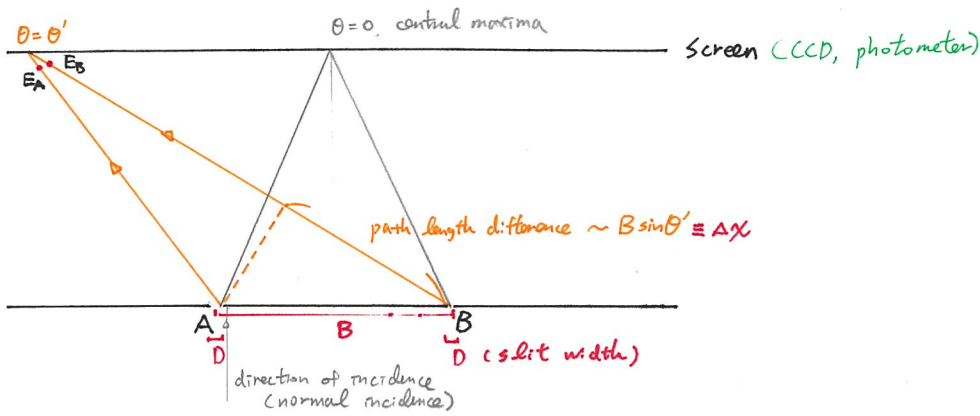


# From double-slit to radio interferometer



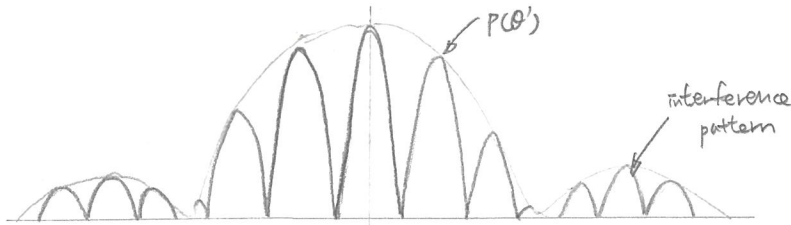
(i) if  $D \sim 0$ , for normal incident light with wavelength  $\lambda$  (making long-term exposure)  
light sources A and B can be viewed as monochromatic light sources that have identical initial phase.  
photon number flux density at  $\theta = \theta'$  on the screen

$$\propto \left( \underbrace{\cos(kx - \omega t + \phi_0)}_{E_A: \text{electric field contributed by light source A}} + \underbrace{\cos(k(x + \Delta x) - \omega t + \phi_0)}_{E_B} \right)^2$$

$$= \underbrace{\cos^2(kx - \omega t + \phi_0)}_{\text{auto-correlation}} + \underbrace{\cos^2(k(x + \Delta x) - \omega t + \phi_0)}_{\text{auto-correlation (long-term average over time is } \frac{1}{2})} + \underbrace{2 \cos(kx - \omega t + \phi_0) \cos(k(x + \Delta x) - \omega t + \phi_0)}_{\text{cross-correlation}}$$

using trigonometric identity:  
 $\cos \theta \cos \phi = \frac{1}{2} (\cos(\theta - \phi) + \cos(\theta + \phi))$

$$\cos(k\Delta x) + \underbrace{\cos(2kx - 2\omega t + 2\phi_0 + k\Delta x)}_{\text{long-term time average is 0}}$$



$$\frac{2\pi}{\lambda} \cdot B \sin \theta = m\pi, \quad m = 1, 3, 5, \dots$$

$$\theta' \sim \frac{m}{2} \frac{\lambda}{D}$$

Amplitude equal to 1. At the minimum  $\cos(k\Delta x) = -1$   
this cross-correlation term (together with the factor 2) can cancel the contribution of the auto-correlation terms.

When  $D$  is not negligibly small, the amplitude of the electric field contributed by each of the two slits is modulated by the factor  $\frac{\sin(\frac{1}{2}kD \sin \theta')}{\frac{1}{2}k \sin \theta'}$ , and carries an extra phase offset  $\frac{1}{2}kD \sin \theta'$ . Since the phase offsets are identical, the calculation of the interference pattern is not affected. The amplitude modulation can be expressed as an envelope factor  $P(\theta)$  in the final interference pattern:

(photon number flux density)  
Interference pattern can be expressed by

$$\frac{I(\theta')}{A} = P(\theta') \left[ 1 + \underbrace{\cos\left(\frac{2\pi}{\lambda} B \theta'\right)}_{\text{contribution of cross-correlation term}} \right] = P(\theta') \left[ 1 + \underbrace{\cos(2\pi u \theta')}_{\text{contribution of auto-correlation terms}} \right]$$

a coefficient that describes how bright is each of the light source

$u \equiv \frac{B}{\lambda}$ : slit separation measured in unit of wavelength

As compared to cross-correlation term auto-correlation terms are less sensitive to angle of emergence  $\theta$

When  $P(\theta')$  and  $u$  are known, making two independent measurements of  $I(\theta')$  (i.e., at two different  $\theta'$  positions) will allow unambiguously determining  $\theta'$  and  $A$

## Radio detector technique

lecture 3

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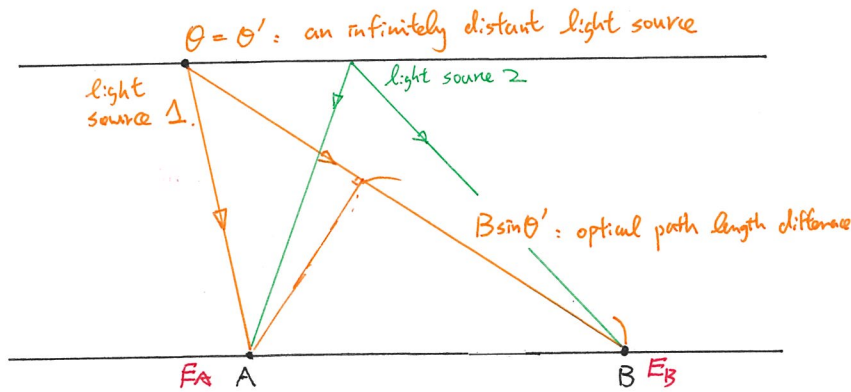
Directly measuring the electric field strength  $E_0 \cos(kx - \omega t + \phi_0)$

instead of photon number density, i.e., directly measuring  $\begin{cases} E_A = \cos(kx - \omega t + \phi_0) \\ E_B = \cos(k(x + \Delta x) - \omega t + \phi_0) \end{cases}$

then use correlation to evaluate

$\langle E_A^2 \rangle$ ,  $\langle E_B^2 \rangle$ ,  $\langle E_A E_B \rangle$   $\rightarrow$  denotes long-term time average.

## Radio Interferometry



(i) when there is only one infinitely distant, monochromatic light source, the  $\langle E_A E_B \rangle$  is exactly the same with the case of double slit

(ii) when there are two light sources, denoted as light sources 1 and 2, then  $\begin{cases} E_A = E_A^{(1)} + E_A^{(2)} \\ E_B = E_B^{(1)} + E_B^{(2)} \end{cases}$

$$\begin{aligned} \langle (E_A + E_B)^2 \rangle &= \langle E_A^{(1)2} \rangle + \langle E_B^{(1)2} \rangle + \langle 2E_A^{(1)} E_B^{(1)} \rangle \quad \text{--- interference pattern of source 1 alone} \\ &+ \langle E_A^{(2)2} \rangle + \langle E_B^{(2)2} \rangle + \langle 2E_A^{(2)} E_B^{(2)} \rangle \quad \text{--- interference pattern of source 2 alone} \\ &+ \langle 2E_A^{(1)} E_A^{(2)} \rangle + \langle 2E_A^{(1)} E_B^{(2)} \rangle + \langle 2E_B^{(1)} E_A^{(2)} \rangle + \langle 2E_B^{(1)} E_B^{(2)} \rangle \end{aligned}$$

if the two sources are not correlated, then the phase of these cross-term will walk randomly over time. In this case, the time average of these cross-term over time is zero

### Incoherent source

thermal noise is also incoherent source, of which the expectation value is 0, but the root-mean-square is positively definite.

if we only look at the cross-correlation terms.

$$\langle E_A E_B \rangle \propto \sum_{i=0}^N P(\theta'_i) A_i \cos(2\pi u \theta'_i)$$

a coefficient to describe how bright is the source

$\frac{B}{\lambda}$

generalizing from discrete sources to continuous sources.

F.T. coefficient  $\rightarrow$   $\int P(\theta) A(\theta) \cos(2\pi u \theta) d\theta$   $\rightarrow$  output of cosine correlator (sensitive to symmetric  $P(\theta)A(\theta)$  function)

the original intensity distribution, modulated by the primary beam response function.

Fourier transform

If there is a way to manually add an  $\frac{1}{2}\pi$  phase to  $E_B$ , (denoting phase shifted case as  $E'_B$ )

$$\begin{aligned} (E_A + E'_B)^2 &= \left( \cos(kx - \omega t + \phi_0) + \cos(k(x + \Delta x) - \omega t + \phi_0 + \frac{\pi}{2}) \right)^2 \\ &= \left( \cos(kx - \omega t + \phi_0) + \cos(k(x + \Delta x) - \omega t + \phi_0) \cos \frac{\pi}{2} - \sin(k(x + \Delta x) - \omega t + \phi_0) \sin \frac{\pi}{2} \right)^2 \\ &= \cos^2(kx - \omega t + \phi_0) + \sin^2(k(x + \Delta x) - \omega t + \phi_0) - 2 \cos(kx - \omega t + \phi_0) \sin(k(x + \Delta x) - \omega t + \phi_0) \end{aligned}$$

using trigonometric identity

$$\cos \theta \sin \phi = \frac{1}{2} (\sin(\theta + \phi) - \sin(\theta - \phi))$$

$$\frac{\sin(2kx - 2\omega t + 2\phi_0 + k\Delta x)}{\text{long-term average is 0}} - \frac{\sin(k\Delta x)}{\text{sine correlation}}$$

generalizing from discrete sources to continuous sources

$\int P(\theta) A(\theta) \sin(2\pi u \theta) d\theta \rightarrow$  output of sine correlator (sensitive to asymmetric  $P(\theta)A(\theta)$  function)

we can combine these measurements to yield the complex visibility

The outputs of cosine/sine correlators are the real/imaginary parts of the complex visibility. A complex visibility includes two independent measurements. The flux density and position of a single point-source can be determined by one complex visibility.

\* Imaging: the inversion from  $\langle E_A E_B \rangle$  to  $P(\theta)A(\theta)$

if we can vary the value of  $u$  continuously (i.e., changing the separation of the two radio detectors), then we can make a complete Fourier transform of  $P(\theta)A(\theta)$ . In this case, imaging can be a simple inverse Fourier transform. Otherwise, we obtain incomplete samplings that requires a more complicated  $\rightarrow$   $\text{reconstruction}$  (3)



Example of complex visibility: point source

$$A(\theta) = \delta(\theta - \theta')$$

⇒ output of cosine correlator

$$\int p(\theta) \delta(\theta - \theta') \cos(2\pi u \theta) d\theta$$

complex gain  
↓  
attributed by atmospheric and instrumental effects.

$$= g p(\theta') \cos(2\pi u \theta') \quad \text{for any } u$$

similarly, the output of sine correlator is  $g p(\theta') \sin(2\pi u \theta')$

complex visibility =  $g p(\theta') (\cos(2\pi u \theta') + i \sin(2\pi u \theta'))$

$$= \underbrace{g p(\theta')}_{\text{uncalibrated}} e^{i \underbrace{2\pi u \theta'}_{\text{phase}}}$$

visibility amplitude (e.g., in unit of volts)

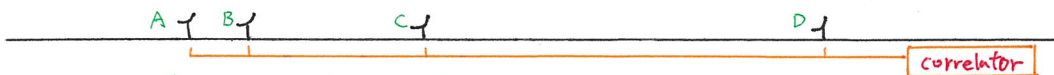
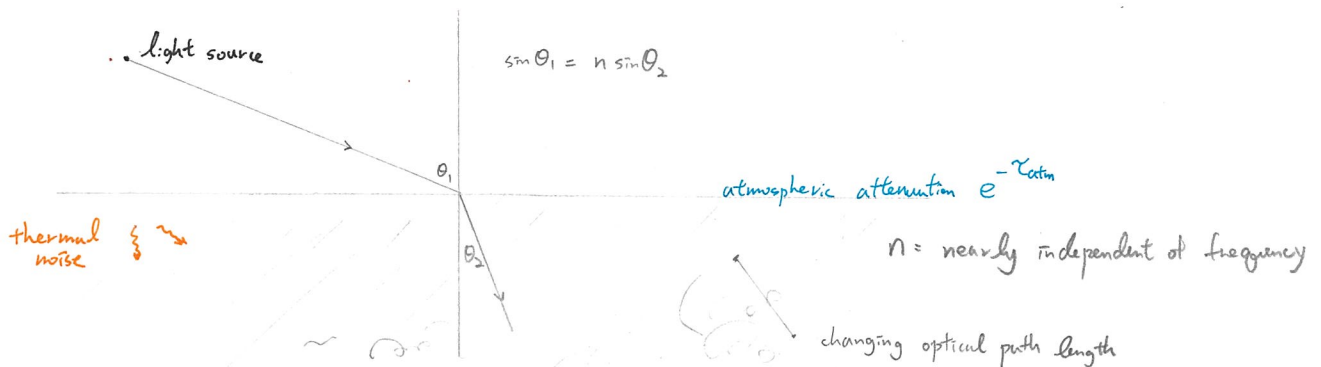
Summary: for a point source, visibility amplitude does not vary with  $|u|$ .

For a fixed source location  $\theta'$ , phase depends linearly on  $u$ .

We can regard  $2\pi u$  as the rate for the phase to vary with  $\theta'$ .

With a longer baseline length  $|u|$ , phase will vary more rapidly with  $\theta'$ .

1D linear radio interferometer



$$U_{AB}, U_{AC}, U_{AD}, U_{BC}, U_{BD}, U_{CD}, \dots$$

The total number of independent baseline is  $\frac{1}{2}N(N-1)$

When designing an array, the less redundant is the sampling  $U_{ij}$  (i.e., less repeats of the measurements at the same baseline length), the better for the image reconstruction.

\* if you cannot understand the derivation, you can read the 1-D case again, and take the conclusion for 2D as granted for a moment)  
Generalization to a 2D array observations towards celestial sources (quasi-monochromatic)  
 (ignoring polarization for this moment)

Electric field contributed from infinitely distant sources:

the electric field detected at the station that is located at  $\vec{r}$

$$\underline{E}_z(\vec{r}) = \int \underline{E}_z(\vec{R}) \frac{e^{i k |\vec{R} - \vec{r}|}}{|\vec{R} - \vec{r}|} dA = \int \underline{E}_z(\vec{R}) \frac{e^{i \frac{2\pi}{\lambda} |\vec{R} - \vec{r}|}}{|\vec{R} - \vec{r}|} dA$$

field strength at the same location

Huygens' Principle: dilution of optical wave field strength during the propagation

a surface area on the celestial sphere. Dimension is  $[m^2]$   
 $dA = |\vec{R}|^2 d\Omega$

if we correlated the signal taken at two different locations

$$V_z(\vec{r}_1, \vec{r}_2) = \langle \underline{E}_z(\vec{r}_1) \underline{E}_z^*(\vec{r}_2) \rangle = \left\langle \iint \underline{E}_z(\vec{R}_1) \underline{E}_z^*(\vec{R}_2) \frac{e^{i k |\vec{R}_1 - \vec{r}_1|}}{|\vec{R}_1 - \vec{r}_1|} \frac{e^{-i k |\vec{R}_2 - \vec{r}_2|}}{|\vec{R}_2 - \vec{r}_2|} dA_1 dA_2 \right\rangle$$

$(\cos\theta_1 + i \sin\theta_1)(\cos\theta_2 - i \sin\theta_2)$

for coherent source,  
 $\langle \underline{E}_z(\vec{R}_1) \underline{E}_z^*(\vec{R}_2) \rangle = 0$  for  $\vec{R}_1 \neq \vec{R}_2$   
 (i.e. this can be expressed as delta function)

\* 1D:

$$\frac{1}{|\vec{R} - \vec{r}|} = \frac{1}{(R-r)^2}^{\frac{1}{2}}$$

1st order Taylor expansion  $R \gg r$

$$\frac{1}{R(1 - \frac{r}{R})^2} \approx \frac{1}{R(1 - \frac{r}{R})} \approx \frac{1}{R} (1 + \frac{r}{R})$$

$$V_z(\vec{r}_1, \vec{r}_2) \approx \left\langle \iint \underline{E}_z(\vec{R}_1) \underline{E}_z^*(\vec{R}_2) \frac{e^{i k R}}{R} \frac{e^{-i k R}}{R} \frac{e^{-i k \vec{s}_1 \cdot \vec{r}_1}}{R} \frac{e^{i k \vec{s}_2 \cdot \vec{r}_2}}{R} |\vec{R}_1|^2 d\Omega_1 |\vec{R}_2|^2 d\Omega_2 \right\rangle$$

$\vec{s}_1 \equiv \frac{\vec{R}_1}{|\vec{R}_1|} = \text{unit vector}$

intensity:  $I_z(\vec{s}) \equiv |\vec{R}|^2 \langle |\underline{E}_z(\vec{s})|^2 \rangle$ ,  $\vec{s} \equiv \frac{\vec{R}}{|\vec{R}|}$

$$\Rightarrow V_z(\vec{r}_1, \vec{r}_2) \approx \int I_z(\vec{s}) e^{-i k \vec{s} \cdot (\vec{r}_1 - \vec{r}_2)} d\Omega$$

$$= \int I_z(\vec{s}) e^{-2\pi i \vec{s} \cdot \frac{(\vec{r}_1 - \vec{r}_2)}{\lambda}} d\Omega$$

- ① for a coplanar array, we can always select the Cartesian coordinate system such that  $\vec{r}_1 - \vec{r}_2$  can be explicitly expressed as  $(\lambda u, \lambda v, \lambda w)$  baseline length measured in units of wavelength
- ② the unit direction vector  $\vec{s}$  can be expressed in a cartesian coordinate system as  $(\ell, m, \sqrt{1 - \ell^2 - m^2})$

$$V_z(u, v, w=0) = \iint I_z(\ell, m) \frac{e^{-2\pi i (\ell u + m v)}}{\sqrt{1 - \ell^2 - m^2}} d\ell dm$$

For target source with very small angular scale

$\vec{s} \equiv \vec{s}_0 + \vec{\delta}$  angular offset from the center  
 $\vec{s}_0$  unit vector in the line-of-sight direction towards the center of the source  
 choosing a Cartesian coordinate system such that  $\vec{s}_0 = (0, 0, 1)$  we call  $\vec{s}_0$  the phase tracking center

$$\Rightarrow V'_z(u, v, w) = e^{-2\pi i w} \iint I_z(\ell, m) e^{-2\pi i (\ell u + m v)} d\ell dm$$

if we redefine  $V_z = e^{2\pi i w} V'_z$  then

$$V_z(u, v) = \iint I_z(\ell, m) e^{-2\pi i (\ell u + m v)} d\ell dm$$

The relation between  $V_z$  and  $I_z$  is a (inverse) Fourier transform