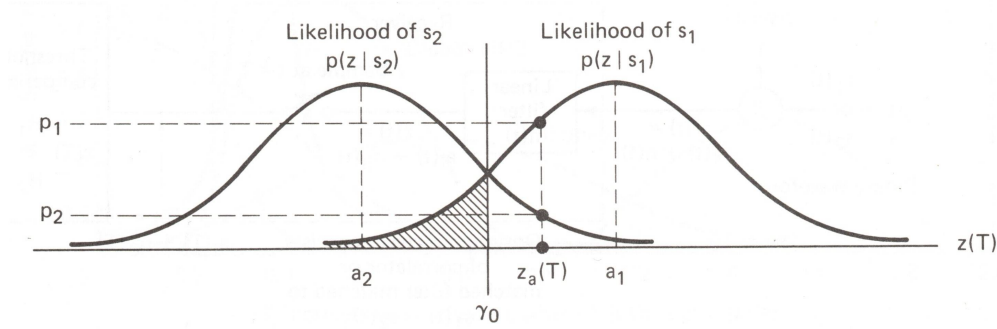


**P3. Error Probability**FIGURE 1. Conditional probability density functions:  $p(z|s_1)$  and  $p(z|s_2)$  .

For the binary example in Fig. 1, there are two ways in which errors can occur. An error,  $e$ , will occur when  $s_1(t)$  is sent, and channel noise results in the receiver output signal  $z(T)$ , being less than  $\gamma_0$ . The probability of such an occurrence is:

$$(1) \quad P(e|s_1) = P(H_2|s_1) = \int_{-\infty}^{\gamma_0} p(z|s_1) dz$$

This is illustrated by the shaded area to the left of  $\gamma_0$  in Fig. 1. Similarly an error occurs when  $s_2(t)$  is sent and the channel noise results in  $z(T)$ , being greater than  $\gamma_0$ . The probability of this occurrence is:

$$(2) \quad P(e|s_2) = P(H_1|s_2) = \int_{\gamma_0}^{\infty} p(z|s_2) dz$$

The probability of an error is the sum of the probabilities of all the ways that an error can occur. For the binary case, we can express the probability of bit error,  $P_B$ , as follows:

$$(3) \quad P_B = \sum_{i=1}^2 P(e, s_i)$$

Combining the previous three equations, we can write:

$$(4) \quad P_B = P(H_2|s_1)P(s_1) + P(H_1|s_2)P(s_2)$$

That is, given that signal  $s_1(t)$  was transmitted, an error results if hypothesis  $H_2$  is chosen, or given that the signal  $s_2(t)$  was transmitted, an error results if hypothesis  $H_1$  is chosen. For the case where the a priori probabilities are equal, that is,  $P(s_1) = P(s_2) = \frac{1}{2}$ ,

$$(5) \quad P_B = \frac{1}{2}P(H_2|s_1) + \frac{1}{2}P(H_1|s_2)$$

and because of the symmetry of the probability density functions

$$(6) \quad P_B = P(H_2|s_1) = P(H_1|s_2)$$

The probability of a bit error,  $P_B$ , is numerically equal to the area under the "tail" of either likelihood function,  $p(z|s_1)$  or  $p(z|s_2)$ , falling on the "incorrect" side of the threshold. We can therefore compute  $P_B$  by integrating  $p(z|s_1)$  between the limits  $-\infty$  and  $\gamma_0$ , or as shown below, by integrating  $p(z|s_2)$  between the limits  $\gamma_0$  and  $\infty$ :

$$(7) \quad P_B = \int_{\gamma_0=(a_1+a_2)/2}^{\infty} p(z|s_2) dz$$

where  $\gamma_0 = (a_1 + a_2)/2$  is the optimum threshold from Fig. 1. Replacing the likelihood  $p(z|s_2)$  with its Gaussian equivalent, we have:

$$(8) \quad P_B = \int_{\gamma_0=(a_1+a_2)/2}^{\infty} \frac{1}{\sigma_0\sqrt{2\pi}} \exp\left[-\frac{1}{2}\left(\frac{z-a_2}{\sigma_0}\right)^2\right] dz$$

where  $\sigma_0^2$  is the variance of the noise out of the correlator.

Let  $u = (z - a_2)/\sigma_0$ . Then  $\sigma_0 du = dz$  and:

$$(9) \quad P_B = \int_{\gamma_0=(a_1-a_2)/2}^{u=\infty} \frac{1}{\sqrt{2\pi}} \exp\left(-\frac{u^2}{2}\right) du = Q\left(\frac{a_1 - a_2}{2\sigma_0}\right)$$

where  $Q(x)$ , called the complementary error function or co-error function, is a commonly used symbol for the probability under the tails of the Gaussian distribution. It is defined as:

$$(10) \quad Q(x) = \frac{1}{\sqrt{2\pi}} \int_x^{\infty} \exp\left(-\frac{u^2}{2}\right) du$$

APT. 805 80 POINT MCKAY CR NW, CALGARY, ALBERTA, CANADA, T3B 4W4