



Chebyshev Filters

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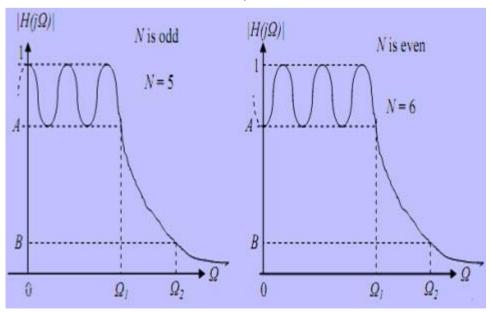


INTRODUCTION

• The name of Chebyshev filters is termed after "Pafnufy Chebyshev" because its mathematical characteristics are derived from his name only. Chebyshev filters are nothing but analog or digital filters. These filters have a steeper roll off & type-1 filter (more pass band ripple) or type-2 filter (stop band ripple) than Butterworth filters. The property of this filter is, it reduces the error between the characteristic of the actual and idealized filter. Because, inherent of the pass band ripple in this filter

Chebyshev Filter

Chebyshev filters are used for distinct frequencies of one band from another. They
cannot match the windows-sink filter's performance and they are suitable for many
applications. The main feature of Chebyshev filter is their speed, normally faster
than the windowed-sinc. Because these filters are carried out by recursion rather
than convolution. The designing of the Chebyshev and Windowed-Sinc filters
depends on a mathematical technique called as the Z-transform.







Types of Chebyshev Filters

 Chebyshev filters are classified into two types, namely type-I Chebyshev filter and type-II Chebyshev filter.

Type-I Chebyshev Filters

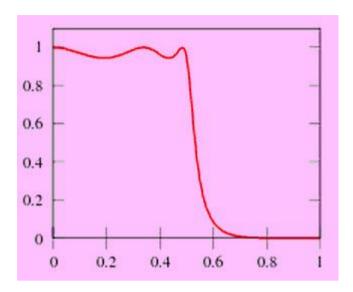
 This type of filter is the basic type of Chebyshev filter. The amplitude or the gain response is an angular frequency function of the nth order of the LPF (low pass filter) is equal to the total value of the transfer function Hn (jw)

Gn(w)=|Hn (j
$$\omega$$
)|=1 $\sqrt{(1+\epsilon 2Tn2() \omega/\omega o)}$

- Where,ε=ripple factor
 ωo= cutoff frequency
 Tn= Chebyshev polynomial of the nth order
- The pass-band shows equiripple performance. In this band, the filter interchanges between -1 & 1 so the gain of the filter interchanges between max at G = 1 and min at $G = 1/\sqrt{(1+\epsilon 2)}$. At the cutoff frequency, the gain has the value of $1/\sqrt{(1+\epsilon 2)}$ and remains to fail into the stop band as the frequency increases. The behavior of the filter is shown below. The cutoff frequency at -3dB is generally not applied to Chebyshev filters.







The order of this filter is similar to the no. of reactive components required for the Chebyshev filter using analog devices. The ripple in dB is $20\log 10 \sqrt{(1+\epsilon 2)}$. So that the amplitude of a ripple of a 3db result from $\epsilon=1$ An even steeper roll-off can be found if ripple is permitted in the stop band, by permitting 0's on the jw-axis in the complex plane. Though, this effect in less suppression in the stop band. The effect is called a Cauer or elliptic filter.

Poles and Zeros of Type-I Chebyshev Filter

The poles and zeros of the type-1 Chebyshev filter is discussed below. The poles of the Chebyshev filter can be determined by the gain of the filter.

$$1 + \varepsilon^2 T_n^2(-js) = 0$$

-js=cos(θ) & the definition of trigonometric of the filter can be written as

$$1 + \varepsilon^2 T_n^2(\cos(\theta)) = 1 + \varepsilon^2 \cos^2(n\theta) = 0$$

Here θ can be solved by





$$\theta = \frac{1}{n}\arccos\left(\frac{\pm j}{\varepsilon}\right) + \frac{m\pi}{n}$$

Where the many values of the arc cosine function have made clear using the number index m. Then the Chebyshev gain poles functions are

$$s_{pm} = j\cos(\theta)$$

$$= j\cos\left(\frac{1}{n}\arccos\left(\frac{\pm j}{\varepsilon}\right) + \frac{m\pi}{n}\right)$$

Using the properties of hyperbolic & the trigonometric functions, this may be written in the following form

$$s_{pm}^{\pm} = \pm \sinh\left(\frac{1}{n}\operatorname{arsinh}\left(\frac{1}{\varepsilon}\right)\right)\sin(\theta_m)$$
$$+j\cosh\left(\frac{1}{n}\operatorname{arsinh}\left(\frac{1}{\varepsilon}\right)\right)\cos(\theta_m)$$

Here m = 1, 2, 3,n

$$\theta_m = \frac{\pi}{2} \, \frac{2m-1}{n}.$$

The above equation produces the poles of the gain G. For each pole, there is the complex conjugate, & for each and every pair of conjugate there are two more negatives of the pair. The TF should be stable, The transfer function (TF) is given by

$$H(s) = \frac{1}{2^{n-1}\varepsilon} \ \prod_{m=1}^n \frac{1}{(s-s_{pm}^-)}$$





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