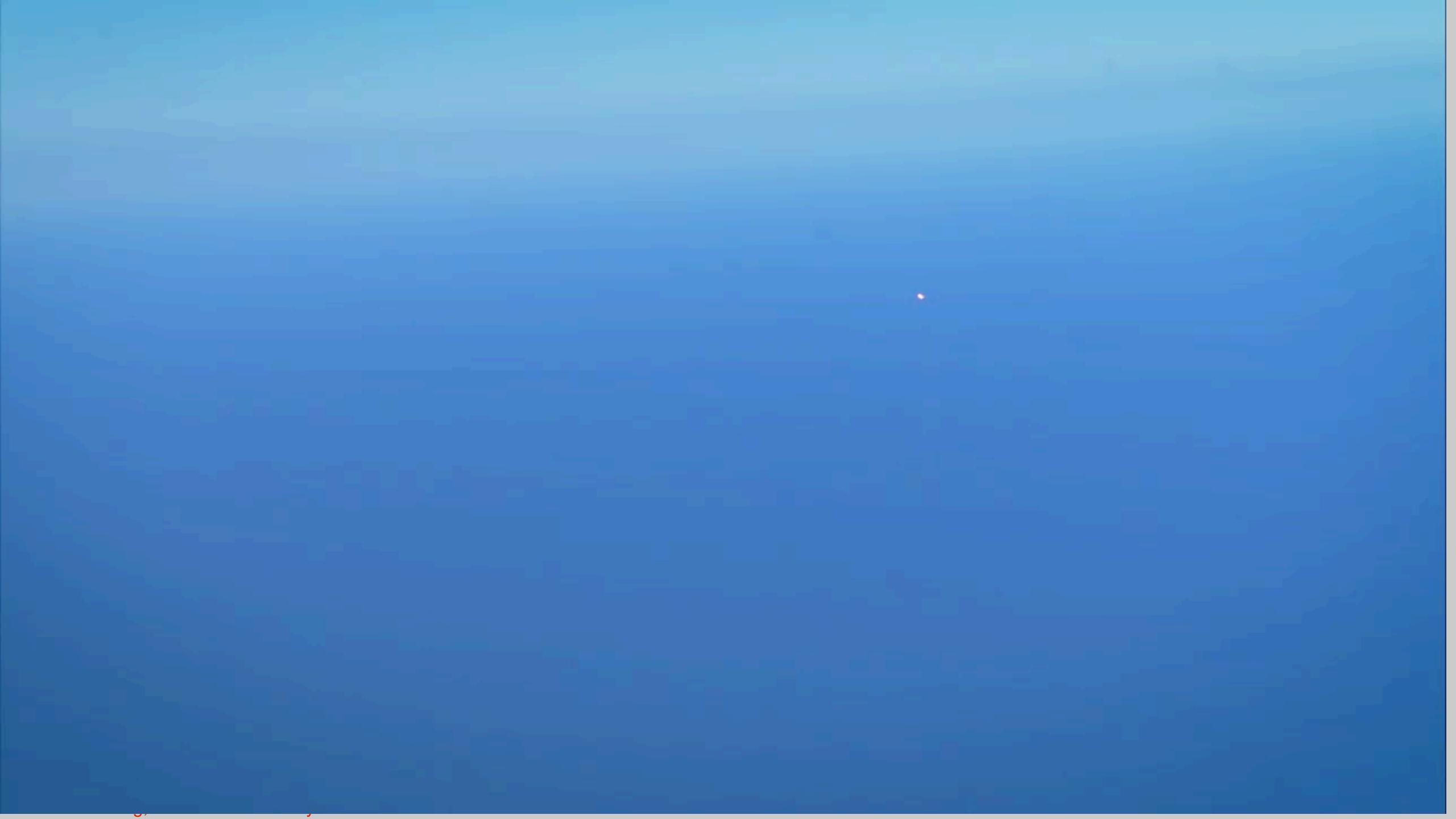
EE16B Designing Information Devices and Systems II

Lecture 6B
Cont. stability of Linear State Models
Controllability

Today

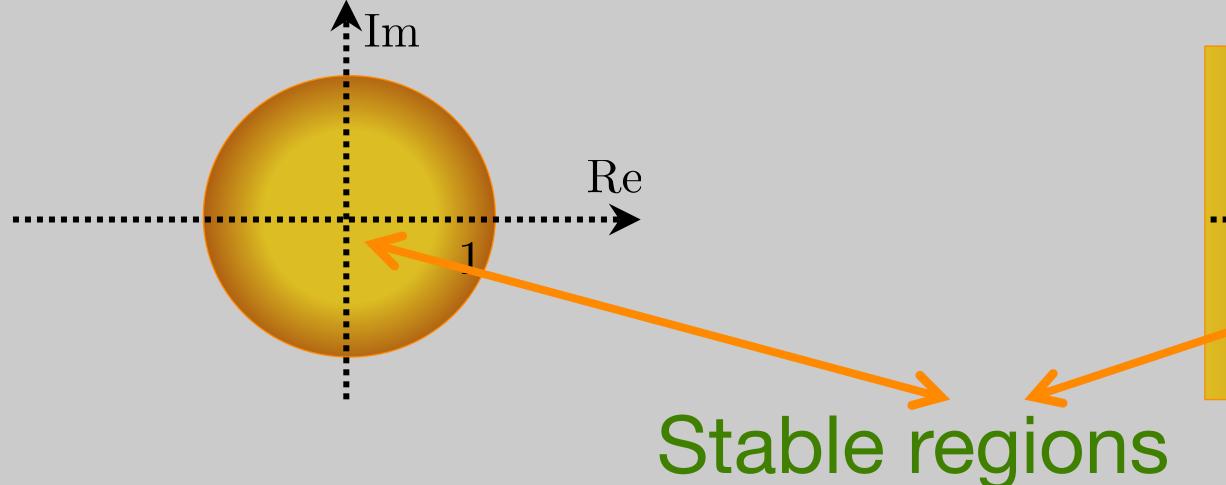
- Last time:
 - Derived stability conditions for disc. and cont.
 systems
 - Easy to analyze using eigenvalues
- Today:
 - -Eigenvalues can predict system behaviour
 - Envelope (decay) and Oscillation (frequency)
 - Controllability of systems



Stability -- Summary

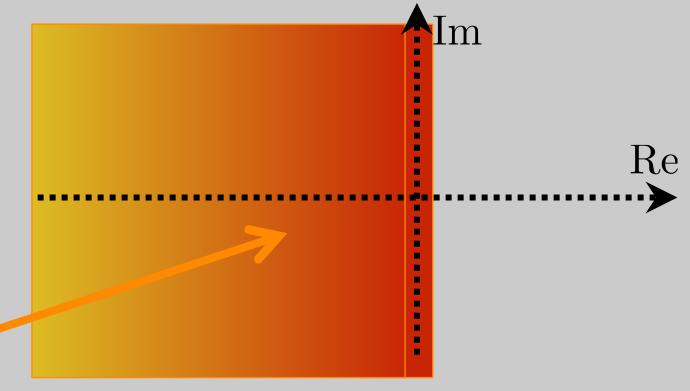
Discrete-Time

$$|\lambda_i(A)| < 1$$



Continuous-Time

$$\text{Real}\{\lambda_i(A)\}<0$$



Stay away from boundaries! System uncertainty can Move you over to unstable region

$$A_{
m down} = \left[egin{array}{cc} 0 & 1 \ -rac{g}{l} & -rac{k}{m} \end{array}
ight] \ A_{
m up} = \left[egin{array}{cc} 0 & 1 \ rac{g}{l} & -rac{k}{m} \end{array}
ight]$$

$$|\lambda I - A_{\text{down}}| = \begin{bmatrix} \lambda & -1 \\ \frac{g}{l} & \lambda + \frac{k}{m} \end{bmatrix} = \lambda^2 + \frac{k}{m}\lambda + \frac{g}{l} = 0$$

$$\lambda_{1,2} = -\frac{k}{2m} \pm \frac{1}{2} \sqrt{\frac{k^2}{m^2} - 4\frac{g}{l}}$$

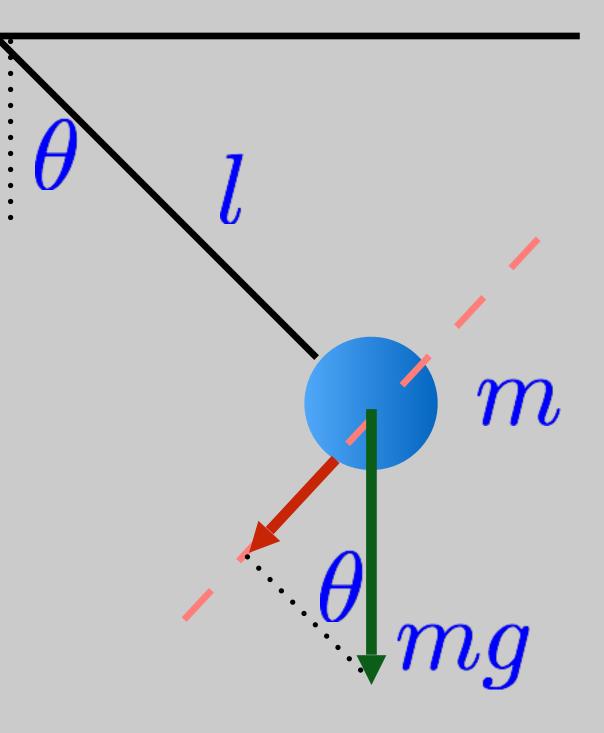
$$\lambda_{1,2} = -\frac{k}{2m} \pm \frac{1}{2} \sqrt{\frac{k^2}{m^2} - 4\frac{g}{l}}$$

If
$$\frac{k^2}{m^2} \ge 4\frac{g}{l}$$
, i.e, sqrt is real, then $\frac{k}{2m} \ge \frac{1}{2}\sqrt{\frac{k^2}{m^2} - 4\frac{g}{l}}$

So, $\lambda_{1,2}$ always negative -- stable!

If
$$\frac{k^2}{m^2} < 4\frac{g}{l}$$
 , i.e, sqrt is imaginary, then $\text{Re}\{\lambda_{1,2}\} = -\frac{k}{2m}$

So, $Re\{\lambda_{1,2}\}$ always negative -- stable!



$$A_{\text{down}} = \begin{bmatrix} 0 & 1 \\ -\frac{g}{l} & -\frac{k}{m} \end{bmatrix}$$
$$A_{\text{up}} = \begin{bmatrix} 0 & 1 \\ \frac{g}{l} & -\frac{k}{m} \end{bmatrix}$$

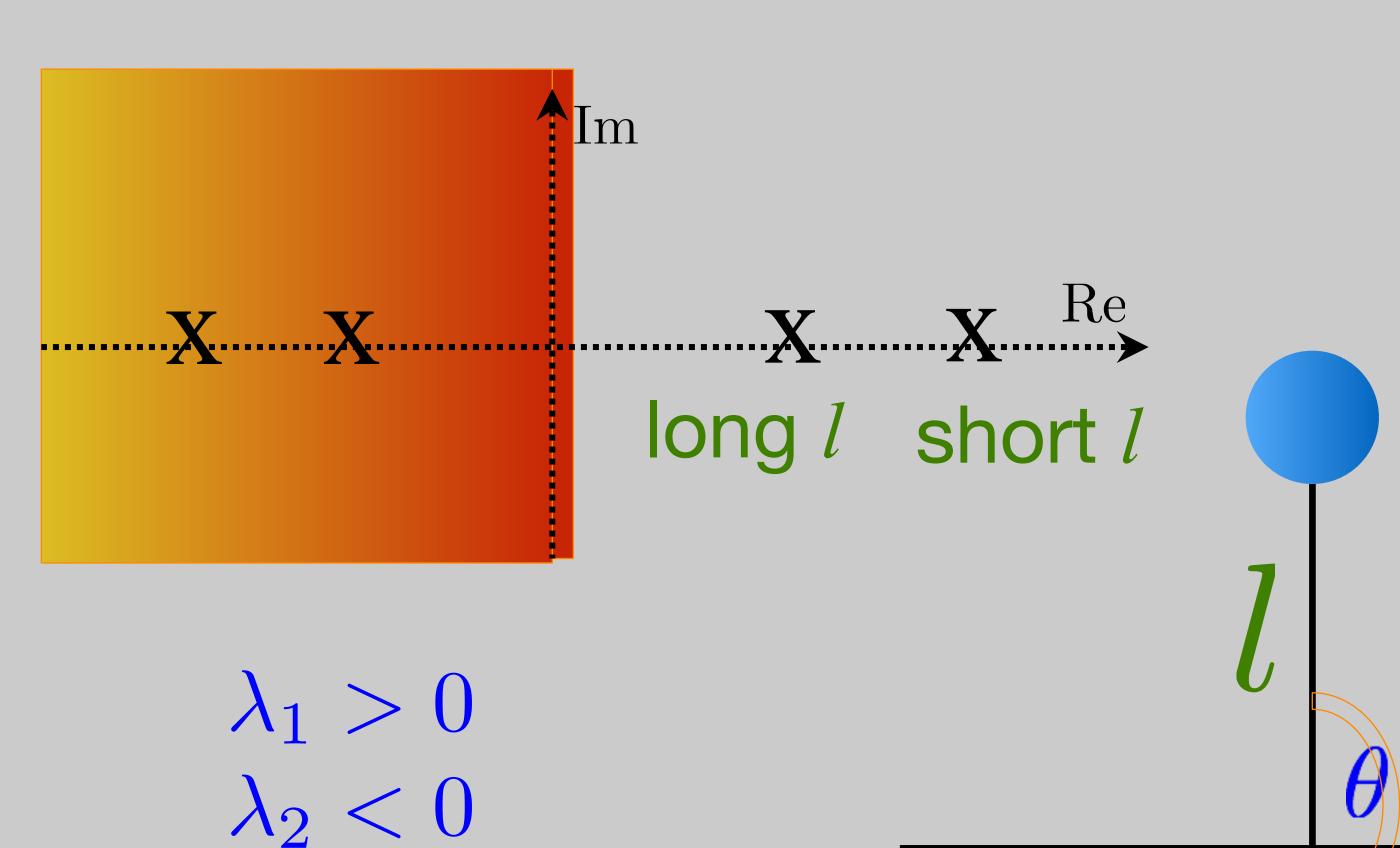
$$|\lambda I - A_{\rm up}| = \begin{bmatrix} \lambda & -1 \\ -\frac{g}{l} & \lambda + \frac{k}{m} \end{bmatrix} = \lambda^2 + \frac{k}{m}\lambda - \frac{g}{l} = 0$$

$$\lambda_{1,2} = -\frac{k}{2m} \pm \frac{1}{2} \sqrt{\frac{k^2}{m^2} + 4\frac{g}{l}}$$

$$\lambda_1 > 0$$

$$\lambda_2 < 0$$

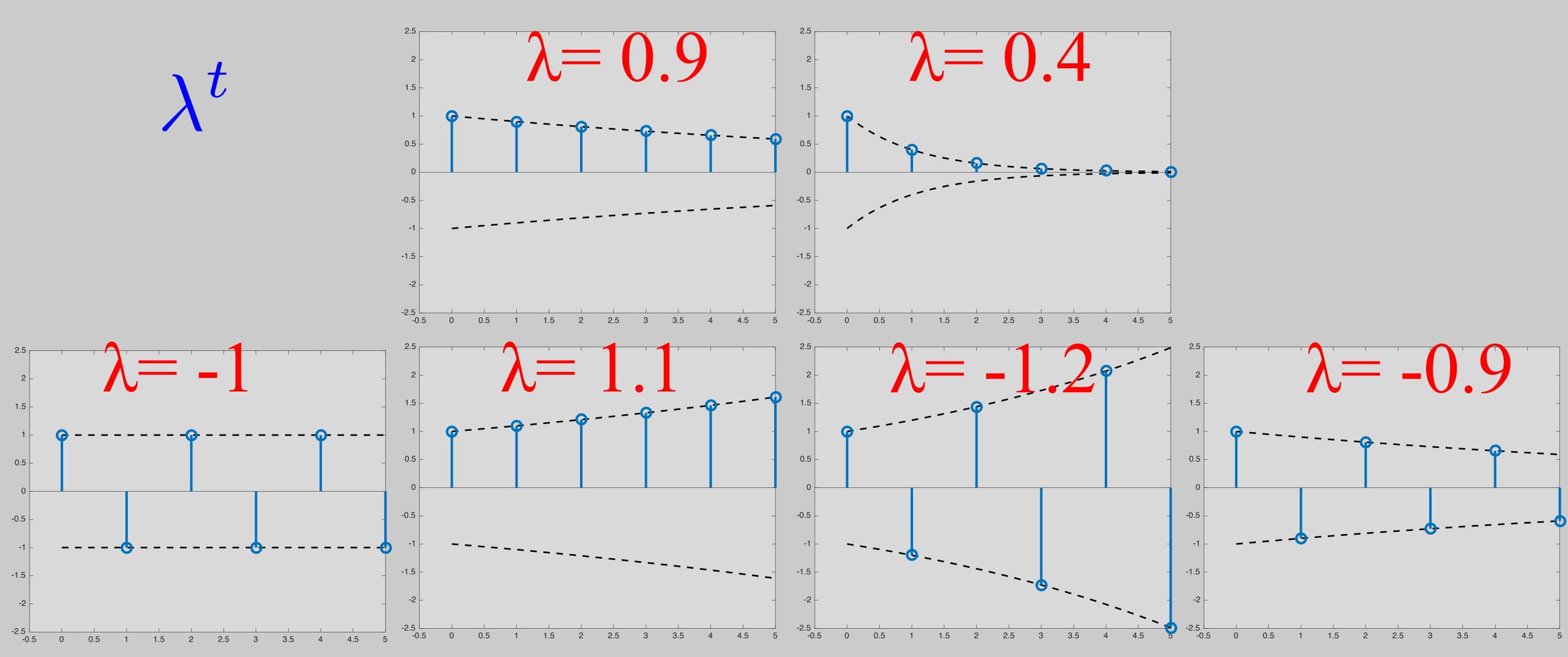
$$\lambda_{1,2} = -\frac{k}{2m} \pm \frac{1}{2} \sqrt{\frac{k^2}{m^2} + 4\frac{g}{l}}$$



Predicting System Behavior

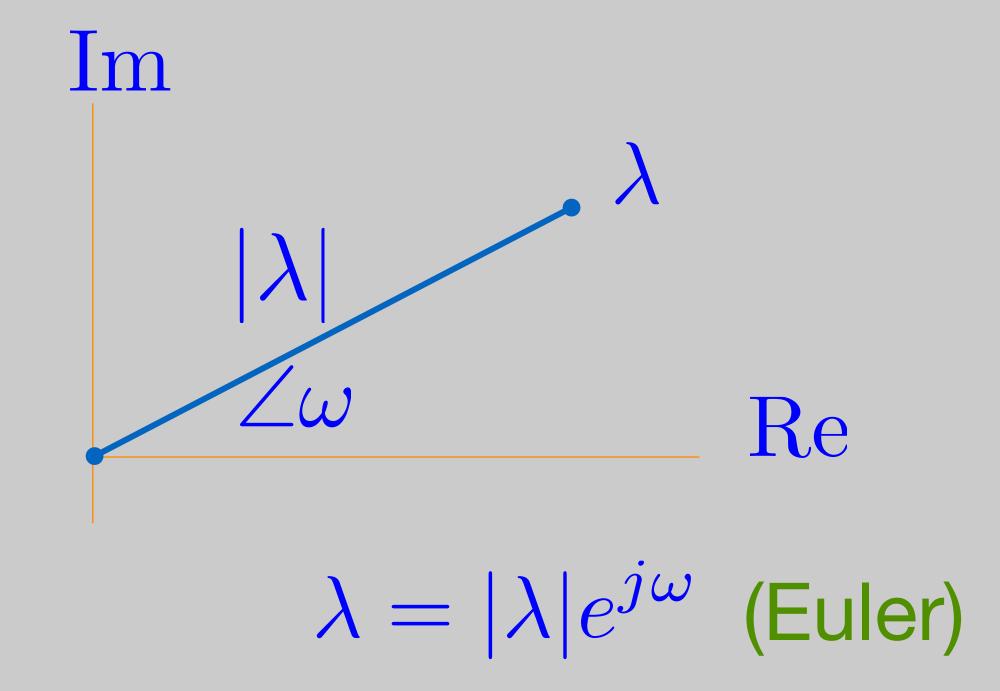
$z(t+1) = \lambda_i z(t)$ Soln: $\lambda_i^t z(0)$

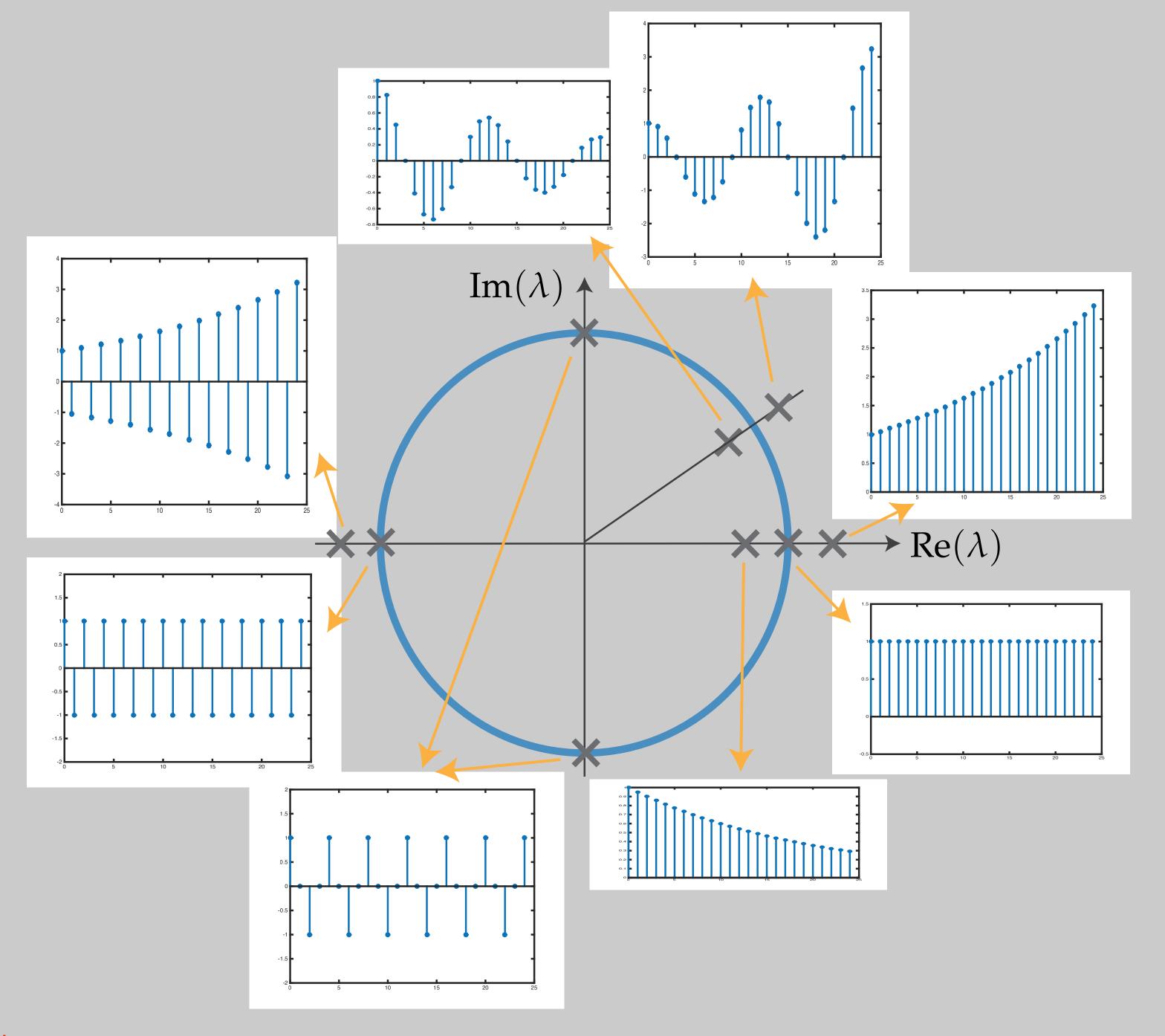
Discrete Time



• If λ is complex

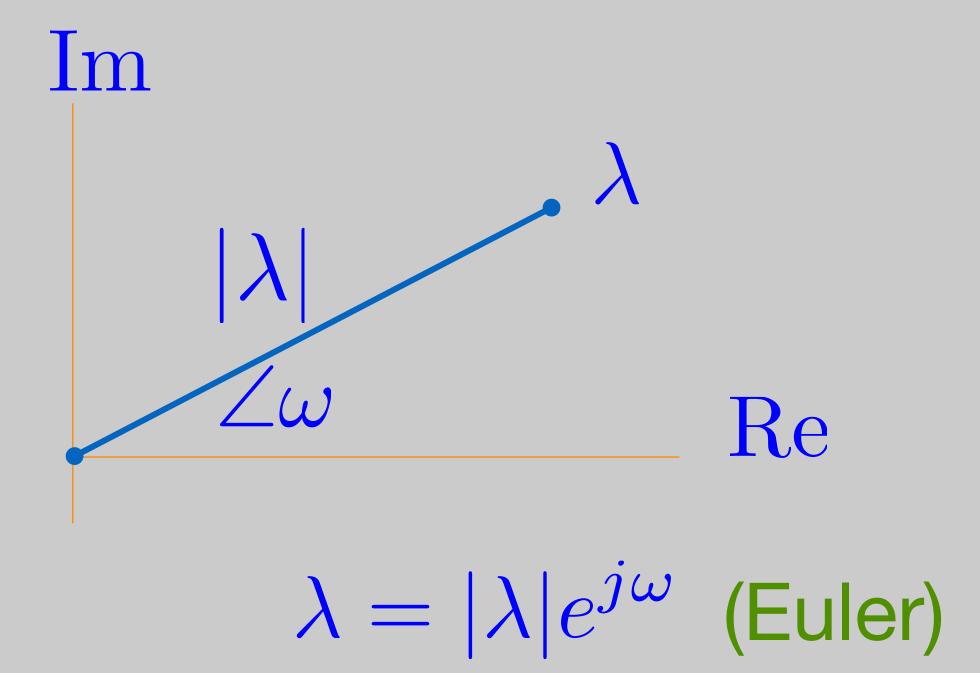
$$\lambda^{t} = (|\lambda|e^{j\omega})^{t}$$
$$= |\lambda|^{t}e^{j\omega t}$$





• If λ is complex

$$\lambda^{t} = (|\lambda|e^{j\omega})^{t}$$
$$= |\lambda|^{t}e^{j\omega t}$$

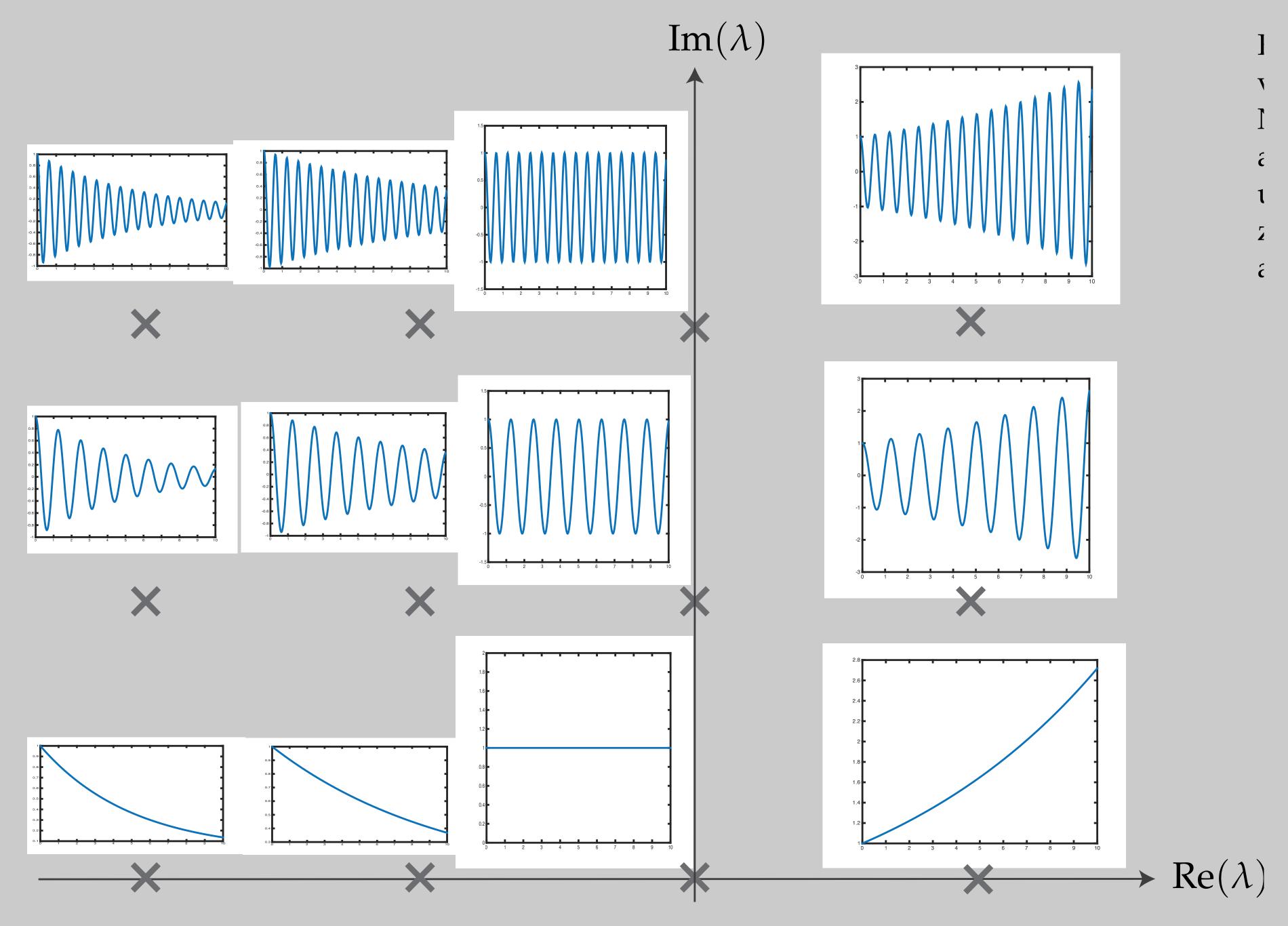


Continuous time:

$$\frac{d}{dt}Z_i(t) = \lambda_i Z_i(t) \Rightarrow e^{\lambda_i t} Z_i(0)$$

Q) What does $e^{\lambda t}$ look like for different choices of λ ?

A)
$$\lambda = v + j\omega$$
 $\Rightarrow e^{\lambda t} = e^{vt}e^{j\omega t}$



Example

Big Picture

- State space modeling is powerful
- Linear state-space are awesome
 - We can say a lot about them!
 - We can approximate non-linear as linear at equilibrium points
- What can we say about linear systems?
 - We can tell if they are stable we have a test!
 - We can can predict system behaviour for initial conditions!
- What about controls?
 - Can test if the system can be controlled to reach all states (controllability)
 - We can control a system to move to a certain state (open loop)
 - We can control a system to stay around a state (feedback)

Discrete-time:

$$\vec{x}(t+1) = A\vec{x}(t) + Bu(t)$$

From last time:

$$\vec{x}(t) = A^t \vec{x}(0) + \sum_{k=0}^{t-1} A^{t-1-k} Bu(k)$$

$$= A^{t}\vec{x}(0) + A^{t-1}Bu(0) + A^{t-2}Bu(1) + \dots + Bu(t-1)$$

$$\vec{x}(t) - A^t \vec{x}(0) = \begin{bmatrix} & & & \\$$

$$\vec{x}(t) - A^t \vec{x}(0) = A^{t-1} B u(0) + A^{t-2} B u(1) + \dots + B u(t-1)$$

$$\vec{x}(t) - A^t \vec{x}(0) = \begin{bmatrix} u(0) \\ u(1) \\ \vdots \\ \vdots \\ u(t-1) \end{bmatrix}$$

$$\vec{x}(t) - A^{t}\vec{x}(0) = A^{t-1}Bu(0) + A^{t-2}Bu(1) + \dots + Bu(t-1)$$

$$\vec{x}(t) - A^{t}\vec{x}(0) = \begin{bmatrix} u(0) \\ A^{t-1}B & A^{t-2}B & \dots & AB & B \end{bmatrix} \begin{bmatrix} u(0) \\ u(1) \\ \vdots \\ u(t-1) \end{bmatrix}$$

- Q) Given any x(0), can we find u(t) s.t. $x(t) = x_{target}$ for some t?
- A) Depends if it is in the span of R_t

$$R_t = \begin{bmatrix} A^{t-1}B & A^{t-2}B & \cdots & AB & B \end{bmatrix}$$

- Q) For t < n? A) No
- Q) At t=n, If columns are independent? A) Absolutely!
- Q) If not independent, does increasing thelps? A) No!

Cayley-Hamilton Theorem: If A is n x n, then Aⁿ can be written as a linear combination of Aⁿ⁻¹,...A, 1

$$A^{n} = \alpha_{n-1}A^{n-1} + \dots + \alpha_{1}A + \alpha_{0}1$$

So does:
$$A^{n}B = \alpha_{n-1}A^{n-1}B + \cdots + \alpha_{1}AB + \alpha_{0}B$$

$$R_n = \begin{bmatrix} A^{n-1}B & A^{n-2}B & \cdots & AB & B \end{bmatrix}$$

What about R_{n+1} ?

Controllability Test

If R_t doesn't have n independent columns at t=n, it never will for t > n either!

Therefore, we need only to examine R_n for

controllability:
$$R_n = \left[\begin{array}{ccccc} A^{n-1}B & A^{n-2}B & \cdots & AB & B \end{array} \right]$$

Conclusion: $\vec{x}(t+1) = A\vec{x}(t) + Bu(t)$

is controllable if and only if

$$\operatorname{rank}\{R_n\} = n$$

Example 1:

$$\vec{x}(t+1) = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} \vec{x}(t) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(t)$$

$$R_2 = \begin{bmatrix} AB & B \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & 0 \end{bmatrix}$$

$$R_3 = \begin{bmatrix} A^2B & AB & B \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 \\ 0 & 0 & 0 \end{bmatrix}$$

Rank
$$\{R2\} = 1$$
, $< n=2 \Rightarrow Not controllable!$

$$x_1(t+1) = x_1(t) + x_2(t) + u(t)$$

$$x_2(t+1) = 2x_2(t)$$
 (not stable)

Can not control x₂, not with u and not with x₁

Example 2:

$$p(t+1) = p(t) + Tv(t) + \frac{1}{2}T^{2}u(t)$$

$$v(t+1) = v(t) + Tu(t)$$

$$\begin{bmatrix} p(t+1) \\ v(t+1) \end{bmatrix} = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix} \begin{bmatrix} p(t) \\ v(t) \end{bmatrix} + \begin{bmatrix} \frac{1}{2}T^{2} \\ T \end{bmatrix} u(t)$$

$$A$$

$$\begin{bmatrix} \frac{3}{2}T^{2} & \frac{1}{2}T^{2} \end{bmatrix}$$

$$R_2 = \begin{bmatrix} AB & B \end{bmatrix} = \begin{bmatrix} \frac{3}{2}T^2 & \frac{1}{2}T^2 \\ T & T \end{bmatrix}$$

Rank = $2 \Rightarrow$ Controllable!

Continuous Time (no derivation here)

The continuous-time system

$$\frac{d}{dt}\vec{x}(t) = A\vec{x}(t) + Bu(t)$$

is controllable if and only if

$$R_n = \left[\begin{array}{cccccc} A^{n-1}B & A^{n-2}B & \cdots & AB & B \end{array} \right]$$

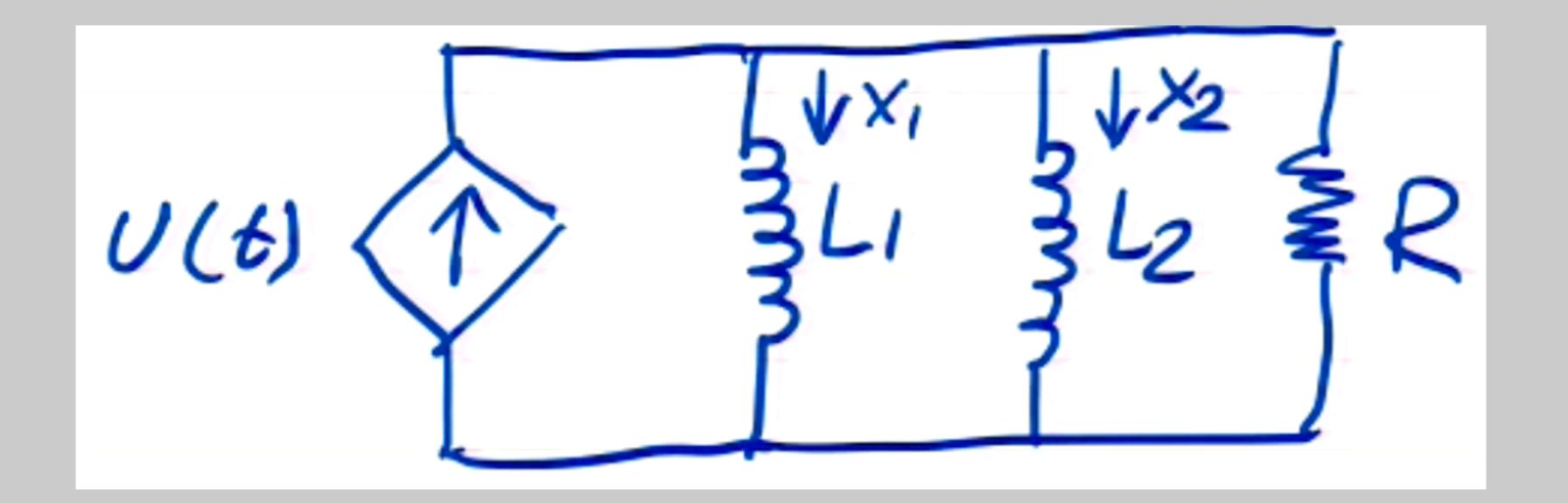
has rank = n

Example 3 + Quiz

Write the state model:

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} u(t) \\ x_2(t) \end{bmatrix}$$

For:

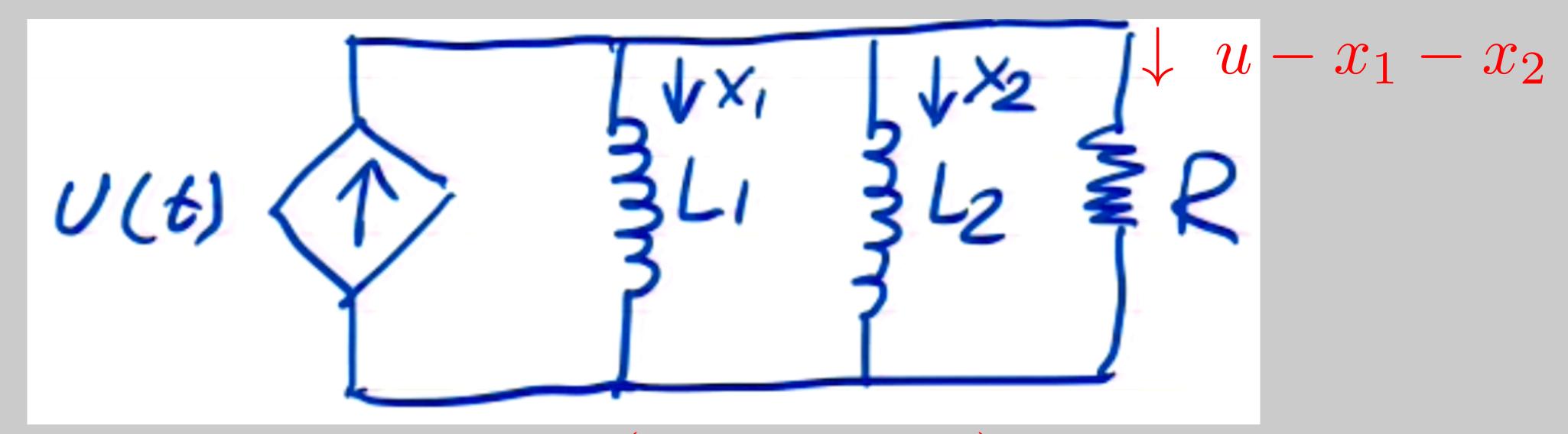


Quiz

Write the state model:

$$\frac{d}{dt} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} -\frac{R}{L_1} & -\frac{R}{L_2} \\ -\frac{R}{L_2} & -\frac{R}{L_2} \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix} u(t)$$

For:



$$V_r = R(u - x_1 - x_2) = L_1 \dot{x}_1 = L_2 \dot{x}_2$$

Example 3

Controllability:

$$B = \begin{bmatrix} \frac{R}{L_1} \\ \frac{R}{L_2} \end{bmatrix}$$

$$AB = \begin{bmatrix} -\frac{R}{L_1} \left(\frac{R}{L_1} + \frac{R}{L_2} \right) \\ -\frac{R}{L_2} \left(\frac{R}{L_1} + \frac{R}{L_2} \right) \end{bmatrix}$$

$$R = [AB \ B]$$

$$AB = \left(\frac{R}{L_1} + \frac{R}{L_2}\right)B$$

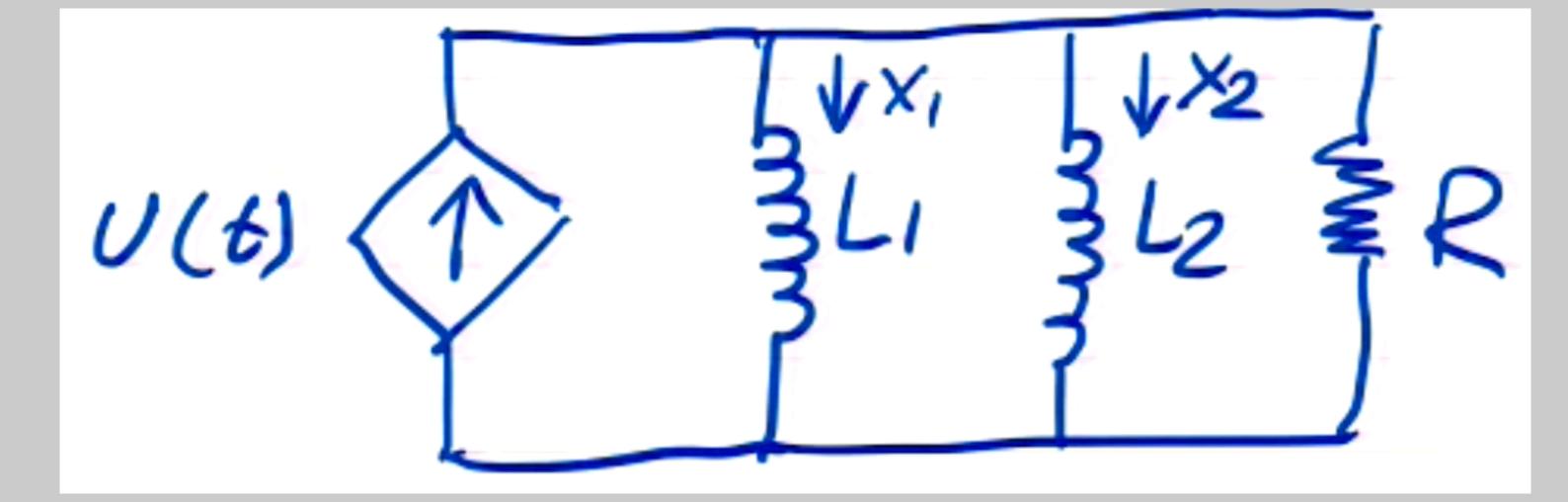
$$Rank = 1$$

Not controllable

Physical explanation

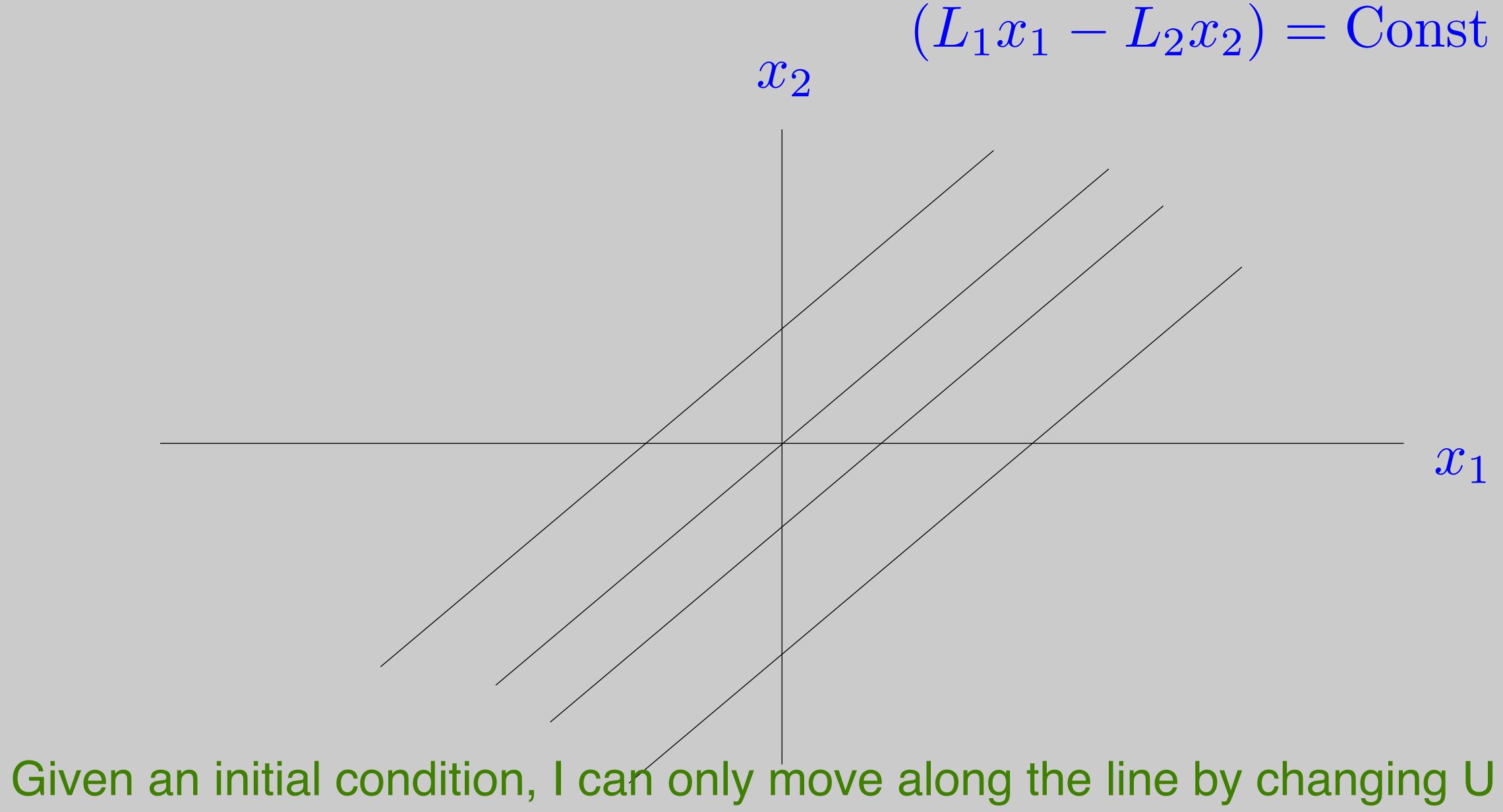
Why can't I drive the currents x1 and x2 freely

using U(t)?



$$L_{1} \frac{dx_{1}}{dt} = L_{2} \frac{dx_{2}}{dt} = R \cdot i_{R} = V_{R} \quad \Rightarrow L_{1} \frac{dx_{1}}{dt} - L_{2} \frac{dx_{2}}{dt} = 0$$

$$\frac{d}{dt} (L_{1}x_{1} - L_{2}x_{2}) = 0 \qquad \Rightarrow (L_{1}x_{1} - L_{2}x_{2}) = \text{Const}$$



Q) What if A = 0? Can the system be controllable?

$$\frac{d}{dt}\vec{x}(t) = Bu(t)$$

$$R = \begin{bmatrix} 0 & 0 & 0 & 0 & \cdots & B \end{bmatrix}$$

A) Only if u(t) is a vector with the same number of elements as the number of states

Summary

- Described and derived conditions for controllability of linear state models.
 - Rank of R_n for both discrete and continuouse
- Showed how to discretize continuous systems
- Showed examples of controllable and noncontrollable systems

- Next time:
 - Open loop and state feedback control
 - Controllers to make systems do what we want!