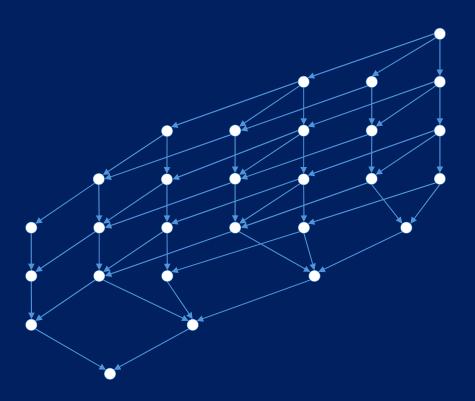
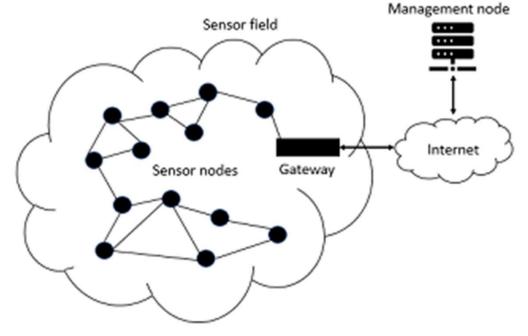
# Graph Signal Processing

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#### **Problem statement**

- Sensors might be expensive to operate
- Sensor/battery lifetime might be increased when used less often
- System might be more efficient with less data transmitted



https://study.com/cimages/multimages/16/c701edbb-2f75-4518-a658-60a5743f66a5\_sensor\_networ.png

Sampling problem: How to infer values of nodes not observed?

## **Sampling**

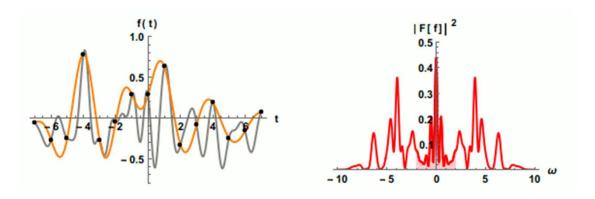
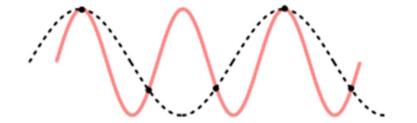
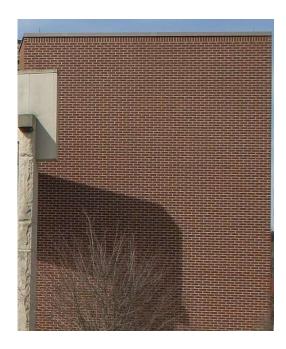


Figure: https://en.wikipedia.org/wiki/Nyquist%E2%80%93Shannon\_sampling\_theorem

## **Down-Sampling (gone wrong)**







Figures: https://en.wikipedia.org/wiki/Nyquist%E2%80%93Shannon\_sampling\_theorem

### Sampling

 $\mathbf{x}_{\mathcal{M}} = \Psi \mathbf{x} \in \mathbb{C}^{M}$ ,  $\mathcal{M} \subset \mathcal{N}$   $\mathcal{N}$  Measured signal

Sampling takes (a linear combination of) samples at choosen measurement points to produce a lower dimensional signal.

#### Interpolation

$$\widetilde{\mathbf{x}} = \Phi \mathbf{x}_{\mathcal{M}} = \Phi \Psi \mathbf{x} \in \mathbb{C}^N$$

Interpolation approximates the original signal based on the sampled signal.

#### **Example**

$$\Psi = \begin{bmatrix} 0 & 1 & 0 & 1 \\ 1 & 0 & 1 & 0 \end{bmatrix}$$

$$\Psi \mathbf{x} = \begin{bmatrix} x_2 + \\ x_1 + \end{bmatrix}$$

$$\hat{Q} = \begin{bmatrix} 0 & \frac{1}{2} \\ \frac{1}{2} & 0 \\ 0 & \frac{1}{2} \\ \frac{1}{2} & 0 \end{bmatrix}$$

$$\Phi \Psi \mathbf{x} = \frac{1}{2} \begin{vmatrix} x_1 + x_3 \\ x_2 + x_4 \\ x_1 + x_3 \\ x_1 + x_3 \end{vmatrix}$$

 $\Psi = \begin{bmatrix} 0 & 1 & 0 & 1 \\ 1 & 0 & 1 & 0 \end{bmatrix} \qquad \Psi \mathbf{x} = \begin{bmatrix} x_2 + x_4 \\ x_1 + x_3 \end{bmatrix} \qquad \tilde{\Phi} = \begin{bmatrix} 0 & \frac{1}{2} \\ \frac{1}{2} & 0 \\ 0 & \frac{1}{2} \\ \frac{1}{2} & 0 \end{bmatrix} \qquad \Phi \Psi \mathbf{x} = \frac{1}{2} \begin{bmatrix} x_1 + x_3 \\ x_2 + x_4 \\ x_1 + x_3 \\ x_2 + x_4 \end{bmatrix}$   $\chi = \begin{pmatrix} \times A \\ \times 2 \\ \times 3 \\ \times 4 \end{pmatrix} \qquad \text{Signal } \text{wit } \mathbf{x}_1 = \mathbf{x}_2 \text{ and } \mathbf{x}_2 = \mathbf{x}_4$ 

Sampling and Interpolation are pseudo-inverse operations to each other

#### **Bandlimited Graph Signals**

Graph signals are called k-bandlimited if they can be expressed as a linear combination of k eigenvectors of the graph shift operator  $\mathbf{x} \in \mathbf{x} = U_k \alpha, \quad \text{with } \alpha \in \mathbb{R}^k$ 

$$\mathbf{x} = U_k \alpha, \quad \text{with } \alpha \in \mathbb{R}^k$$

Die ktregnenzen können belichis Note that this is equivalent to k-sparsity in the

graph Fourier domain

urier domain (Hint: Spage 
$$||Fx||_0 = k$$
 Fourier transformation)

#### **Perfect Reconstruction Condition**

k-bandlinited

A graph signal can be perfectly recovered if

$$rank(\Psi U_k) = k$$

The interpolation operator which achieves perfect recovery is

$$\Phi = U_k V$$

where

$$V\Psi U_k = 1_{\mathsf{K}}$$

is an identity matrix (i.e., a pseudo-inverse)

更中×=重中UK«=UKVPUKα=UK=X

Perfect reconstruction is possible for bandlimited signals

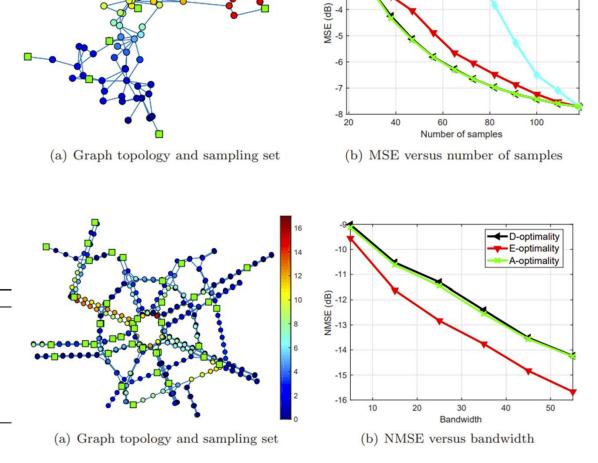
#### **Sampling strategies**

- **D-optimal design**: Minimizes the log-determinant of the interpolation error covariance to achieve maximum overall confidence in parameter estimation
- A-optimal design: Minimizes the mean squared error across all estimated coefficients to give the best average estimation accuracy
- E-optimal design: Minimize the maximum (worst-case) variance in any direction, ensuring robust reconstruction under the least favorable scenario

```
Algorithm 1: Greedy selection of graph samples

Input Data: U_{\mathcal{F}}, M;
Output Data: \mathcal{S}, the sampling set.

Function: initialize \mathcal{S} \equiv \emptyset
while |\mathcal{S}| < M
s = \arg\max_{j} f(\mathcal{S} \cup \{j\});
\mathcal{S} \leftarrow \mathcal{S} \cup \{s\};
```



D-optimalityE-optimalityRandom

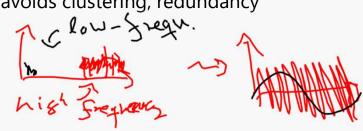
A-optimality

Figures: Sampling and Recovery of Graph Signals, Lorenzo et al.





- •Human perception prefers **even but non-uniform** sample distributions
- •Blue noise = no low-frequency components → avoids clustering, redundancy



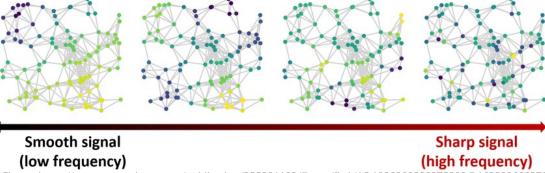


Figure: https://www.researchgate.net/publication/355901199/figure/fig1/AS:1086260500275200@1635996093701 graph-signal-becomes-sharper-going-from-left-to-right-Smooth-graph-signals-assign.ppm

### Blue noise sampling

- •Blue noise sampling patterns have energy concentrated in high frequencies
- Select nodes so that the frequency spectrum of the sampling pattern avoids low frequencies, preserving signal diversity and reducing aliasing

Blue noise sampling spreads samples to avoid low-frequency redundancy and clustering

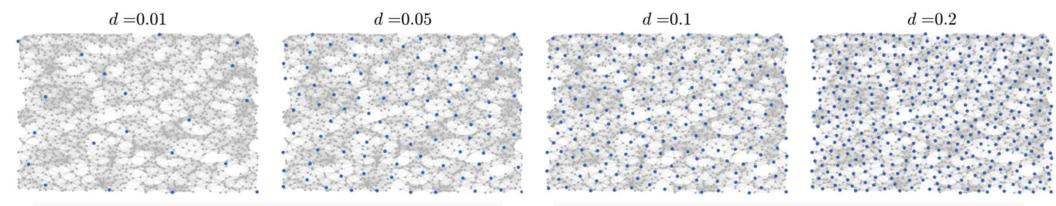
#### **Energy Penalization Method**

Blue-noise sampling can be framed as minimizing overlap with low-frequency eigenspaces:

$$\min_{s \in \{0,1\}^n} \|U_{low}^{\top} s\|^2$$

#### **Spectral Spreading Principle**

- Maximize energy outside the low-frequency subspace by choosing sampling sets that "look like noise" in the spectral domain
- Ensure samples are well-distributed across the graph topology



Figures: Blue-Noise Sampling on Graphs, Parada-Mayorga et al.

Blue-noise sampling penalizes low-frequency concentration in sampling vector.

#### **Algorithm**

Idea: Sequential computation of distances between sampling points and relocation of points with short geodesic distance to put them far from each other.



Figures: Blue-Noise Sampling on Graphs, Parada-Mayorga et al.

```
Algorithm 1: Void and Cluster Algorithm for Graphs.
 Input: m: number of samples, \sigma, NumIter.
 Output: s: sampling pattern
      Initialisation: s = 0, IndA = -1, IndB = -1.
  Calculate \mathbf{K}(i,j) = e^{-\frac{\mathbf{\Gamma}(i,j)^2}{\sigma}} for all 1 \leq i, j \leq N.
  2: \mathbf{c} = \mathbf{K} \mathbf{1}_{N \times 1}.
     Get \mathcal{M} as m nodes selected at random.
     for r = 1:1: NumIter do
   \mathbf{c}(\operatorname{supp}(s)) = \sum \mathbf{K}(\operatorname{supp}(s), \operatorname{supp}(s)).
        if IndA = \arg \max_{i} \{ \mathbf{c}(i) \} and IndB = \arg \min_{i} \{ \mathbf{c}(i) \}
         \{\mathbf{c}(i)\}
         then
            break
 12:
         else
            IndA= \arg \min_{i} \{ \mathbf{c}(i) \}.
           IndB= \arg \max_{i} \{ \mathbf{c}(i) \}.
 14:
```

end if

16: end for return s

Algorithm is computationally more efficient than greedy sampling