

11.09.2020

Digital Image Processing (CSE/ECE 478)

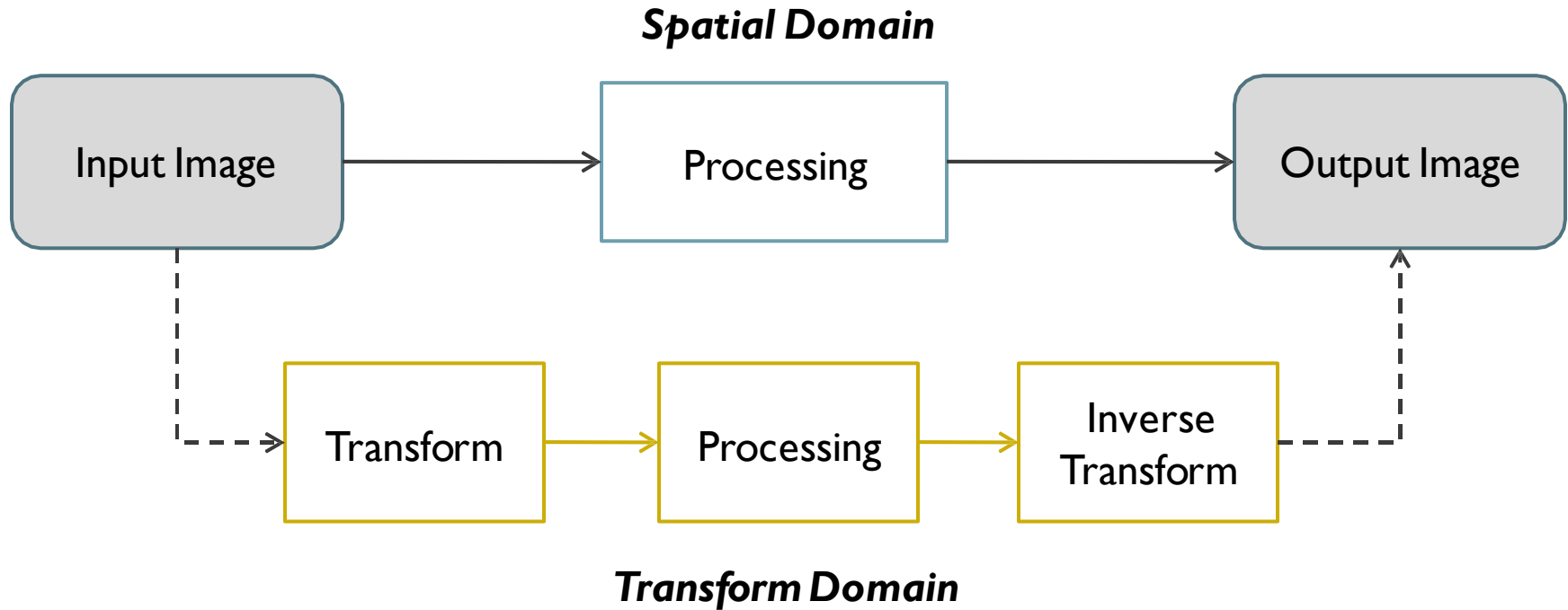
Lecture-10: Frequency Domain Processing

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Spatial vs. Transform Domain Processing

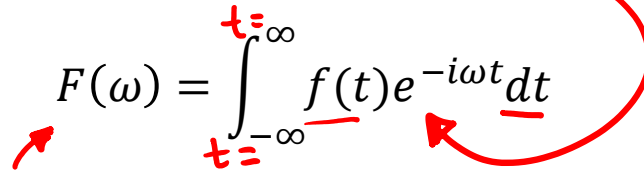


Fourier Transform

Approximate non-periodic signals with complex sinusoids

Intuition for FT

- $f(t)$ = Single number
- How much of frequency ω signal is present for all values of t ?



The diagram shows the Fourier Transform integral $F(\omega) = \int_{-\infty}^{\infty} f(t)e^{-i\omega t} dt$. Red annotations include: a red arrow pointing to $F(\omega)$; a red arrow pointing to the lower limit $-\infty$ with a red $t =$ written below it; a red arrow pointing to the upper limit ∞ with a red $t =$ written above it; a red underline under the t in the integrand $f(t)$; and a large red curved arrow starting from the ω in the exponent and pointing back towards the text 'frequency ω signal' in the list above.

$$F(\omega) = \int_{-\infty}^{\infty} f(t)e^{-i\omega t} dt$$

$$F(\omega) = \mathcal{F}[f(t)]$$

Fourier Transform and Inverse Fourier Transform

- Fourier Transform

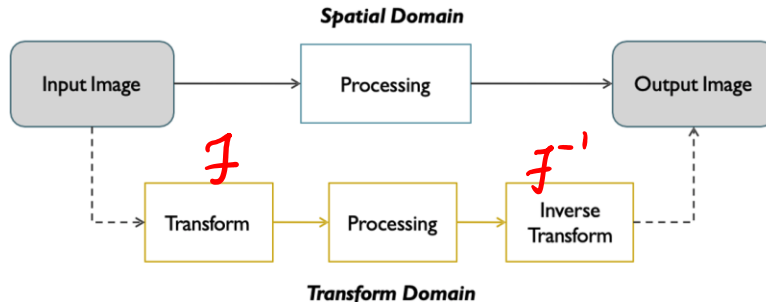
$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-i\omega t} dt$$

$$F(\omega) = \mathcal{F}[f(t)]$$

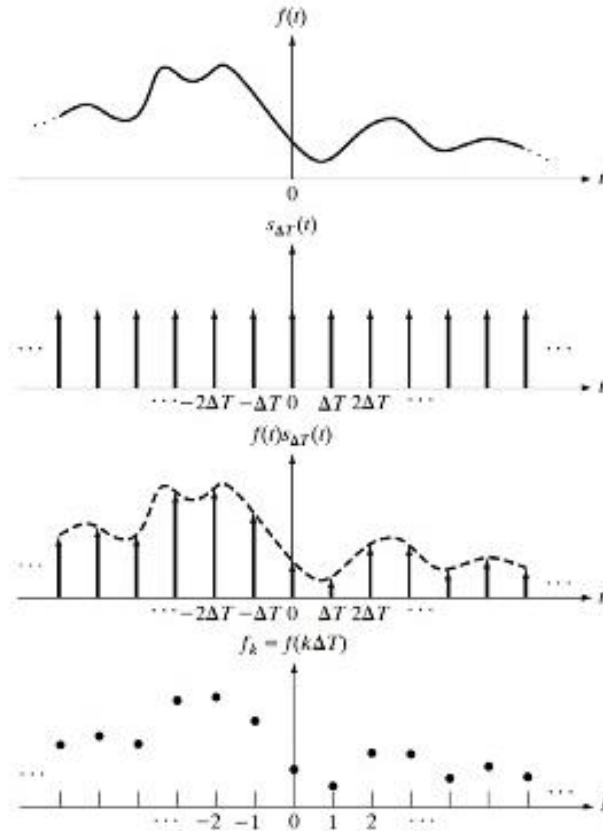
- Inverse Fourier Transform

$$f(t) = \int_{\omega=-\infty}^{\omega=\infty} \frac{1}{2\pi} F(\omega) e^{i\omega t} d\omega$$

$$f(t) = \mathcal{F}^{-1}[F(\omega)]$$



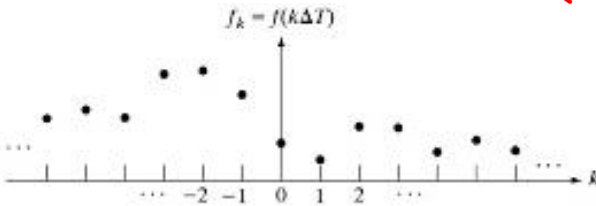
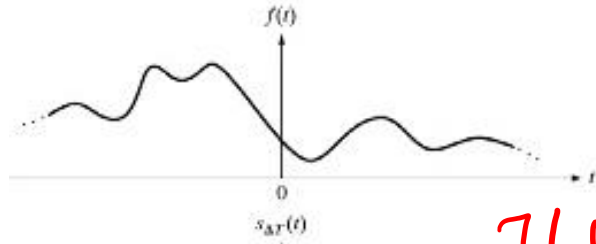
Sampling = $f(t)$ x Impulse Train



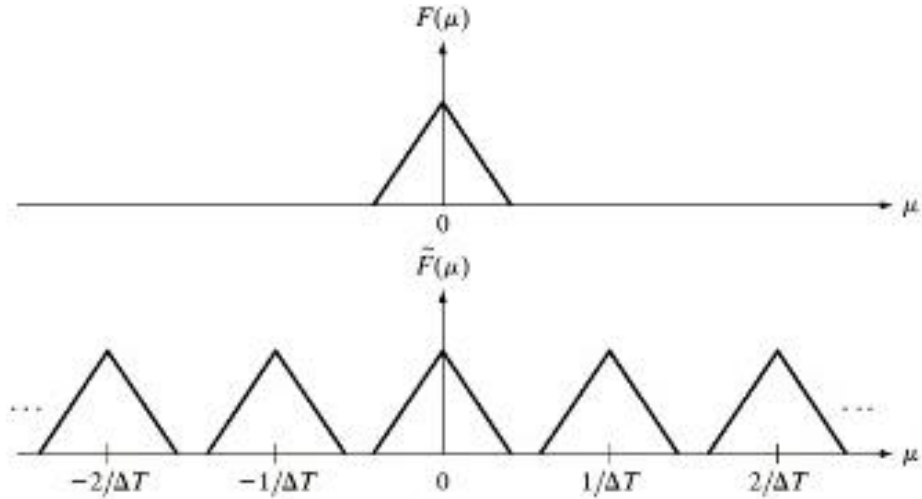
$$s_{\Delta T}(t) = \sum_{n=-\infty}^{n=\infty} \delta(t - n\Delta T)$$

$$\underline{\tilde{f}(t)} = \sum_{n=-\infty}^{n=\infty} \underbrace{f_n}_{\text{red}} \delta(t - n\Delta T)$$

FT of sampled function (G&W 4.2.4)



$$f(f(t) s_{\Delta}(t))$$

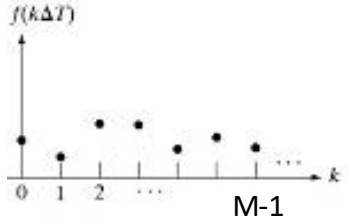


- Continuous
- Periodic (copies of $f(t)$'s FT)

$$\tilde{f}(t) = \sum_{n=-\infty}^{n=\infty} f_n \delta(t - n\Delta T)$$

$$\tilde{F}(\mu) = \frac{1}{\Delta T} \sum_{n=-\infty}^{n=\infty} F\left(\mu - \frac{n}{\Delta T}\right) \quad \frac{1}{\Delta T}$$

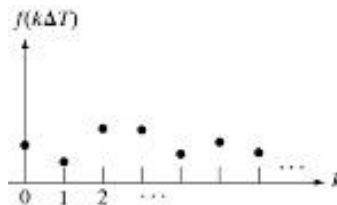
FT of sampled function



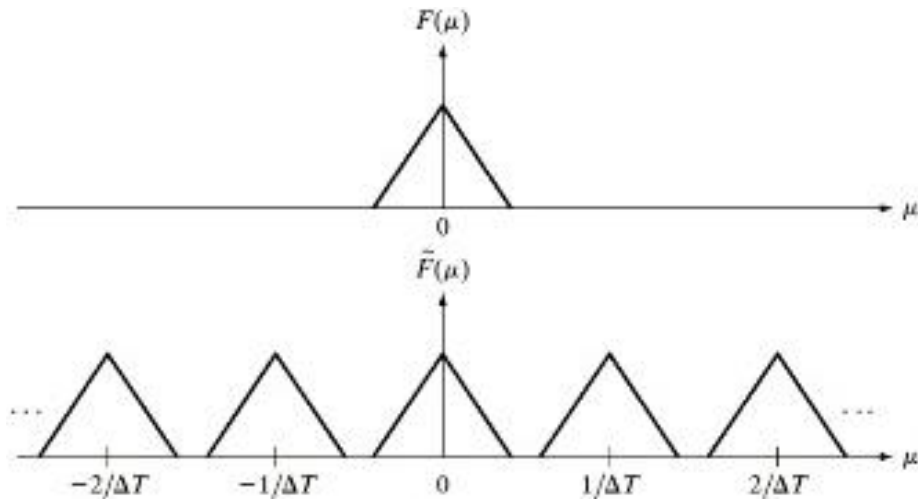
$$\tilde{f}(t) = \sum_{n=0}^{n=(M-1)\Delta T} f_n \delta(t - n\Delta T)$$

$$\underline{\tilde{F}(\mu)} = \sum_{n=0}^{n=(M-1)} f_n e^{-j2\pi\mu n\Delta T}$$

Digital processing of frequencies



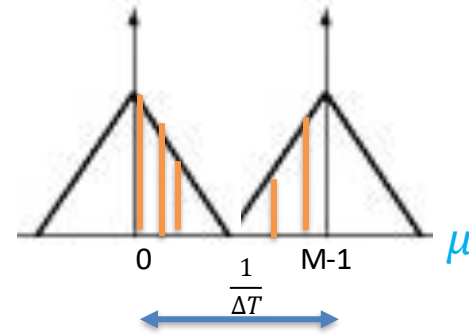
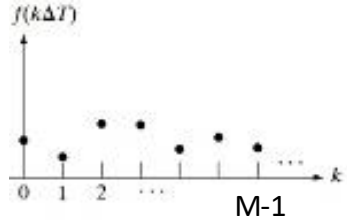
$$\tilde{f}(t) = \sum_{n=-\infty}^{n=\infty} f_n \delta(t - n\Delta T)$$



$$\tilde{F}(\mu) = \sum_{n=0}^{n=(M-1)} f_n e^{-j2\pi\mu n\Delta T} \quad \mu \in \mathbb{R}$$

- Need discrete frequency samples, but FT of sampled function is continuous
- OBSERVATION: Characterizing one period ($\frac{1}{\Delta T}$) is enough
- How do we get frequency 'samples' ?

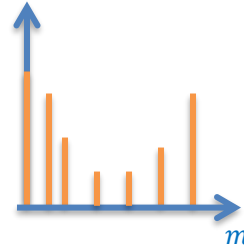
FT of sampled function (G&W 4.4.1)



$$\tilde{f}(t) = \sum_{n=0}^{n=(M-1)\Delta T} f_n \delta(t - n\Delta T)$$

$$\tilde{F}(\mu) = \sum_{n=0}^{n=(M-1)} f_n e^{-j2\pi\mu n\Delta T}, \mu \in R$$

$F[m]$



• Substituting $\mu = \frac{m}{M\Delta T}$ $m = 0, 1, 2, \dots, M-1$

$$F[m] = \sum_{n=0}^{n=(M-1)} f_n e^{-j\frac{2\pi mn}{M}}, m = 0, 1, \dots, (M-1)$$

NOTE: No direct dependence on ΔT



DFT and IDFT

$$\underline{F[m]} = \sum_{n=0}^{n=(M-1)} f_n e^{\frac{-j2\pi nm}{M}}, m = 0, 1, \dots (M-1)$$

$$f_n = \frac{1}{M} \sum_{m=0}^{m=(M-1)} F_m e^{\frac{j2\pi nm}{M}}, n = 0, 1, \dots (M-1)$$

$F[m]$

$\text{Re}\{F[m]\}$ $\text{Im}\{F[m]\}$

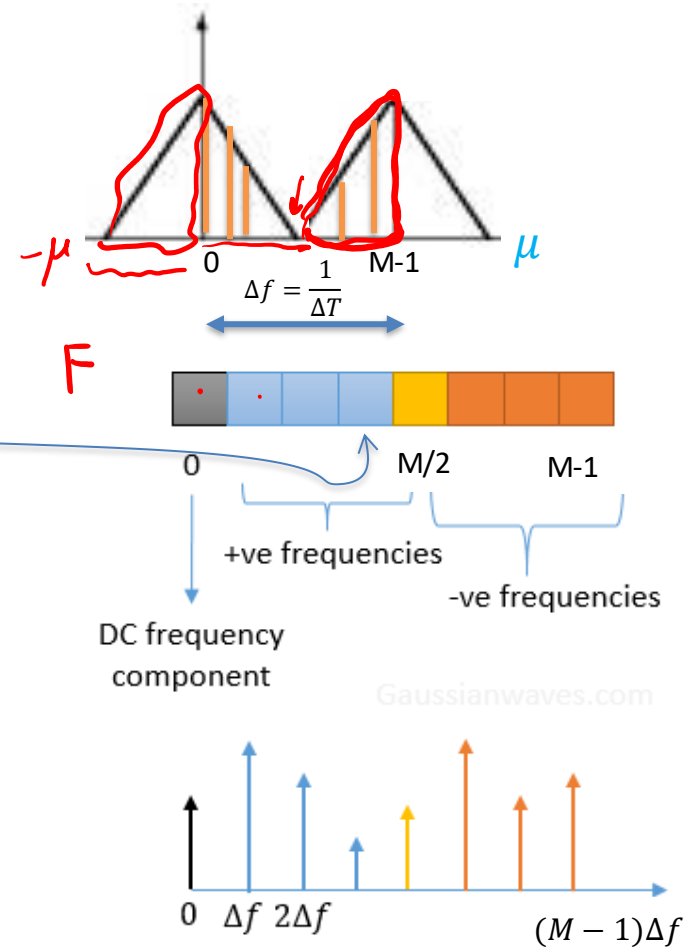
- A complex value
- Represents amplitude, phase of function $f[.]$'s content at angular frequency $2\pi m/M$

DFT: Record of 'energy' portion at various frequency bands present in input function $f[.]$

$$C = A + iB \quad |C| = \sqrt{A^2 + B^2}$$
$$\phi = \tan^{-1}\left(\frac{B}{A}\right)$$

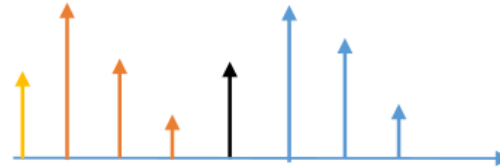
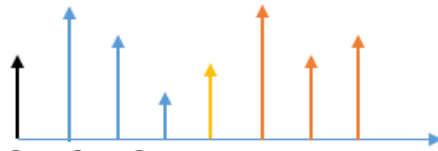
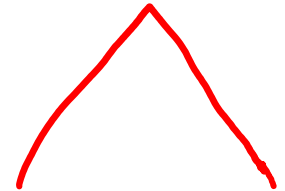
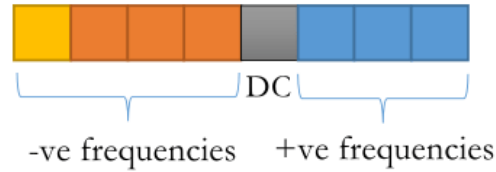
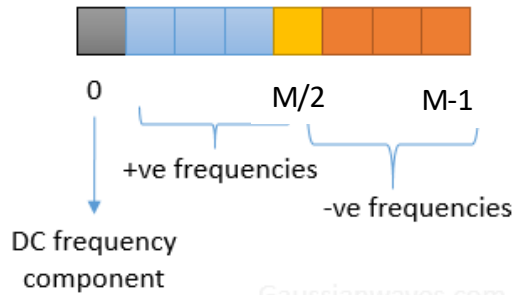
DFT (in practice)

$$F[m] = \sum_{n=0}^{n=(M-1)} \underline{f[n]} e^{\frac{-j2\pi mn}{M}}, m = 0, 1, \dots (M-1)$$



DFT – center shifted (for plotting)

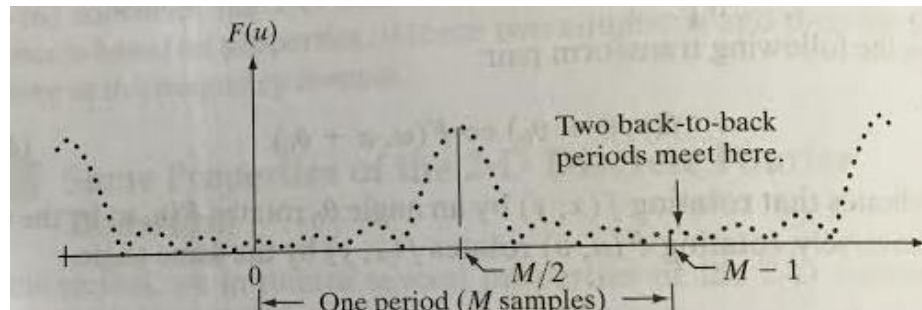
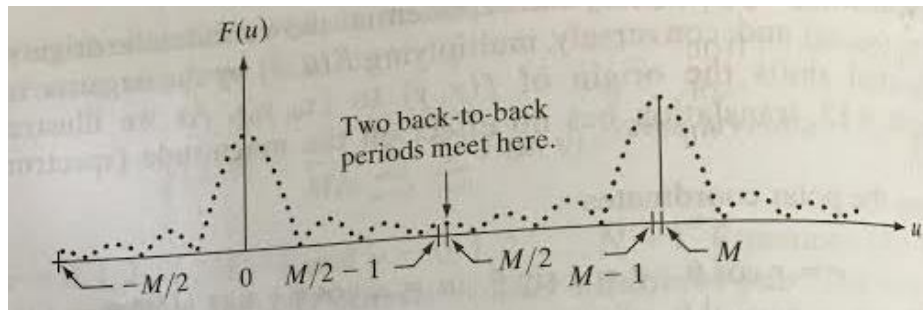
$$F[m] = \sum_{n=0}^{n=(M-1)} f[n] e^{\frac{-j2\pi mn}{M}}, m = 0, 1, \dots (M-1)$$



Gaussianwaves.com

Shifting origin

1-D



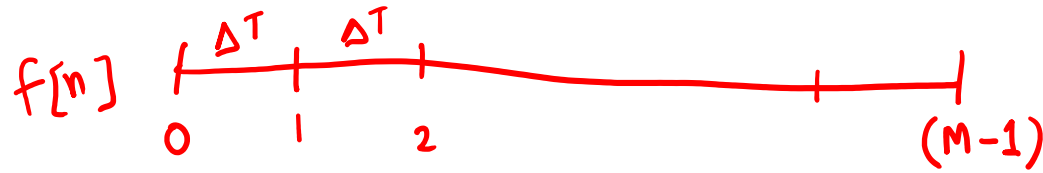
$$F(u) \quad F(u - u_o)$$

$$\underbrace{f[x] e^{\frac{j2\pi u_o x}{M}}}_{\text{red underline}} \leftrightarrow \underbrace{F(u - u_o)}_{\text{red underline}}$$

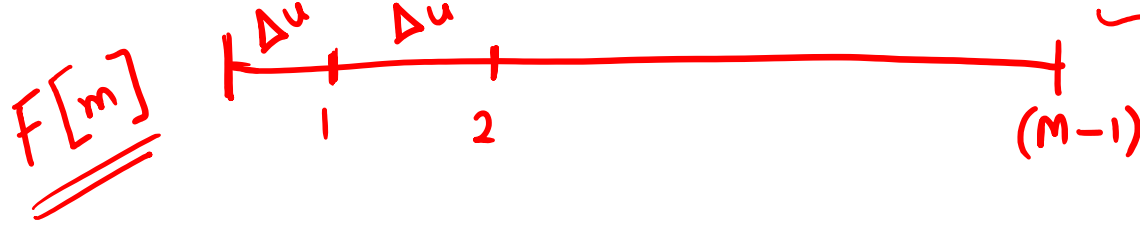
$$\underbrace{f[n](-1)^n}_{\text{red underline}}$$

$$u_o = \frac{M}{2}$$

Relationship between Sampling and Frequency Intervals



$$T = (M-1) \Delta T$$

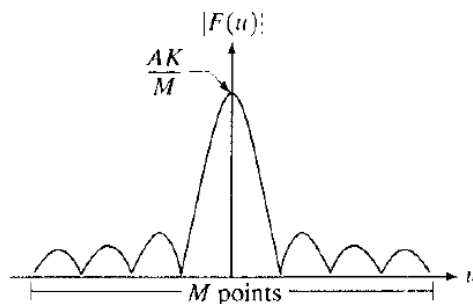
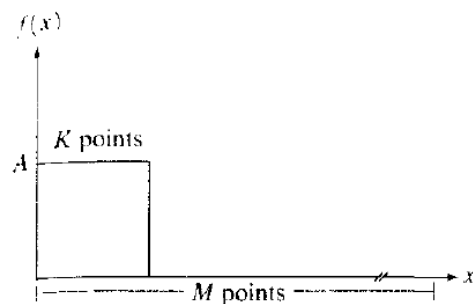


$$\Omega = (M-1) \Delta u = \frac{1}{\Delta T}$$

$$\Delta u = \frac{1}{(M-1) \Delta T} = \frac{1}{T}$$

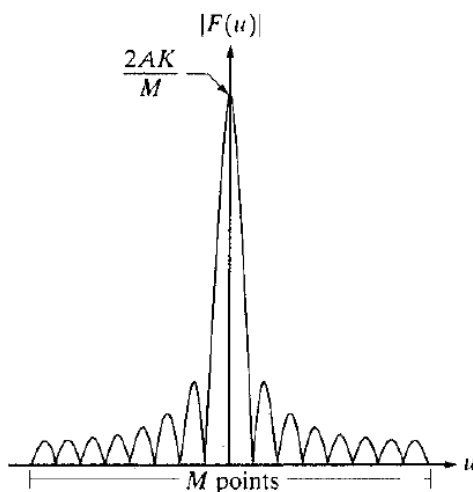
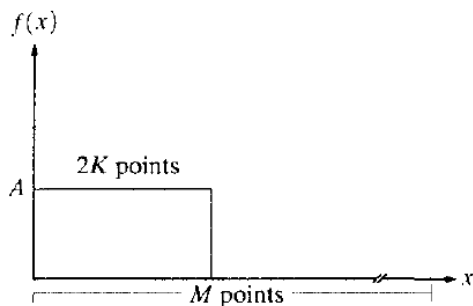
- Ω (Range of frequencies) depends inversely on sampling interval ΔT
- Δu (Frequency Resolution of DFT) depends inversely on duration T over which $f(t)$ is sampled

Relationship between u and x



a b
c d

FIGURE 4.2 (a) A discrete function of M points, and (b) its Fourier spectrum. (c) A discrete function with twice the number of nonzero points, and (d) its Fourier spectrum.



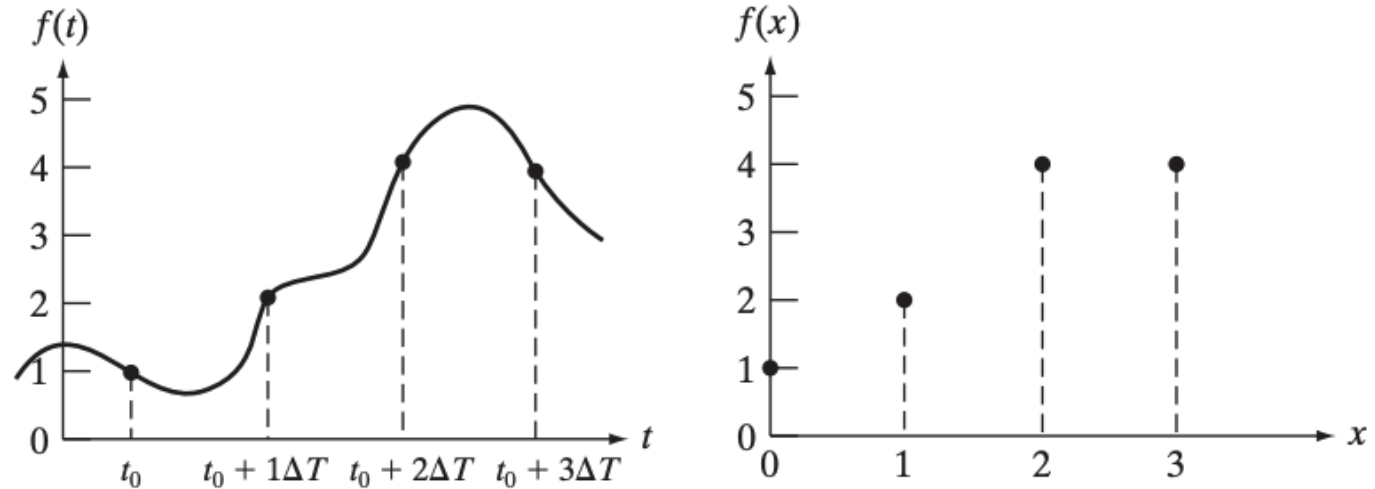
$$\Delta u = \frac{1}{M \Delta x}$$

1-D DFT example

a b

FIGURE 4.11

(a) A function, and (b) samples in the x -domain. In (a), t is a continuous variable; in (b), x represents integer values.



$$F[m] = \sum_{n=0}^{n=(M-1)} f[n] e^{-j \frac{2\pi mn}{M}}, m = 0, 1, \dots, (M-1)$$

$$F[1] = \sum_{n=0}^3 f[n] e^{-j 2\pi n/4} = -3 + 2j$$

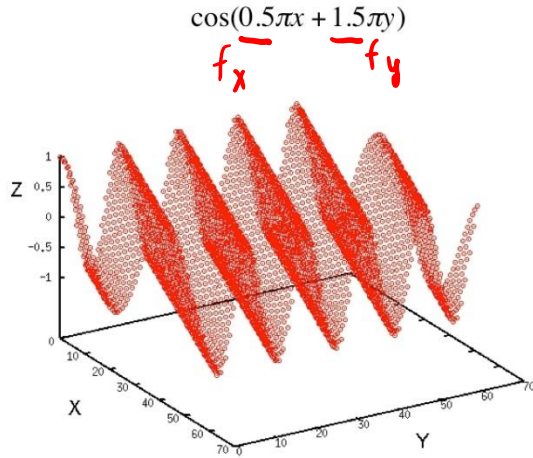
$$F[0] = 1 + 2 + 4 + 4$$

$$\begin{aligned} 0 &\rightarrow e^0 \rightarrow 1 \times 1 \\ 1 &\rightarrow e^{-j 2\pi/4} \rightarrow -j \times 2 \\ 2 &\rightarrow e^{-j 4\pi/4} \rightarrow -1 \times 4 \\ 3 &\rightarrow e^{-j 6\pi/4} \rightarrow +j \times 4 \end{aligned}$$

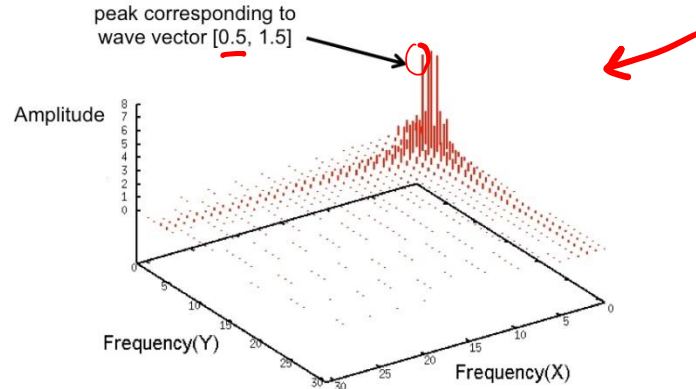
2D DFT and IDFT

$$\underline{F[m,n]} = \sum_{y=0}^{y=(N-1)} \sum_{x=0}^{x=(M-1)} \underline{f[x,y]} \underbrace{e^{-2\pi j\left(\frac{mx}{M} + \frac{ny}{N}\right)}}_{\substack{\downarrow \quad \downarrow \\ f_x \quad f_y}}$$

$$F[3,3] =$$



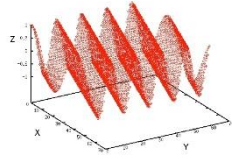
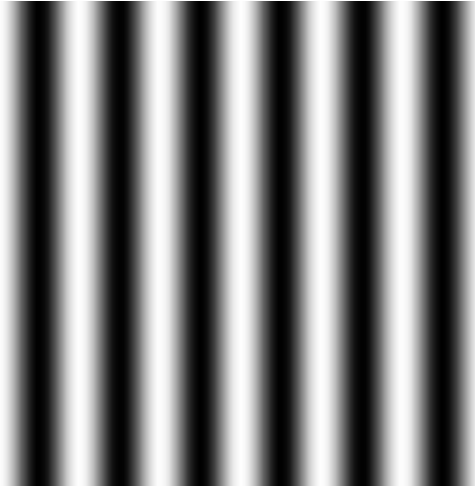
$$\text{FFT}[\cos(0.5\pi x + 1.5\pi y)]$$



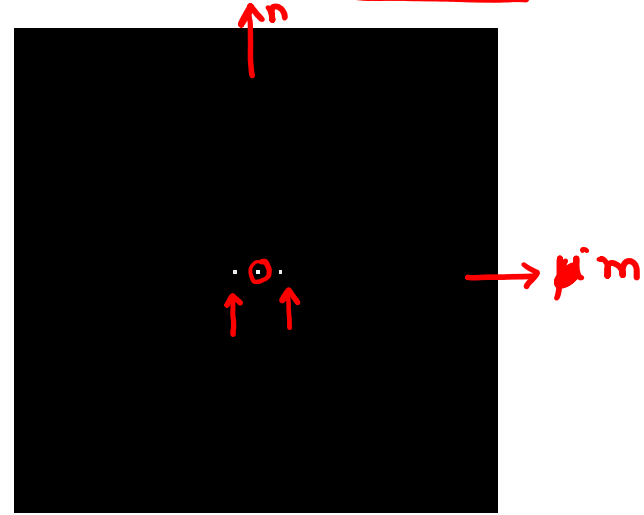
$$\underline{f[x,y]} = \frac{1}{MN} \sum_{n=0}^{n=(N-1)} \sum_{m=0}^{m=(M-1)} \underline{F[m,n]} e^{2\pi j\left(\frac{mx}{M} + \frac{ny}{N}\right)}$$

DFT for simple spatial patterns

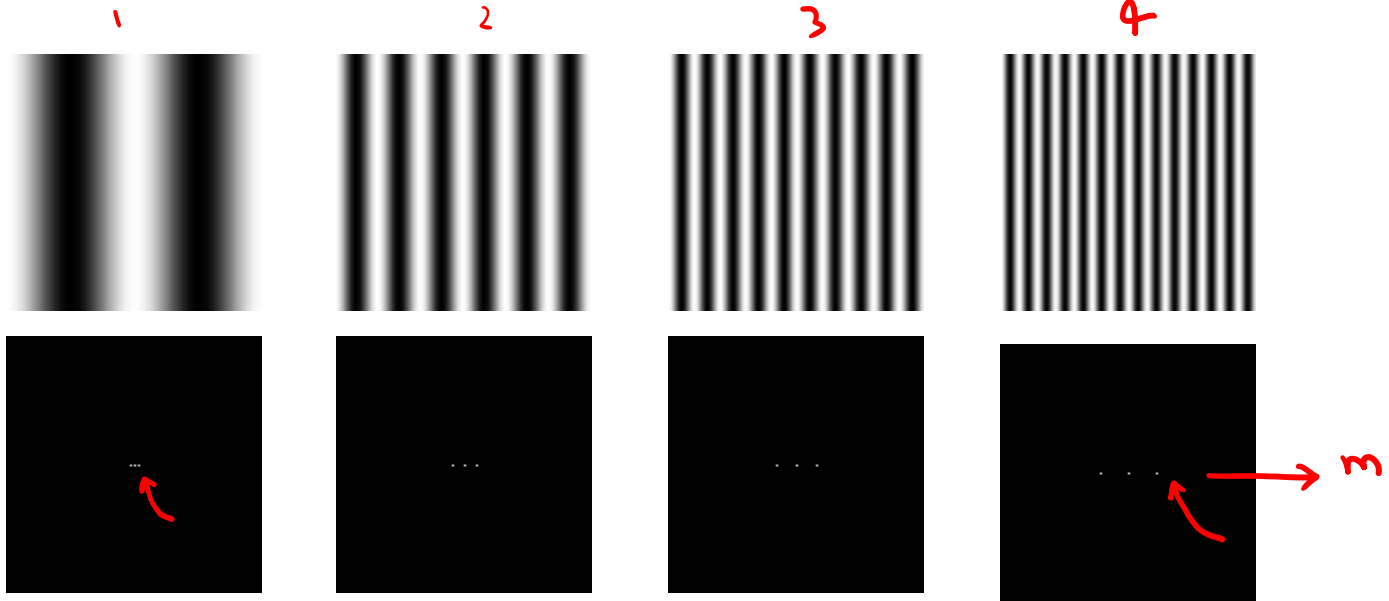
Brightness Image



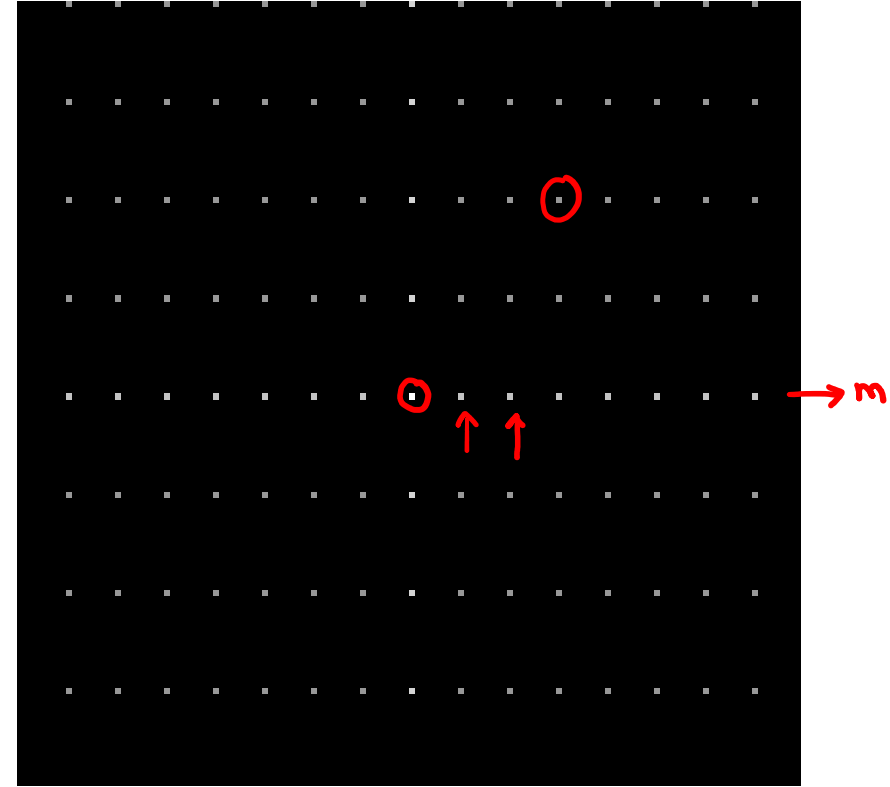
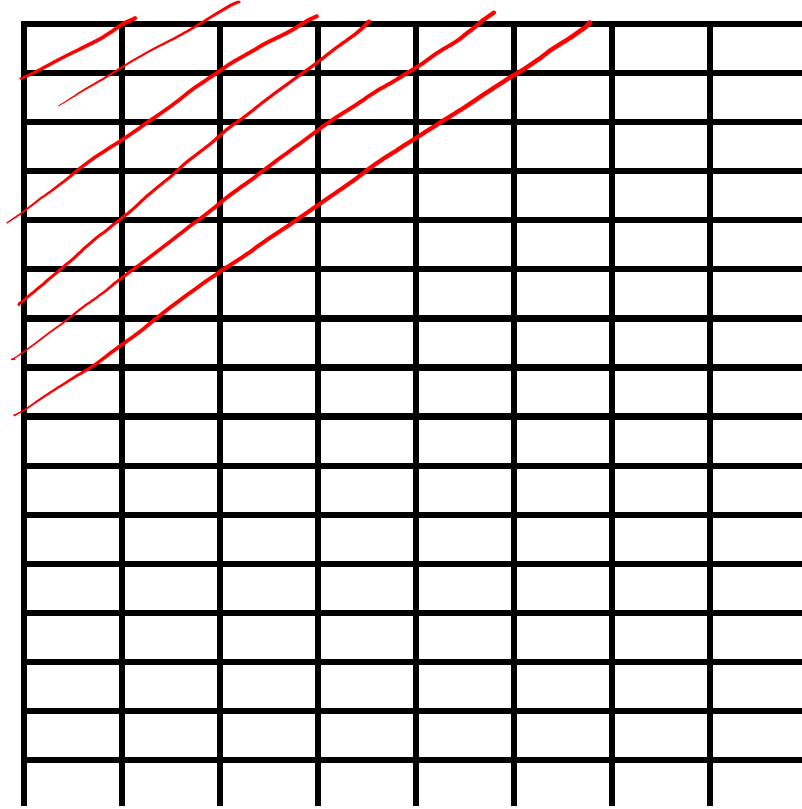
$|F(m,n)|$
Fourier transform spectrum



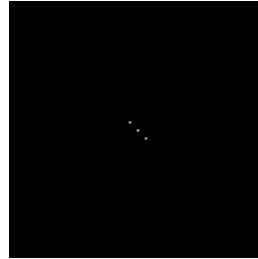
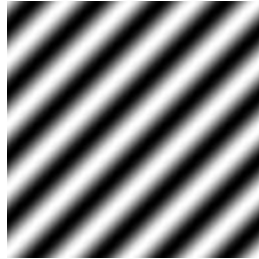
DFT Example



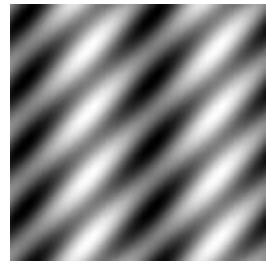
Example



DFT Example (Rotation)



DFT Example (Sum of Signals)



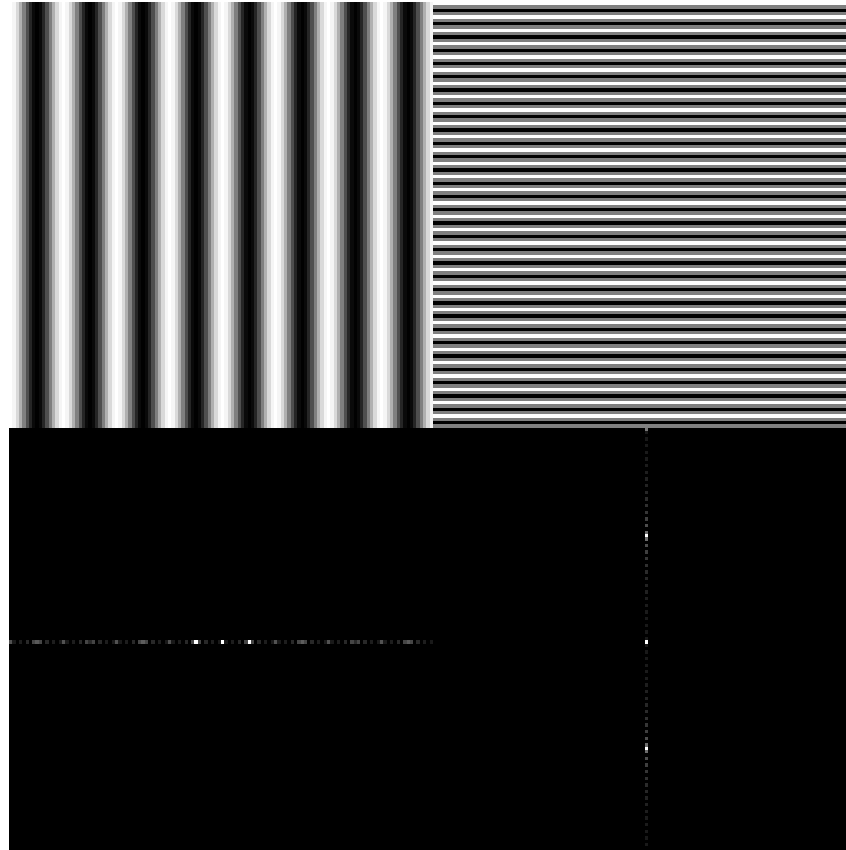
$D + X$



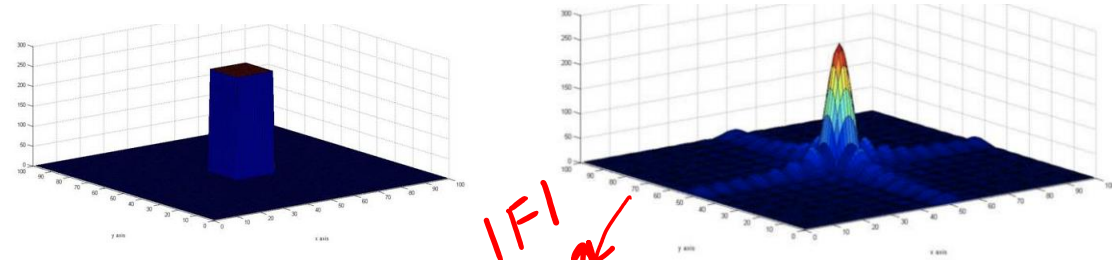
$F(f_1) + F(f_2)$



DFT for simple 'spatial' patterns



DFT Example (Rect, Log Transformation)



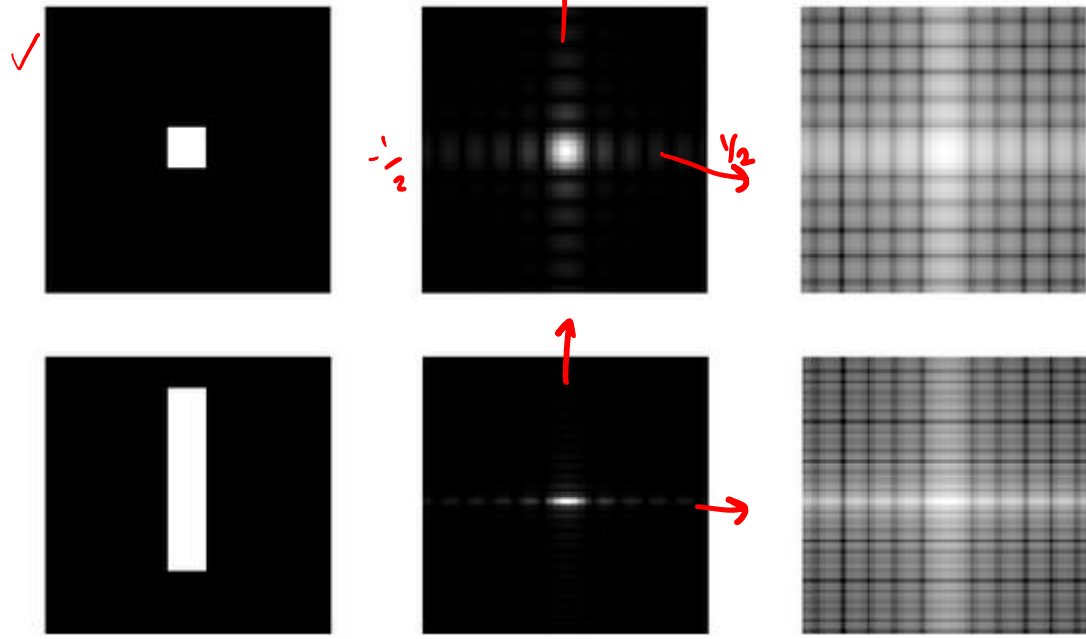
$|F|$

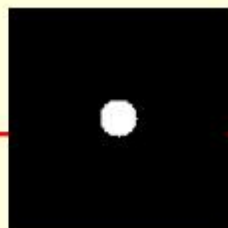
$$\frac{10^6 \text{ to } 10^{-2} : V}{\log(1+V)}$$

$$\log(1+V)$$

m, n

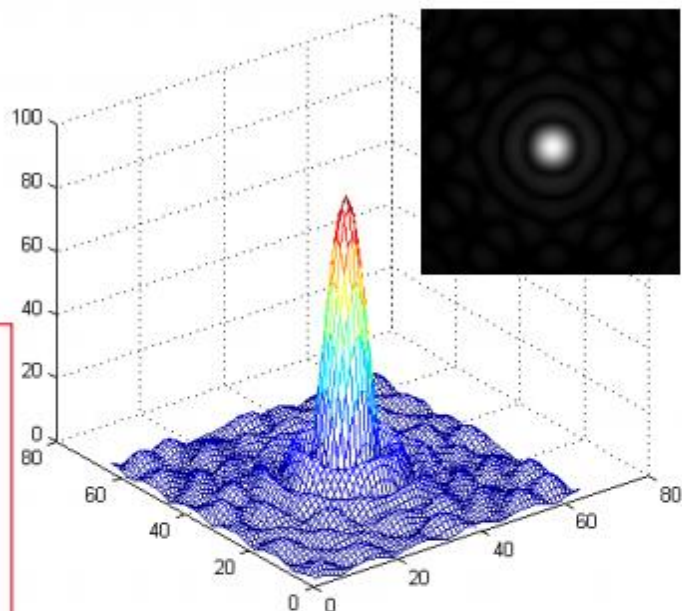
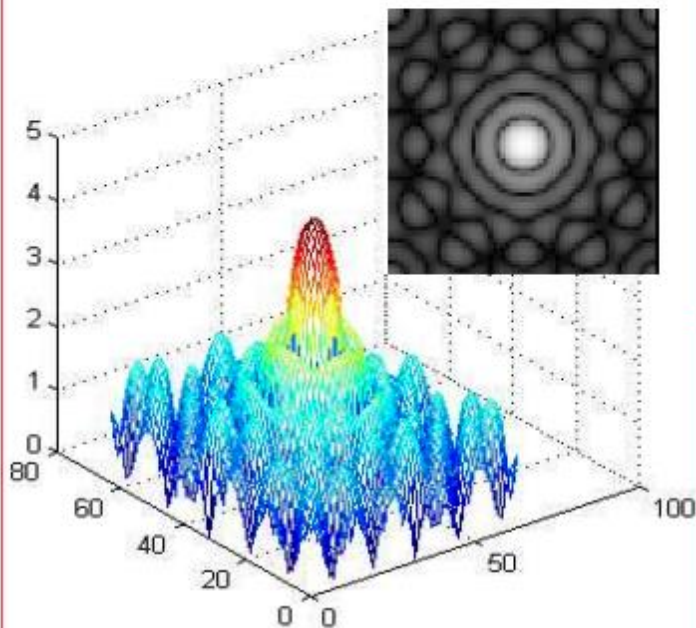
$$\begin{matrix} 0 \cdots m-1 \\ 0 \cdots n-1 \end{matrix}$$





$f(x,y)$

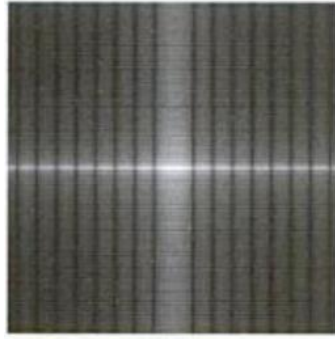
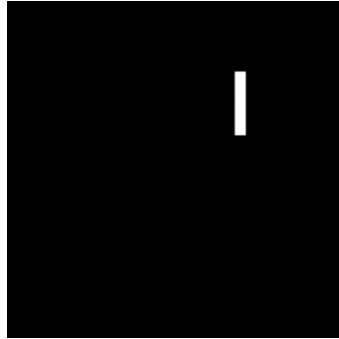
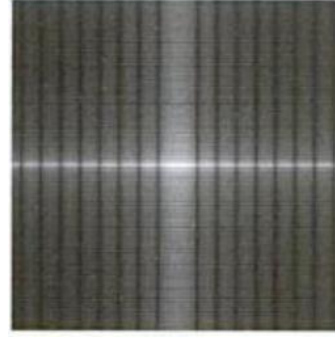
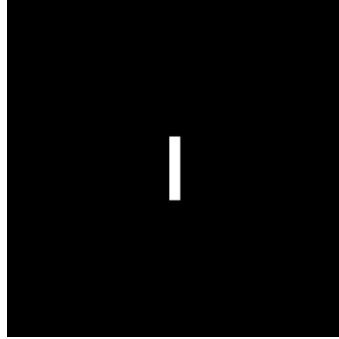
(64x64)



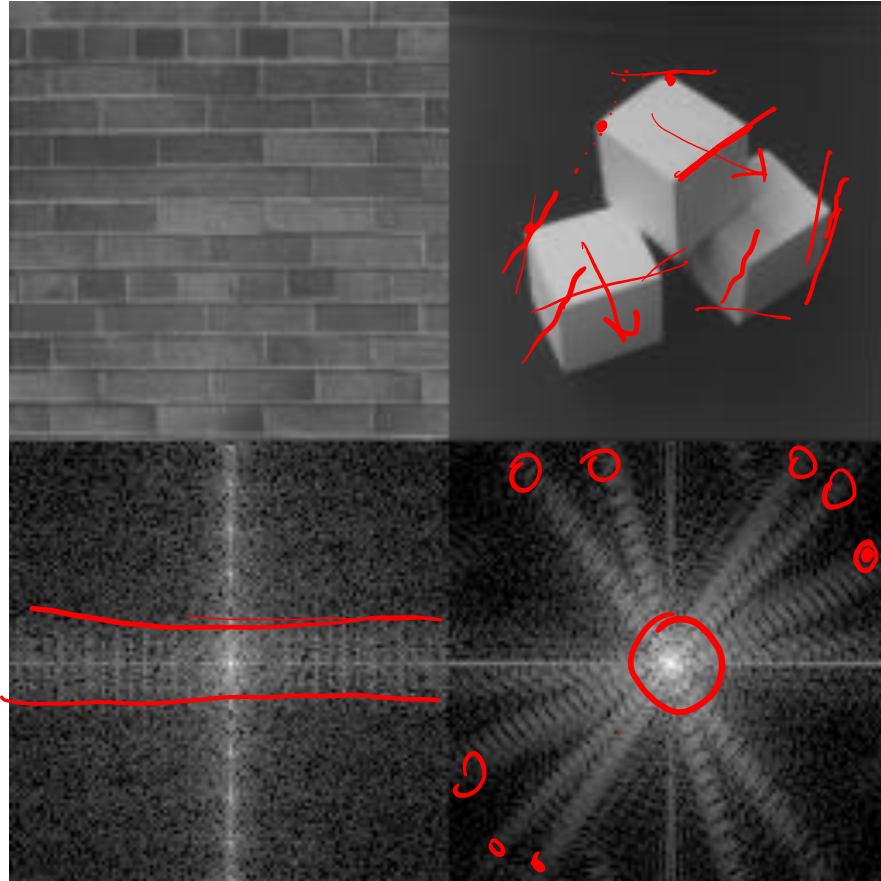
$|F(u,v)|$

$\log(1+|F(u,v)|)$

DFT Example (Translation - Magnitude)



Some examples of images and spectra

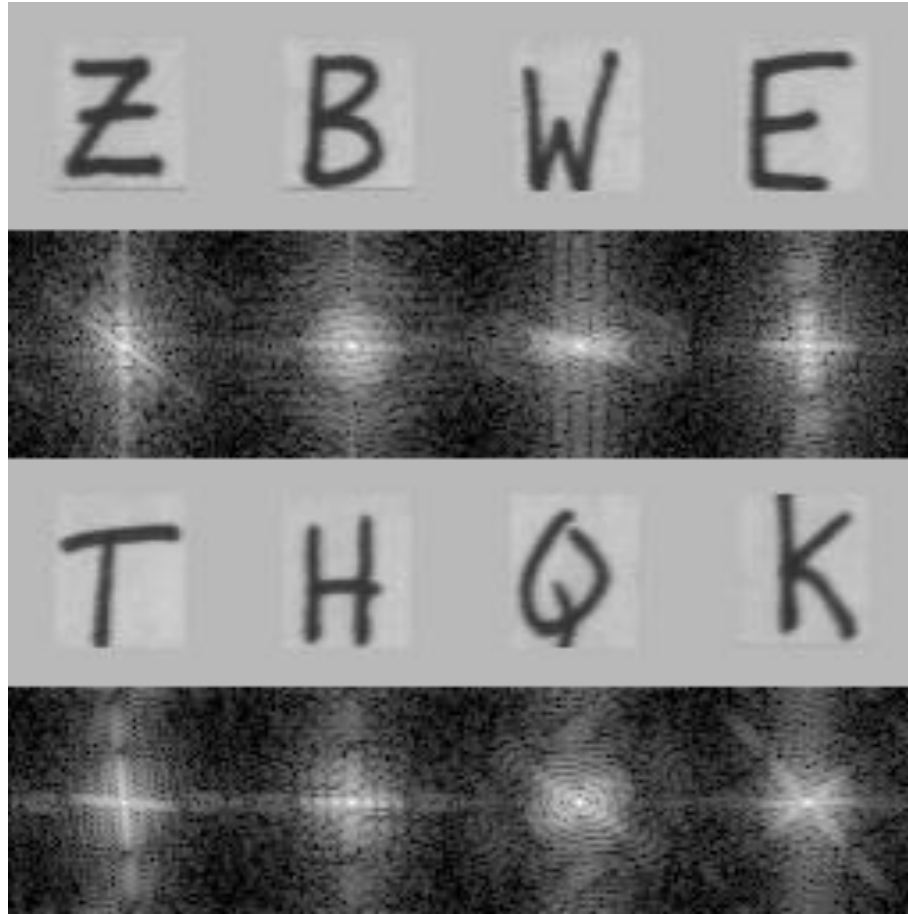


Handwritten red annotations and a formula:

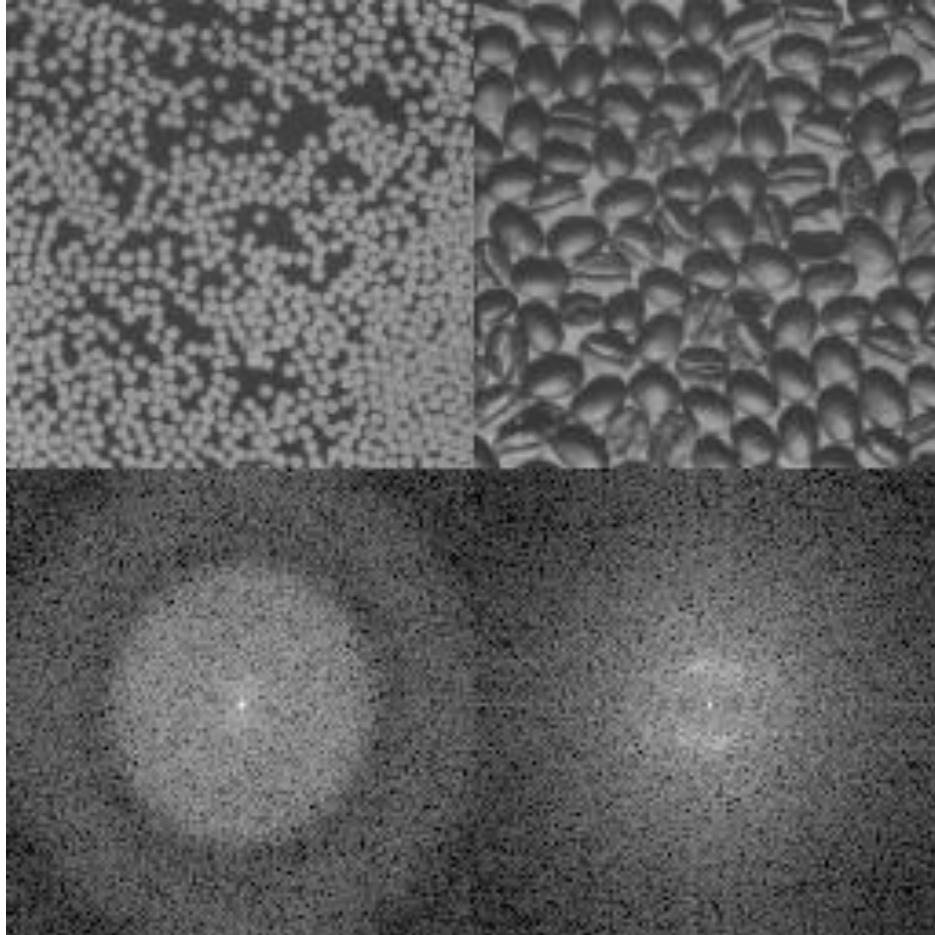
$$F[m,n] = \frac{1}{f} \int \int f(x,y) e^{-j2\pi(mx+ny)} dx dy$$

The formula is written in red ink, with the original formula crossed out and replaced with the simplified version. Above the formula, there is a red wavy arrow pointing right and a red curly brace.

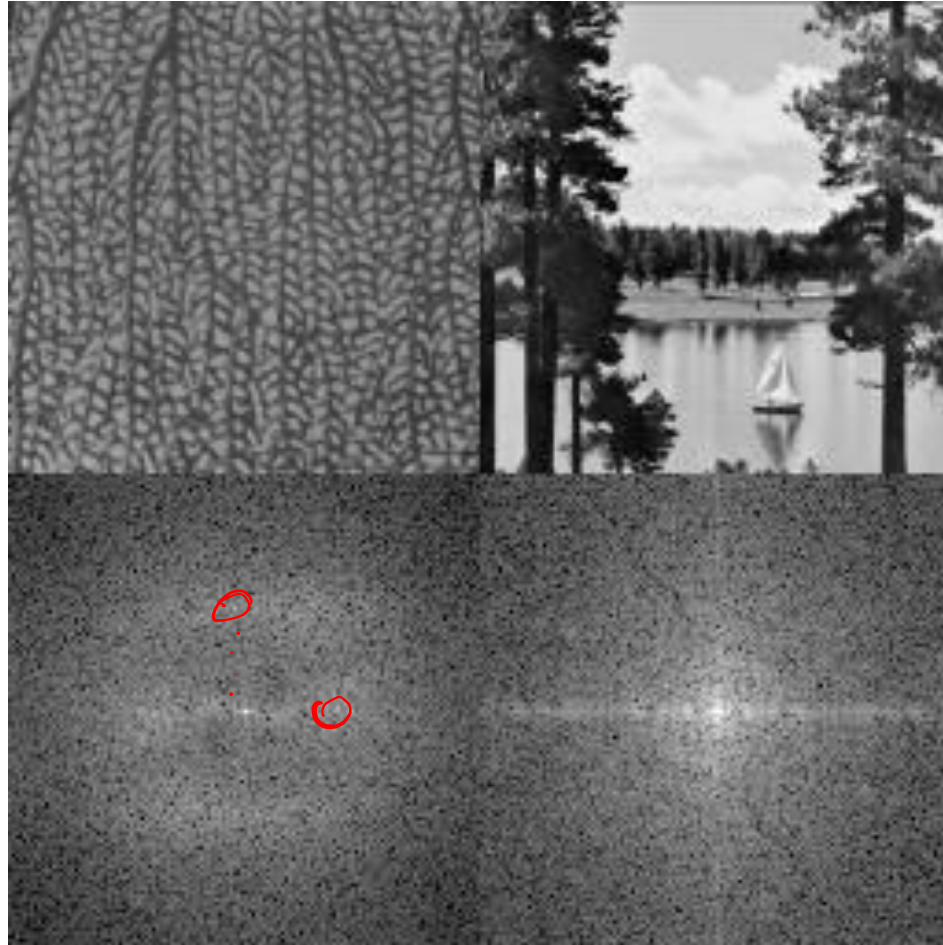
Some examples of images and spectra



Some examples of images and spectra



Some examples of images and spectra



Important Terms

- Magnitude spectrum

$$|F(\omega)| = \left[R^2(\omega) + I^2(\omega) \right]^{1/2}$$

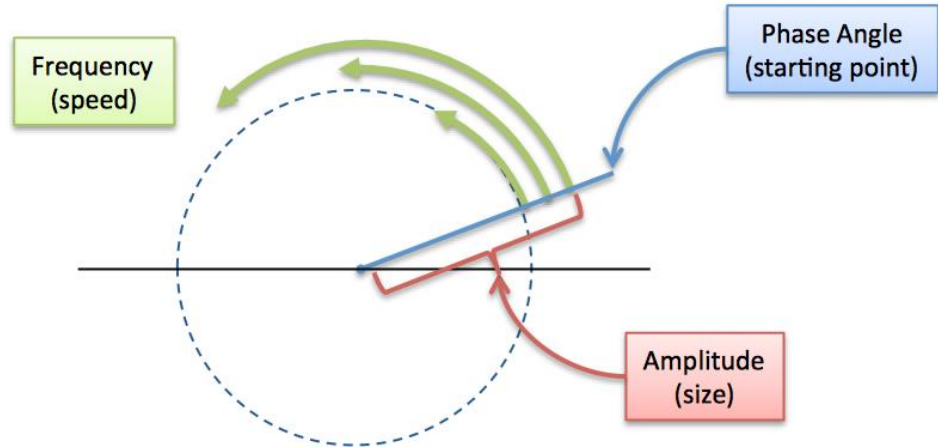
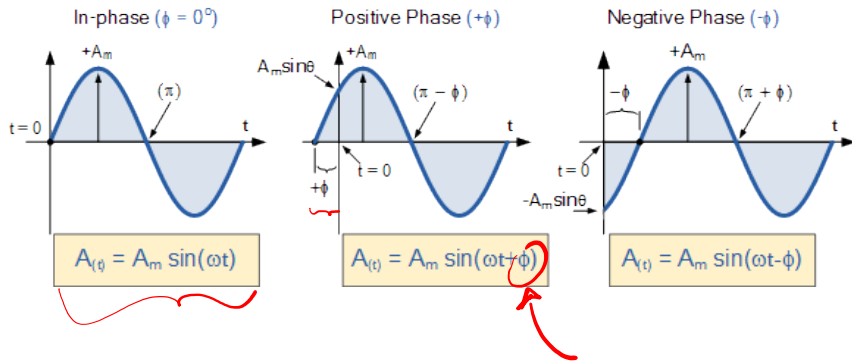
- Phase Spectrum

$$\phi(\omega) = \tan^{-1} \left[\frac{I(\omega)}{R(\omega)} \right]$$

- Power Spectrum

$$P(\omega) = |F(\omega)|^2$$

Phase



Magnitude and Phase Spectra



Figure 4a
Original

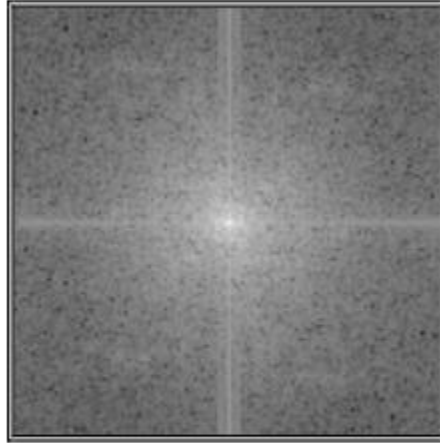


Figure 4b
 $\log(|A(\Omega, \Psi)|)$

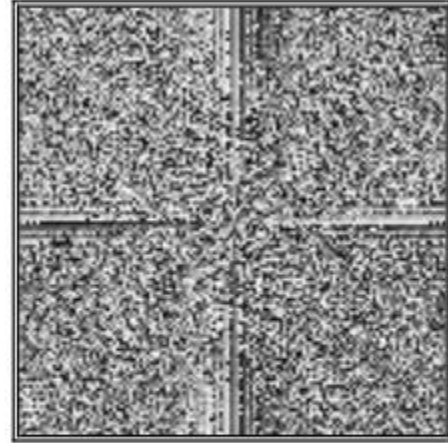


Figure 4c
 $\phi(\Omega, \Psi)$

$$\tan^{-1}\left(\frac{\text{Im}}{\text{Re}}\right)$$

We generally do not display PHASE images because most people who see them shortly thereafter succumb to hallucinogenics or end up in a Tibetan monastery – John Brayer

Magnitude and Phase Spectra

Both matter for reconstruction



Figure 4a
Original

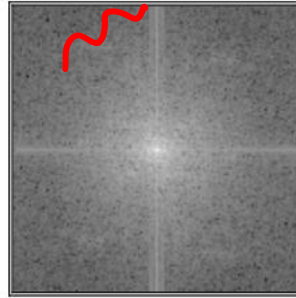


Figure 4b
 $\log(|A(\Omega, \Psi)|)$

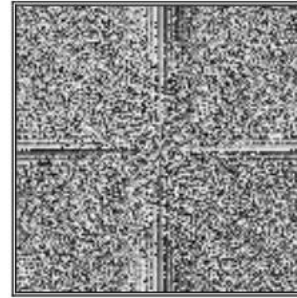


Figure 4c
 $\phi(\Omega, \Psi)$

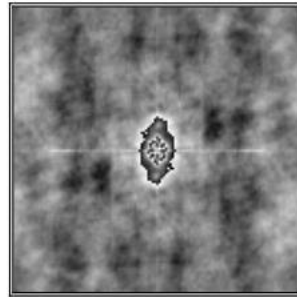
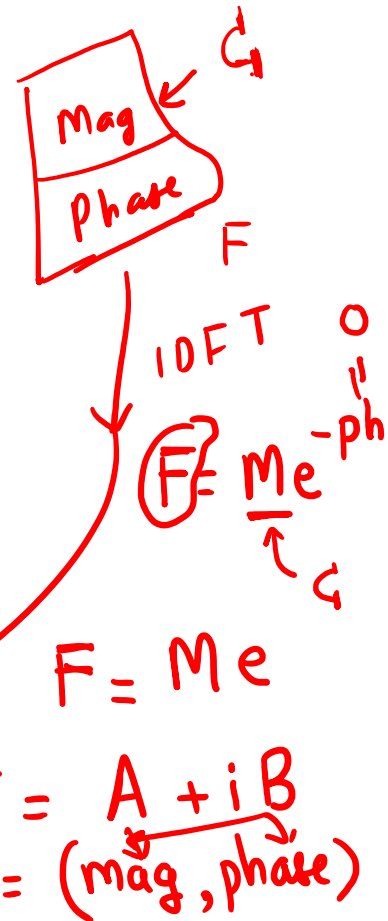
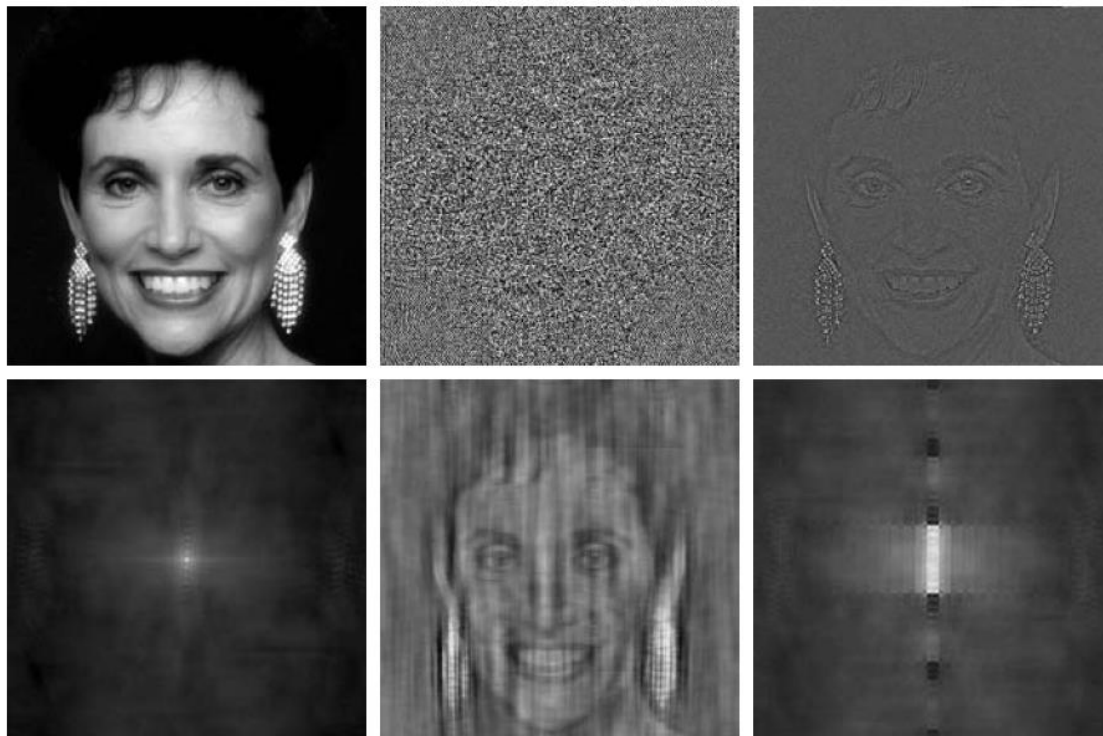


Figure 5a
 $\phi(\Omega, \Psi) = 0$



Figure 5b
 $|A(\Omega, \Psi)| = \text{constant}$



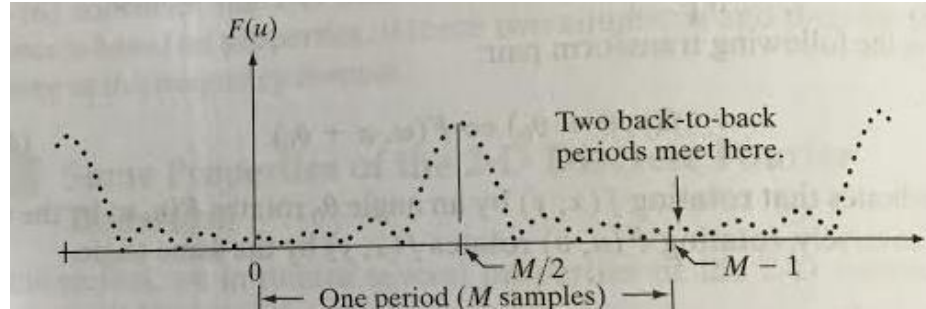
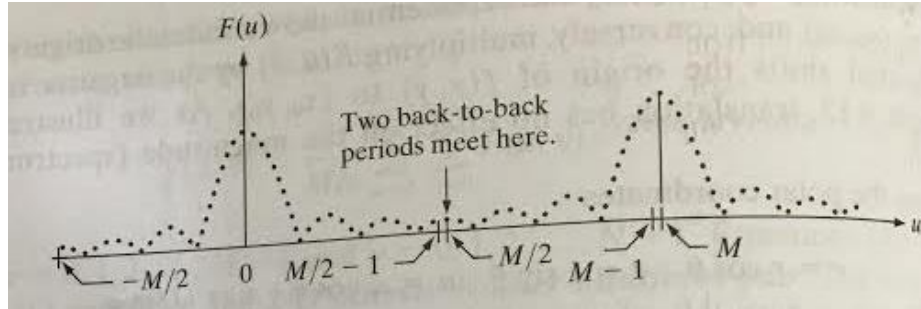


| | | |
|---|---|---|
| a | b | c |
| d | e | f |

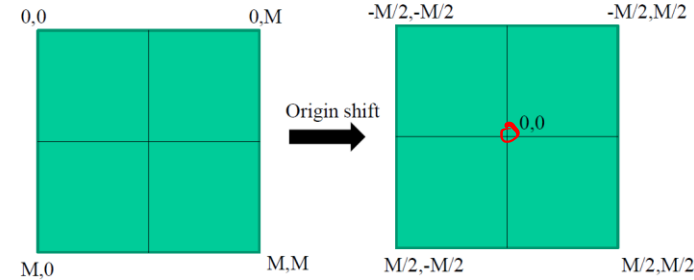
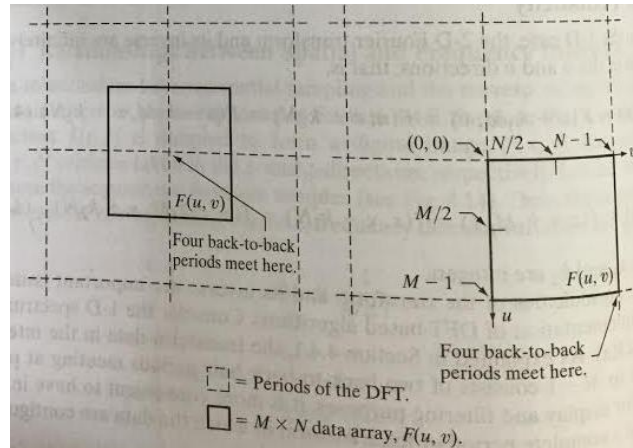
FIGURE 4.27 (a) Woman. (b) Phase angle. (c) Woman reconstructed using only the phase angle. (d) Woman reconstructed using only the spectrum. (e) Reconstruction using the phase angle corresponding to the woman and the spectrum corresponding to the rectangle in Fig. 4.24(a). (f) Reconstruction using the phase of the rectangle and the spectrum of the woman.

Shifting origin

1-D

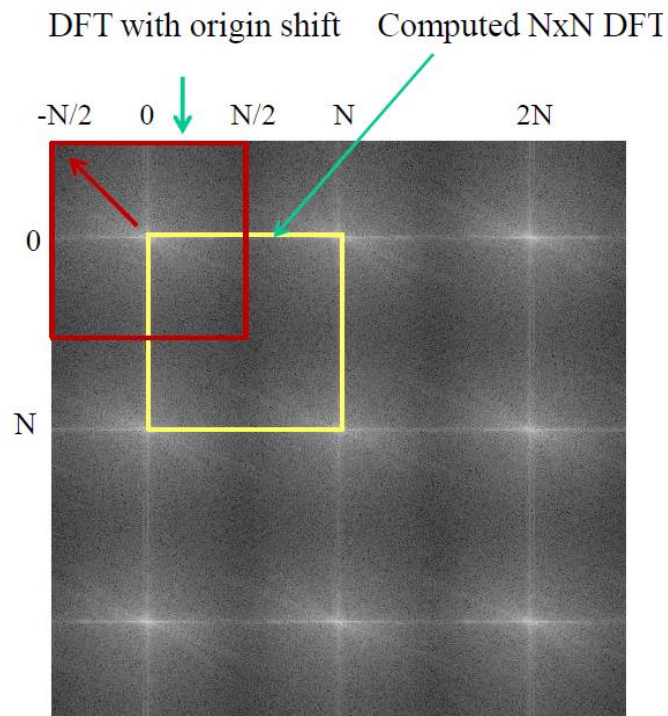
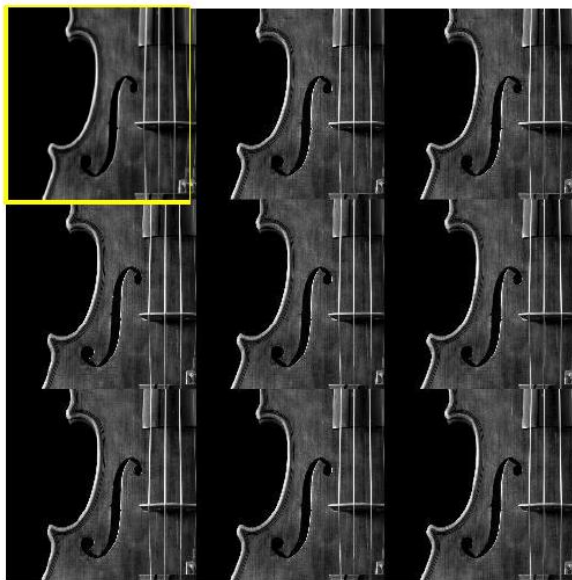


2-D

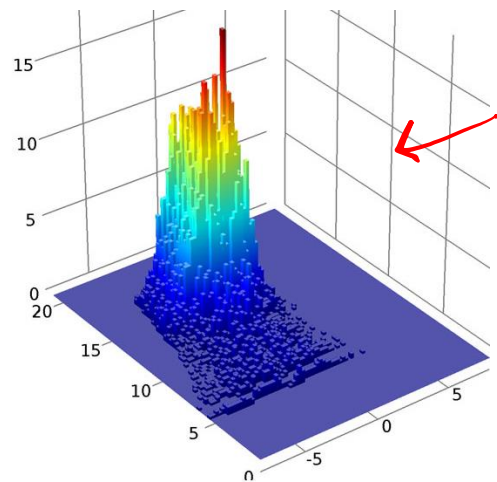


$$\underline{f[x, y]e^{j2\pi(\frac{u_0x}{M} + \frac{v_0y}{M})}} \leftrightarrow F(u - u_0, v - v_0)$$

Shifting origin



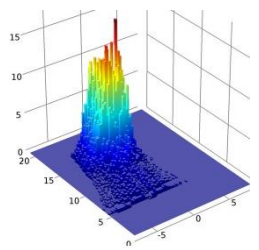
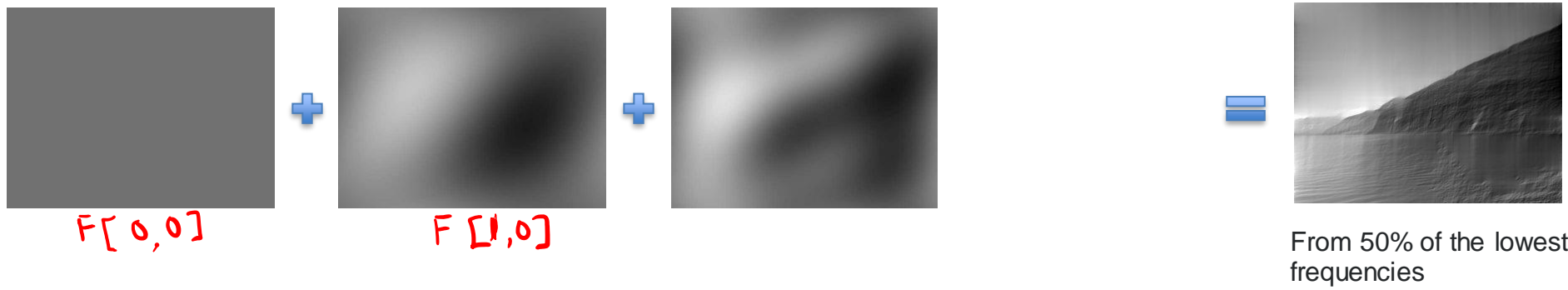




$$F[m,n] = \sum_{y=0}^{y=(N-1)} \sum_{x=0}^{x=(M-1)} f[x,y] e^{-2\pi j \left(\frac{mx}{M} + \frac{ny}{N} \right)}$$

$$f[x,y] = \frac{1}{MN} \sum_{n=0}^{n=(N-1)} \sum_{m=0}^{m=(M-1)} F[m,n] e^{2\pi j \left(\frac{mx}{M} + \frac{ny}{N} \right)}$$

$$f[x,y] = \frac{1}{MN} \sum_{n=0}^{n=(N-1)} \sum_{m=0}^{m=(M-1)} F[m,n] e^{2\pi j \left(\frac{mx}{M} + \frac{ny}{N} \right)}$$



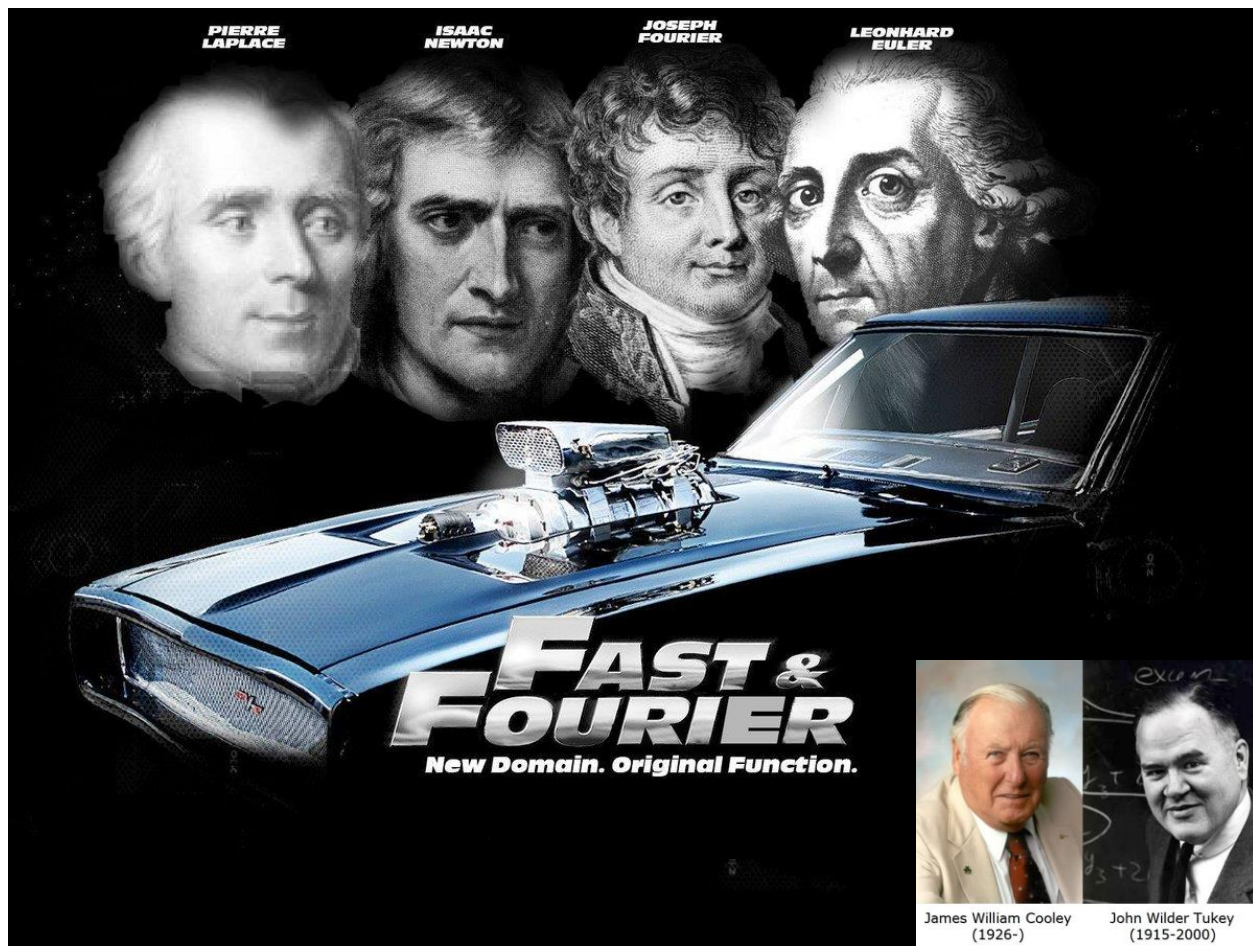
Adding up to 50% lowest frequencies



(Some) Properties of FT

| | Name: | Condition: | Property: |
|----|--------------------|---|---|
| 1 | Amplitude scaling | $f(t) \leftrightarrow F(\omega)$, constant K | $Kf(t) \leftrightarrow KF(\omega)$ |
| 2 | Addition | $f(t) \leftrightarrow F(\omega)$, $g(t) \leftrightarrow G(\omega)$, \dots | $f(t) + g(t) + \dots \leftrightarrow F(\omega) + G(\omega) + \dots$ |
| 3 | Hermitian | Real $f(t) \leftrightarrow F(\omega)$ | $F(-\omega) = F^*(\omega)$ |
| 4 | Even | Real and even $f(t)$ | Real and even $F(\omega)$ |
| 5 | Odd | Real and odd $f(t)$ | Imaginary and odd $F(\omega)$ |
| 6 | Symmetry | $f(t) \leftrightarrow F(\omega)$ | $F(t) \leftrightarrow 2\pi f(-\omega)$ |
| 7 | Time scaling | $f(t) \leftrightarrow F(\omega)$, real s | $f(st) \leftrightarrow \frac{1}{ s } F(\frac{\omega}{s})$ |
| 8 | Time shift | $f(t) \leftrightarrow F(\omega)$ | $f(t - t_o) \leftrightarrow F(\omega)e^{-j\omega t_o}$ |
| 9 | Frequency shift | $f(t) \leftrightarrow F(\omega)$ | $f(t)e^{j\omega_o t} \leftrightarrow F(\omega - \omega_o)$ |
| 10 | Modulation | $f(t) \leftrightarrow F(\omega)$ | $f(t) \cos(\omega_o t) \leftrightarrow \frac{1}{2}F(\omega - \omega_o) + \frac{1}{2}F(\omega + \omega_o)$ |
| 11 | Time derivative | Differentiable $f(t) \leftrightarrow F(\omega)$ | $\frac{df}{dt} \leftrightarrow j\omega F(\omega)$ |
| 12 | Freq derivative | $f(t) \leftrightarrow F(\omega)$ | $-jtf(t) \leftrightarrow \frac{d}{d\omega} F(\omega)$ |
| 13 | Time convolution | $f(t) \leftrightarrow F(\omega)$, $g(t) \leftrightarrow G(\omega)$ | $f(t) * g(t) \leftrightarrow F(\omega)G(\omega)$ |
| 14 | Freq convolution | $f(t) \leftrightarrow F(\omega)$, $g(t) \leftrightarrow G(\omega)$ | $f(t)g(t) \leftrightarrow \frac{1}{2\pi} F(\omega) * G(\omega)$ |
| 15 | Compact form | Real $f(t)$ | $f(t) = \frac{1}{2\pi} \int_0^\infty 2 F(\omega) \cos(\omega t + \angle F(\omega)) d\omega$ |
| 16 | Parseval, Energy W | $f(t) \leftrightarrow F(\omega)$ | $W \equiv \int_{-\infty}^\infty f(t) ^2 dt = \frac{1}{2\pi} \int_{-\infty}^\infty F(\omega) ^2 d\omega$ |

- DFT has an efficient version called FFT (Fast Fourier Transform)



DFT vs FFT computation times

$$F[m] = \sum_{n=0}^{n=(M-1)} f[n] e^{\frac{-j2\pi mn}{M}}, m = 0, 1, \dots (M-1)$$

$O(N^2)$

| n | $N = 2^n$ | N^2 | $N \log N$ |
|-----|-----------|---------------|------------|
| 10 | 1 024 | 1 048 576 | 10 240 |
| 12 | 4 096 | 16 777 216 | 49 152 |
| 14 | 16 384 | 268 435 456 | 229 376 |
| 16 | 65 536 | 4 294 967 296 | 1 048 576 |

FFT(n , [a_0, a_1, \dots, a_{n-1}]):

if $n=1$: return a_0

$F_{\text{even}} = \text{FFT}(n/2, [a_0, a_2, \dots, a_{n-2}])$

$F_{\text{odd}} = \text{FFT}(n/2, [a_1, a_3, \dots, a_{n-1}])$

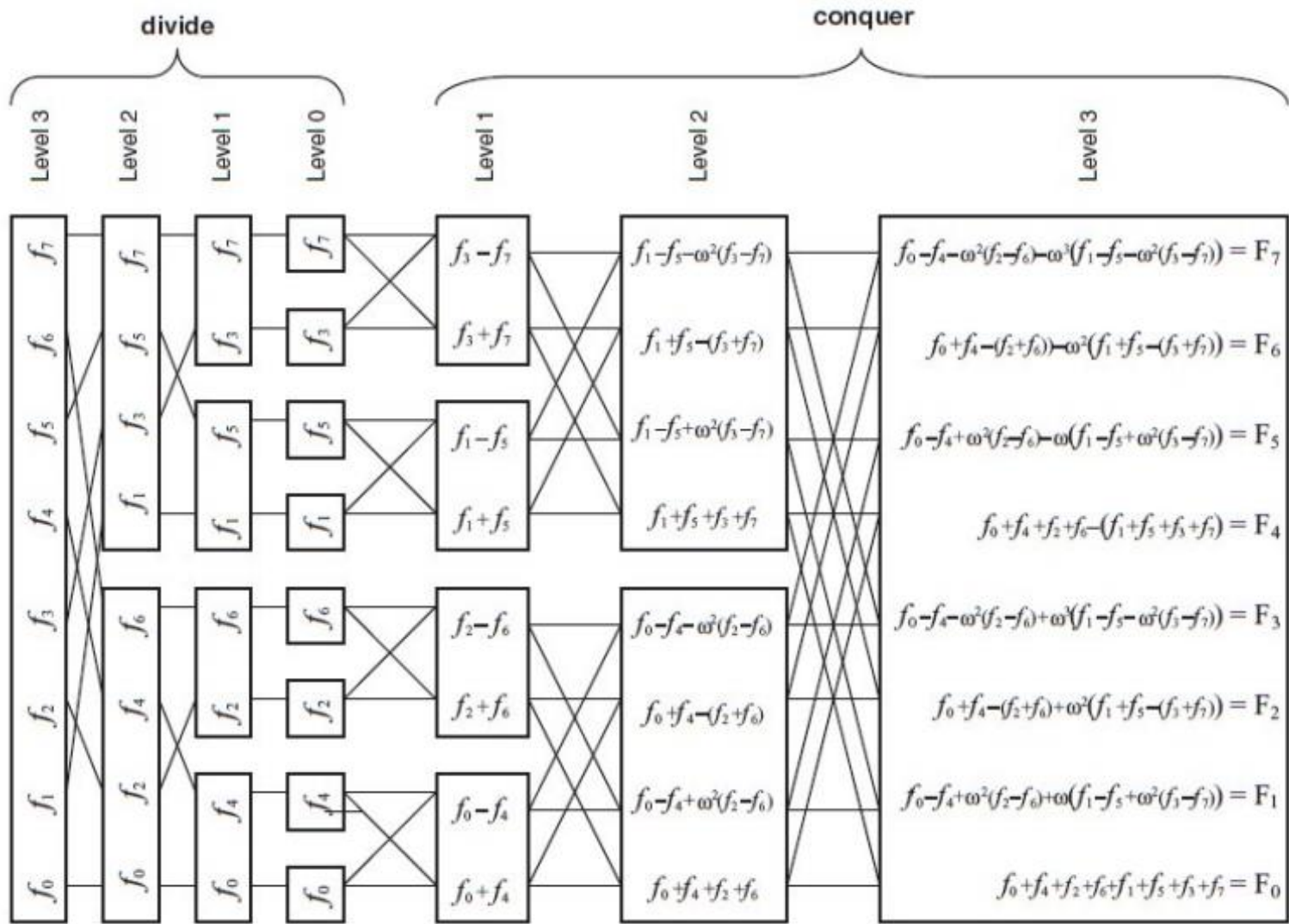
for $k = 0$ to $n/2 - 1$:

$\omega^k = e^{2\pi i k/n}$

$y^k = F_{\text{even } k} + \omega^k F_{\text{odd } k}$

$y^{k+n/2} = F_{\text{even } k} - \omega^k F_{\text{odd } k}$

return [y_0, y_1, \dots, y_{n-1}]



References

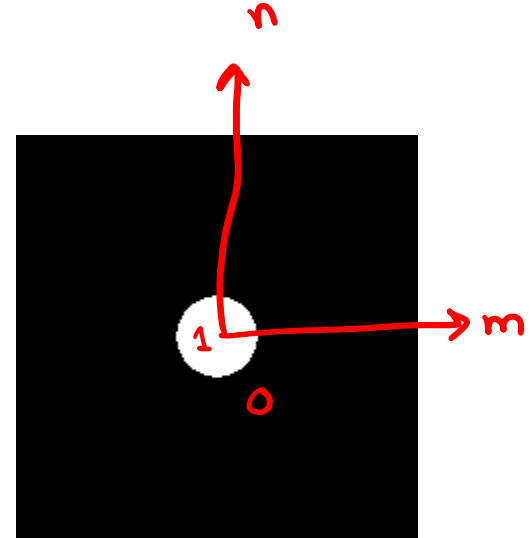
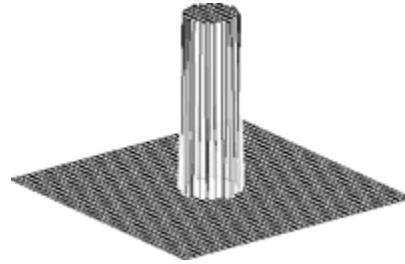
- <http://cns-alumni.bu.edu/~slehar/fourier/fourier.html>
- <https://slideplayer.com/slide/5665338/>
- <https://2e.mindsmachine.com/asf07.02.html>
- <https://radiologykey.com/a-walk-through-the-spatial-frequency-domain/>
- <https://blogs.mathworks.com/steve/2009/12/04/fourier-transform-visualization-using-windowing/>
- <https://photo.stackexchange.com/questions/40401/what-does-frequency-mean-in-an-image>
- <http://homepages.inf.ed.ac.uk/rbf/HIPR2/fourier.htm>
- <http://paulbourke.net/miscellaneous/imagefilter>
- <https://www.cs.unm.edu/~brayer/vision/fourier.html>

Image Enhancement and Filtering in Frequency Domain

Ideal Low Pass Filters

$$I \longrightarrow \underbrace{F \cdot H}_{\text{Filter}} \longrightarrow I$$

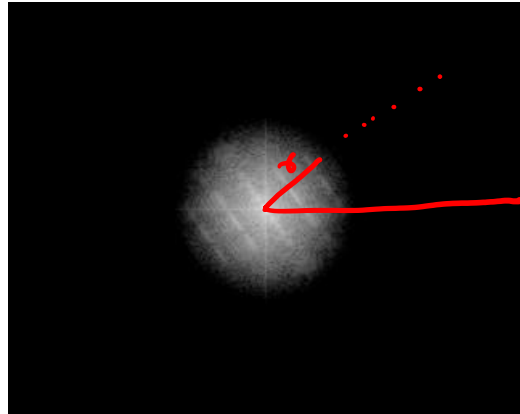
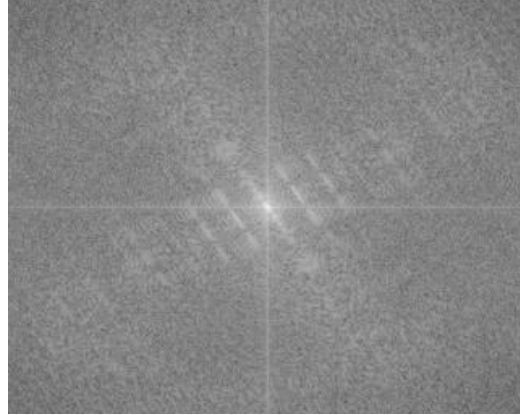
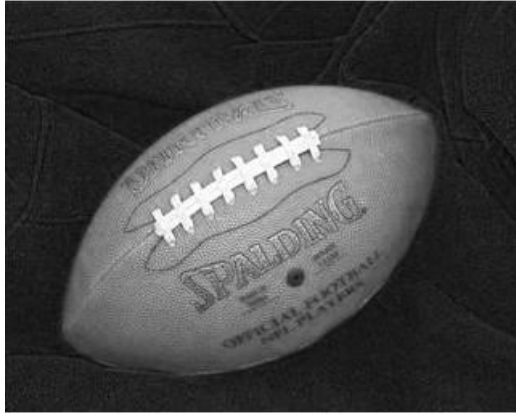
$$H(u, v) = \begin{cases} 1 & \text{if } D(u, v) \leq D_0 \\ 0 & \text{if } D(u, v) > D_0 \end{cases}$$



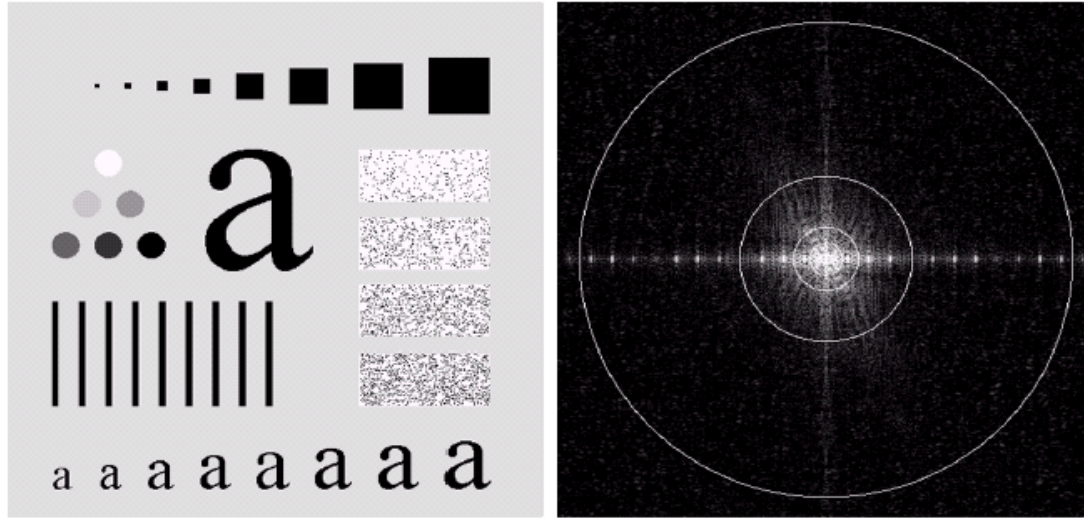
where $D(u, v) = [(u - M / 2)^2 + (v - N / 2)^2]^{1/2}$

$D_0 \rightarrow$ cut off frequency

Ideal Low Pass Filters

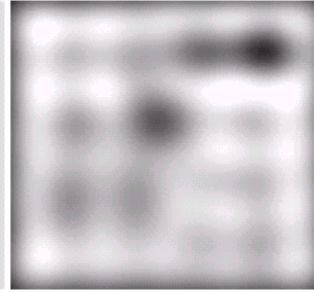
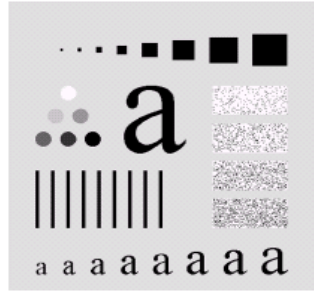


Ideal Low Pass Filters



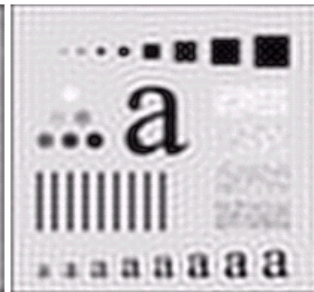
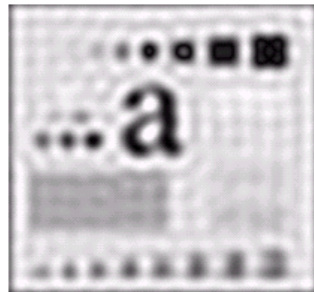
Radii 10,30,60,160 and 460 \rightarrow power 87, 93.1, 95.7, 97.8 and 99..2

Ideal Low Pass Filters



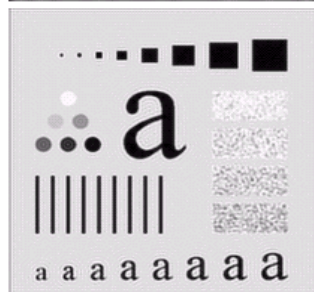
ILPF radius 10

ILPF radius 30



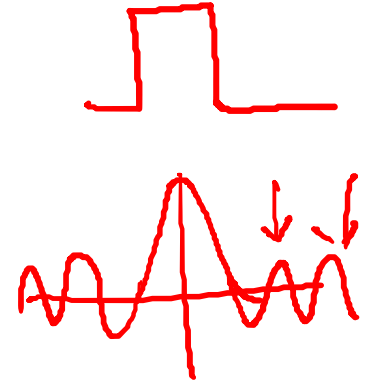
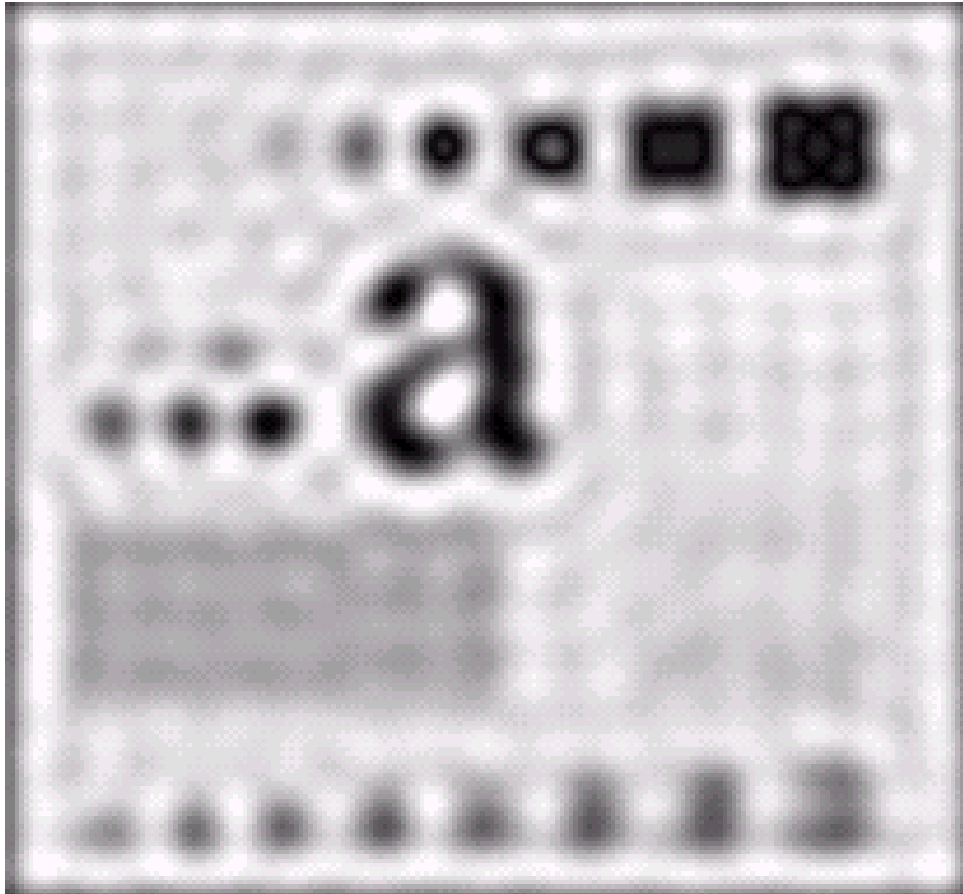
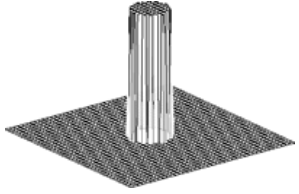
ILPF radius 60

ILPF radius 160



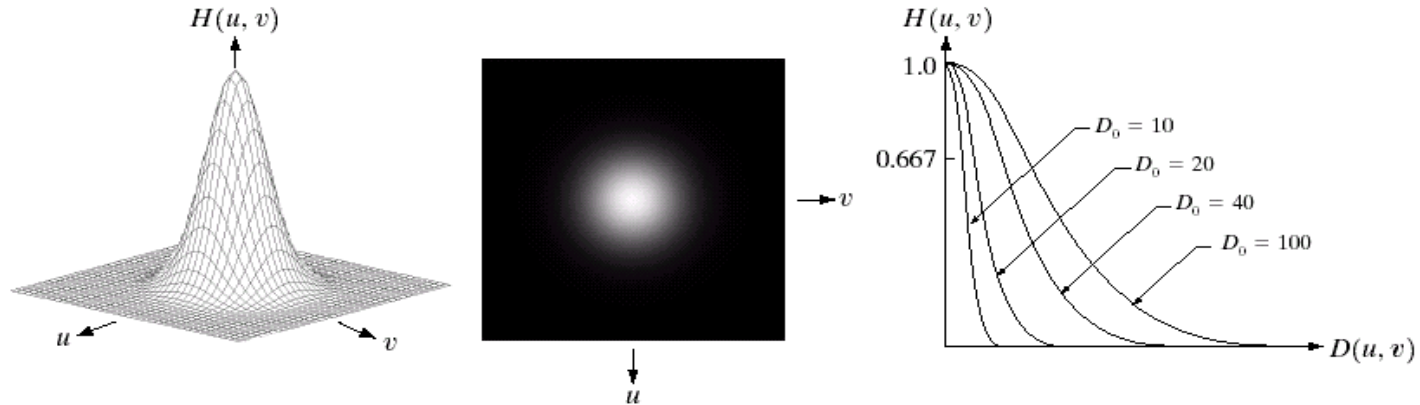
ILPF radius 460

Ideal Low Pass Filters



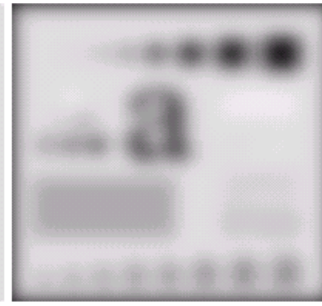
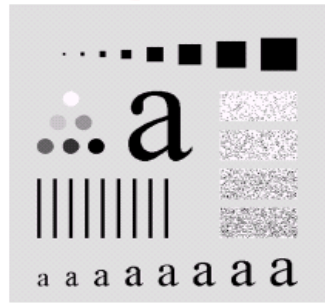
ILPF radius 30

Gaussian Low Pass Filters

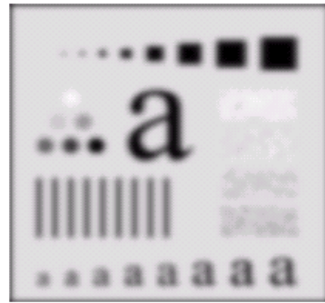


$$\underline{H(u, v) = e^{-D^2(u, v) / 2D_0^2}}$$

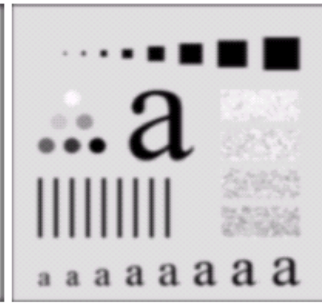
Gaussian Low Pass Filters (GLPF)



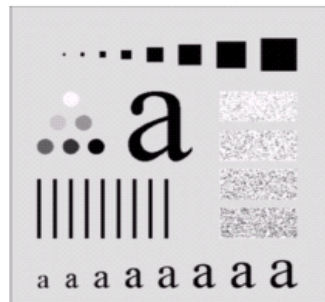
GLPF cut off
frequency 10



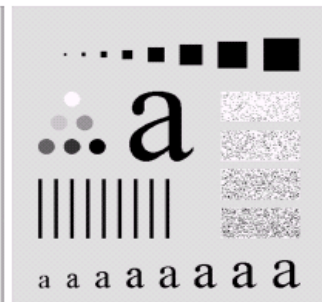
GLPF cut off
frequency 30



GLPF cut off
frequency 60

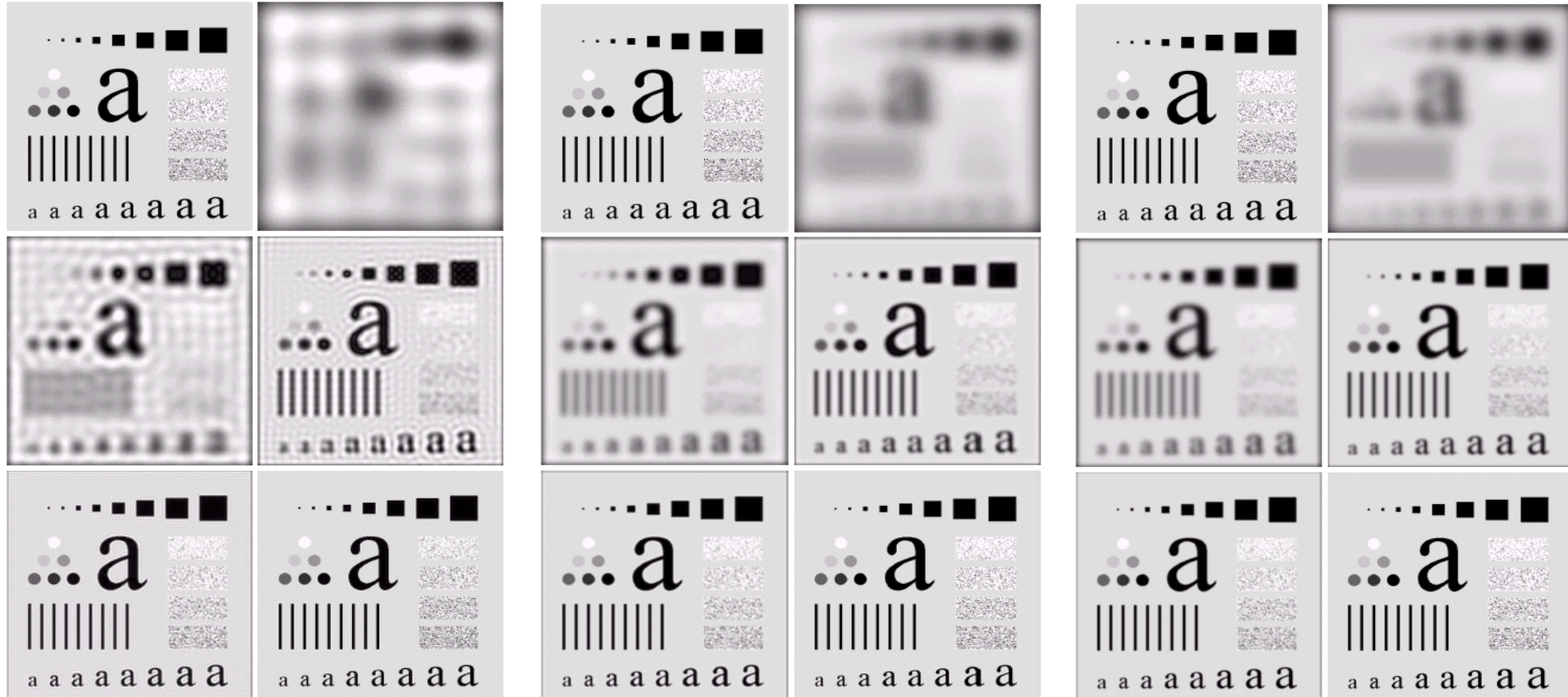


GLPF cut off
frequency 160



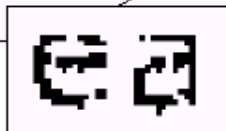
GLPF cut off
frequency 460

Comparison (ILPF, BLPF, GLPF)



Low pass filtering application

Historically, certain computer programs were written using only two digits rather than four to define the applicable year. Accordingly, the company's software may recognize a date using "00" as 1900 rather than the year 2000.



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References

- G & W (4.5.1, 4.5.2, 4.5.5, 4.6 – 4.11)
- http://mstrzel.eletel.p.lodz.pl/mstrzel/pattern_rec/fft_ang.pdf

Scribe List

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