

**Question #1 (Points: 20, 5 + 5 + 10, CLO - 2)**

Calculate Q-point parameters i.e.,  $I_C$ , and  $V_{CE}$ , and  $V_{CB}$  for the following cases. Assume  $V_{BE} = 0.7V$  for the analysis purposes. Also, identify whether the transistor is working in forward active region i.e., as an amplifier or not? Assume common-emitter topology for the analysis. Draw the circuits as well.

- a) An emitter degeneration-based biasing scheme having  $V_{CC} = 20V$ ,  $R_B = 430k\Omega$ ,  $R_C = 2k\Omega$ ,  $R_E = 1k\Omega$ ,  $\beta = 50$ .

Applying KVL on the base-emitter loop gives,

$$V_{CC} - I_B R_B - V_{BE} - I_E R_E = 0 \quad \text{--- (1)}$$

We know that,

$$I_C = \alpha I_E$$

$$\beta I_B = \left( \frac{\beta}{\beta+1} \right) I_E$$

$$I_E = (1+\beta) I_B$$

Putting value of  $I_E$  in eqn (1)

$$V_{CC} - I_B R_B - V_{BE} - (1+\beta) R_E = 0$$

$$V_{CC} - V_{BE} = I_B ((1+\beta) R_E + R_B)$$

$$I_B = \frac{V_{CC} - V_{BE}}{(1+\beta) R_E + R_B}$$

$$I_B = \frac{20 - 0.7}{(1+50)1000 + 430 \times 10^3}$$

$$I_B = 40.12 \mu A$$

Now we know,

$$I_C = \beta I_B$$

$$I_C = 50 \times (40.12 \mu A)$$

$$I_C = 2.006 \text{ mA}$$

Applying KVL on collector-emitter loop,

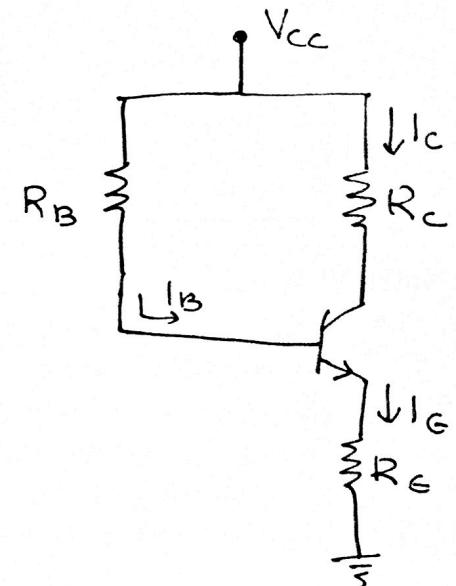
$$V_{CE} = V_{CC} - I_C R_C + I_E R_E$$

As  $I_C \approx I_E$

$$V_{CE} = V_{CC} - I_C R_C - I_C R_E$$

$$V_{CE} = V_{CC} - I_C (R_C + R_E)$$

$$V_{CC} = 13.982 \text{ V}$$



$$\text{Now, } V_{CB} = V_{CE} - V_{BE}$$

$$V_{CB} = 13.982 - 0.7$$

$$V_{CB} = 13.282 \text{ V}$$

for active region  $V_{BE} > 0$  and

$$V_{BC} \gg 0$$

$$V_{BC} = V_{BE} - V_E$$

$$= 0.7 - 13.982 = -13.282$$

which is proved hence  $^2$

in forward active  
region

A self-bias (feedback bias) scheme having  $V_{CC} = 10V$ ,  $R_B = 180k\Omega$ ,  $R_C = 1.5k\Omega$ ,  $\beta = 100$ .

Applying KVL to the base-emitter loop gives,

$$V_{CC} - I_c R_C - I_B R_B - V_{BE} = 0$$

$$I_B = \frac{V_{CC} - V_{BE}}{R_B + \beta R_C}$$

$$I_B = \frac{10 - 0.7}{180k\Omega + (100)(1.5k)}$$

$$\boxed{I_B = 2.818mA}$$

$$I_c = \beta I_B$$

$$\boxed{I_c = 2.818mA}$$

Applying KVL on collector-emitter loop,

$$V_{CE} = V_{CC} - I_c R_C$$

$$V_{CE} = 10 - (2.818mA)(1.5k)$$

$$\boxed{V_{CE} = 5.773}$$

We know,

$$V_{CB} = V_{CE} - V_{BE}$$

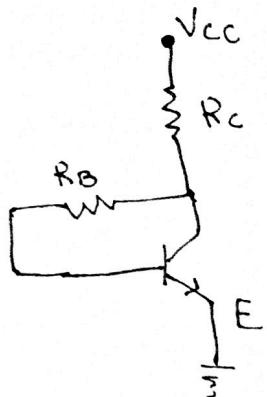
$$V_{CB} = 5.773 - 0.7$$

$$\boxed{V_{CB} = 5.073V}$$

for forward region,

$$V_{BC} = -5.073 \quad V_{BE} = 0.7 \quad \left. \begin{array}{l} \text{forward} \\ \text{region} \end{array} \right\}$$

$$V_{BC} < 0 \quad \text{and} \quad V_{BE} > 0$$



- c) A voltage divider network-based biasing scheme having  $V_{CC} = 10V$ ,  $R_1 = 18k\Omega$ ,  $R_2 = 4.7k\Omega$ ,  $R_E = 1.1k\Omega$ ,  $R_C = 3k\Omega$ ,  $\beta = 50$ .

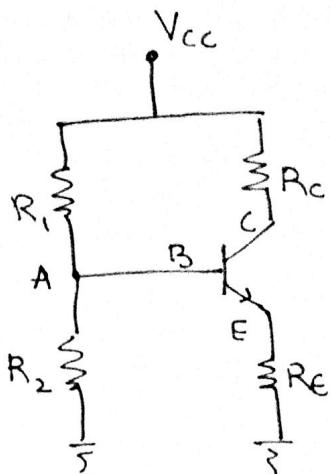
**1. Use approximate approach for analysis**

Voltage divider at node A,

$$V_A = \frac{R_2 V_{CC}}{R_1 + R_2}$$

$$= \frac{(4.7k)(10)}{18k + 4.7k}$$

$$\boxed{V_A = 2.07V}$$



Now as we can see  $V_B = V_A$ ,

$$\boxed{V_B = 2.07V}$$

$$V_B = V_{BE} + V_{RE}$$

$$V_{RE} = V_B - V_{BE}$$

$$\boxed{V_{RE} = 1.37V}$$

$$V_{RE} = I_{RE} R_E$$

$$I_{RE} = \frac{1.37}{1.1k}$$

$$\boxed{I_{RE} = 1.245mA}$$

$$\text{As } I_{RE} = I_e \approx I_c$$

$$\boxed{I_c = 1.245mA}$$

Applying KVL on collector-emitter loop,

$$V_{CC} - I_c R_C - V_{CE} - I_e R_E = 0$$

$$V_{CE} = V_{CC} - I_c (R_C + R_E) \quad : I_c \approx I_e$$

$$V_{CE} = 10 - (1.245m)(3k + 1.1k)$$

$$\boxed{V_{CE} = 4.896V}$$

$$\text{Now, } V_{CB} = V_{CE} - V_{BC}$$

$$V_{CB} = 4.896 - 0.7$$

$$\boxed{V_{CB} = 4.196V}$$

$$\text{As } V_{BC} < 0$$

$$V_{BC} = -V_{CB}$$

$$= -4.196V$$

$$\text{and } V_{BE} > 0$$

$$V_{BE} = 0.7$$

so in forward active region

Thevenin equivalent circuit,

## 2. Use accurate approach for analysis

Now using NL,

$$V_{TH} - I_B R_{TH} - V_{BE} - I_E R_E = 0$$

$$\Rightarrow V_{TH} - I_B R_{TH} - V_{BE} - (1+\beta) I_B R_E = 0$$

$$I_B = \frac{V_{TH} - V_{BE}}{R_{TH} + (1+\beta) R_E}$$

$$I_B = \frac{2.07 - 0.7}{3.73K + (1+50)(1.1K)}$$

$$I_B = 22.9 \mu A$$

$$\text{Now, } I_C = \beta I_B$$

$$I_C = 1.145 \text{ mA}$$

$$\text{so, } I_E = (1+\beta) I_B$$

$$I_E = 1.168 \text{ mA}$$

Now,

$$V_{CC} - I_C R_C - V_{CE} - I_E R_E = 0$$

$$V_{CE} = 10 - (1.145 \text{ mA})(3000) - (1.168 \text{ mA})(1100)$$

$$V_{CE} = 5.28 \text{ V}$$

$$V_{CB} = V_{CE} - V_{BE} \\ = 5.28 - 0.7$$

$$V_{CB} = 4.58 \text{ V}$$

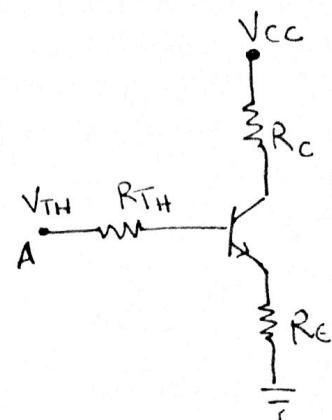
$$\text{Now } V_{BC} < 0$$

$$V_{BC} = -4.58 \text{ V}$$

and  $V_{BE} > 0$

$$V_{BE} = 0.7$$

so in forward active region



$$V_{TH} = V_A$$

from previous part,

$$V_{TH} = 2.07 \text{ V}$$

for  $R_{TH}$  removing the sources  $R_1$  and  $R_2$  in parallel

$$R_{TH} = R_1 || R_2$$

$$R_{TH} = 3.73 \text{ K}\Omega$$

**Question # 2: (Points: 40, 10 + 20 + 10, CLO - 3)**

a) Design an emitter-bias BJT configuration with the following specifications:

$I_{CQ} = (\frac{1}{2}) I_{sat}$ ,  $I_{Csat} = 8 \text{ mA}$ ,  $V_{CC} = 28V$ ,  $V_C = 18V$ ,  $V_{BE} = 0.7V$  and  $\beta = 110$ . Determine  $R_C$ ,  $R_E$ , and  $R_B$ . Assume  $V_E$  as 10% of  $V_{CC}$ .

$$I_{CQ} = \frac{1}{2} I_{sat} = \frac{1}{2} \times 8 \text{ mA}$$

$$I_{CQ} = 4 \text{ mA}$$

$V_E = 10\% \text{ of } V_{CC}$

$$V_E = 2.8V$$

$$I_E = \frac{I_C}{\beta} = \frac{4 \text{ mA}}{(110/11)} = 4.036 \text{ mA}$$

$$I_E = 4.036 \text{ mA}$$

Now as,

$$V_E = R_E I_E$$

$$R_E = \frac{2.8}{4.036 \text{ mA}}$$

$$R_E = 693.756 \Omega$$

Applying KVL at base-emitter,

$$V_{CC} - I_B R_B - V_{BE} - I_E R_E = 0$$

$$R_B = \frac{V_{CC} - V_{BE} - V_0}{(1/\beta)}$$

$$R_B = 673.75 \text{ k}\Omega$$

Applying KVL at the collector-emitter,

$$V_{CE} = V_{CC} - I_C R_C - I_E R_E$$

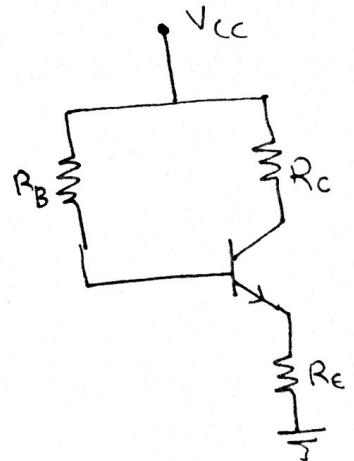
$$V_{CE} = V_C - V_E \\ = 18 - 2.8$$

$$V_{CE} = 15.2V$$

$$\text{So, } R_C = \frac{V_{CC} - V_{CE} - I_E R_E}{I_C}$$

$$= \frac{28 - 15.2 - 2.8}{4 \text{ mA}}$$

$$R_C = 2500 \Omega$$



Design an voltage divider-bias BJT configuration with the following specifications:

$I_{CQ} = 10\text{mA}$ ,  $V_{CC} = 20\text{V}$ ,  $V_{CEQ} = 8\text{V}$ ,  $V_{BE} = 0.7\text{V}$ , and  $\beta = 80$ . Determine  $R_C$ ,  $R_E$ , and  $R_B$ . Use both approaches (use notes uploaded on LMS) to design the biasing stage. Assume  $V_E$  as 10% of  $V_{CC}$ .

### 1. Accurate Approach

$$V_E = 10\% \text{ of } V_{CC} = 2\text{V}$$

$$I_E = I_C/\alpha = 10\text{m}/80/80$$

$$I_E = 10.125\text{mA}$$

$$R_E = \frac{V_E}{I_E} = \frac{2}{10.125\text{m}}$$

$$R_E = 197.53\Omega$$

$$V_B = V_{BEC} + V_E \\ = 0.7 + 2$$

$$V_B = 2.7\text{V}$$

$$R_B = \frac{(1+\beta) R_E}{10} = \frac{(1+80)(197.53)}{10}$$

$$R_B = 1600\Omega$$

as  $R_B = R_{TH}$ ,

$$V_{CC} * \frac{R_B}{V_B} = R_1$$

$$R_1 = 20 * \frac{1600}{2.7}$$

$$R_1 = 11.852\text{k}\Omega$$

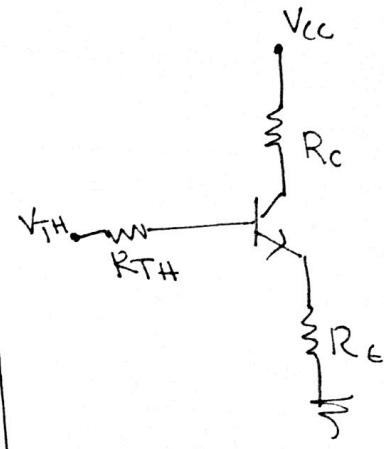
$$\text{As } R_{TH} = \frac{R_1 R_2}{R_1 + R_2}$$

$$R_{TH} R_1 + R_{TH} R_2 = R_1 R_2$$

$$R_2 = \left[ -\frac{R_{TH} R_1}{(R_{TH} - R_1)} \right]$$

$$R_2 = \left[ -\frac{(1600)(11.852\text{k})}{1600 - 11.852\text{k}} \right]$$

$$R_2 = 1.85\text{k}\Omega$$



$$V_{TH} = \frac{R_2 V_{CC}}{R_1 + R_2}$$

$$R_{TH} = R_1 || R_2$$

$$V_{CC} - I_C R_C - V_E - V_E = 0$$

$$R_C = \frac{V_{CC} - V_E - V_E}{I_C}$$

$$R_C = \frac{20 - 8 - 2}{10\text{m}}$$

$$R_C = 1000\Omega$$

## 2. Approximate Approach

from previous part,

$$V_E = 2V$$

$$\Rightarrow I_E = 10.125 \text{ mA}$$

and

$$R_E = 197.53 \Omega$$

$$\text{let, } I_I = 10 I_B$$

$$I_B = I_C / \beta$$

$$I_B = 12.5 \mu\text{A}$$

$$I_I = \frac{V_{CC}}{R_1 + R_2}$$

$$R_1 + R_2 = 16 \text{ k}\Omega$$

Now,

$$V_B = V_{BE} + V_E$$

$$\frac{V_{CC} R_2}{R_1 + R_2} = 0.7 + 2$$

$$2.7(16k) = 20 R_2$$

$$R_2 = 2160 \Omega$$

$$R_1 = 16000 - 2160$$

$$R_1 = 13840 \Omega$$

Now,

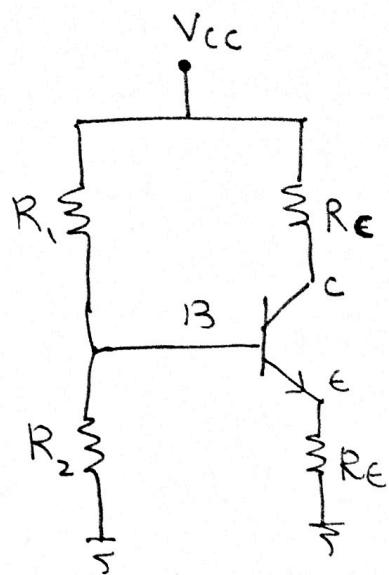
$$V_{CC} - I_C R_C - V_{CE} - V_E = 0$$

$$R_C = \frac{V_{CC} - V_{CE} - V_E}{I_C}$$

$$R_C = 1k$$

$$R_{IB} = R_1 // R_2$$

$$R_{IB} = 187 \text{ k}\Omega$$



c) Design a self-bias (collector feedback bias) BJT configuration with the following specifications:

$I_{CQ} = 10\text{mA}$ ,  $V_{CC} = 20\text{V}$ ,  $V_{CEQ} = 8\text{V}$ ,  $V_{BE} = 0.7\text{V}$ , and  $\beta = 80$ . Determine  $R_C$ ,  $R_E$ , and  $R_B$ . Assume  $V_E$  as 10% of  $V_{CC}$

$$V_E = 2\text{V}$$

$$I_E = I_C/\alpha = 10.125\text{mA}$$

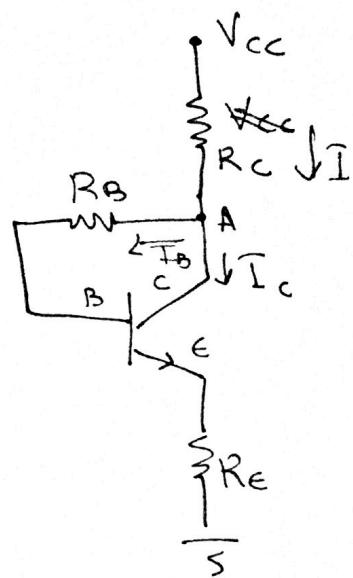
$$R_C = 197.53\Omega$$

Applying KCL at node A,

$$I = I_C + I_B$$

$$I_B = I_C/\beta$$

$$\Rightarrow I_B = 0.125\text{mA}$$



$$\text{Now, } V_{CC} - I_C R_C - V_{CE} - I_E R_E = 0$$

$$\text{So, } R_C = \frac{V_{CC} - V_{CE} - V_E}{I_C + I_B}$$

$$R_C = 987.654\Omega$$

$$V_A = V_{CC} - R_C I$$

$$V_A = I_B R_B + V_{BE} + V_E$$

$$\text{So, } V_{CC} - R_C I = I_B R_B + V_{BE} + V_E$$

$$R_B = \frac{V_{CC} - R_C(I_C + I_B) - V_{BE} - V_E}{I_B}$$

$$R_B = 58.4\text{k}\Omega$$