Austin Barrilleaux Whiting School of Engineering Johns Hopkins University

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#### MODULE 5 — Practice Assignment

#### Problem 1

Solve the following practice problems in the 9th edition textbook.

- Chapter 2:
  - -2-33 (a-f)
  - -2-39
- 2-33. Without using the Routh-Hurwitz criterion, determine if the following systems are asymptotically stable. marginally stable, or unstable. In each case, the closed-loop system transfer function is given.

$${\bf (a)} \quad {\bf M(s)} = \frac{10(s+2)}{s^3+3s^2+5s}$$

Using the roots () function MATLAB, we get that the roots are:

roots([1,3,4,0]) = 
$$\begin{pmatrix} 0 \\ -1.5000 + 1.6583i \\ -1.5000 - 1.6583i \end{pmatrix}$$

There is a real pole at s = 0.

The system is marginally stable.

 $\longrightarrow \mathcal{A}$ nswer

(b) 
$$M(s) = \frac{(s-1)}{(s+5)(s^2+2)}$$

Using the roots () function MATLAB, we get that the roots are:

roots(conv([1,5],[1,0,2])) = 
$$\begin{pmatrix} -5\\ 0.0+1.4142i\\ 0.0-1.4142i \end{pmatrix}$$

There are complex conjugate poles on the imaginary axis (real parts of s equal to zero).

The system is marginally stable.

 $\longrightarrow \mathcal{A}$ nswer

$$\mathbf{(c)} \quad \mathbf{M(s)} = \frac{\mathbf{K}}{\mathbf{s^3 + 5s + 5}}$$

Using the roots () function MATLAB, we get that the roots are:

roots([1,0,5,5]) = 
$$\begin{pmatrix} 0.4344 + 2.3593i \\ 0.4344 - 2.3593i \\ -0.8688 \end{pmatrix}$$

There are complex conjugate poles in the right-hand plane (RHP).

The system is **unstable**.

 $\longrightarrow \mathcal{A}$ nswer

$$\mathbf{(d)} \quad \mathbf{M(s)} = \frac{100(s-1)}{(s+5)(s^2+2s+2)}$$

Using the roots () function MATLAB, we get that the roots are:

roots(conv([1,5],[1,2,2])) = 
$$\begin{pmatrix} -5\\ -1+1i\\ -1-1i \end{pmatrix}$$

All poles exist in the left-hand plane (LHP).

The system is **stable**.

 $o \mathcal{A}$ nswer

(e) 
$$M(s) = \frac{100}{s^3 - 2s^2 + 3s + 10}$$

Using the roots () function MATLAB, we get that the roots are:

roots([1,-2,3,10]) = 
$$\begin{pmatrix} 1.6694 + 2.1640i \\ 1.6694 - 2.1640i \\ -1.3387 \end{pmatrix}$$

There are complex conjugate poles in the right-hand plane (RHP).

The system is **unstable**.

 $\longrightarrow \mathcal{A}$ nswer

$$(f) \quad M(s) = \frac{10(s+12.5)}{s^4+s^3+50s^2+s+10^6}$$

Using the roots () function MATLAB, we get that the roots are:

$$\operatorname{roots}([1,3,50,1,10^{6}]) = \begin{pmatrix} -22.8487 + 22.6376i \\ -22.8487 - 22.6376i \\ 21.3487 + 22.6023i \\ 21.3487 - 22.6023i \end{pmatrix}$$

There are complex conjugate poles in the right-hand plane (RHP).

The system is **unstable**.

 $\longrightarrow \mathcal{A}$ nswer

# 2-39. The loop transfer function of a single-loop feedback control system is given as

$$\mathbf{G}(\mathbf{s})\mathbf{H}(\mathbf{s}) = \frac{\mathbf{K}(\mathbf{s}+\mathbf{5})}{\mathbf{s}(\mathbf{s}+\mathbf{2})(\mathbf{1}+\mathbf{T}\mathbf{s})}$$

The parameters K and T may be represented in a plane with K as the horizontal axis and T as the vertical axis. Determine the regions in the T-versus-K parameter plane where the closed-loop system is asymptotically stable and where it is unstable. Indicate the boundary on which the system is marginally stable.

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The closed loop transfer function of the above loop transfer function is:

$$\frac{\mathbf{G}(\mathbf{s})\mathbf{H}(\mathbf{s})}{\mathbf{1} + \mathbf{G}(\mathbf{s})\mathbf{H}(\mathbf{s})} = \frac{\frac{\mathbf{K}(\mathbf{s} + \mathbf{5})}{\mathbf{s}(\mathbf{s} + 2)(\mathbf{1} + \mathbf{T}\mathbf{s})}}{\mathbf{1} + \frac{\mathbf{K}(\mathbf{s} + \mathbf{5})}{\mathbf{s}(\mathbf{s} + 2)(\mathbf{1} + \mathbf{T}\mathbf{s})}} = \frac{K(s + 5)}{s(s + 2)(1 + Ts) + K(s + 5)}$$

This makes the characteristic equation:

$$s(s+2)(1+Ts) + K(s+5) = 0$$

Which can be written as:

$$Ts^3 + (1+2T)s^2 + (2+K)s + 5K = 0$$

$s^3$	Т	(2 + K)
$s^2$	(1+2T)	5K
$s^1$	(2+K)(1+2T) - 5TK	0
	(1+2T)	
$\mathbf{s^0}$	5K	0

The third left-most row is calculated as:

$$-\frac{\left|\begin{array}{cc} T & (2+K) \\ (1+2T) & 5K \end{array}\right|}{(1+2T)} = \frac{(2+K)(1+2T) - 5TK}{(1+2T)}$$

The fourth left-most row is simply equal to the coefficient of  $s^0$ .

Given this Routh array, the following must be true so that no signs change occours in the left-most column.

$$T > 0$$

$$(T + \frac{1}{2}) > 0$$

$$(2 + K)(1 + 2T) - 5TK > 0$$

$$K > 0$$

Taking (2+K)(1+2T)-5TK>0, we can write this as:

$$2 + K + 4T + 2TK - 5TK > 0$$

Which simplifies to:

$$2 + K + 4T + -3TK > 0$$

Solving for K:

$$2 + 4T + K(1 - 3T) > 0$$

$$2 + 4T + > K(3T - 1)$$

$$\frac{(2+4T)}{(3T-1)} > K$$

The condition of stability exists when T > 0, K > 0 and  $K < \frac{(2+4T)}{(3T-1)}$ . The following figure shows the stability region:

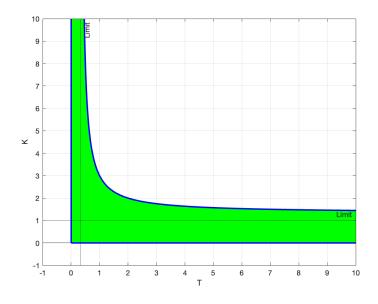


Figure 1: Stability Region

Populating these three boundaries with test values, I can evaluate the roots of the characteristic equation:

If I use values of T = 0 and K = 1:

roots(subs([T,(1 + 2\*T),(2 + K),5\*K],[T,K],[0,1])) = 
$$\begin{pmatrix} -2.5000 - 2.9580i \\ -2.5000 + 2.9580i \end{pmatrix}$$

Indicating that the system is **stable** along the boundary T = 0, since both poles are in the left-hand plane (LHP).

 $\longrightarrow \mathcal{A}$ nswer

If I use values of T = 1 and K = 0:

roots(subs([T,(1 + 2\*T),(2 + K),5\*K],[T,K],[1,0])) = 
$$\begin{pmatrix} -3\\0.0-2.2361i\\0.0+2.2361i \end{pmatrix}$$

Indicating that the system is **marginally stable** along the boundary K = 0, as there are complex conjugate poles with real components equal to zero.

 $\longrightarrow \mathcal{A}$ nswer

If we evaluate  $K = \frac{(2+4T)}{(3T-1)}$  with the value T=2, we get K=2. Evaluating the characteristic equation with these values:

roots(subs([T,(1 + 2\*T),(2 + K),5\*K],[T,K],[2,2])) = 
$$\begin{pmatrix} -2.5000\\0.0-1.4142i\\0.0+1.4142i \end{pmatrix}$$

Indicating that the system is **marginally stable** along the boundary K = 0, as there are complex conjugate poles with real components equal to zero.

 $\longrightarrow \mathcal{A}$ nswer

If I use values of T = 0 and K = 0:

roots(subs([T,(1 + 2\*T),(2 + K),5\*K],[T,K],[0,0])) = 
$$\begin{pmatrix} 0 \\ -2 \end{pmatrix}$$

Indicating that the system is marginally stableat the point K = 0 and T = 0, as there is a pole with a value of zero.

 $\longrightarrow \mathcal{A}$ nswer

#### Problem 2

Consider the following transfer function and do the following

$$G(s) = \frac{Y(s)}{R(s)} = \frac{3s+2}{2s^3+4s^2+5s+1}$$

### (a) Employ the Routh-Hurwitz criterion to determine if this system stable or unstable?

I wrote the following MATLAB function to solve for a Routh-Array given the coefficients of a characteristic equation:

```
for iterCol = 1:num_col-1
    % Intialize matrix for determinant
    matrix = zeros(2,2);
    % Populate matrix for determinant
    matrix(1,1) = table(iterRow-2,1);
    matrix(2,1) = table(iterRow-1,1);
    matrix(1,2) = table(iterRow-2,iterCol+1);
    matrix(2,2) = table(iterRow-1,iterCol+1);
    % Compute table value
    table(iterRow,iterCol) = ...
        -det(matrix)/table(iterRow-1,1);
    end
end
% Last table value is equal to coefficient of
% s^0
table(end,1) = CE(1,end);
```

If I run it using the command routhHurwitz([2,4,5,1]), I get the following result:

$s^3$	2	5
$\mathbf{s^2}$	4	1
$\mathbf{s^1}$	4.5	0
$s^0$	1	0

Since there are no sign changes in the left-most column, the system is **stable**.

 $\longrightarrow \mathcal{A}$ nswer

If I evaluate the command roots ([2, 4, 5, 1]) in MATLAB, I get the following:

roots([2,4,5,1]) = 
$$\begin{pmatrix} -0.8796 + 1.1414i \\ -0.8796 - 1.1414i \\ -0.2408 \end{pmatrix}$$

All poles exist in the left-hand plane (LHP).

The system is **stable**.

 $\longrightarrow \mathcal{A}$ nswer

\* Note that the function I wrote is not rubust, as if any of the determinants evaluate zero, the resulting table will be incorrect.

# (b) Define the numerator and denominator in MATLAB, and use the STEP command to plot the system unit step response

I evaluated the following in MATLAB:

```
% Problem 2 (b)
num = [3,2];
den = [2,4,5,1];
sys = tf(num,den);
step(sys)
```

This produced the following plot:

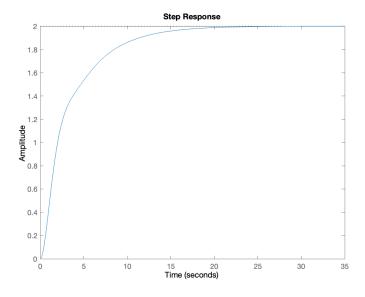


Figure 2: Step Response

# (c) Use the final value theorem to determine the steady state value of the system - does this agree with the step response?

From the final value theorem we know that:

$$\lim_{t \to +\infty} f(t) = \lim_{s \to 0} sF(s)$$

Therefore:

$$\lim_{t\to +\infty}y(t)=\lim_{s\to 0}sY(s)$$

For the open loop case, given  $r(t) = u_s(t)Y$ :

$$R(s) = \frac{1}{s}$$

If we substitute this into the closed loop transfer function, we get:

$$sY(s) = \frac{3s+2}{2s^3+4s^2+5s+1}$$

Therefore:

$$\lim_{t \to +\infty} y(t) = \lim_{s \to 0} sY(s) = \lim_{s \to 0} \frac{3s+2}{2s^3+4s^2+5s+1} = \frac{2}{1} = 2$$

 $\longrightarrow \mathcal{A}$ nswer

The steady state value of the system is **2**, and this **agrees** with the plot that I produced in part (c).

 $\longrightarrow \mathcal{A}$ nswer

Submitted by Austin Barrilleaux on September 30, 2023.