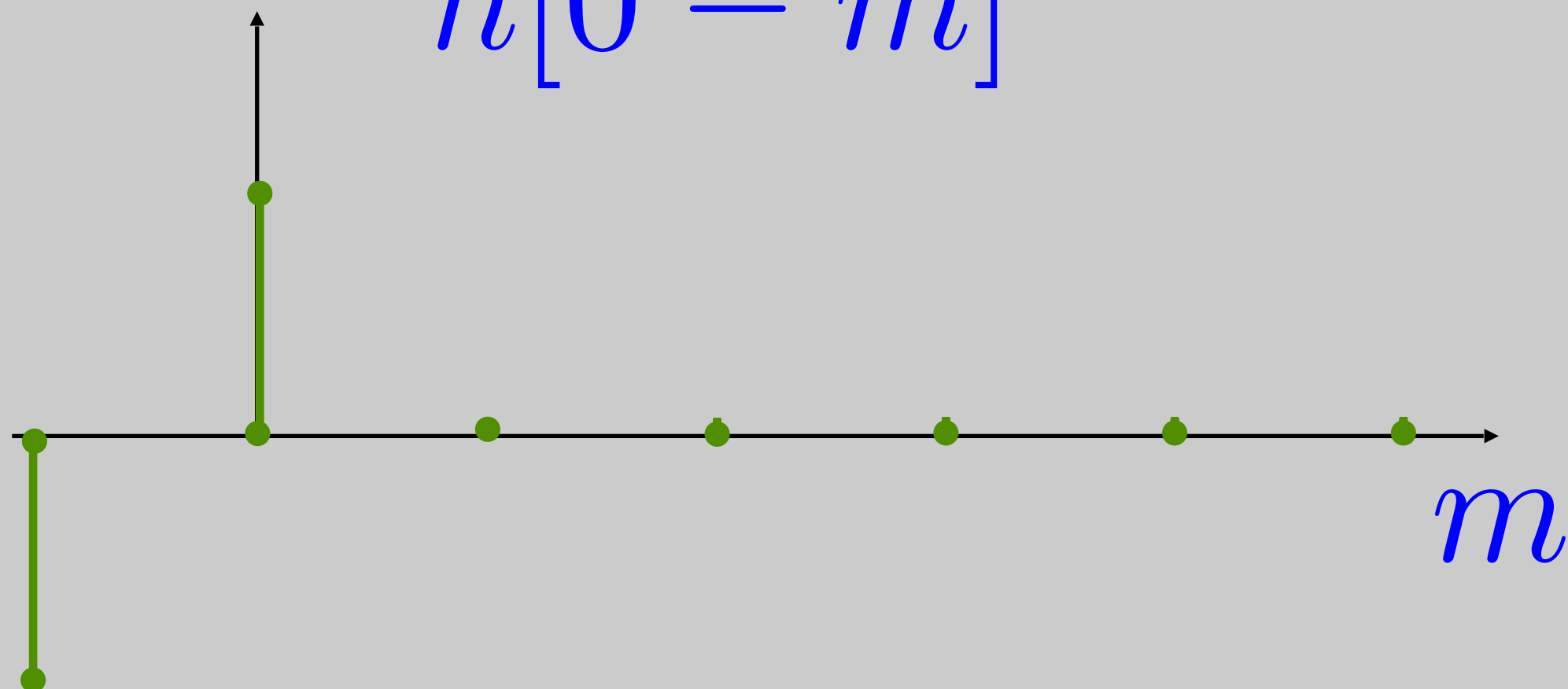


Convolution Sum

$$y[n] = \sum_{m=-\infty}^{\infty} x[m]h[n-m]$$

- What is $h[n-m]$ for different n 's?

$$h[0-m]$$



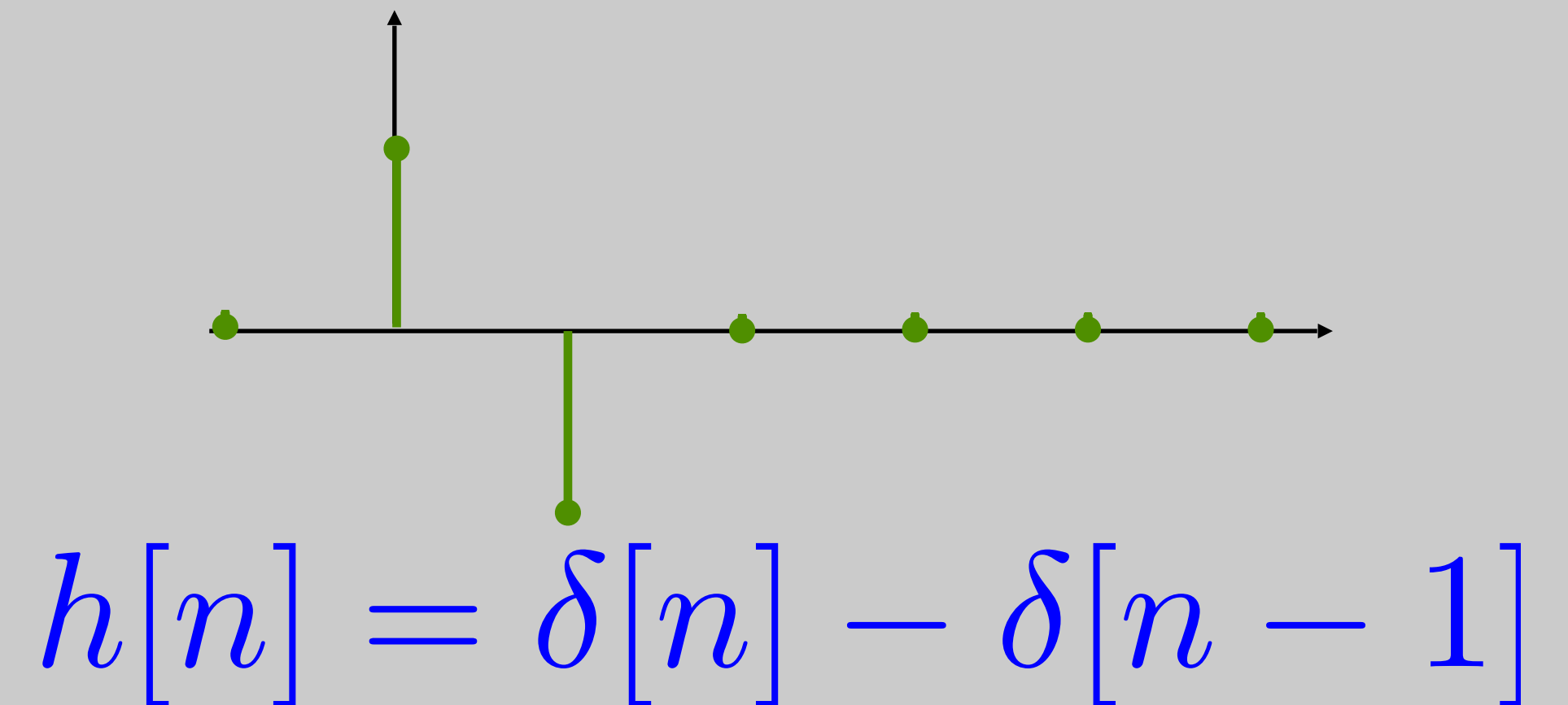
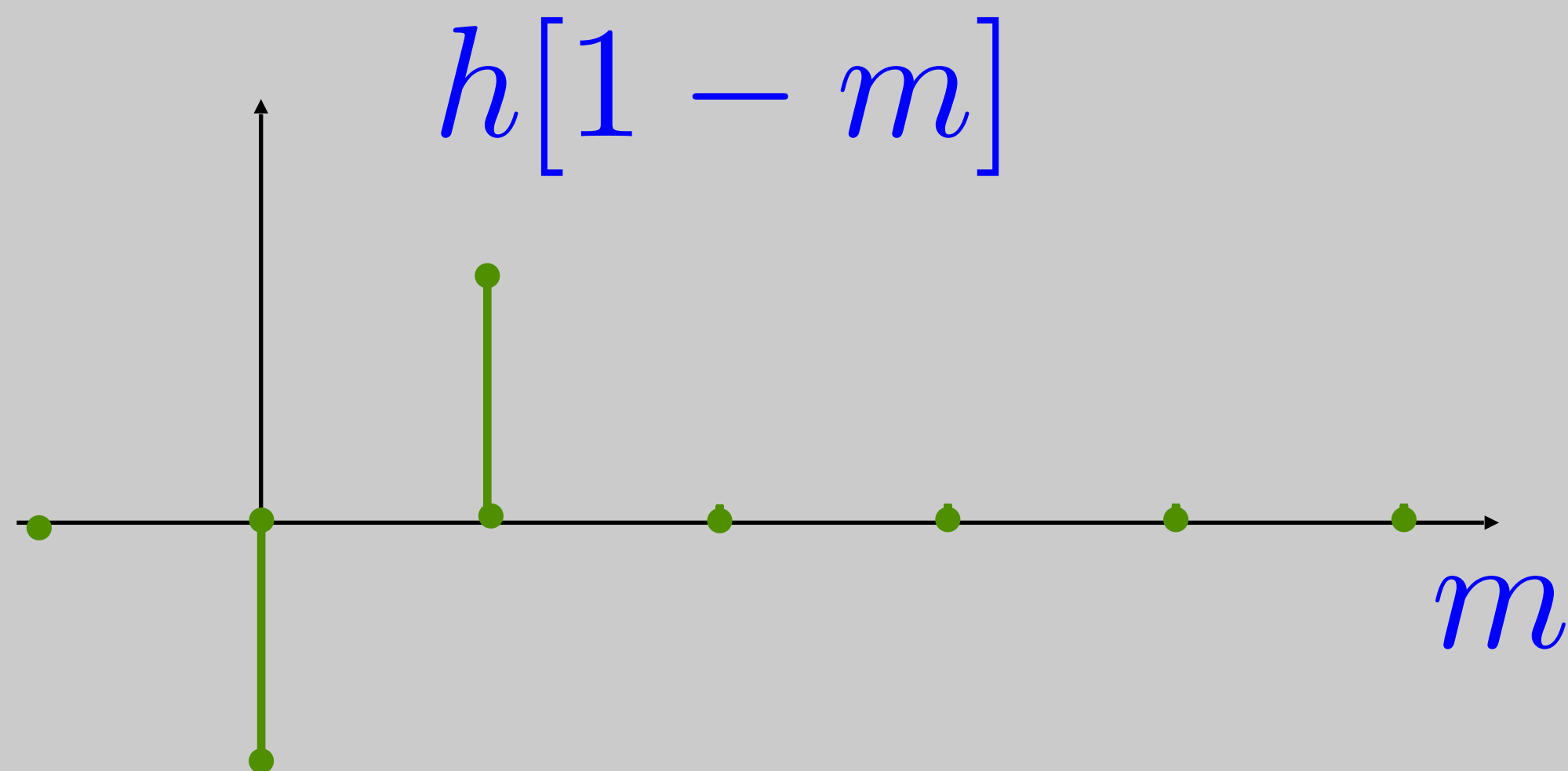
A discrete-time plot of the signal $h[n]$ versus n . The horizontal axis is indicated by a rightward arrow. The vertical axis is indicated by an upward arrow. The signal is plotted as green dots connected by vertical lines. It has a value of 1 at $n = 0$ and a value of -1 at $n = 1$. For all other integer values of n (-1, 2, 3, 4, 5), the signal value is 0.

$$h[n] = \delta[n] - \delta[n-1]$$

Convolution Sum

$$y[n] = \sum_{m=-\infty}^{\infty} x[m]h[n-m] = x[n] * h[n]$$

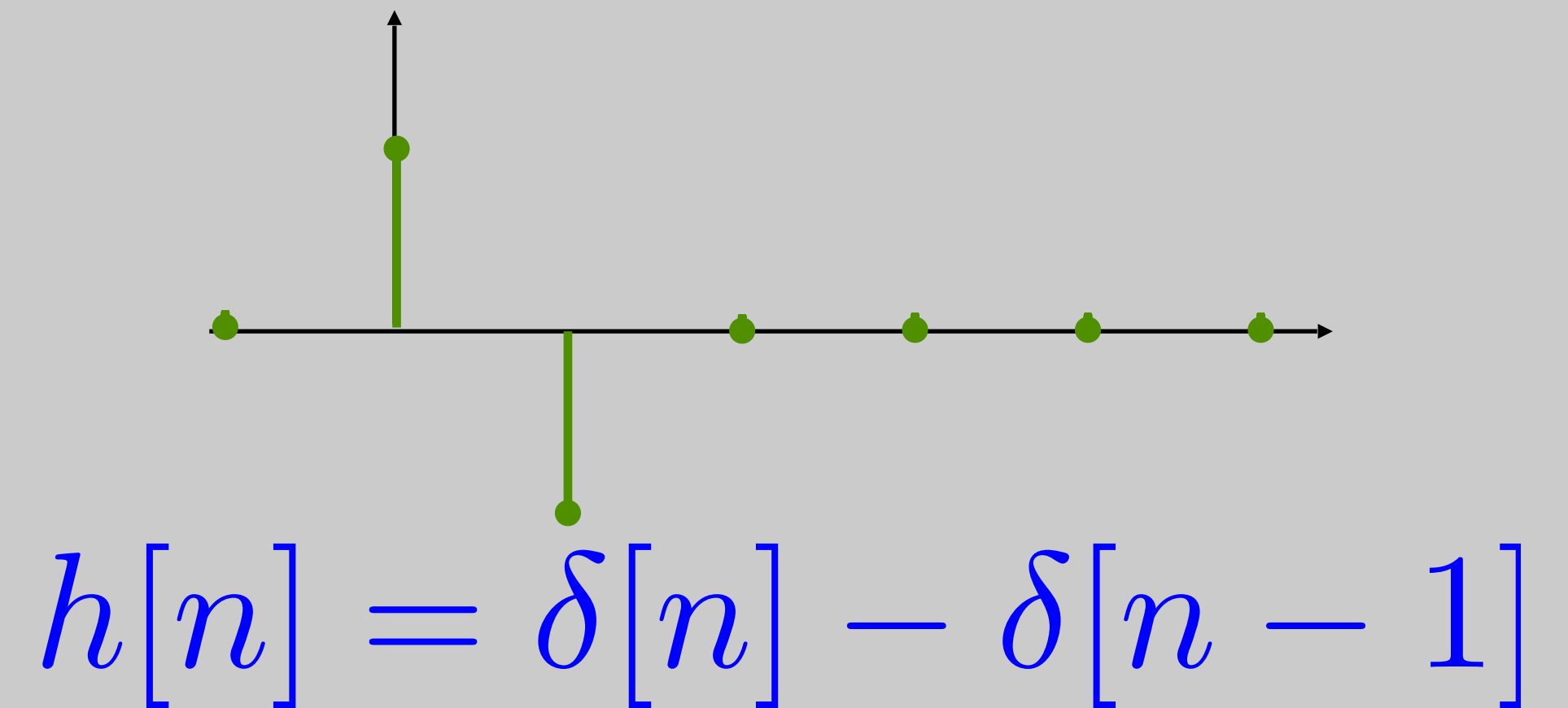
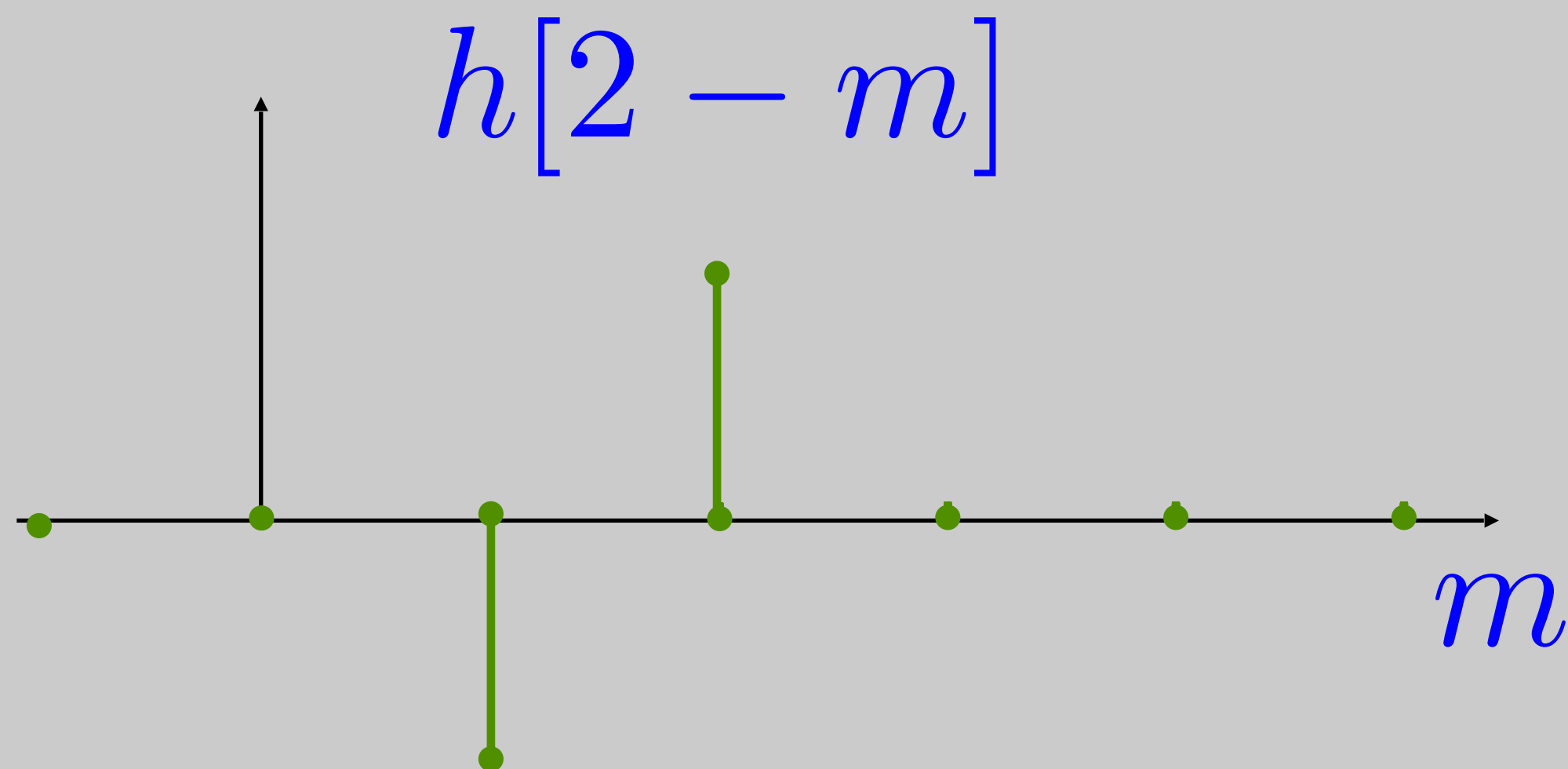
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Convolution Sum

$$y[n] = \sum_{m=-\infty}^{\infty} x[m]h[n-m]$$

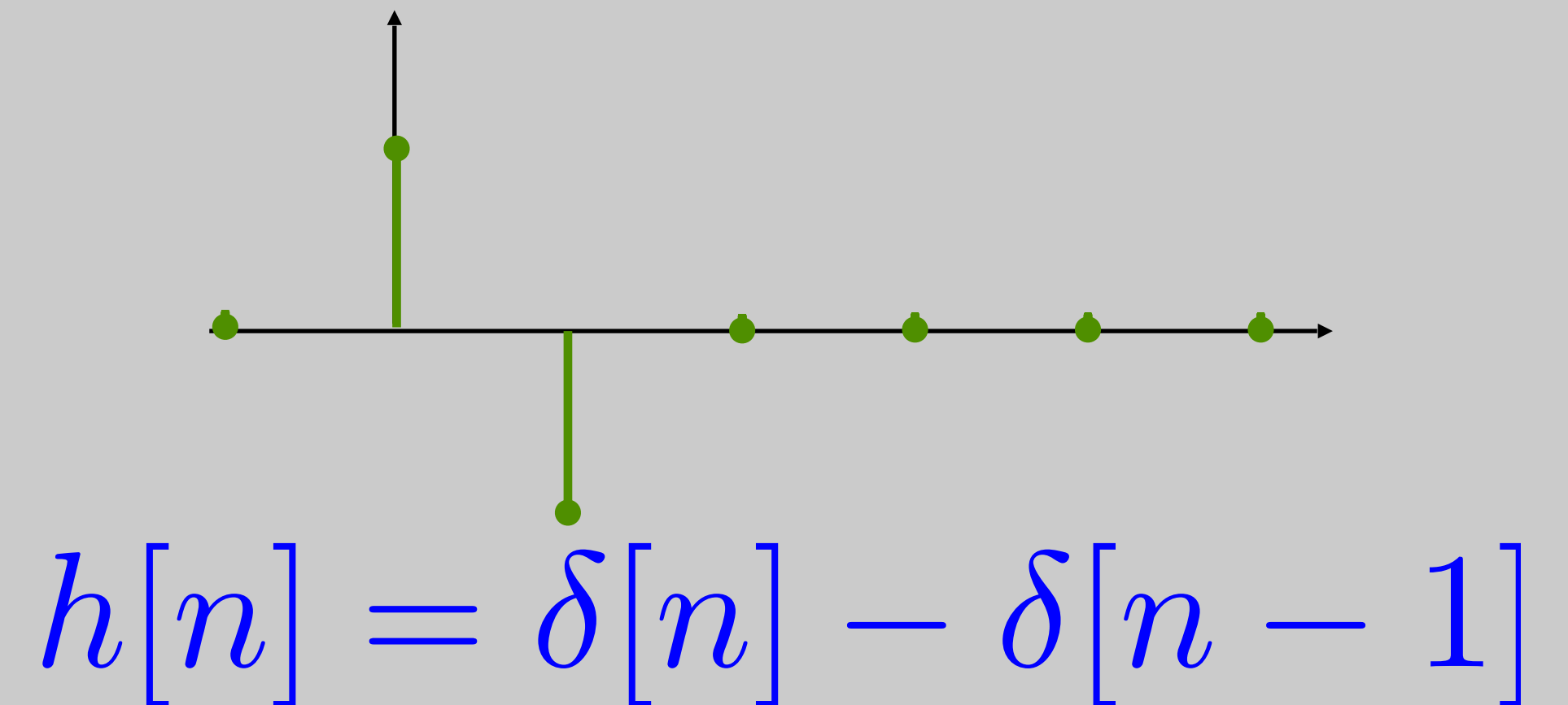
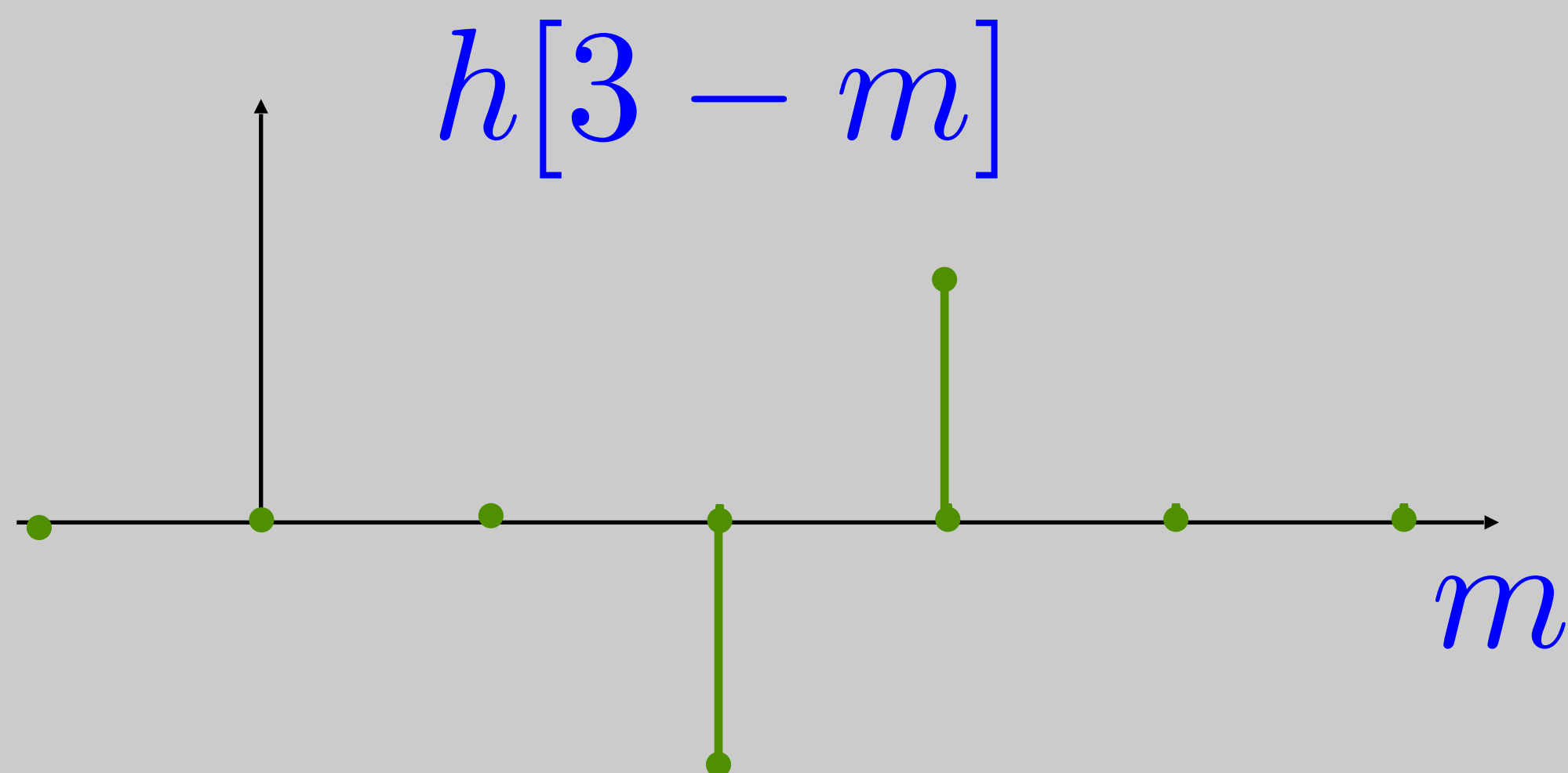
- What is $h[n-m]$ for different n 's?



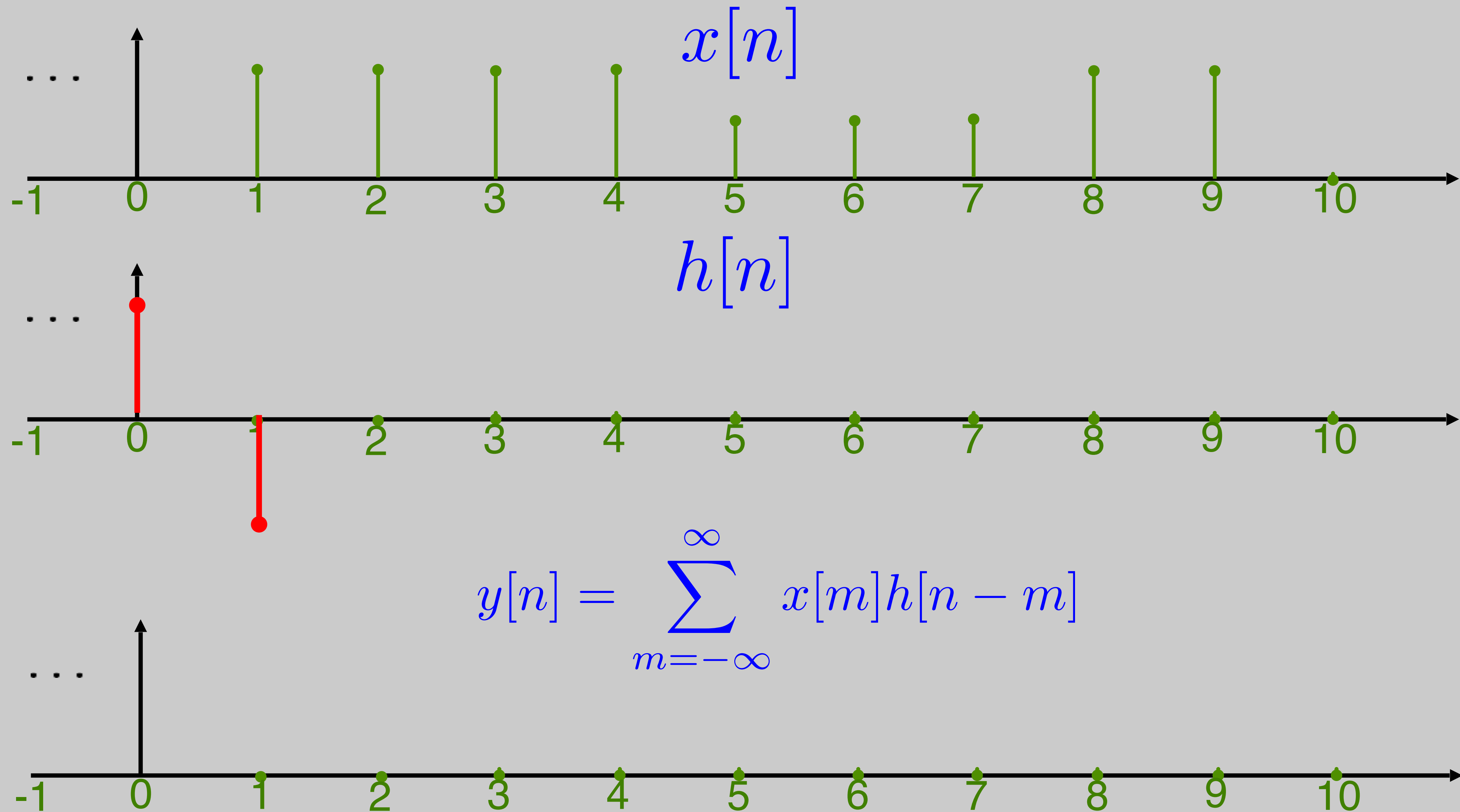
Convolution Sum

$$y[n] = \sum_{m=-\infty}^{\infty} x[m]h[n-m]$$

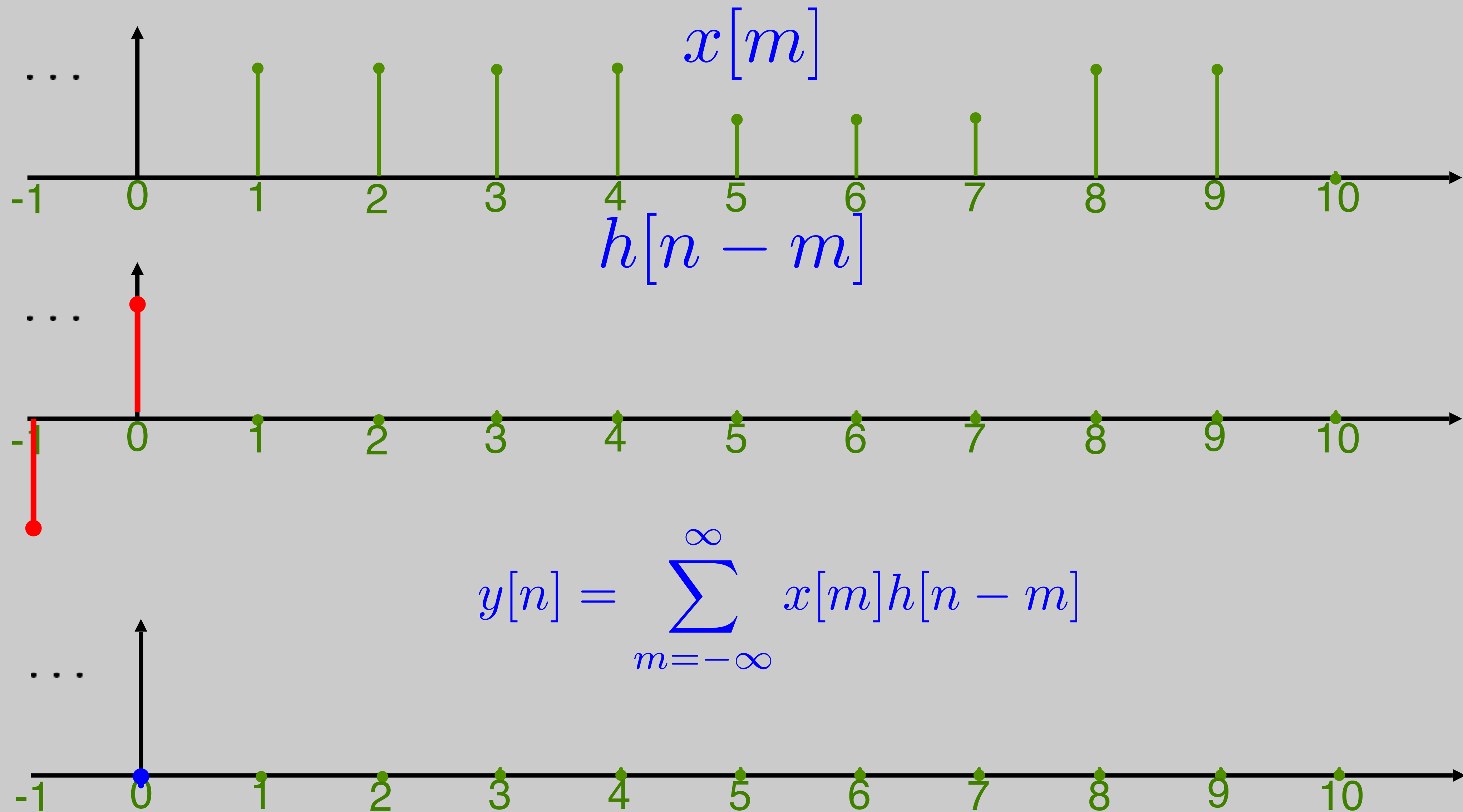
- What is $h[n-m]$ for different n 's?



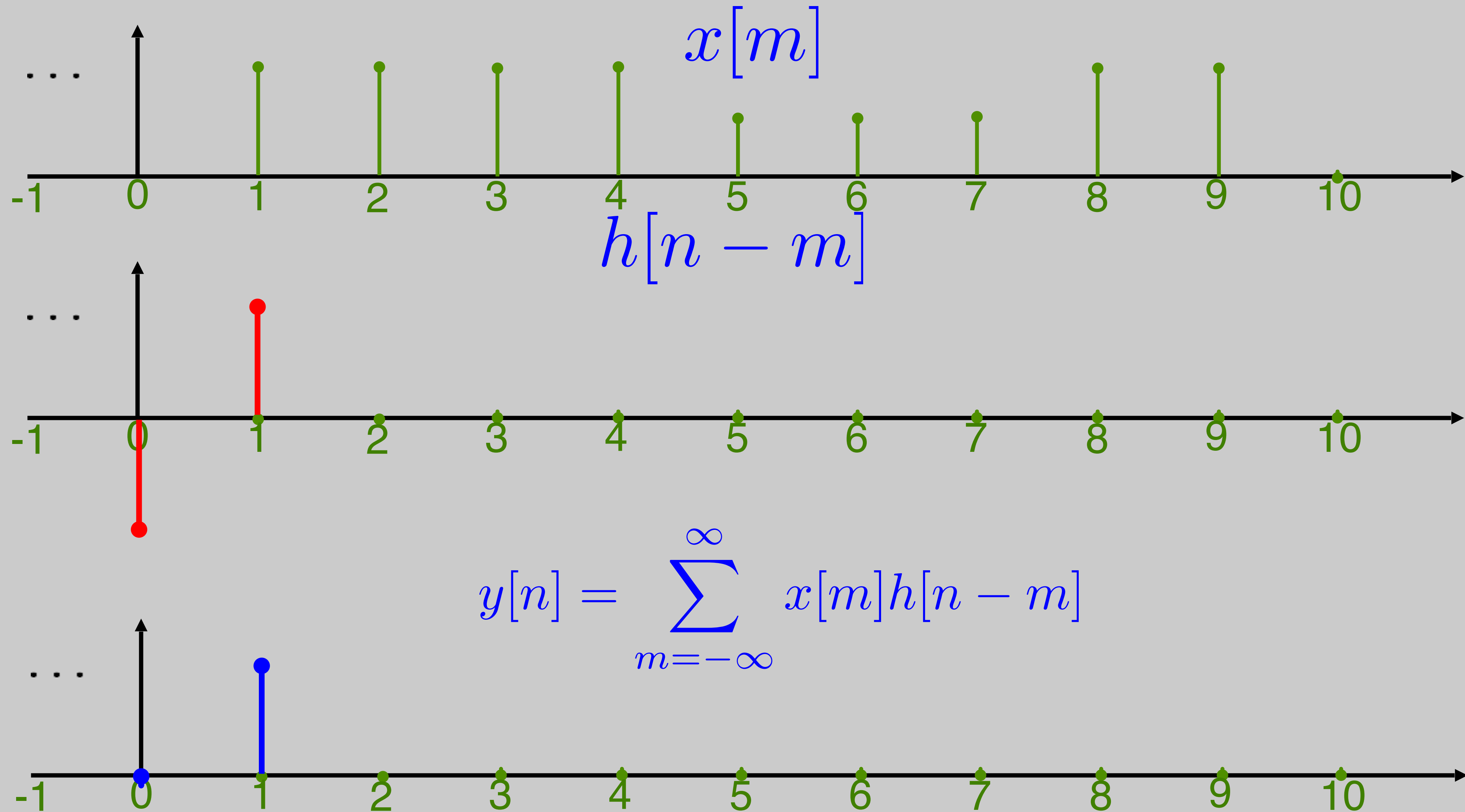
Graphical Example of Convolution



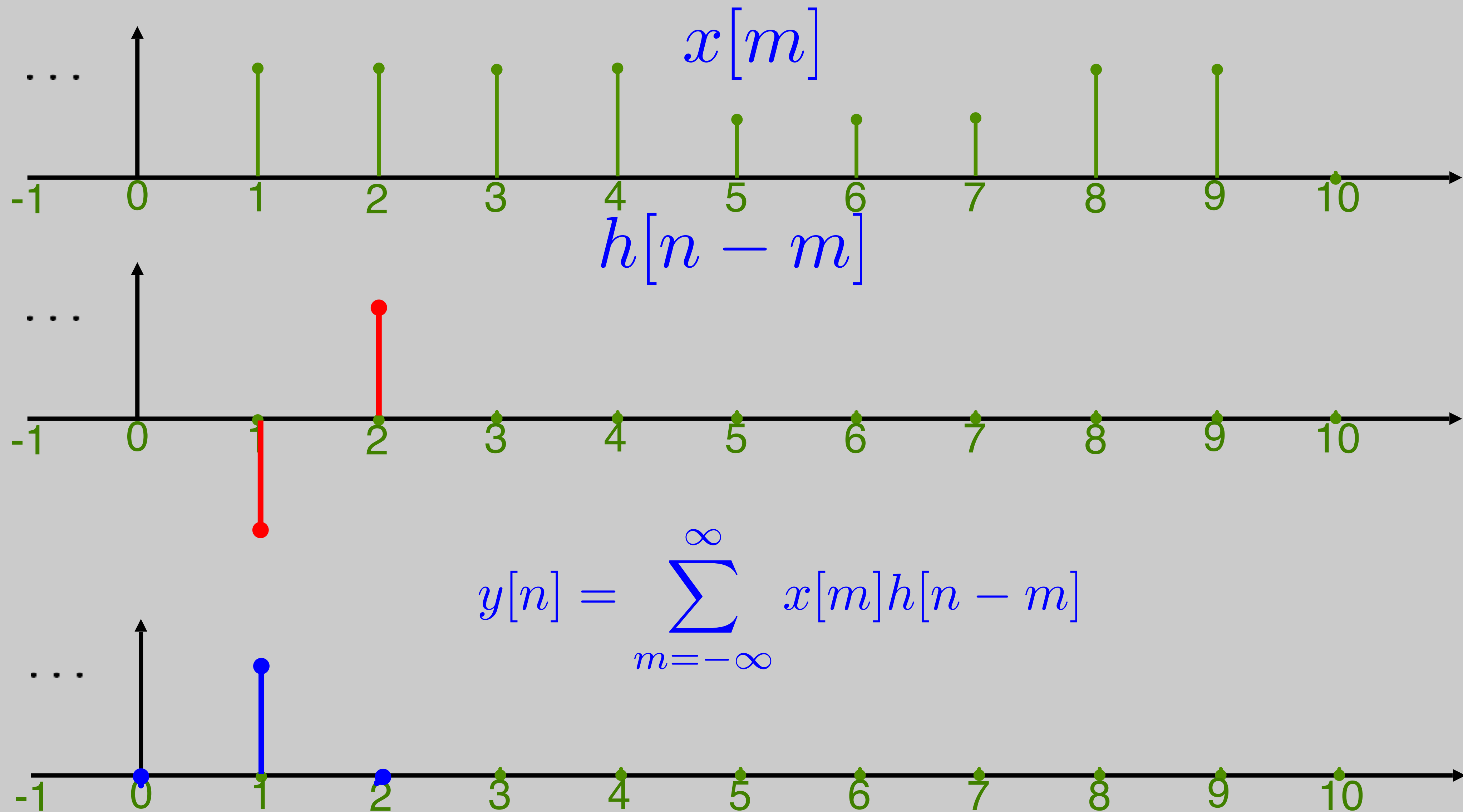
Graphical Example of Convolution



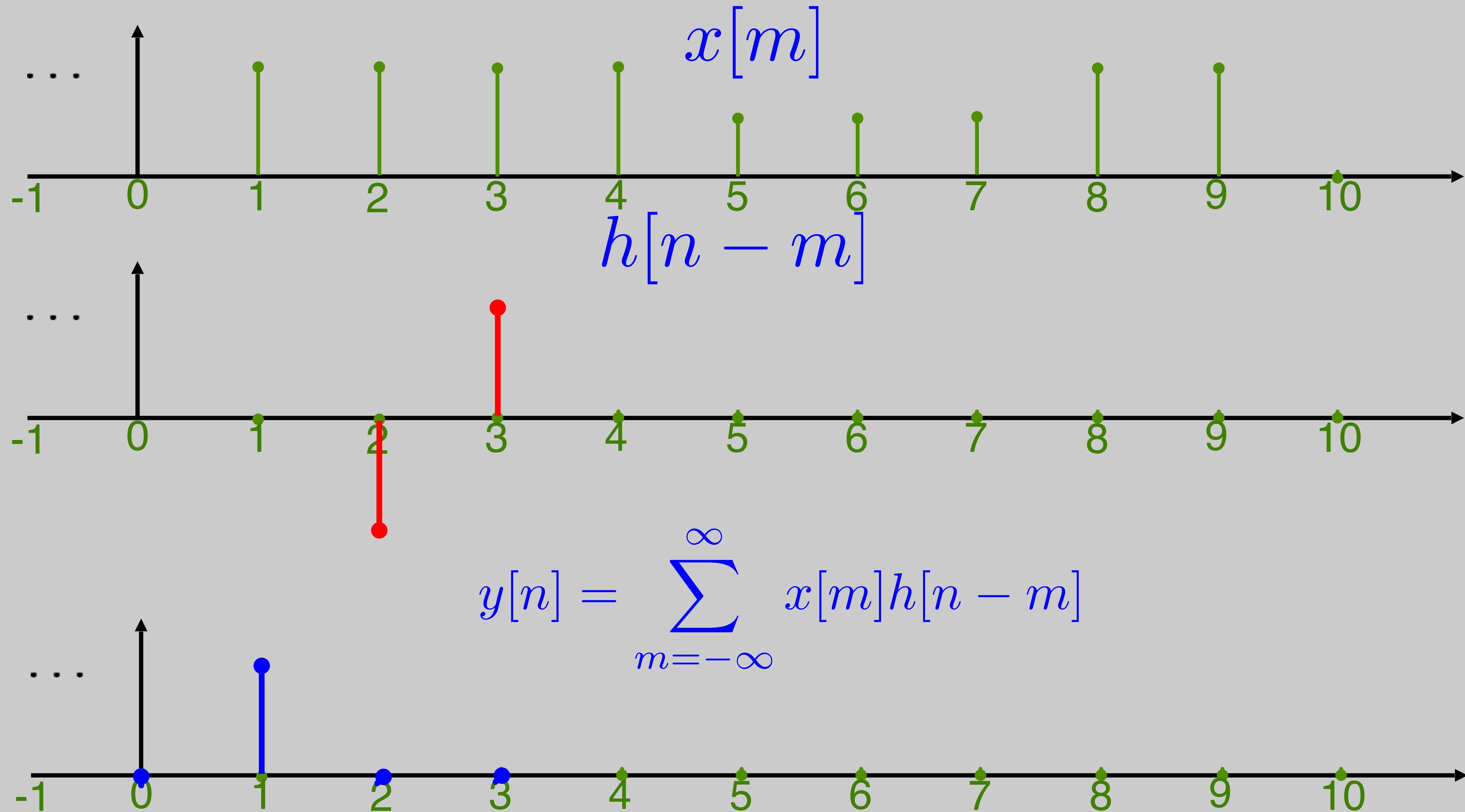
Graphical Example of Convolution



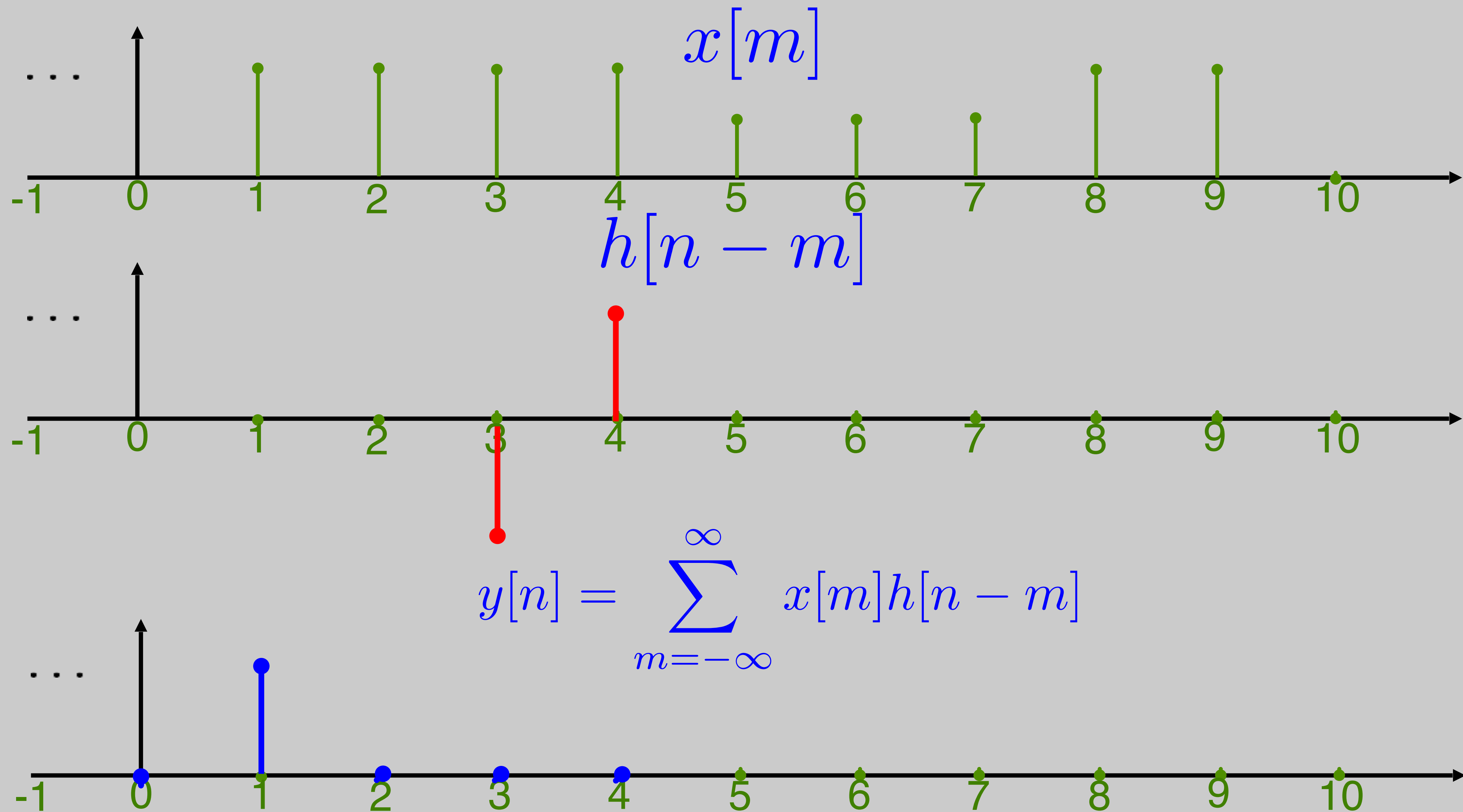
Graphical Example of Convolution



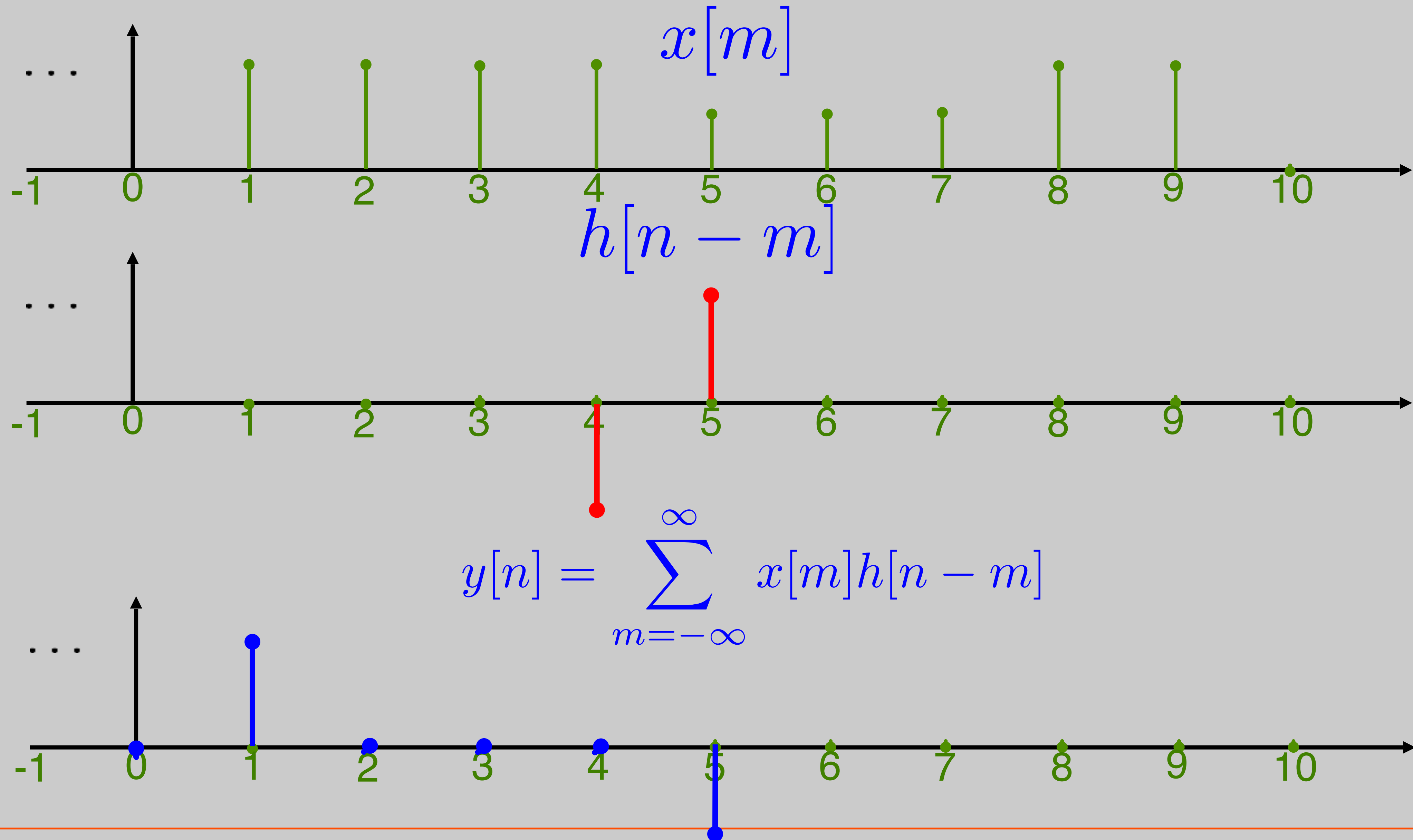
Graphical Example of Convolution



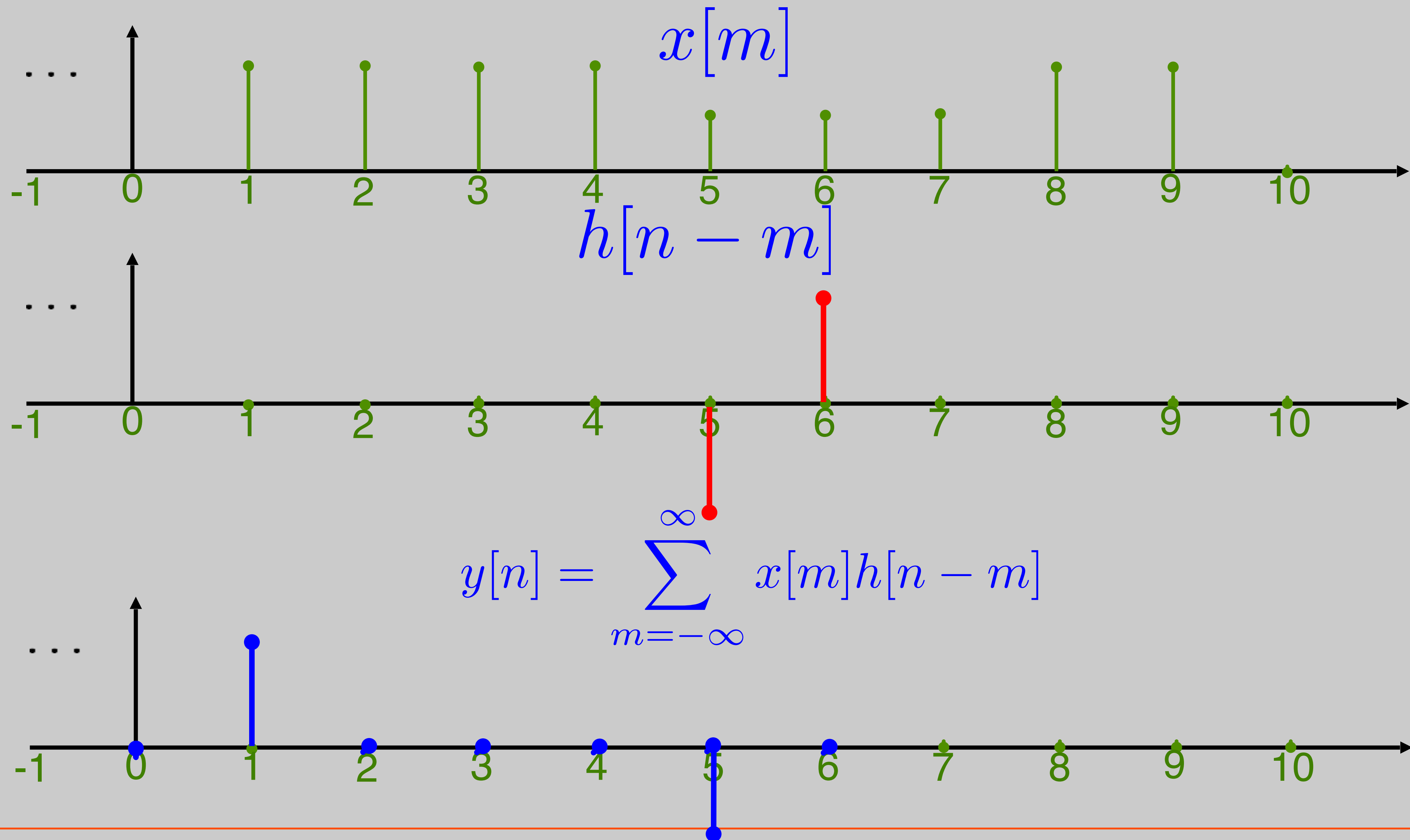
Graphical Example of Convolution



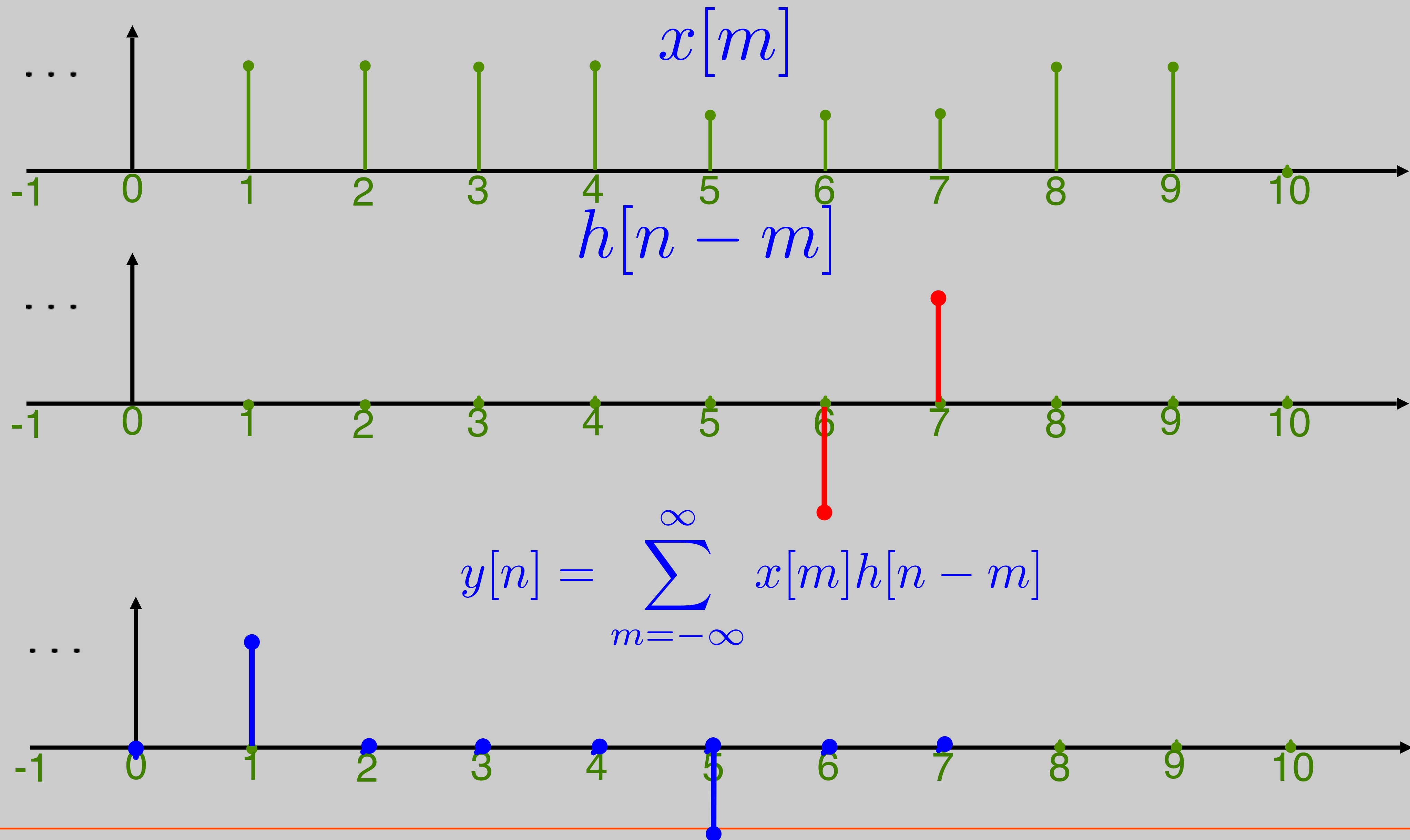
Graphical Example of Convolution



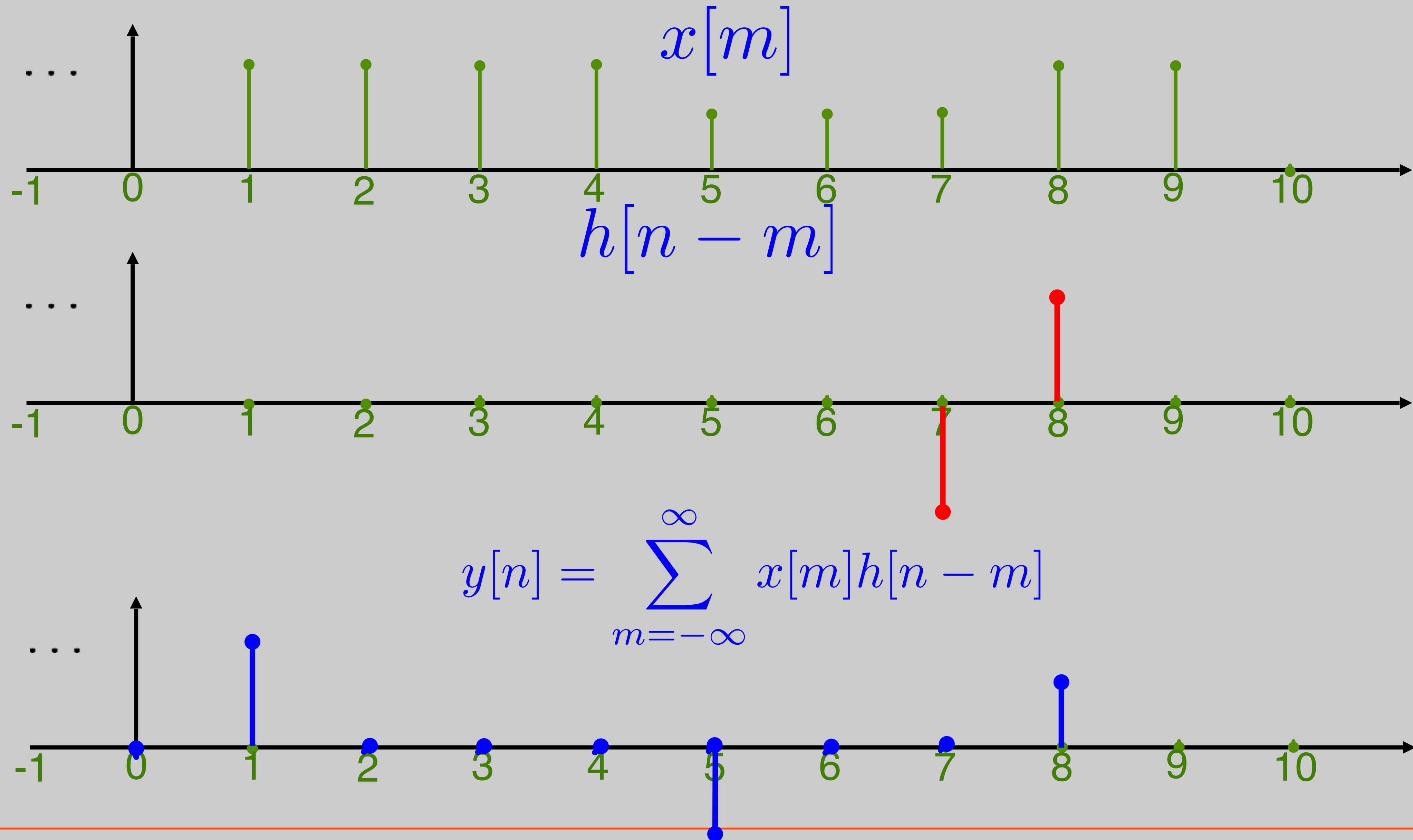
Graphical Example of Convolution



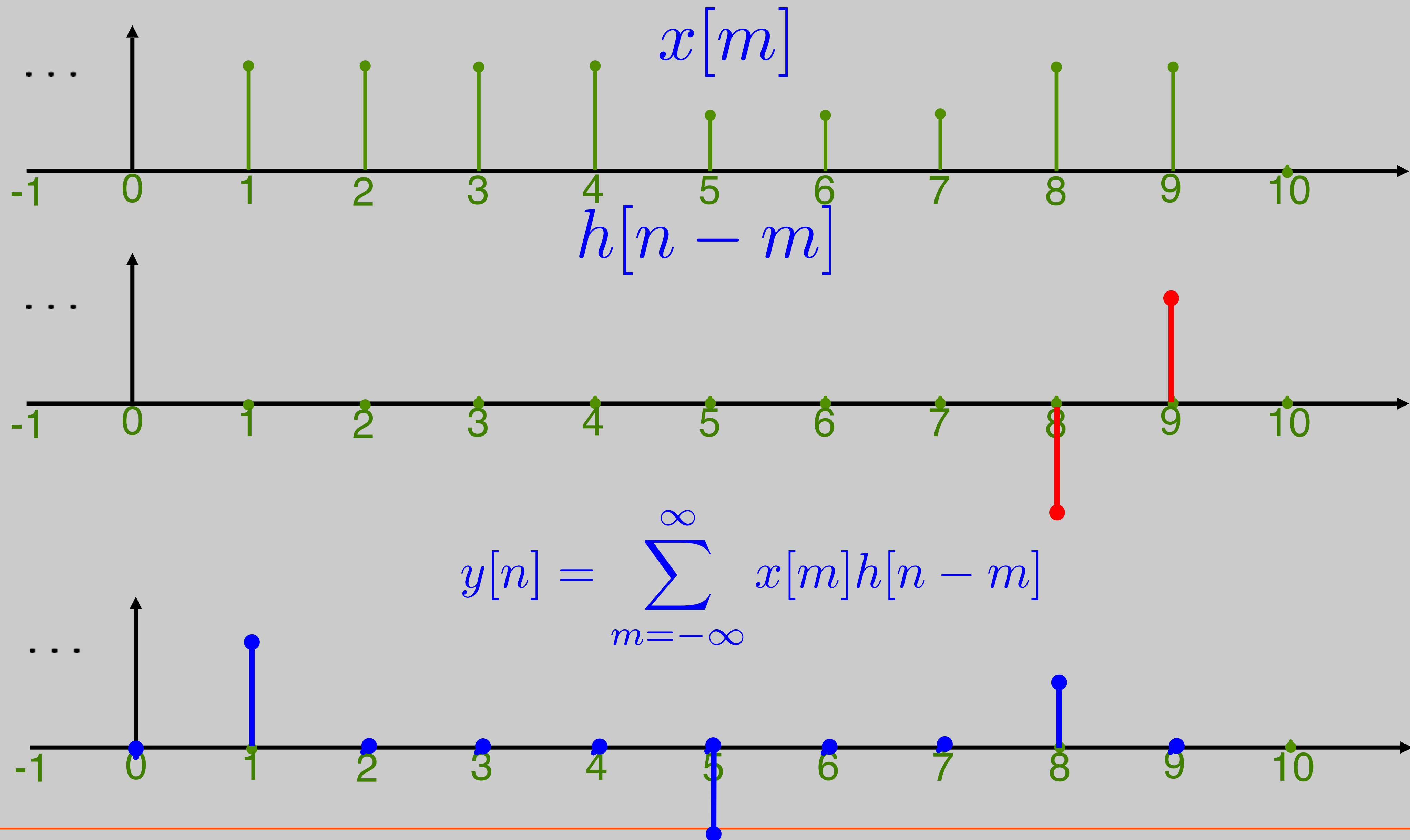
Graphical Example of Convolution



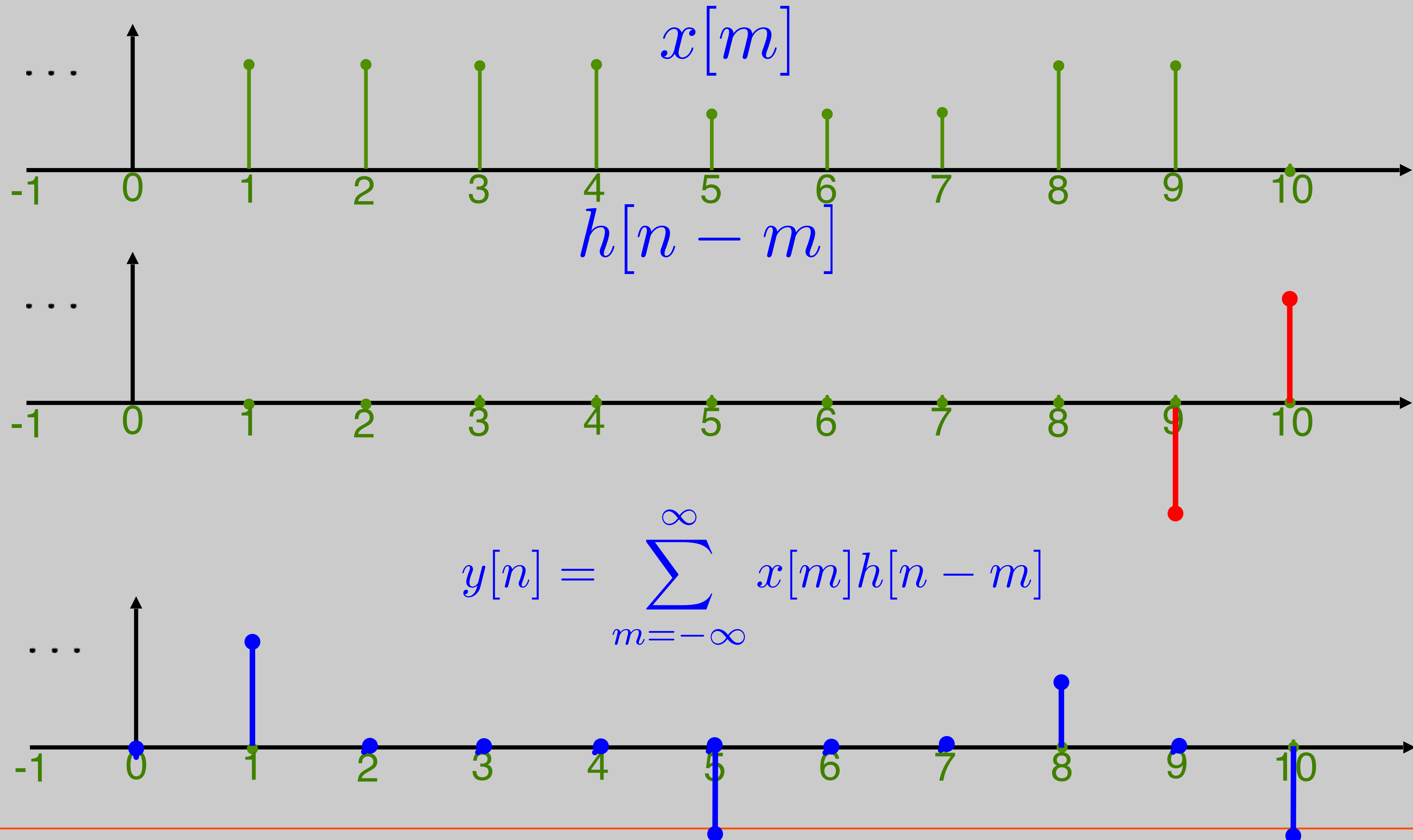
Graphical Example of Convolution



Graphical Example of Convolution



Graphical Example of Convolution



Example

Awkward



Friendly

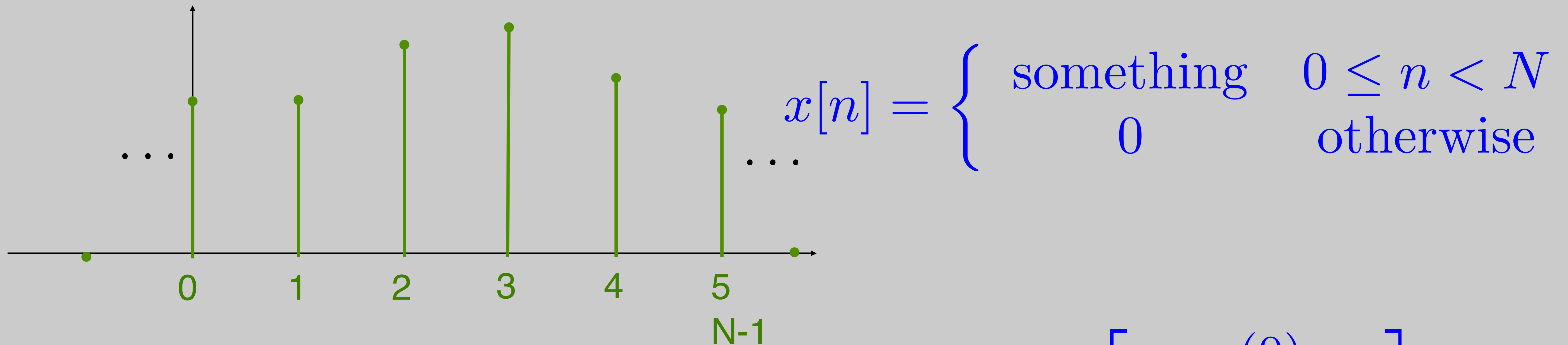


Mean



Finite Sequences

- Consider a finite sequence of length N



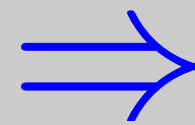
- Can also be written as a vector

$$\vec{x} = \begin{bmatrix} x(0) \\ x(1) \\ \vdots \\ x(N-1) \end{bmatrix}$$

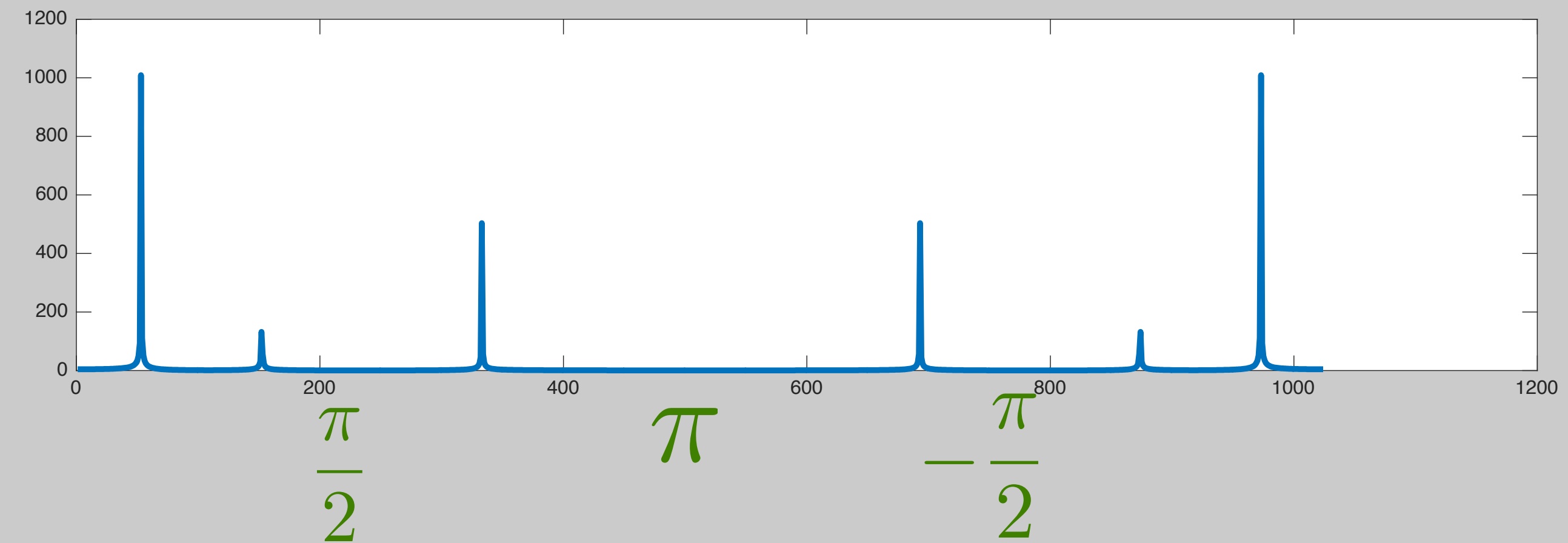
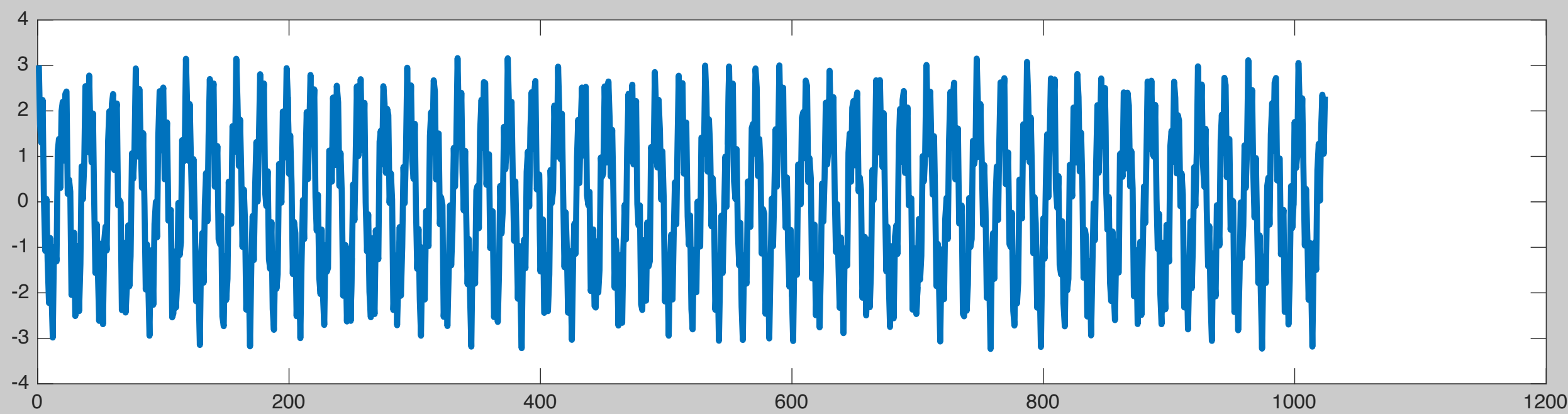
Why?

- To compute this:

$x[n]$



$X[k]$



Finite Sequences as Vectors

- Define an inner-product (for \mathbb{R}^N):

$$\begin{aligned}\langle \vec{x}, \vec{y} \rangle &= \vec{x} \cdot \vec{y} = \sum_{n=0}^{N-1} x[n]y[n] = \\ &= \vec{x}^T \vec{y}\end{aligned}$$

So,

$$\begin{aligned}\langle \vec{x}, \vec{x} \rangle &= \sum_{n=0}^{N-1} x[n]x[n] = \sum_{n=0}^{N-1} x^2[n] = ||\vec{x}||^2 \\ \Rightarrow \vec{x}^T \vec{x} &= ||\vec{x}||^2\end{aligned}$$

Finite Sequences as Vectors

- What about complex?

$$x \cdot x = x^2 = (x_r + jx_i)(x_r + jx_i) = x_r^2 - x_i^2 + 2jx_rx_i \neq ||x||^2$$

but,

$$x^* \cdot x = (x_r - jx_i)(x_r + jx_i) = x_r^2 + x_i^2 = ||x||^2$$

- Transpose vs Transpost conjugate

$$\vec{x} = \begin{bmatrix} 1 \\ j \\ 1+j \end{bmatrix}$$

$$\vec{x}^T = \begin{bmatrix} 1 & j & 1+j \end{bmatrix}$$

$$\vec{x}^* = \begin{bmatrix} 1 & -j & 1-j \end{bmatrix}$$

Finite Sequences as Vectors

- Define Complex inner product

$$\langle \vec{x}, \vec{y} \rangle = \overline{\vec{x}} \cdot \vec{y} = \sum_{n=0}^{N-1} x[n]^* y[n] = \vec{x}^* \vec{y} = \vec{x}^H \vec{y}$$

$$\vec{x} = \begin{bmatrix} 1 \\ j \end{bmatrix} \Rightarrow \vec{x}^* x = \begin{bmatrix} 1 & -j \end{bmatrix} \begin{bmatrix} 1 \\ j \end{bmatrix} = 2$$

Projections

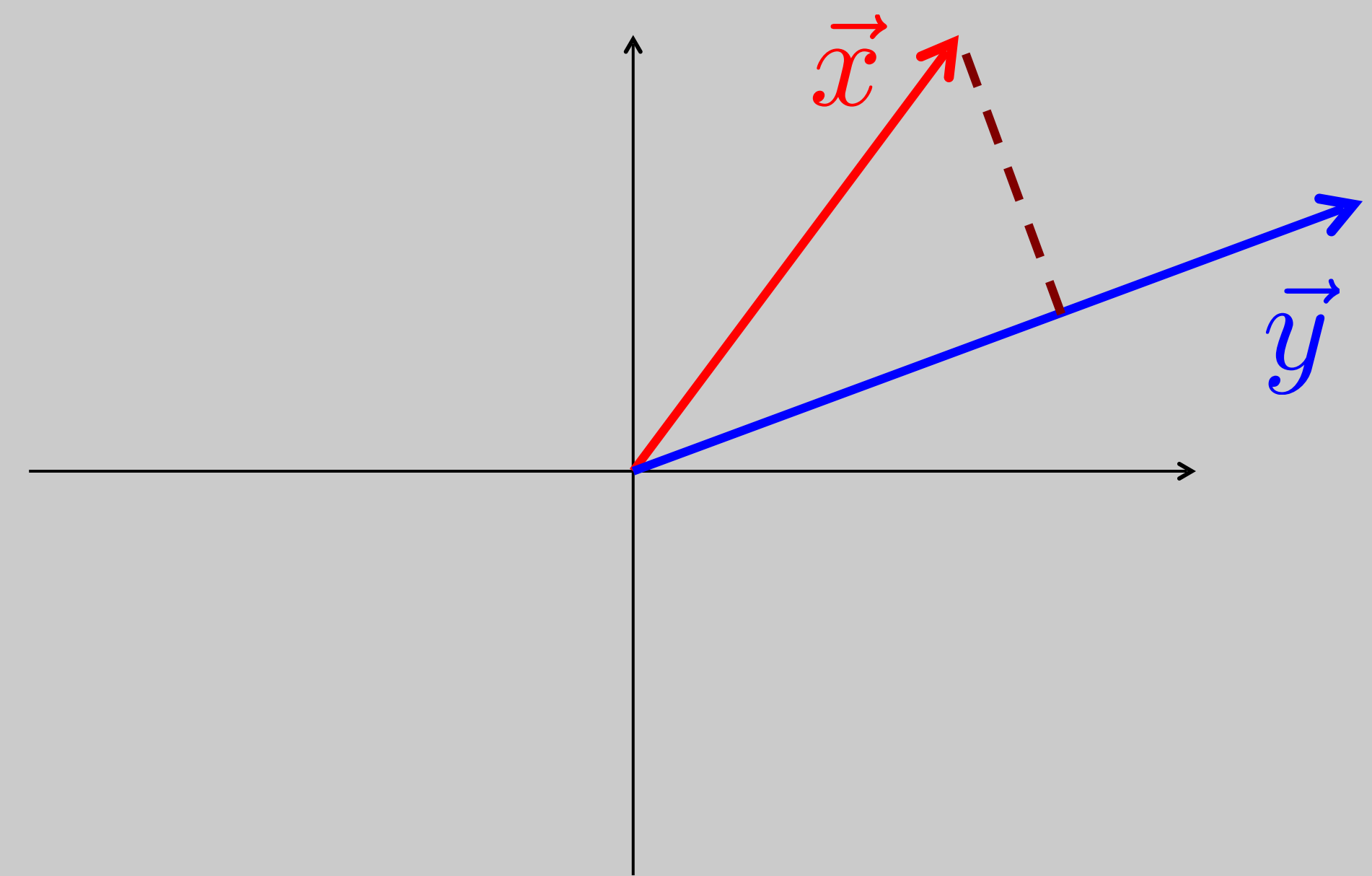
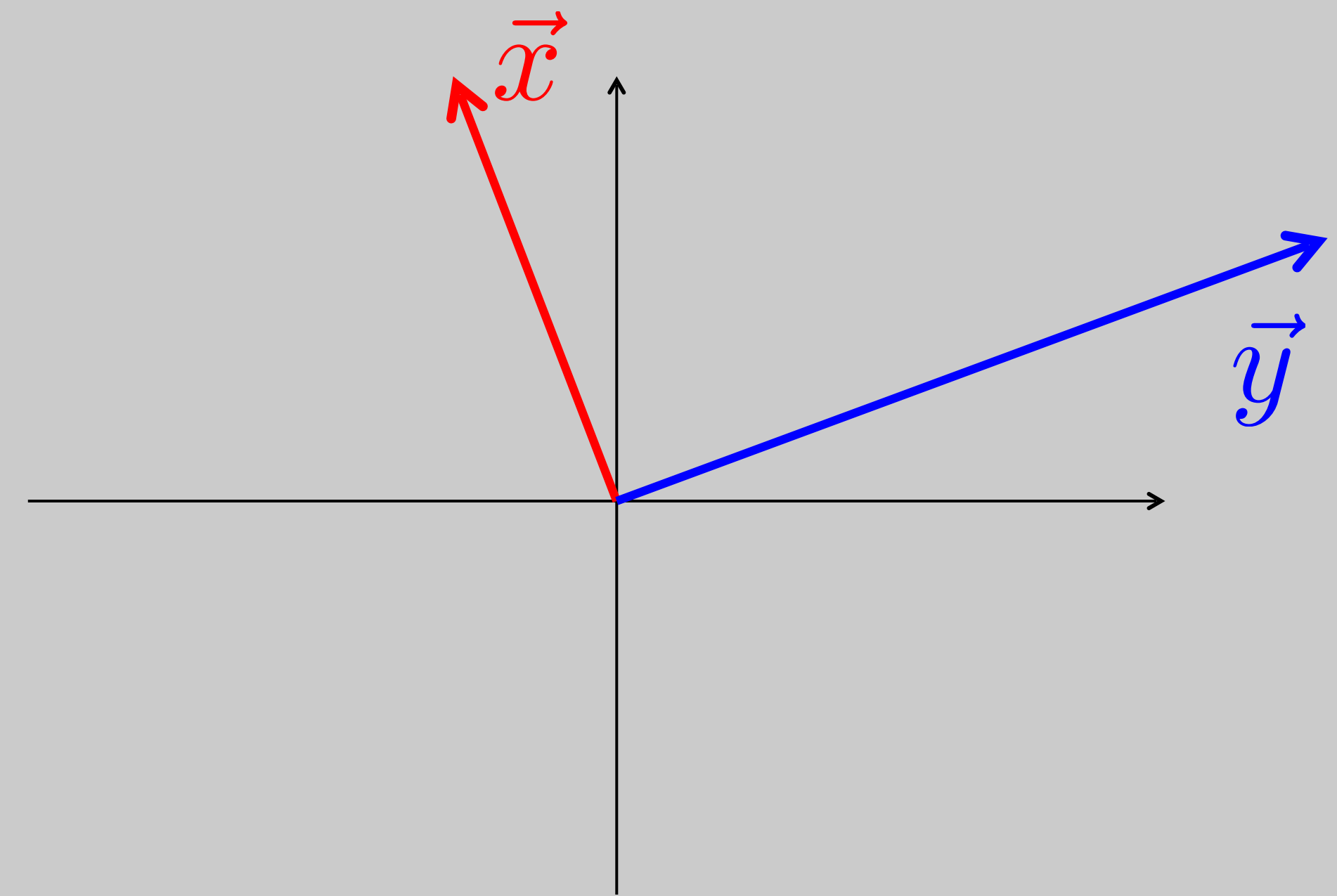
- Orthogonality:

$$\vec{x}^* \vec{y} = \sum_{n=0}^{N-1} x[n]^* y[n] = 0$$

- Unit vector: $||\hat{x}|| = 1$

$$\hat{x} = \frac{\vec{x}}{||\vec{x}||}$$

- Define projection as: $\frac{\vec{y}^* x}{||\vec{y}||}$

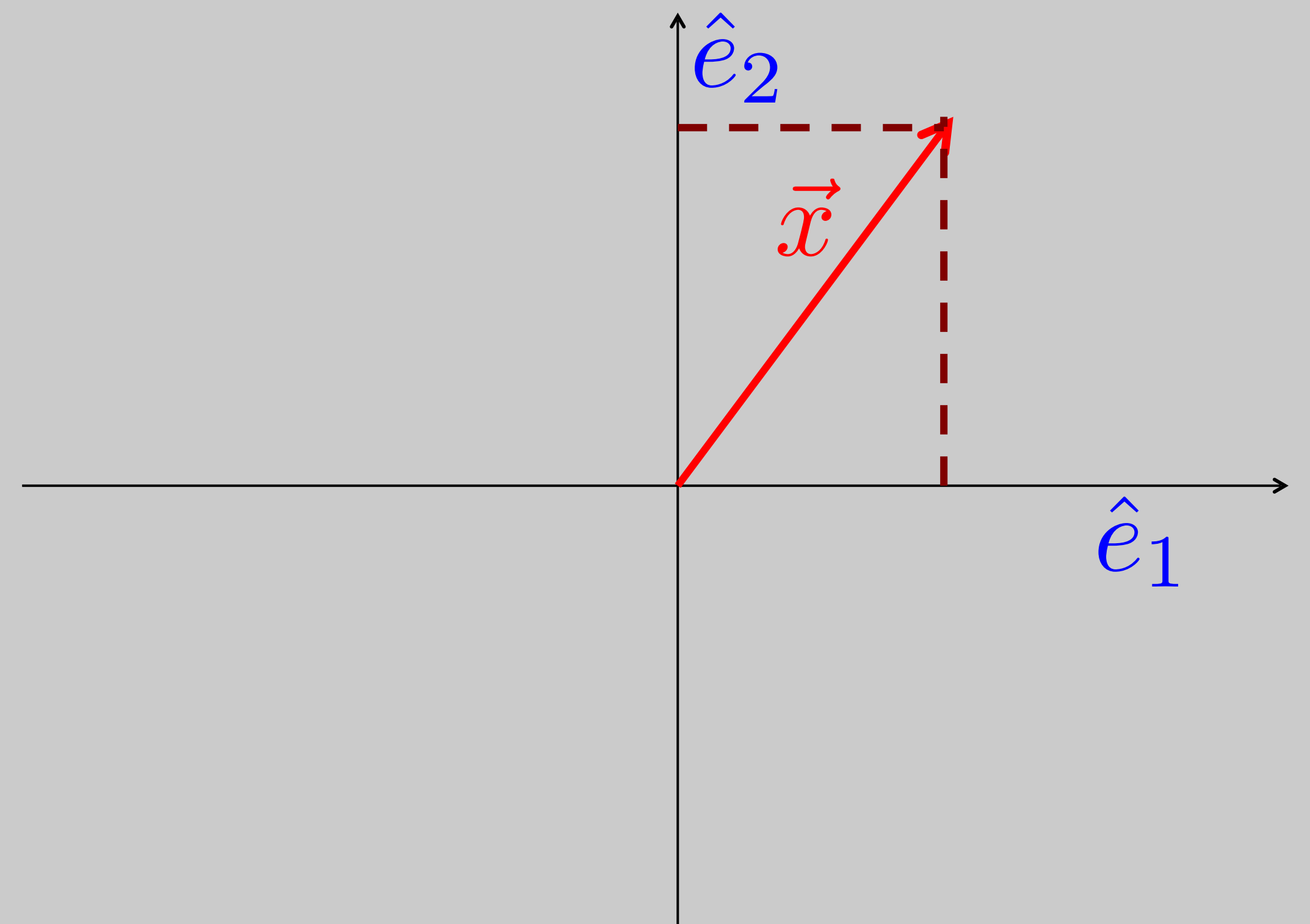


Change of Coordinates (Basis)

- We can compute new coordinates by projections onto orthonormal basis vectors

$$\hat{e}_1^* \vec{x} = \begin{bmatrix} 1 & 0 \end{bmatrix} \vec{x} = x_1$$

$$\hat{e}_2^* \vec{x} = \begin{bmatrix} 0 & 1 \end{bmatrix} \vec{x} = x_2$$



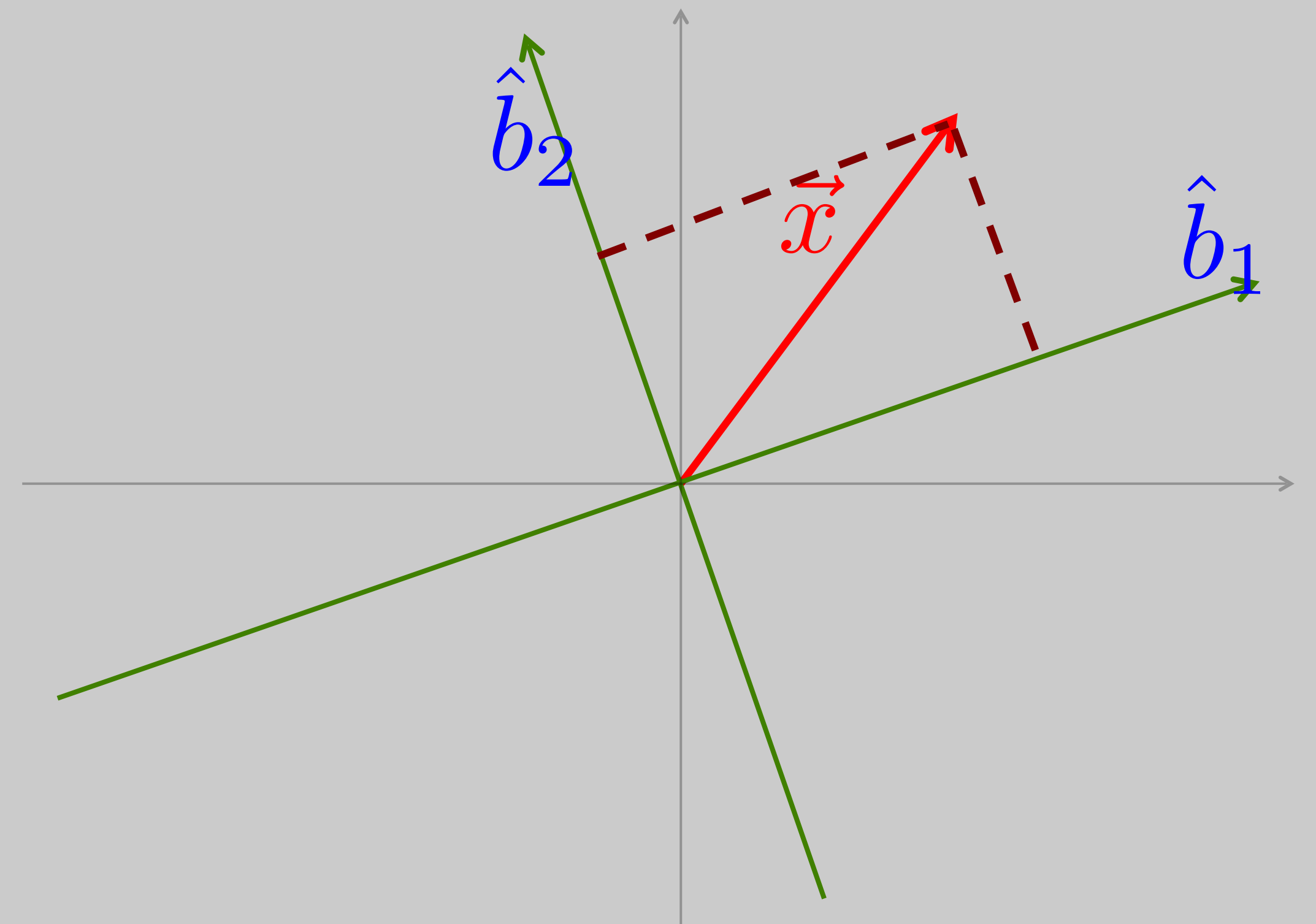
Change of Coordinates (Basis)

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New coordinates:

$$\begin{bmatrix} \hat{b}_1^* \vec{x} \\ \hat{b}_2^* \vec{x} \end{bmatrix} = \begin{bmatrix} \hat{b}_1 & \hat{b}_2 \end{bmatrix}^* \vec{x}$$

$$\Rightarrow \vec{x} = (\hat{b}_1^* \vec{x}) \hat{b}_1 + (\hat{b}_2^* \vec{x}) \hat{b}_2$$



Change of basis

