

# OPTIMAL CONTROL OF AN AXIAL DISPERSION TUBULAR REACTOR WITH DELAYED RECYCLE

## Authors

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# Presentation Outline

1. Introduction and Literature Review
2. Open-loop System Design
3. Optimal Control
4. Results and Discussion
5. Conclusion
6. References

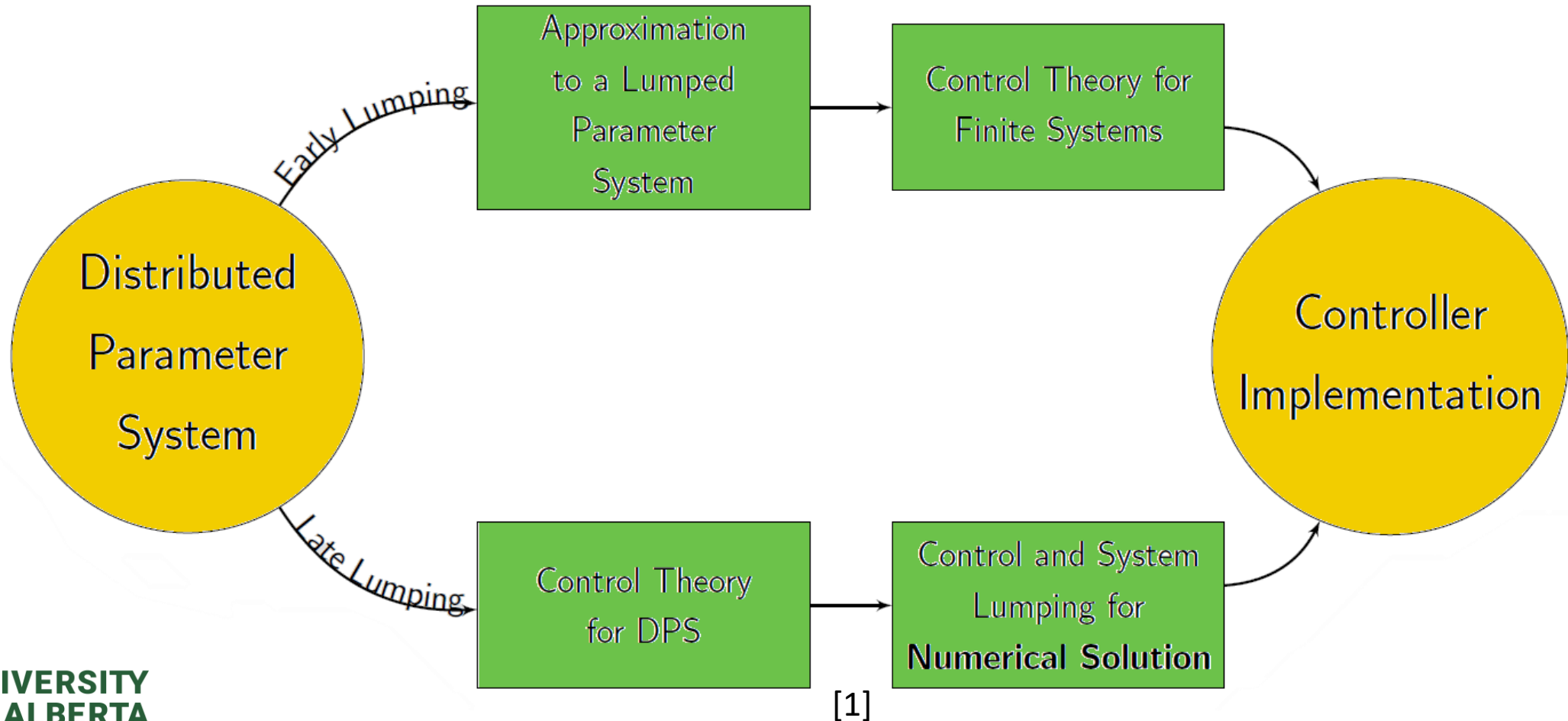
# 1- Introduction and Literature Review

- Distributed Parameters System
- Axial Dispersion Tubular Reactor with Recycle
- Notion of Delay in Recycle Flow
- Problem Statement
- Literature Review

# Distributed Parameter Systems (DPS)

- Spatially varying states in model
- Partial Differential Equations (PDE)
- Infinite-dimensional systems
- Regulator / Observer design becomes challenging

# DPS: Early Lumping vs. Late Lumping



# Axial Dispersion Tubular Reactor

In general:

$$x = x(\zeta, t), \zeta \in \Omega \text{ and } t \in \mathbb{R}^+ \quad (1)$$

A linear PDE:

$$\frac{\partial x}{\partial t}(\zeta, t) = A(\zeta, t)x(\zeta, t) + B(\zeta, t)u(t) \quad (2)$$

with:

$$A(\zeta, t) = \underbrace{\frac{\partial}{\partial \zeta} \left( D(\zeta, t) \frac{\partial}{\partial \zeta} \right)}_{\text{Diffusion}} + \underbrace{v(\zeta, t) \frac{\partial}{\partial \zeta}}_{\text{Convection}} + \underbrace{k(\zeta, t)}_{\text{Gen./Cons.}} \quad (3)$$

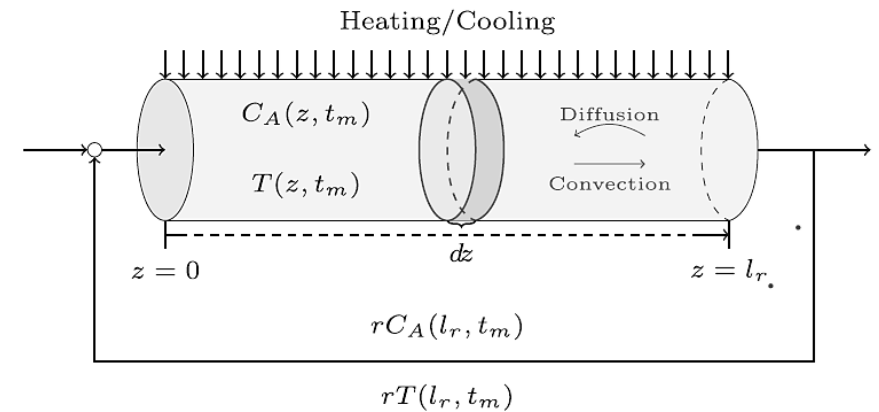


Figure 1 - Schematic view an axial dispersion tubular reactor with recycle [2]

# Notion of Delay

- Recycle flow: Pure Transport (Plug Flow)
- Equivalent to Notion of Delay:

$$\frac{\partial x}{\partial t}(\zeta, t) = -v \frac{\partial}{\partial \zeta} x(\zeta, t)$$

with:  $\begin{cases} x(\zeta = 0, t) = u(t) & \text{B.C} \\ x(\zeta, t = 0) = x_0(\zeta) & \text{I.C} \\ y(t) = x(\zeta = l, t) \end{cases} \xRightarrow{\tau_d = \frac{l}{v}} y(t) = u(t - \tau_d)$  (4)

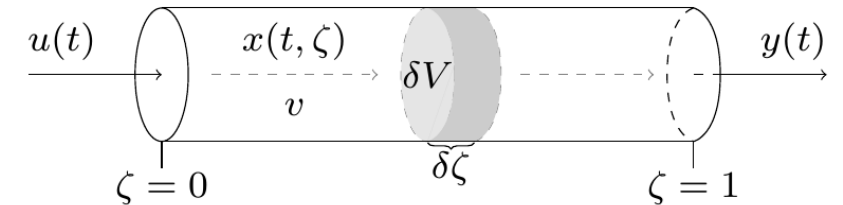


Figure 2 – Pure Transport Representation

# Problem Statement: Model

- Dispersion Tubular Reactor: Second Order Parabolic PDE
- Recycle Flow: First Order PDE
  - Considering Danckwerts boundary conditions
  - Input applied at the Boundary (Boundary-Control Problem)

$$\begin{cases} \partial_t x(\zeta, t) = D\partial_{\zeta\zeta}x(\zeta, t) - v\partial_{\zeta}x(\zeta, t) + kx(\zeta, t) \\ D\partial_{\zeta}x(0, t) - vx(0, t) = -v[Rx(1, t - \tau) + (1 - R)u(t)] \\ \partial_{\zeta}x(1, t) = 0 \\ y(t) = x(1, t) \end{cases} \quad (5)$$



# Problem Statement: Methodology and Goals

- Open-loop modeling using Late Lumping Methods
- Spectrum (Eigenvalue) Analysis
- Determining Dominant Modes
- Continuous Infinite-dimensional Optimal Controller
- Confirm Stability of the Closed-loop System

# Literature Review

- Optimal in-domain Control of Scalar Infinite-dimensional Systems [3]
- Optimal Boundary Control of Infinite-dimensional Spectral Systems [4]
- Model Predictive Control of Axial Dispersion Tubular Reactors with Instantaneous Recycle [2]
- Optimal Boundary Control of First-order Hyperbolic Systems [5]

## 2- Open-loop System Design

- Eigenvalue Problem
- Adjoint Operator
- Characteristic Equation
- Eigenvalue Distribution (Spectrum Analysis)
- Eigenfunctions (Bi-orthogonal Basis)
- Dominant Modes

# Eigenvalue Problem

$$\left\{ \begin{array}{l} \lambda\phi = D\partial_{\zeta\zeta}\phi - v\partial_{\zeta}\phi + k\phi \\ \lambda\psi = \frac{1}{\tau}\partial_{\zeta}\psi \\ D\partial_{\zeta}\phi(0) - v\phi(0) = -Rv\psi(0) \\ \partial_{\zeta}\phi(1) = 0 \\ \psi(1) = \phi(1) \end{array} \right. \quad (6)$$

$$\left\{ \begin{array}{l} \partial_{\zeta}X = \begin{bmatrix} 0 & 1 & 0 \\ \frac{\lambda-k}{D} & \frac{v}{D} & 0 \\ 0 & 0 & \tau\lambda \end{bmatrix} X = \Lambda X \\ DX_2(0) - vX_1(0) = -RvX_3(0) \\ X_2(1) = 0 \\ X_3(1) = X_1(1) \end{array} \right. \quad (7)$$

where  $X = [\phi, \partial_{\zeta}\phi, \psi]^T$

# Finding the Adjoint Operator

$$\left\{ \begin{array}{l} \hat{A}(\cdot) = \begin{bmatrix} D\partial_{\zeta\zeta}(\cdot) - v\partial_{\zeta}(\cdot) + k(\cdot) & 0 \\ 0 & \frac{1}{\tau}\partial_{\zeta}(\cdot) \end{bmatrix} \\ B.C. : \begin{cases} D\partial_{\zeta}\phi(0) - v\phi(0) = -Rv\psi(0) \\ \partial_{\zeta}\phi(1) = 0 \\ \psi(1) = \phi(1) \end{cases} \end{array} \right. \quad (8)$$

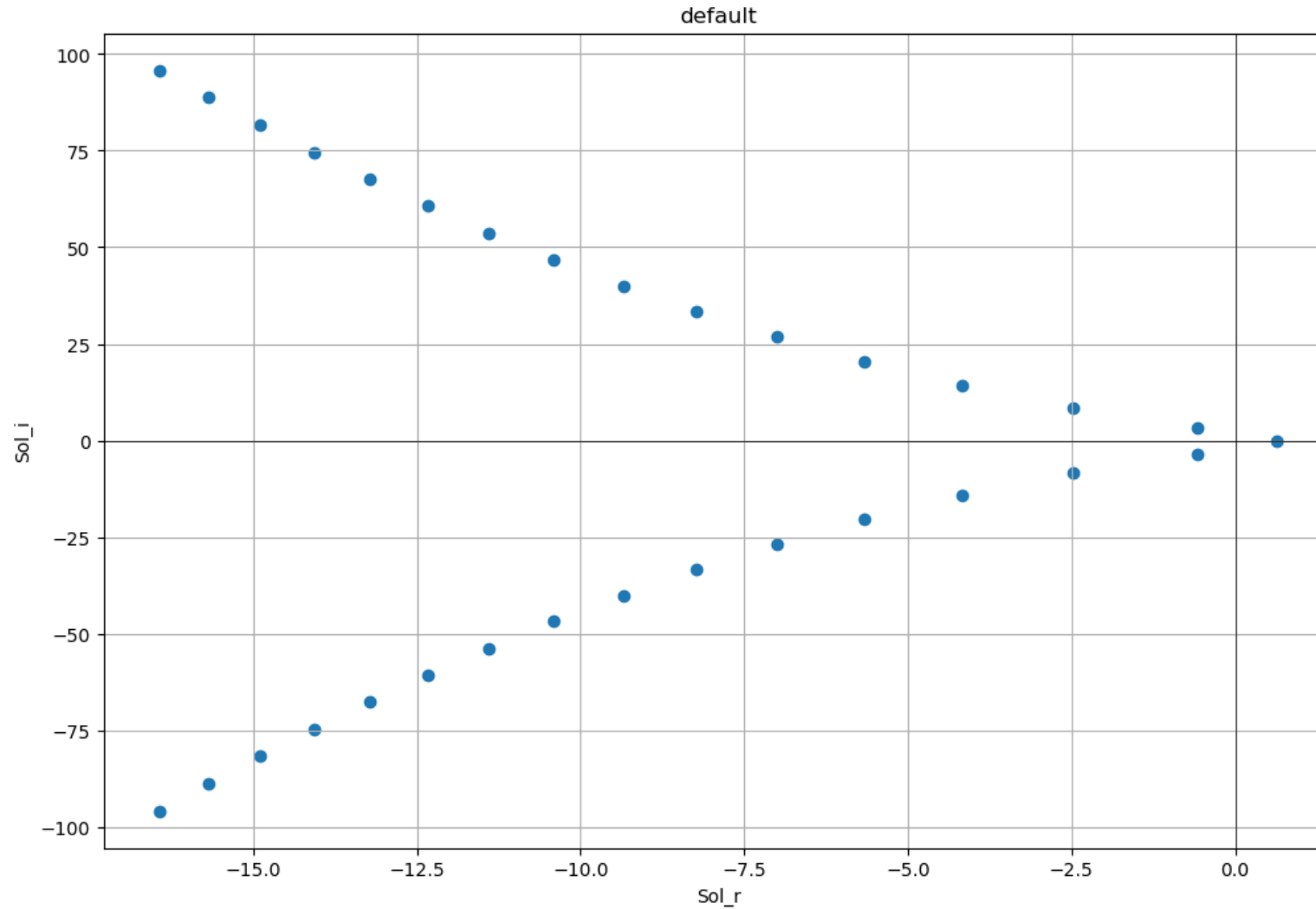
$$< \hat{A}\Phi, \Psi > = < \Phi, \hat{A}^*\Psi > \Rightarrow \left\{ \begin{array}{l} \hat{A}^*(\cdot) = \begin{bmatrix} D\partial_{\zeta\zeta}(\cdot) + v\partial_{\zeta}(\cdot) + k(\cdot) & 0 \\ 0 & -\frac{1}{\tau}\partial_{\zeta}(\cdot) \end{bmatrix} \\ B.C. : \begin{cases} D\partial_{\zeta}\phi^*(1) + v\phi^*(1) = Rv\psi^*(1) \\ \partial_{\zeta}\phi^*(0) = 0 \\ \psi^*(0) = \phi^*(0) \end{cases} \end{array} \right. \Rightarrow \hat{A} \neq \hat{A}^*$$

# Characteristic Equation

- Breaking 2<sup>nd</sup> order PDE in a system of two 1<sup>st</sup> order PDEs
- Solving for the eigenvalues:

$$\det(\bar{A}) = \det \left( \begin{bmatrix} -v & D & Rv \\ q_4 & q_5 & 0 \\ q_1 & q_2 & -q_9 \end{bmatrix} \right) = 0 \quad \text{where} \quad \begin{bmatrix} q_1 & q_2 & q_3 \\ q_4 & q_5 & q_6 \\ q_7 & q_8 & q_9 \end{bmatrix} = e^\Lambda \quad (9)$$

# Eigenvalue Distribution



# Eigenfunctions (Bi-orthogonal Basis)

$$\begin{cases} \phi_i(\zeta) = b_i \left[ \left( -\frac{r_{i,2} e^{r_{i,2}}}{r_{i,1} e^{r_{i,1}}} \right) e^{r_{i,1}\zeta} + e^{r_{i,2}\zeta} \right] \\ \psi_i(\zeta) = b_i \left( 1 - \frac{r_{i,2}}{r_{i,1}} \right) e^{r_{i,2}-\tau\lambda} e^{\tau\lambda\zeta} \end{cases}$$

$$\begin{cases} \phi_i^*(\zeta) = b_i^* \left[ -\frac{r_{i,2}^*}{r_{i,1}^*} e^{r_{i,1}^*\zeta} + e^{r_{i,2}^*\zeta} \right] \\ \psi_i^*(\zeta) = b_i^* \left( 1 - \frac{r_{i,2}^*}{r_{i,1}^*} \right) e^{-\tau\lambda\zeta} \end{cases}$$

- Normalizing coefficients may be found utilizing **Bi-orthogonality Theorem [2]** as follows:

If  $\{\Phi_n^*, n \geq 1\}$  are the eigenvectors of the adjoint of A corresponding to the eigenvalues  $\{\lambda_n, n \geq 1\}$ , then the eigenvectors can be suitably scaled such that  $\langle \Phi_n, \Phi_m^* \rangle = \delta_{mn}$

$$r_{1,2_i} = \frac{v \pm \sqrt{v^2 + 4D(\lambda_i - k)}}{2D} \quad (10)$$



# Dominant Modes (Eigenvalues)

- Using the following Corollary [2] of the previous Theorem makes way for determining the **Dominant Modes** of the system dynamics, by trying to reconstruct any arbitrary function that belongs to our function space with sufficient accuracy using a finite sum, instead of the infinite sum:

Every function  $z \in Z$  can be represented by the following infinite sum:

$$z = \sum_{n=1}^{\infty} \langle z, \Phi_n^* \rangle \Phi_n \quad (11)$$

# Dominant Modes (Eigenvalues)

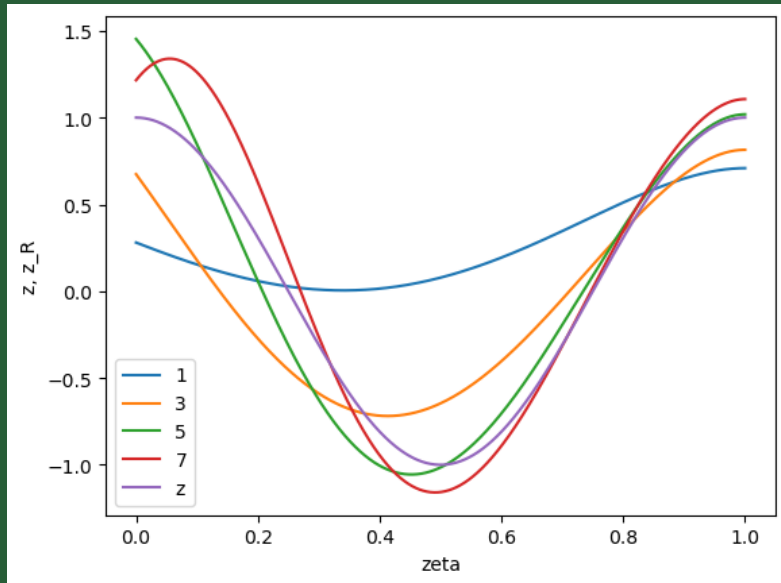


Figure 3 – Arbitrarily chosen function  $z(\zeta)$  and its reconstructed approximations  $z_R(\zeta)$

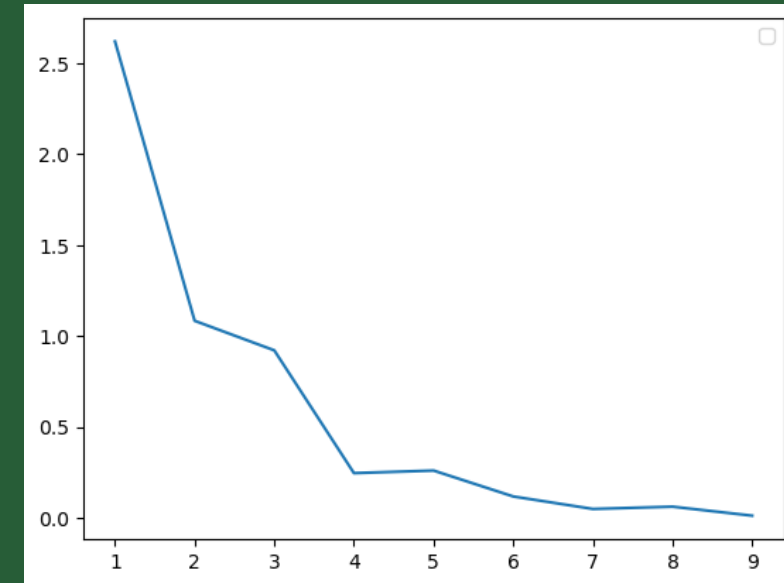


Figure 4 – Reconstruction error  $[z(\zeta) - z_R(\zeta)]^2$  against the number of modes used

7 modes chosen:  $(\lambda_i \text{ for } i = 1, \dots, 7)$

# 3- Controller Design

- Linear Quadratic (LQ) State Feedback Optimal Controller
- Continuous Algebraic Riccati Equation (ARE)

# LQ State Feedback Optimal Controller

- The goal is to find the input that minimizes the following cost function:

$$\mathbf{J}(u) = \int_0^{\infty} \langle \mathbf{x}, Q\mathbf{x} \rangle + \langle u, Ru \rangle dt \quad (12)$$

- The solution of this optimal control problem can be obtained by solving **Algebraic Riccati Equation (ARE)**:

$$\langle A^*P\mathbf{x}, \mathbf{y} \rangle + \langle PA\mathbf{x}, \mathbf{y} \rangle + \langle R^{-1}B^*P\mathbf{x}, B^*P\mathbf{y} \rangle + \langle Q\mathbf{x}, \mathbf{y} \rangle = 0 \quad (13)$$

- The quadratic cost function is minimized by the following optimal control law:

$$K\mathbf{x} = R^{-1}B^*P\mathbf{x} = R^{-1} \int_0^1 [k_1(\zeta), k_2(\zeta)] \cdot \mathbf{x}(\zeta) d\zeta \quad (14)$$

# Continuous Algebraic Riccati Equation

- For this system with boundary control, we can write:

$$\hat{B} = v(1 - R)[\delta_0(\zeta), 0]^T \quad (15)$$

$$\hat{B}^* = v(1 - R) \int_0^1 [\delta_0(\zeta), 0] \cdot \begin{bmatrix} (\cdot) \\ (\cdot) \end{bmatrix} \quad (16)$$

- Q** and **R** are also assumed to be as follows:

$$\mathbf{Q} = \begin{bmatrix} R & 0 \\ 0 & 1 - (1 - R)\zeta \end{bmatrix}; \quad \mathbf{R} = \mathbf{R}^{-1} = 1 \quad (17)$$

## 4- Results and Discussion

- Infinite Dimensional Feedback Gain
- Closed-loop System Input
- Input Response of the System (Stability)

# Infinite Dimensional Feedback Gain

- ARE is solved numerically to get  $P$
- $K(\zeta) = [k_1(\zeta), k_2(\zeta)]$  will be obtained accordingly as functions of space
- $u(t) = \langle K(\zeta) \cdot X(\zeta, t) \rangle$
- Feedback Gain shown in Figure 5:

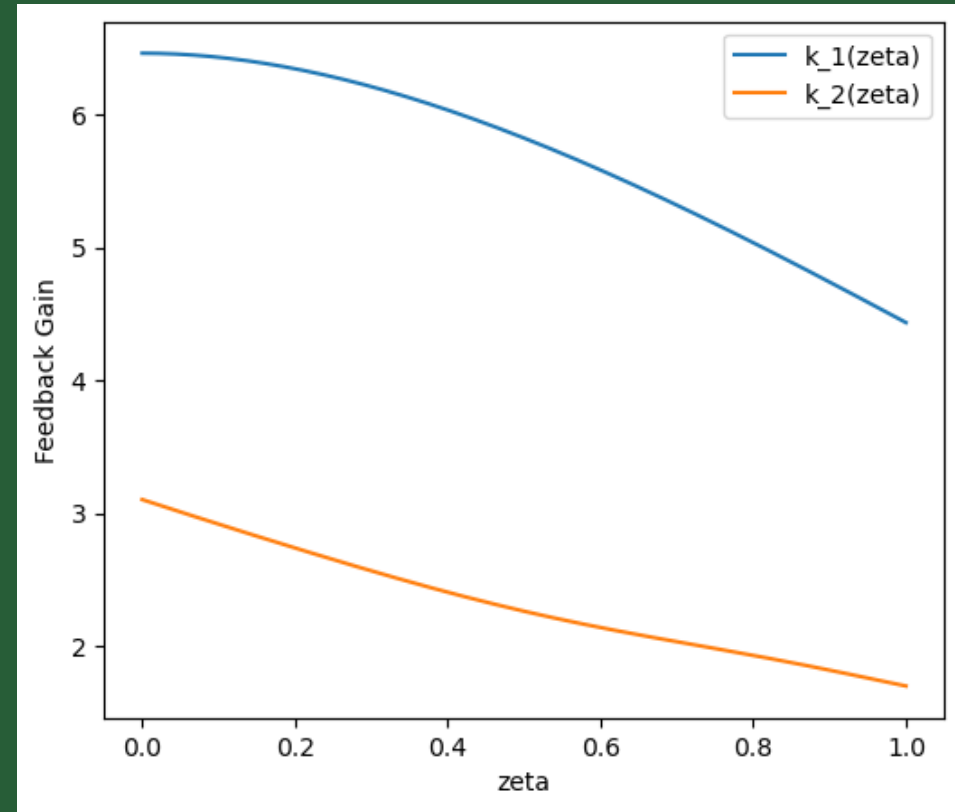


Figure 5 – Feedback Gain  $K(\zeta) = [k_1(\zeta), k_2(\zeta)]$  versus  $\zeta$

# Input Response (Stability)

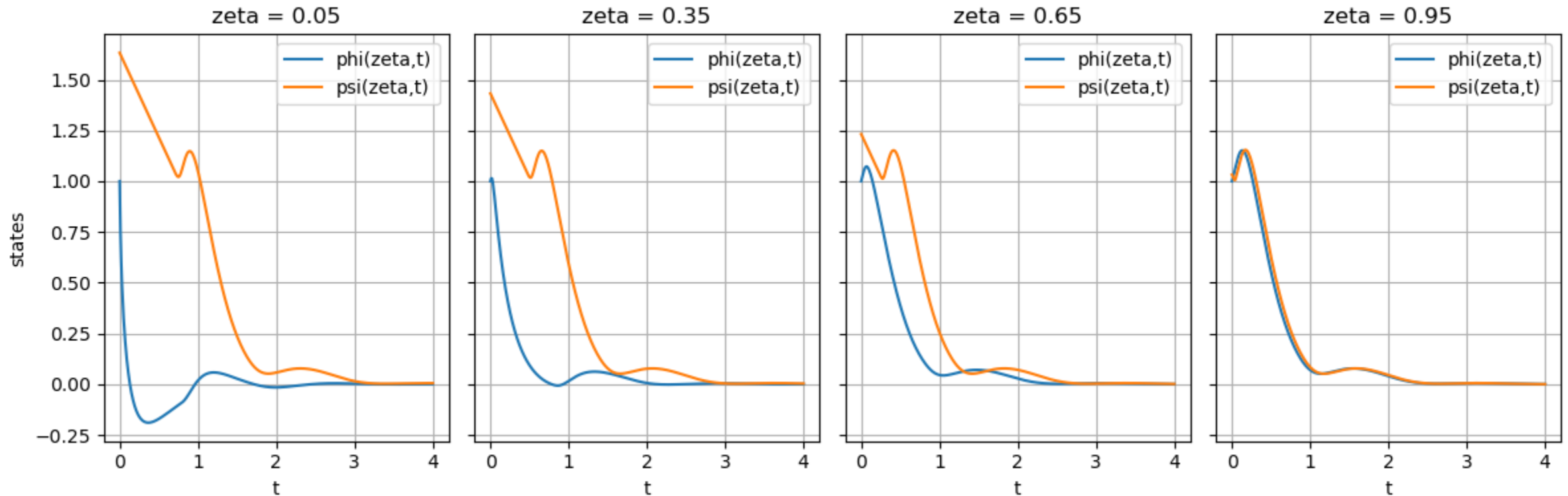


Figure 6 – Input response vs. time at four different locations of the reactor and the recycle stream – i.e.  $\phi(zeta, t)$  and  $\psi(zeta, t)$ , respectively



# Input Response (Stability)

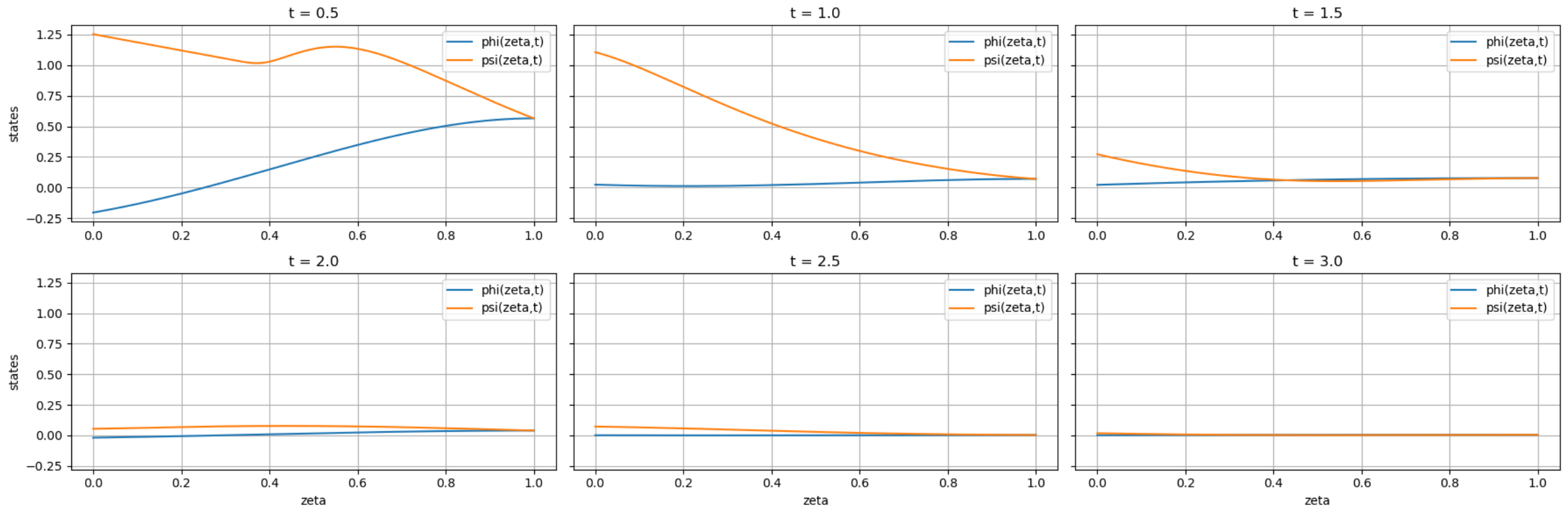


Figure 7 – Input response vs. zeta at six different times for  $\phi(zeta,t)$  and  $\psi(zeta,t)$

# Conclusion

- **Open-loop System:**

- Notion of Delay and Pure Transport
- Axial Dispersion Tubular Reactor with Delayed Recycle
- A 3x3 system of coupled PDEs
- Late Lumping

- **Closed-loop System:**

- Infinite-horizon Linear Quadratic State Feedback Optimal Control
- Boundary Control
- Stability is achieved

# Further Considerations and Future Work

- Study the effect of Parameters on Eigenvalue Distribution
- State Reconstruction to address Full State Feedback issue
- Study the limitations of addressing the system through Early Lumping Methods
- Considering Constraints and Designing MPC

# References

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# THANK YOU!

Any Questions?



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