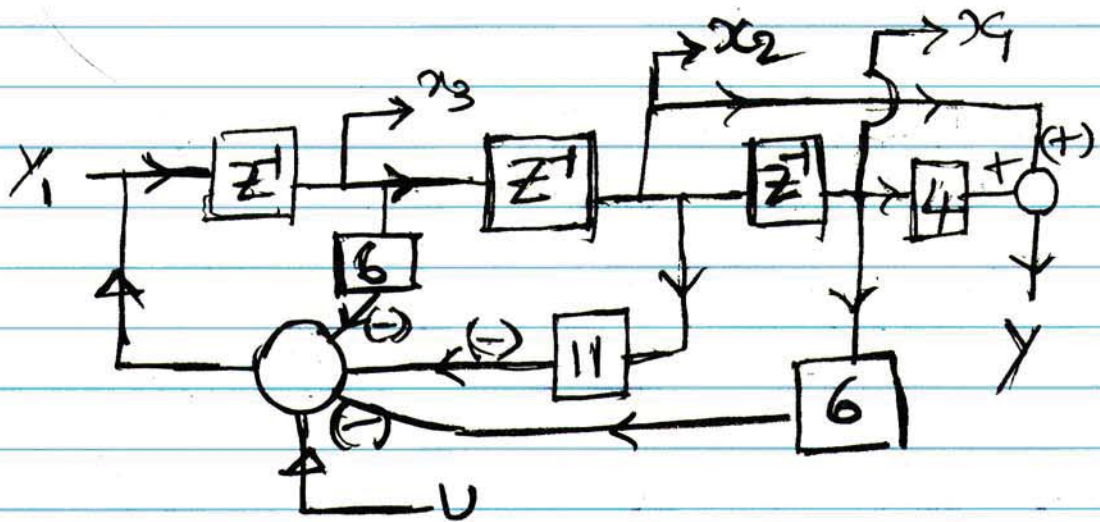


①

① ② $\frac{Y(z)}{U(z)} = \frac{z^{-2} + 4z^{-3}}{1 + 6z^{-1} + 11z^{-2} + 6z^{-3}}$

Call,

$$\frac{Y}{U} = \frac{1}{1 + 6z^{-1} + 11z^{-2} + 6z^{-3}}$$



⑥ Phase Variable Form

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \\ x_3(k+1) \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -11 & -6 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u(k)$$

$$y = \begin{bmatrix} 4 & 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

2. Given: $x[k+1] = Gx[k] + Hu[k] = \begin{bmatrix} 1 & 0.2 \\ 0.5 & 0 \end{bmatrix} x[k] + \begin{bmatrix} 0.4 \\ -0.5 \end{bmatrix} u[k]$

a. Find transfer function $\frac{X[z]}{U[z]}$

$$\begin{aligned} \frac{X[z]}{U[z]} &= (zI - G)^{-1} H = \begin{bmatrix} z-1 & -0.2 \\ -0.5 & z \end{bmatrix}^{-1} \begin{bmatrix} 0.4 \\ -0.5 \end{bmatrix} \\ &= \frac{1}{z(z-1) - (0.5)(0.2)} \begin{bmatrix} z & 0.2 \\ 0.5 & z-1 \end{bmatrix} \begin{bmatrix} 0.4 \\ -0.5 \end{bmatrix} = \begin{bmatrix} \frac{0.4z - (0.5)(0.2)}{z(z-1) - (0.5)(0.2)} \\ \frac{(0.4)(0.5) + (z-1)(-0.5)}{z(z-1) - (0.5)(0.2)} \end{bmatrix} \end{aligned}$$

$$\frac{X[z]}{U[z]} = \begin{bmatrix} \frac{0.4z - 0.1}{z^2 - z - 0.1} \\ \frac{-0.5z + 0.7}{z^2 - z - 0.1} \end{bmatrix}$$

b. Is the system BIBO stable?

Poles of TF: $z^2 - z - 0.1 = 0$

$$z = 1.092, -0.092$$

$z = 1.092$ is outside the unit circle, so the system is

NOT BIBO stable

Prob 3 starts:

$$(3) H(s) = \frac{1}{(s+\alpha)^2}$$

$$\Rightarrow h(t) = \boxed{e^{-\alpha t} t} \quad - (1)$$

$$\Rightarrow h(kT) = e^{-\alpha kT} kT \quad - (2)$$

$$\Rightarrow H(z) = \sum_{k=0}^{\infty} (e^{-\alpha kT} kT) z^{-k}$$

$$\Rightarrow H(z) = \sum_{k=0}^{\infty} (kT) (ze^{\alpha T})^{-k} \quad - (1)$$

$$\Rightarrow zH(z) - zh(0) = \sum_{k=0}^{\infty} (k+1)T (ze^{\alpha T})^{-k} e^{-\alpha T} \quad L(2)$$

$$\Rightarrow \left(\frac{H(z)}{T} \right) = \sum_{k=0}^{\infty} (k+1) (ze^{\alpha T})^{-k} \quad - (3)$$

$$\left(\frac{zH(z) - zh(0)}{Te^{-\alpha T}} \right) = \sum_{k=0}^{\infty} (k+1) (ze^{\alpha T})^{-k} \quad L(4)$$

(4) - (3) :

Prob 3 continues. :

(4) - (3) :

$$\frac{zH(z) - zh(0) - \frac{H(z)}{T}}{Te^{-\alpha T}} = \sum_{k=0}^{\infty} (ze^{\alpha T})^{-k}$$

$$h(0) = 0 \Rightarrow$$

$$\left[\frac{zH(z)}{e^{-\alpha T}} - H(z) \right] \left(\frac{1}{T} \right) = \frac{1}{1 - (ze^{\alpha T})^{-1}}$$

$$= \frac{z}{z - e^{-\alpha T}}$$

$$\Rightarrow H(z) [z - e^{-\alpha T}] = \frac{Te^{-\alpha T} z}{(z - e^{-\alpha T})}$$

$$H(z) = \frac{Te^{-\alpha T} z}{(z - e^{-\alpha T})^2}$$

4. $F(s) = \frac{k}{s+10}$

a. $e(\infty) = \frac{1}{1+k_p} < 0.1$

$k_p > 9$

$k_p = \lim_{s \rightarrow 0} F(s) = \lim_{s \rightarrow 0} \frac{k}{s+10} = \frac{k}{10} > 9$

$k_p > 90$

$F_{cl}(s) = \frac{k_p}{s+10+k_p}$

closed loop pole:

$s < -100$

b. $\dot{x} = -10x + u$

$y = x$

$u = k_p(r - y) = k_p(r - x)$

since $y=x$



$x[k+1] = e^{AT} x[k] + \left(\int_0^T e^{A\lambda} d\lambda \right) B u[k]$

$G = e^{AT} = e^{-10T}$

$H = \left(\int_0^T e^{A\lambda} d\lambda \right) B = \left(\int_0^T e^{-10\lambda} d\lambda \right) (1) = \left. -\frac{1}{10} e^{-10\lambda} \right|_{\lambda=0}^T = \frac{1}{10} (1 - e^{-10T})$

$x[k+1] = e^{-10T} x[k] + \frac{1}{10} (1 - e^{-10T}) u[k]$

Let $u[k] = k_p(r[k] - x[k])$

$x[k+1] = e^{-10T} x[k] + \frac{1}{10} (1 - e^{-10T}) k_p (r[k] - x[k])$

$= \left(e^{-10T} - \frac{k_p}{10} (1 - e^{-10T}) \right) x[k] + \frac{k_p}{10} (1 - e^{-10T}) r[k]$

$x[k+1] = \left(\left(\frac{k_p}{10} + 1 \right) e^{-10T} - \frac{k_p}{10} \right) x[k] + \frac{k_p}{10} (1 - e^{-10T}) r[k]$

C. For stability $G_k = (G - K_p H)$ must be inside unit circle

$$G = \left(\left(\frac{K_p}{10} + 1 \right) e^{-10T} - \frac{K_p}{10} \right) \Rightarrow \left| \left(\frac{K_p}{10} + 1 \right) e^{-10T} - \frac{K_p}{10} \right| < 1$$

$$\left(\frac{K_p}{10} + 1 \right) e^{-10T} - \frac{K_p}{10} < 1$$

$$\left(\frac{K_p}{10} + 1 \right) e^{-10T} < 1 + \frac{K_p}{10}$$

$$e^{-10T} < 1$$

$$-10T < \ln(1)$$

$$T > \frac{-\ln(1)}{10}$$

$$T > 0$$

$$-1 < \left(\frac{K_p}{10} + 1 \right) e^{-10T} - \frac{K_p}{10}$$

$$\frac{\left(\frac{K_p}{10} - 1 \right)}{\left(\frac{K_p}{10} + 1 \right)} < e^{-10T}$$

$$\ln\left(\frac{K_p - 10}{K_p + 10}\right) < -10T$$

$$\frac{\ln\left(\frac{K_p - 10}{K_p + 10}\right)}{-10} > T$$

For SSE < 0.1:

$x[k+1] = x[k]$ in steady state $r[k] = 1$ since input is a step

$$x[k] = (G - K_p H) x[k] + K_p H$$

$$x[k] = (1 - (G - K_p H))^{-1} K_p H \quad \text{at steady state}$$

$$x[k] = \frac{K_p \left(\frac{1}{10} \right) (1 - e^{-10T})}{\left(1 - \left[\left(\frac{K_p}{10} + 1 \right) e^{-10T} - \frac{K_p}{10} \right] \right)} = \frac{\frac{K_p}{10} (1 - e^{-10T})}{\left(1 - \left(\frac{K_p}{10} + 1 \right) e^{-10T} + \frac{K_p}{10} \right)}$$

$$= \frac{\frac{K_p}{10} (1 - e^{-10T})}{\left(\frac{K_p}{10} + 1 \right) - \left(\frac{K_p}{10} + 1 \right) e^{-10T}} = \frac{\frac{K_p}{10} (1 - e^{-10T})}{\left(\frac{K_p}{10} + 1 \right) (1 - e^{-10T})} = \frac{K_p}{K_p + 10}$$

$$x[k] = \frac{K_p}{K_p + 10}$$

So $x[k]$ at steady state only depends on K_p , not T .

$$\text{For SSE} < 0.1, 0.9 < x[k] < 1.1 \Rightarrow 0.9 < \frac{K_p}{K_p + 10} < 1.1$$

So $90 < K_p$ for this condition to hold

For SSE: $90 < K_p$

For stability: $0 < T < \frac{\ln(\frac{K_p-10}{K_p+10})}{-10}$

Since $\ln(\frac{K_p-10}{K_p+10})$ decreases with increasing K_p , maximal T is at $K_p = 90$

$$\text{So maximal } T = \frac{\ln(\frac{90-10}{90+10})}{-10} = \frac{\ln(0.8)}{-10} = 0.0223$$

$$T_{\max} = 0.0223$$

Prob 5 starts ...

$$(5) \quad G(s) = \frac{k}{s(s+1)}$$

$$(a) \quad G(s) = k \left[\frac{1}{s} - \frac{1}{s+1} \right]$$

$$\Rightarrow g(t) = (k) - k e^{-t} \quad - (1)$$

$$\Rightarrow g(mT) = k - k e^{-mT} \quad - (2)$$

$$G(z) = \sum_{m=0}^{\infty} g(mT) z^{-m}$$

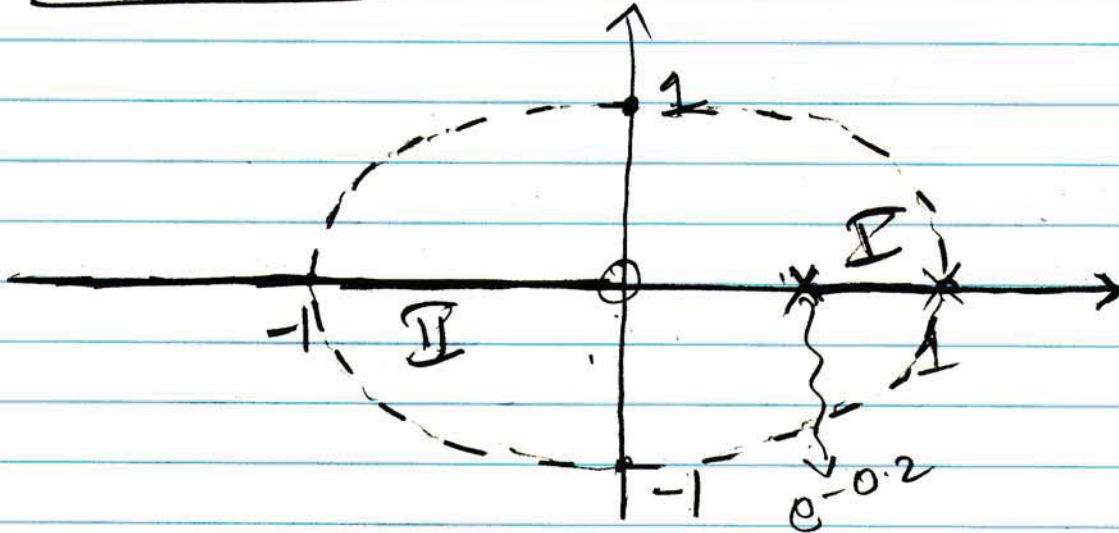
$$= k \sum_{m=0}^{\infty} z^{-m} - k \sum_{m=0}^{\infty} (z e^{-T})^{-m}$$

$$\Rightarrow \boxed{G(z) = \left(\frac{kz}{z-1} \right) - k \frac{z}{z - e^{-T}}}$$

$$T = 0.2 \Rightarrow G(z) = \left(\frac{kz}{z-1} \right) - \frac{kz}{z - e^{-0.2}}$$

$$(b) \quad \boxed{G(z) = \frac{kz(1 - e^{-0.2})}{(z-1)(z - e^{-0.2})}}$$

Root Locus :



(i) $RAS = \{I, II\}$.

(ii) Breakaway / Break in points :

$$1 \leq \frac{1}{\sigma_B + pi} = \leq \frac{1}{\sigma_B + 2i}$$

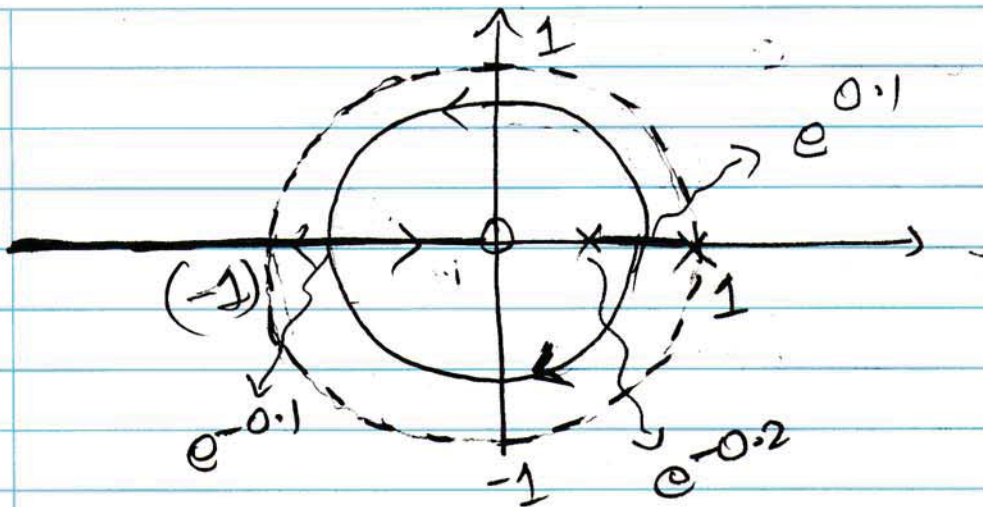
$$\Rightarrow \frac{1}{(\sigma_B - 1)} + \frac{1}{\sigma_B - e^{-0.2}} = \frac{1}{\sigma_B}$$

$$\Rightarrow (\sigma_B - e^{-0.2})\sigma_B + \sigma_B(\sigma_B - 1) = (\sigma_B - 1)(\sigma_B - e^{-0.2})$$

$$\Rightarrow 2\sigma_B^2 - \sigma_B(1 + e^{-0.2}) = \sigma_B^2 - \sigma_B(1 + e^{-0.2}) + e^{-0.2}$$

$$\Rightarrow \sigma_B^2 = e^{-0.2}$$

$$\boxed{\sigma_B = \pm e^{-0.1}}$$



closed loop characteristic polynomial:

$$z^2 + [(k-1) - e^{-0.2}(k+1)]z + e^{-0.2} = CP(z)$$

When $z = -1$ is a pole,

$$CP(-1) = 0 \Rightarrow 1 - [(k-1) - (k+1)e^{-0.2}] + e^{-0.2} = 0$$

$$\Rightarrow k(1 - e^{-0.2}) = 2(1 + e^{-0.2})$$

$$\Rightarrow k = \frac{2(1 + e^{-0.2})}{(1 - e^{-0.2})}$$

\Rightarrow Range of k for stability:

$$0 < k < \frac{2(1 + e^{-0.2})}{1 - e^{-0.2}}$$

$$c) \quad 0.20 = e^{-\varepsilon\pi/\sqrt{1-\varepsilon^2}}$$

$$+ \frac{\varepsilon\pi}{\sqrt{1-\varepsilon^2}} = -\ln(0.2) = \ln 5$$

$$\frac{\varepsilon^2\pi^2}{(1-\varepsilon^2)} = (\ln 5)^2 \Rightarrow \varepsilon^2 = \frac{(\ln 5)^2}{(\ln 5)^2 + \pi^2}$$

$$\Rightarrow \varepsilon = \frac{\sqrt{(\ln 5)^2}}{\sqrt{(\ln 5)^2 + \pi^2}} = \frac{\ln 5}{\sqrt{\ln^2 5 + \pi^2}}$$

$$\Rightarrow \boxed{\varepsilon \approx 0.46}$$

Using matlab :

$$\boxed{k \approx 0.23}$$

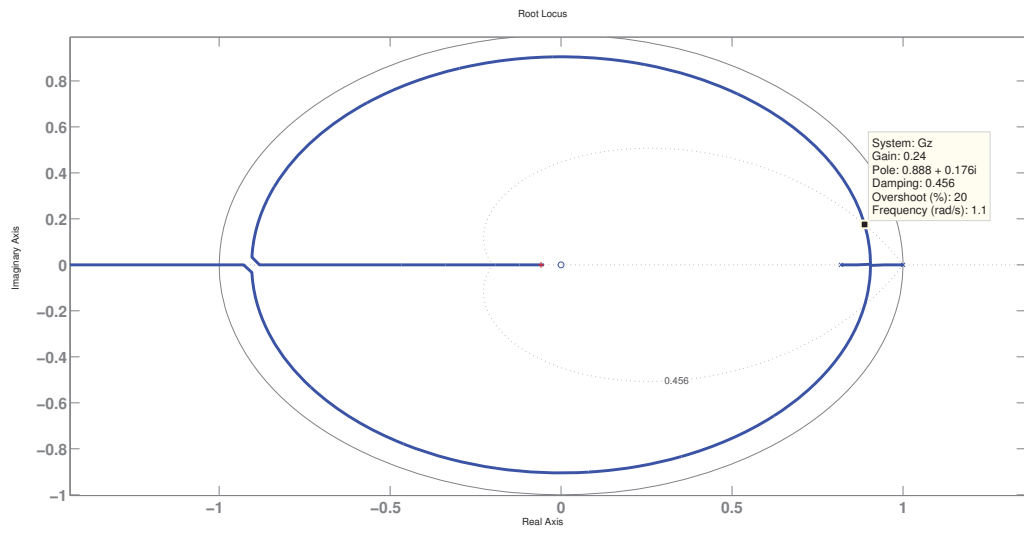


Figure 1: Root Locus showing Gain at OS % = 20%

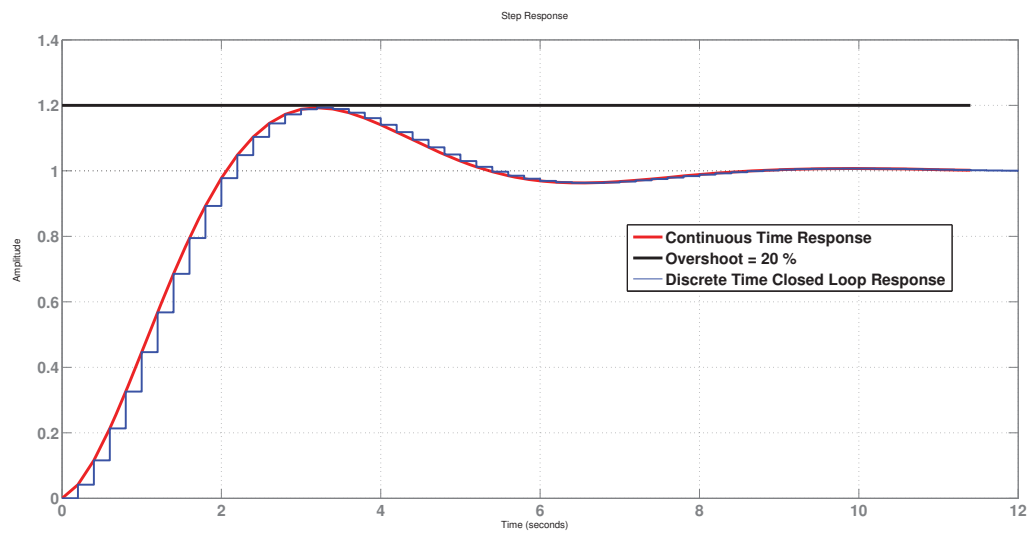


Figure 2: Step Response of Closed Loop System

$$6. \quad x[k+1] = G_1 x[k] + H_1 u[k] = \begin{bmatrix} 0 & 1 \\ -0.64 & 1.6 \end{bmatrix} x[k] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u[k] \quad y = \begin{bmatrix} 1 & 0 \end{bmatrix} x[k]$$

$$T_s = 1$$

$$a. \quad e[k] = r[k] - y[k]$$

$$u[k] = -K_1 x[k] + r[k]$$

$$x[k+1] = [G_1 - H_1 K] x[k] + H_1 r[k]$$

$$\begin{aligned} \text{characteristic equation: } \det(zI - (G_1 - H_1 K)) \\ &= \det\left(zI - \begin{bmatrix} 0 & 1 \\ -0.64 - k_1 & 1.6 - k_2 \end{bmatrix}\right) \\ &= \det\begin{pmatrix} z & -1 \\ 0.64 + k_1 & z - 1.6 + k_2 \end{pmatrix} \\ &= z^2 + (-1.6 + k_2)z + (0.64 + k_1) \end{aligned}$$

desired characteristic equation:

$$\begin{aligned} \chi(z) &= (z - (0.3 + 0.3j))(z - (0.3 - 0.3j)) \\ &= z^2 - 0.6z + 0.18 \end{aligned}$$

$$\begin{aligned} -1.6 + k_2 &= -0.6 & 0.64 + k_1 &= 0.18 \\ k_2 &= 1 & k_1 &= -0.46 \end{aligned}$$

$$K_1 = \begin{bmatrix} -0.46 & 1 \end{bmatrix}$$

$$G_k = (G_1 - H_1 K) = \begin{bmatrix} 0 & 1 \\ -0.64 - (-0.46) & 1.6 - 1 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -0.18 & 0.6 \end{bmatrix}$$

Closed loop system:

$$x[k+1] = G_k x[k] + H_1 r[k] = \begin{bmatrix} 0 & 1 \\ -0.18 & 0.6 \end{bmatrix} x[k] + \begin{bmatrix} 0 \\ 1 \end{bmatrix} r[k], y = \begin{bmatrix} 1 & 0 \end{bmatrix} x[k]$$

At steady state: $x[k+1] = x[k]$, $r[k] = 1$

$$x[k] = G_k x[k] + H_1 r[k] = G_k x[k] + H_1$$

$$x[k] = (I - G_k)^{-1} H_1$$

$$e[k] = r[k] - y[k] = 1 - Cx[k] = 1 - C(I - G_k)^{-1} H_1$$

$$= 1 - [1 \ 0] \begin{bmatrix} 1 & -1 \\ 0.18 & 0.4 \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$= 1 - [1 \ 0] \frac{1}{0.58} \begin{bmatrix} 0.4 & 1 \\ -0.18 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$= 1 - \frac{1}{0.58} [1 \ 0] \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

$$= 1 - \frac{1}{0.58} (1)$$

$$e[k] = -0.724$$

b. Add integrator: $x_N[k+1] = x_N[k] + e[k]$

$$= x_N[k] + r[k] - Cx[k]$$

$$\begin{bmatrix} x_1[k+1] \\ x_2[k+1] \\ x_N[k+1] \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ -0.64 & 1.6 & 0 \\ -1 & 0 & 1 \end{bmatrix} \begin{bmatrix} x_1[k] \\ x_2[k] \\ x_N[k] \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} u[k] + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} r[k]$$

$$y[k] = [1 \ 0 \ 0] \begin{bmatrix} x_1[k] \\ x_2[k] \\ x_N[k] \end{bmatrix}$$

This representation:

$$\bar{x}[k+1] = \bar{G} \bar{x}[k] + \bar{H} u[k] + \bar{F} r[k]$$

$$y[k] = \bar{C} \bar{x}[k]$$

$$c. \quad u[k] = -Kx[k] + k_e x_w[k] = -\overbrace{[k \quad -k_e]}^{\bar{K}} \begin{bmatrix} x[k] \\ x_w[k] \end{bmatrix}$$

$$x[k+1] = (\bar{G} - \bar{H}\bar{K})x[k] + \bar{F}r[k]$$

$$x[k+1] = \begin{bmatrix} x_1[k+1] \\ x_2[k+1] \\ x_w[k+1] \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ -0.64 - k_1 & 1.6 - k_2 & k_e \\ -1 & 0 & 1 \end{bmatrix} \begin{bmatrix} x_1[k] \\ x_2[k] \\ x_w[k] \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} r[k]$$

$$y[k] = [1 \quad 0 \quad 0] \begin{bmatrix} x_1[k] \\ x_2[k] \\ x_w[k] \end{bmatrix}$$

Characteristic polynomial:

$$\det(zI - (\bar{G} - \bar{H}\bar{K})) = z^3 + (k_2 - \frac{13}{5})z^2 + (k_1 - k_2 + \frac{56}{25})z + (-k_1 + k_e - \frac{16}{25})$$

Desired characteristic polynomial:

$$\chi(z) = (z - (0.3 + 0.3j))(z - (0.3 - 0.3j))(z - 0.1)$$

$$= z^3 - 0.7z^2 + 0.24z - 0.018$$

$$k_2 - \frac{13}{5} = -0.7 \quad (k_1 - k_2 + \frac{56}{25}) = 0.24 \quad -k_1 + k_e - \frac{16}{25} = -0.018$$

$$k_2 = 1.9$$

$$k_1 = -0.1$$

$$k_e = 0.522$$

$$\bar{K} = [-0.1 \quad 1.9 \quad -0.522]$$

At steady state: $x[k+1] = x[k]$, $r[k] = 1$

$$x[k] = (\bar{G} - \bar{H}\bar{K})x[k] + \bar{F}r[k]$$

$$x[k] = (I - (\bar{G} - \bar{H}\bar{K}))^{-1} \bar{F}$$

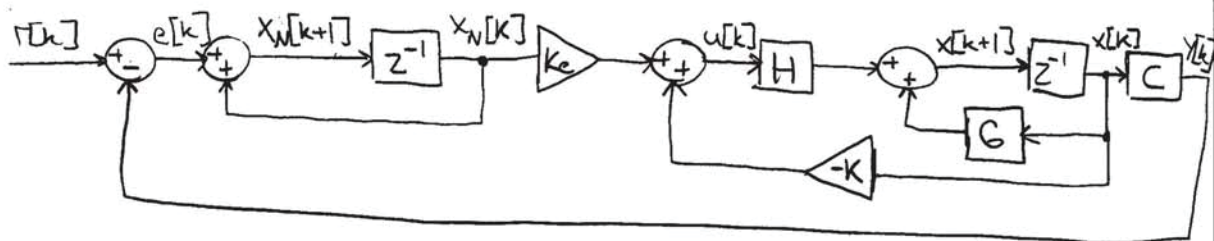
$$e[k] = r[k] - y[k] = 1 - \bar{C}x[k] = 1 - \bar{C}(I - (\bar{G} - \bar{H}\bar{K}))^{-1} \bar{F}$$

$$= 1 - [1 \quad 0 \quad 0] \begin{bmatrix} 1 & -1 & 0 \\ 0.54 & 0.7 & -0.522 \\ 1 & 0 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

$$= 1 - 1$$

$$e[k] = 0$$

Block diagram of system with integrator



Step Response

