

The Equivalent ckt. of an Induction Motor

Recap: • Relative speed of rot. mag. field & rotor bar.

• slip speed; $n_{\text{slip}} = n_{\text{sync}} - n_m$

• slip $s = \frac{n_{\text{slip}}}{n_{\text{sync}}} \times 100\%$

• $f_r = s f_e$

I.M. \Rightarrow equ. ckt. Model.

- Since the "Induction" of emf/current in the rotor bars (winding) of an I.M is in principle a "TRANSFORMER ACTION".

Therefore, the equivalent ckt. model of I.M.

corresponds to equivalent ckt. model of a Transformer.

- Also, it is noted that the I.M. is a "Singly excited" machine. That is, there is NO FIELD CURRENT

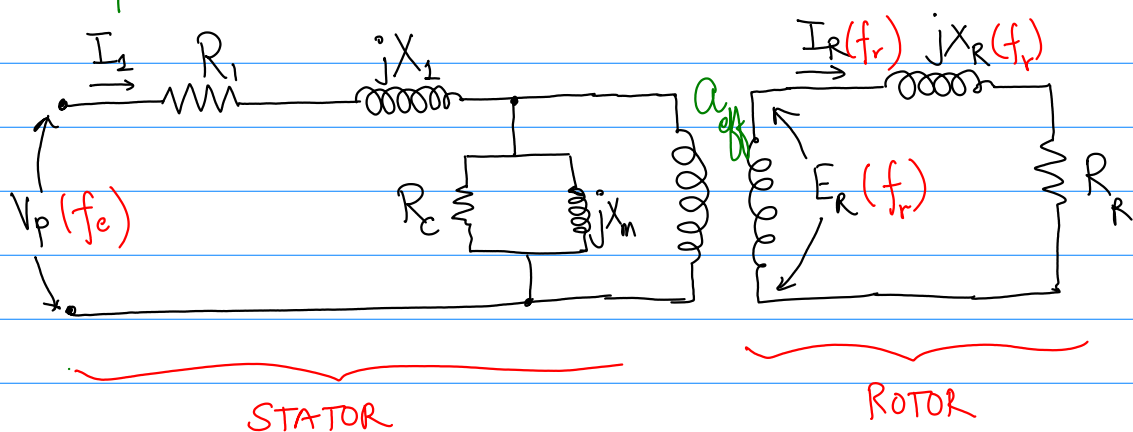
the electrical ckt. model does not have an internal voltage source such as E_A in case sync. machine.

— Thus the equivalent ckt. model of the I.M. will be developed considering :

(i) the equivalent ckt. model of the transformer.

(ii) the variation in the rotor's electrical frequency with the relative mechanical speed (ie, slip)

Per-phase equivalent ckt. model.



a_{eff} = effective turn-ratio

— An I.M. equivalent ckt. differs from the transformer equivalent ckt. model primarily in effects of varying rotor electrical frequency on the

rotor voltage E_R and the rotor impedance X_R

— Hence modeling the rotor ckt. is important.

Rotor Circuit Model :

In general, the greater the relative motion between the stator magnetic field (\vec{B}_s) and rotor magnetic field (\vec{B}_R), the greater the resulting rotor voltage & frequency.

Case (i) Largest relative motion

— When the rotor is stationary i.e., the rotor is mechanically blocked (or locked).

$$S = 1$$

Case (ii) Smallest relative motion (No relative motion)

— When the rotor is moving with sync. speed.

$$S = 0$$

Hence, if the magnitude of the induced rotor voltage in case (i) is E_{R0} , then the magnitude of induced rotor voltage at any slip value is

given as

$$E_R = s E_{R0}$$

Further, we also know that

$$f_r = s f_e$$

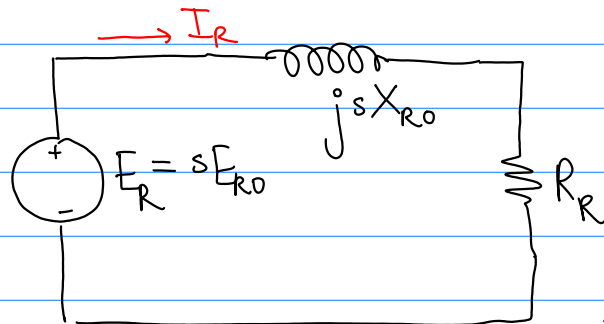
and the reactance $X_R = \underbrace{2\pi f_r}_{\omega} L$

therefore,

$$X_R = s X_{R0}$$

where X_{R0} is the reactance of the rotor when it is mechanically blocked.

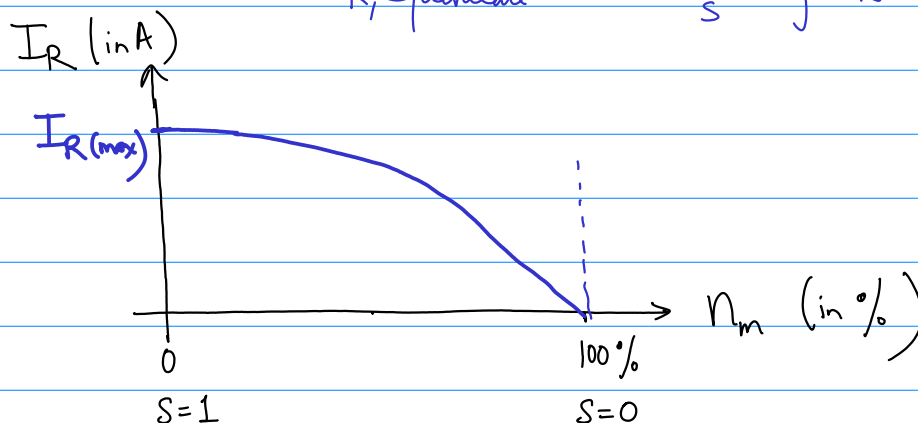
The rotor equivalent ckt. will be



$$I_R = \frac{E_R}{R_R + j X_R}$$

$$\Rightarrow I_R = \frac{s E_{R0}}{R_R + j s X_{R0}}$$

$$Z_{R, \text{equivalent}} = \frac{R_R}{s} + j X_{R0}$$

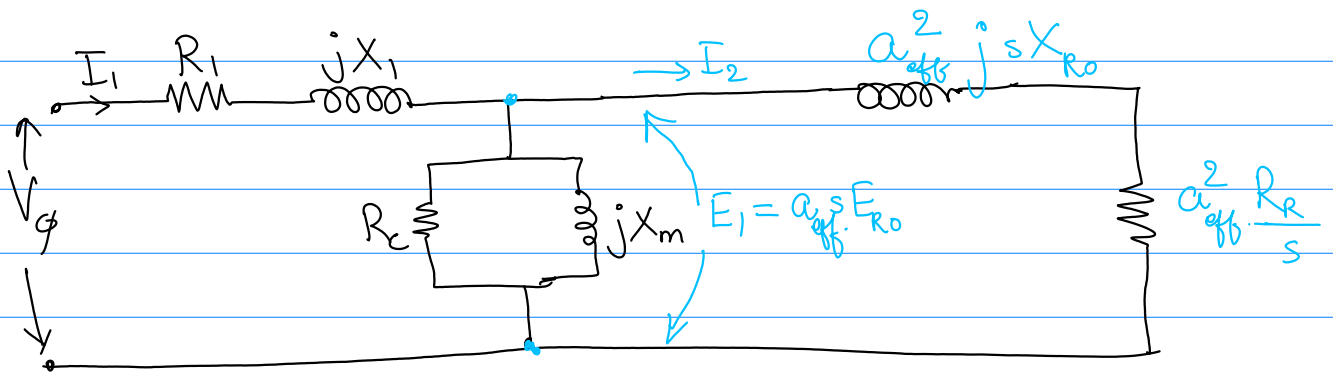


From the discussion of transformer, we know that

$$\frac{V_p}{V_s} = a \quad ; \quad \frac{I_p}{I_s} = \frac{1}{a} \quad ; \quad Z'_s = a^2 Z_s$$

Referred impedance
to the primary

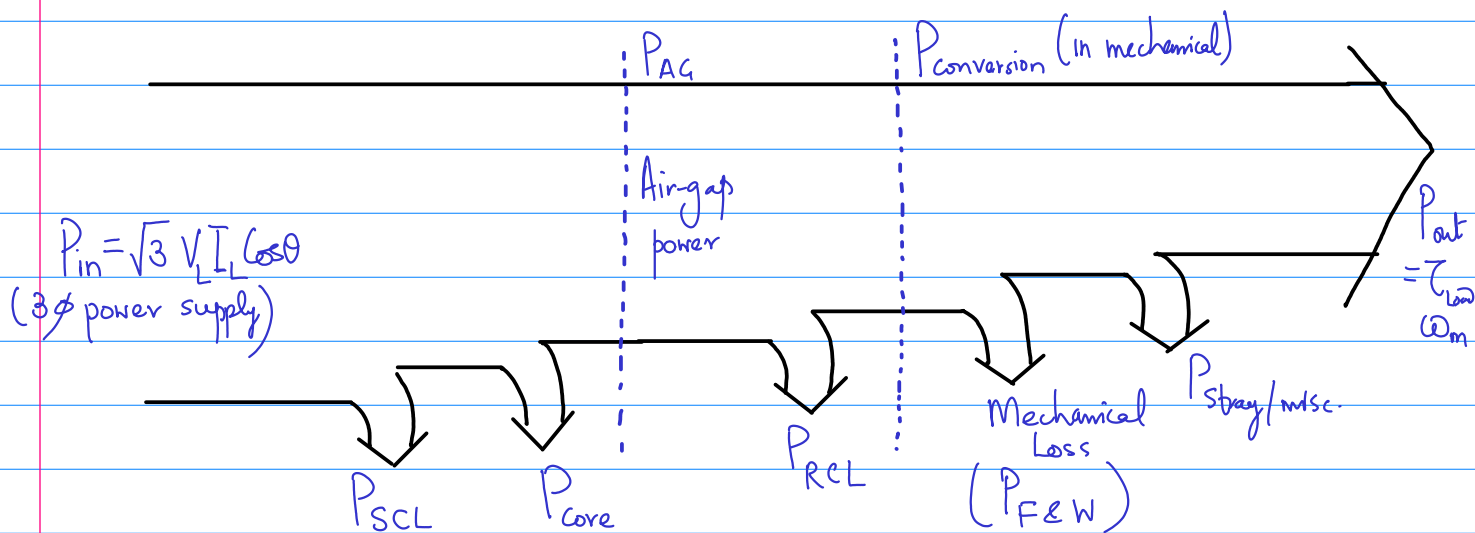
Per-phase equivalent ckt. model of an I.M. referred
to the stator (primary)



Power and Torque in Induction Motor

- Recap:
- (i) Equivalent ckt. model of I.M. corresponds to the equivalent ckt. model of a transformer.
 - (ii) Stator ckt. (primary ckt.) remains the same.
 - (iii) Rotor ckt. (secondary ckt.) is 'SLIP' dependent.
(ie, relative motion)

Power flow diagram of I.M. :



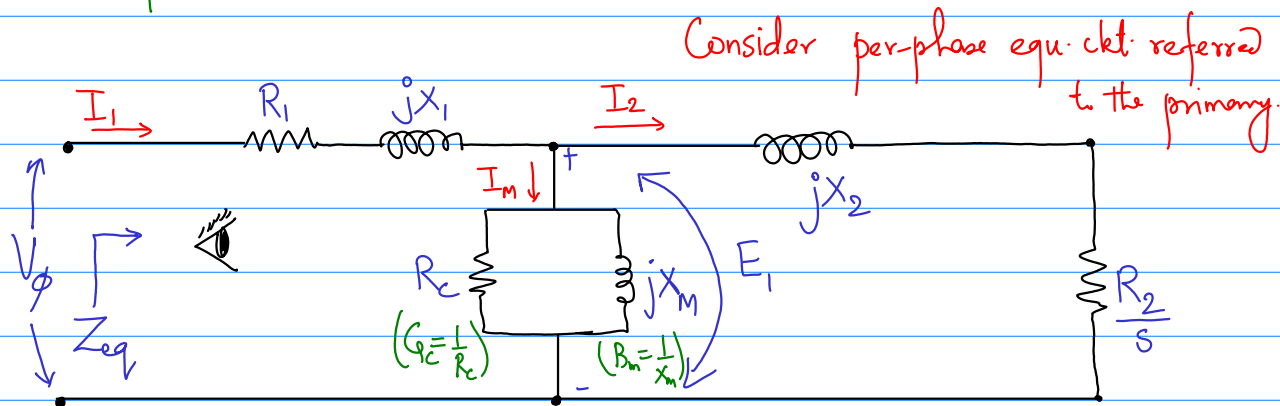
- Input electrical power, $P_{in} = 3\phi$ power supply $= \sqrt{3} V_L I_L \cos \theta$
- This input power is given to the stator winding, therefore, there is a copper loss in stator windings (P_{SCL})
- Since the stator winding is done over the laminated core, hence there are some amount of power loss due to hysteresis & eddy current. (P_{core})

- The electrical power remaining after copper & core losses is transferred to the rotor through the air-gap (P_{Ag}).
- Since the power is transferred in form of induced emf in the rotor, and since the rotor is electrically shorted, there is some amount power lost as $I^2 R_r$ in the rotor (P_{REL} , rotor copper loss).
- After P_{REL} , there is no electrical power loss. Therefore, the remaining power is converted into mechanical power ($P_{conv.}$).
- As a mechanical power loss, we have to account any loss due to friction & windage. ($P_{F\&W}$).
- Any other losses incorporated as $P_{\text{stray/misc.}}$.
- The remaining power is available to the load.

$$P_{\text{out}} = T_{\text{load}} \cdot \omega_m$$

$$P_{\text{out (mech.)}} = P_{\text{in (elec.)}} - P_{\text{scL}} - P_{\text{core}} - P_{\text{REL}} - P_{\text{F\&W}} - P_{\text{stray/misc.}}$$

Relate power & torque. %



here, $X_2 = a_{\text{eff}}^2 X_{R0}$

$R_2 = a_{\text{eff}}^2 R_R$

$E_1 = a_{\text{eff}} E_{R0}$

$$\left\{ I_R = \frac{E_{R0}}{\frac{R_R}{s} + jX_{R0}} \right.$$

- From the equivalent ckt., one can determine the input current (per phase) in the motor.

$$I_1 = \frac{V_\phi}{Z_{\text{eq}}}$$

here, $Z_{\text{eq}} = R_1 + jX_1 + \left[\frac{1}{G_c - jB_m + \left\{ \frac{1}{\frac{R_2}{s} + jX_2} \right\}} \right]$

Therefore, knowing the value of I_1 , we can determine, the stator copper loss & core loss.

$$P_{SCL} = 3 I_1^2 R_1$$

$$P_{core} = \frac{3 E_1^2}{R_c} = 3 G_c E_1^2$$

So, from this we get the power available in air-gap

$$P_{AG} = \underbrace{P_{in}}_{\text{input}} - \underbrace{P_{SCL}}_{3 I_1^2 R_1} - \underbrace{P_{core}}_{3 G_c E_1^2}$$

Now, out of available air-gap power (P_{AG}) the only power consumed is ⁱⁿ the resistor R_2/s .

So, one can say. $P_{AG} = 3 I_2^2 \frac{R_2}{s}$

however, the rotor copper loss is expressed as

$$P_{RCL} = 3 I_2^2 R_2$$

Comparing P_{AG} & P_{RCL} we get

$$P_{RCL} = s P_{AG}$$

$$P_{conv.} = P_{AG} - P_{RCL} = (1-s) P_{AG}$$

$$P_{conv.} = (1-s) P_{AG}$$

Finally,
$$P_{out} = P_{conv.} - \underbrace{P_{F\&W}}_{\text{if given}} - \underbrace{P_{\text{stray/misc.}}}_{\text{if given}}$$

What is Induced Torque (& Developed Torque) in the I.M. ?

- Basically it is the torque generated by the internal electric to mechanical power conversion.
- This internal induced torque is higher than the output torque at the shaft of the rotor.

$$\Rightarrow T_{ind} = \frac{P_{conv.}}{\omega_m} = \frac{(1/s) P_{AG}}{(1/s) \omega_{sync}}$$

$$\Rightarrow \boxed{T_{ind} = \frac{P_{AG.}}{\omega_{sync}}}$$

Note: P_{AG} & ω_{sync} does not vary once the input to the stator is fixed.