

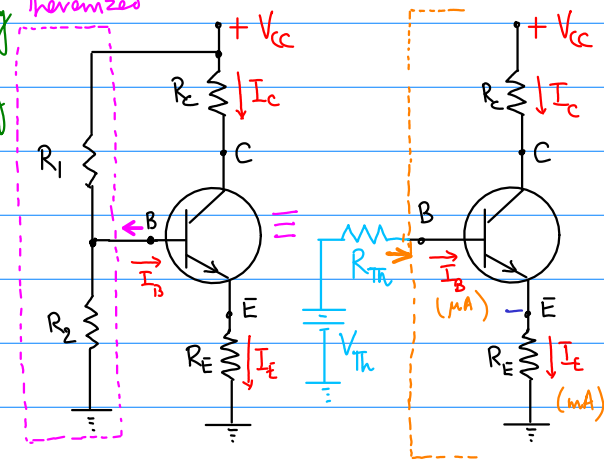
# BJT Biasing Circuits

Reference: Chapter 6 & 7  
Book: Malvino & Paty

Recap: ① Base-Bias ckt.: Fix  $I_B$ ;  $\beta$ -dependency *Therunized*

② Emitter-Bias ckt.:  $R_E$ ;  $\beta$ -independent

③ Voltage-Divider Bias ckt.:

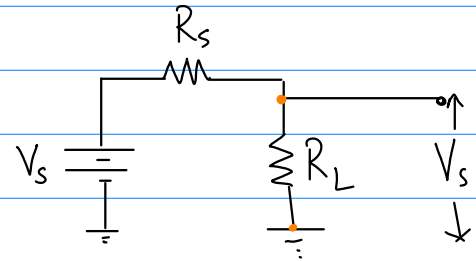


Effectively, one can say that

$V_{TH}$  is the source which bias the base-emitter junction, and  $R_{TH}$  is the source resistance.

where,  $V_{TH} = \frac{R_2}{R_1 + R_2} \cdot V_{CC}$ ;  $R_{TH} = R_1 \parallel R_2$

The condition under which all source voltage to be dropped across the load is called "Stiff".



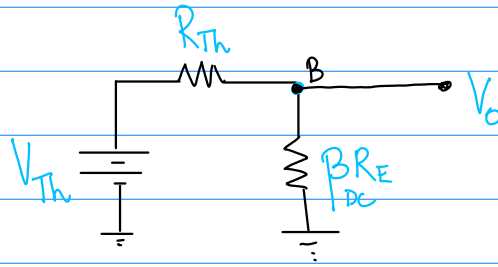
$$R_L \geq 100 R_s$$

----- Refer to chap 1.

Here, to estimate the effective load to the base terminal, we can compare the base-current ( $I_B$ ) with emitter current ( $I_E$ ).

Since  $I_E \approx I_C = \beta_{DC} I_B$

Since  $I \propto \frac{1}{R} \Rightarrow$  Effective load as seen from the base terminal is  $\beta_{DC} R_E$



$$V_o = \frac{\beta R_{E_{DC}}}{\beta R_{E_{DC}} + R_{Th}} \cdot V_{Th}$$

$$\beta R_{E_{DC}} \geq 100 R_{Th} \quad \dots \text{stiff condition.}$$

$$\Rightarrow \boxed{\beta R_{E_{DC}} \geq 100 (R_1 \parallel R_2)}$$

$$\rightarrow V_o \approx V_{Th}$$

Summary: For designing a voltage-divider bias ckt.

- Verify the condition for 'stiff' voltage source is

$$\beta R_{E_{DC}} \geq 100 (R_1 \parallel R_2)$$

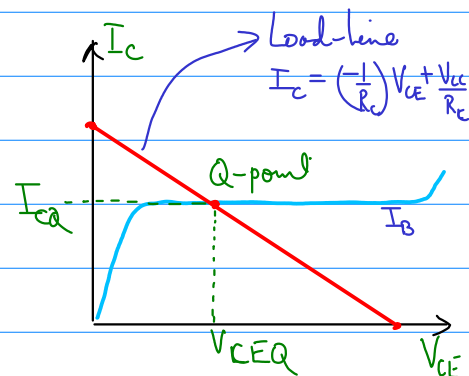
- $I_E = \frac{V_E}{R_E}$

- $I_C = I_E$

- $V_C = V_{CC} - I_C R_C$

- $V_{CE} = V_C - V_E$

- Q-point ( $V_{CEQ}, I_{CQ}$ )



#### ④ Two-supply Emitter Bias :

- $V_B = 0V$

- $V_E = -0.7V$

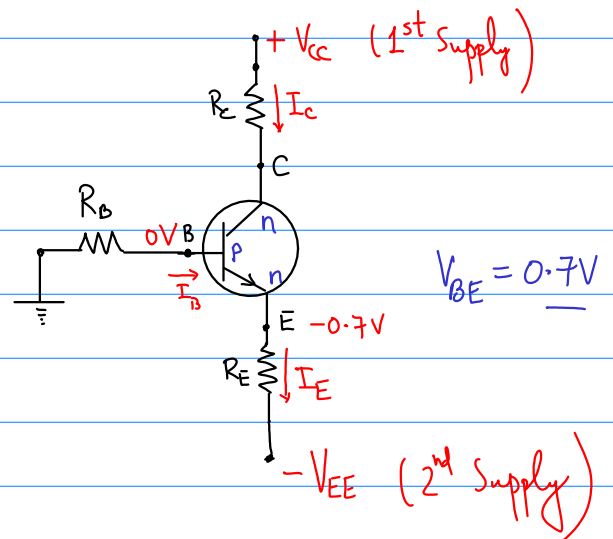
- $I_E = \frac{V_{EE} - 0.7V}{R_E}$  ✓✓

- $I_C \approx I_E$

- $V_C = V_{CC} - I_C R$

- $V_{CE} = V_C - V_E = V_C + 0.7V$

- Operating point (Q-point) :  $V_{CEQ}, I_{CQ}$



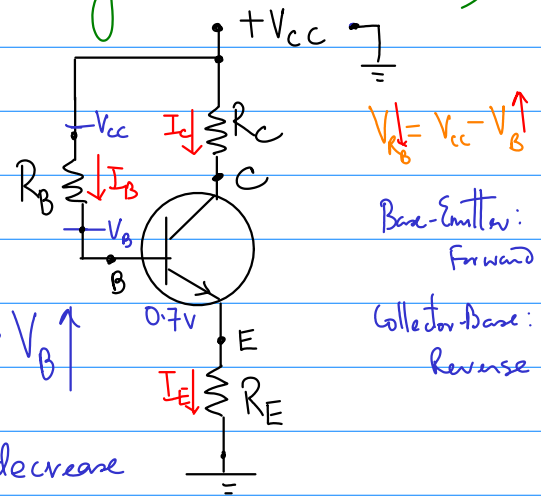
#### ⑤ Feedback Bias Ckt. :

(i) Emitter-Feedback ckt. (Negative Feedback)

In this ckt., suppose while operation, the collector-current increases.

i.e.,  $I_C \uparrow \Rightarrow I_E \uparrow \Rightarrow V_E \uparrow \Rightarrow V_B \uparrow$

effectively, the voltage drop across  $R_B$  decrease



$$V_{R_B} \downarrow \Rightarrow I_B \downarrow \Rightarrow I_C \downarrow \quad \dots \text{Since } I_C = \beta I_B$$

ie, any increase in the value of  $I_C$ , the ckt. behaves as "negative feedback", and effectively

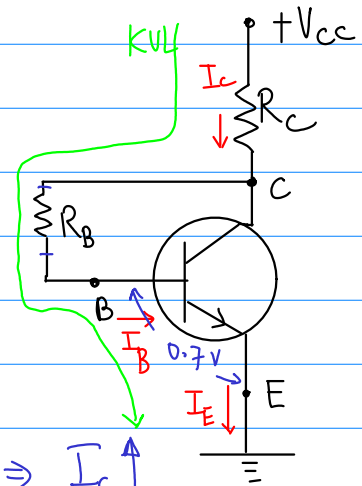
reduces the value of  $I_C$ .  $I_C \uparrow \Rightarrow Q \uparrow \Rightarrow I_C \downarrow$  or  $I_C \downarrow \Rightarrow Q \downarrow \Rightarrow I_C \uparrow$

(ii) Collector-feedback ckt:

$$\begin{aligned} I_C \uparrow &\Rightarrow V_C \downarrow \Rightarrow V_{R_B} \downarrow \Rightarrow I_B \downarrow \\ &\Rightarrow I_C \downarrow \end{aligned}$$

or,

$$I_C \downarrow \Rightarrow V_C \uparrow \Rightarrow V_{R_B} \uparrow \Rightarrow I_B \uparrow \Rightarrow I_C \uparrow$$



$$V_C = V_{CC} - I_C R_C$$

$$V_{CC} = I_C R_C + I_B R_B + V_{BE}$$

$$V_{CC} - V_{BE} = I_C R_C + \frac{I_C R_B}{\beta_{DC}}$$

$$\Rightarrow I_C = \frac{V_{CC} - V_{BE}}{R_C + \frac{R_B}{\beta_{DC}}}$$

Therefore,  $V_C = V_{CC} - I_C R_C$

Q-point ( $V_{CEQ}, I_{CQ}$ )

(iii) Emitter-Collector feedback ckt.

$$V_{CC} = I_C R_C + I_B R_B + V_{BE} + I_E R_E$$

Since,  $I_C \approx I_E$  ;  $I_B = \frac{I_C}{\beta_{DC}}$

$$\Rightarrow V_{CC} = I_C R_C + \frac{I_C R_B}{\beta_{DC}} + V_{BE} + I_C R_E$$

$$\Rightarrow I_C = \frac{V_{CC} - V_{BE}}{R_C + \frac{R_B}{\beta_{DC}} + R_E}$$

Therefore,  $V_C = V_{CC} - I_C R_C$

$$V_E = I_E R_E$$

$$V_{CE} = V_C - V_E$$

Q-point ( $V_{CEQ}, I_{CQ}$ )

