

**THESIS TITLE**

by

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# Abstract

This guide is a comprehensive resource crafted for University of Alberta students tackling their theses with LaTeX. It is designed to simplify the thesis-writing process by offering a detailed walkthrough of a custom LaTeX template tailored to meet the university's formatting requirements. The guide goes beyond mere template usage, providing step-by-step instructions, tips, and best practices for creating a well-structured thesis that adheres to academic standards.

It also delves into advanced customization techniques, allowing users to achieve specific stylistic elements and formatting nuances that suit their individual needs. Alongside these, the guide introduces JabRef, a robust reference management tool, and explains how to seamlessly integrate it with LaTeX to streamline citation management. Whether you are a novice or an experienced LaTeX user, this guide equips you with the knowledge and tools necessary to produce a polished, professional thesis. By following its clear and concise instructions, students can confidently navigate the complexities of thesis writing, ensuring their work is both technically sound and perfectly laid out. Test.

# Preface

Writing a thesis is no small task, and as a graduate student at the University of Alberta, I quickly realized just how challenging it can be to meet all the formatting requirements while also producing a document that looks professional. Like many others, I started out using traditional word processors, but it didn't take long before I ran into the usual headaches—crashes, file corruption, and formatting issues that seemed to have a mind of their own.

These frustrations led me to explore  $\text{\LaTeX}$  as an alternative. I discovered that  $\text{\LaTeX}$  not only provided a way to keep my content and formatting separate but also offered a more reliable and consistent way to produce a high-quality thesis. The learning curve was steep, but once I got the hang of it, I found it to be a game-changer.

The original template document was born out of my experience with  $\text{\LaTeX}$  and the desire to make the thesis-writing process a bit less daunting for others who might be in the same boat. My goal here with this new document is to provide a comprehensive guide that not only walks you through the basics of  $\text{\LaTeX}$  but also gives you practical examples and best practices<sup>1</sup> to follow.

I built this template and class from the ground up with the aim of reducing the typical  $\text{\LaTeX}$  learning curve. I have tried to keep it as simple as possible while still making it powerful enough to handle everything you'll need for a thesis. I hope this

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<sup>1</sup>These might not be the “best” practices, however, these are practices that I follow to make my work more constant.

document makes your life a little easier and that you find L<sup>A</sup>T<sub>E</sub>X as useful as I have.

**Disclaimer:** This guide provides practical insights and tips for writing a thesis, based on the information available at the time of writing. However, it remains *your responsibility* to verify the accuracy of all content, conclusions, and interpretations. Official university guidelines, requirements, and advice *always* takes precedence over this guide. I assume *no responsibility* for any potential inaccuracies, misrepresentations, or unintended consequences resulting from the use of any material presented in this guide or the supporting documentation and files.

To...

*“Etiam ac leo a risus tristique nonummy. Donec dignissim tincidunt nulla. Vestibulum rhoncus molestie odio. Sed lobortis, justo et pretium lobortis, mauris turpis condimentum augue, nec ultricies nibh arcu pretium enim. Nunc purus neque, placerat id, imperdiet sed, pellentesque nec, nisl. Vestibulum imperdiet neque non sem accumsan laoreet. In hac habitasse platea dictumst. Etiam condimentum facilisis libero. Suspendisse in elit quis nisl aliquam dapibus. Pellentesque auctor sapien. Sed egestas sapien nec lectus. Pellentesque vel dui vel neque bibendum viverra. Aliquam porttitor nisl nec pede. Proin mattis libero vel turpis. Donec rutrum mauris et libero. Proin euismod porta felis. Nam lobortis, metus quis elementum commodo, nunc lectus elementum mauris, eget vulputate ligula tellus eu neque. Vivamus eu dolor.”*

*- Author of the Quote*

# Acknowledgements

While I would love to acknowledge every individual who has influenced this document, I realize that doing so might mean this document never gets finished.

I want to extend my deepest gratitude to my friends and family for their unwavering support throughout all my accomplishments.

A special thank-you goes out to the developers and other L<sup>A</sup>T<sub>E</sub>X users who have contributed to the L<sup>A</sup>T<sub>E</sub>X community. Without their work, this guide wouldn't exist in its current form.

I am also incredibly thankful to everyone who provided feedback and support along the way, including those new to L<sup>A</sup>T<sub>E</sub>X. Your fresh perspectives were invaluable in motivating the transformation from the previous version of this template into a full-fledged class, template, and this newly added guide to writing a thesis, complete with examples of the most requested elements to include.

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# List of Symbols

## Constants

$\epsilon_0$	Permittivity of Free Space.	$\epsilon_0 = 8.854 \times 10^{-12} \text{ F/m}$
$\hbar$	Reduced Planck Constant.	$\hbar = 1.055 \times 10^{-34} \text{ Js}$
$\mu_0$	Permeability of Free Space.	$\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$
$\pi$	Mathematical Constant Pi.	$\pi \approx 3.14159$
$R_e$	Rankine Number.	$R_e = \frac{Lv\rho}{\mu}$
$c$	Speed of light in a vacuum.	$299,792,458 \text{ m/s}$
$g$	Acceleration due to Gravity.	$g = 9.81 \text{ m/s}^2$
$h$	Planck constant.	$6.62607015E-34 \text{ Js}$
$k$	Boltzmann Constant.	$k = 1.380649 \times 10^{-23} \text{ J/K}$
$R$	Gas Constant.	$R = 8.314 \text{ J/(mol}\cdot\text{K)}$

## Latin

$A$	Cross-sectional Area.
$a$	Acceleration
$D$	Diameter.
$d$	Distance
$E$	Young's Modulus
$F$	Force.



$G$	Shear Modulus.
$I$	Area Moment of Inertia.
$K$	Elastic Constant
$L$	Length.
$M$	Moment.
$m$	Mass
$P$	Pressure.
$T$	Temperature.
$T$	Torque
$t$	Thickness.
$V$	Volume.
$v$	Velocity

**Greek**

$\alpha$	Primary Angle
$\delta$	Deflection.
$\lambda$	Wavelength.
$\sigma$	Normal Stress.
$\tau$	Shear Stress.
$\varepsilon$	Strain

# Abbreviations

**AC** Armor Class.

**AL** Adventurers League.

**AoE** Area of Effect.

**AoO** Attack of Opportunity.

**BAB** Base Attack Bonus.

**BBEG** Big Bad Evil Guy.

**CHA** Charisma.

**CON** Constitution.

**CoS** Curse of Strahd.

**CR** Challenge Rating.

**CRPG** Computer Role-Playing Game.

**D&D** Dungeons & Dragons.

**DEX** Dexterity.

**DM** Dungeon Master.

**DMG** Dungeon Master's Guide.

**DMPC** Dungeon Master Player Character.

**DnD** Dungeons & Dragons.

**DPR** Damage Per Round.

**DR** Damage Reduction.

**ECL** Effective Character Level.

**HD** Hit Dice.

**HP** Hit Points.

**IC** In Character.

**INT** Intelligence.

**LA** Level Adjustment.

**LFG** Looking For Group.

**LoS** Line of Sight.

**LR** Long Rest.

**MM** Monster Manual.

**NPC** Non-Player Character.

**OGL** Open Game License.

**OOO** Out Of Character.

**PB** Proficiency Bonus.

**PC** Player Character.

**PHB** Player's Handbook.

**PP** Passive Perception.

**RAI** Rules As Intended.

**RAW** Rules As Written.

**RNG** Random Number Generator.

**RP** Roleplaying.

**SR** Spell Resistance.

**SRD** System Reference Document.

**STR** Strength.

**THAC0** To Hit Armor Class 0.

**TPK** Total Party Kill.

**TTRPG** Tabletop Role-Playing Game.

**VTT** Virtual Tabletop.

**WIS** Wisdom.

**WotC** Wizards of the Coast.

**XP** Experience Points.

# Glossary of Terms

**Ability Score** One of six numbers (Strength, Dexterity, Constitution, Intelligence, Wisdom, Charisma) that represent a character's physical and mental attributes.

**AC (Armor Class)** A number representing how difficult it is to hit a character in combat.

**Advantage/Disadvantage** A mechanic where a player rolls two d20s and takes the higher (advantage) or lower (disadvantage) result.

**Alignment** A character's ethical and moral perspective, such as Lawful Good or Chaotic Evil.

**Arcane** A type of magic derived from study, such as wizardry.

**Backstory** The history and background of a character before the campaign begins.

**Bonus Action** An additional action a character can take during their turn, often granted by class features or spells.

**Cantrip** A spell that can be cast at will without using a spell slot.

**Chaotic** Free-spirited and sometimes unpredictable. Can also be reckless or reactionary.

**Charisma (CHR)** Social skills and sometimes physical appearance.

**Class** The primary archetype of a character, such as Fighter, Wizard, or Rogue, which determines abilities and progression.

**Combat** A structured sequence where characters and enemies take turns performing actions like attacking or casting spells.

**Concentration** A mechanic where certain spells require ongoing focus, and taking damage can force a concentration check to maintain the spell.

**Constitution (CON)** Physical resilience. This affects hit points and some physical resistances.

**d20** A 20-sided die, the primary die used in D&D for most rolls.

**Damage Types** One of the thirteen (13) categories of damage: acid, bludgeoning, cold, fire, force, lightning, necrotic, piercing, poison, psychic, radiant, slashing, and thunder.

**Dexterity (DEX)** Agility and accuracy. This affects ranged attacks and dodging.

**Dungeon Master (DM)** The person who runs the game, narrates the story, and controls the world and NPCs.

**Encounter** Any situation where players must overcome a challenge, such as combat, puzzles, or social interaction.

**Equipment** The gear and items a character carries, including weapons, armor, and adventuring tools.

**Evil** Wicked and often selfish or oppressive.

**Experience Points (XP)** Points gained from overcoming challenges, used to level up a character.

**Familiar** A magical creature that assists a spellcaster, often summoned by the spell *Find Familiar*.

**Feat** A special ability or skill a character can choose instead of an ability score increase at certain levels.

**Flanking** A tactical position where a character attacks an enemy from the opposite side of an ally, often granting a combat advantage (this rule is optional and varies by DM).

**Good** Having a respect for life, altruism, and selflessness.

**Grapple** A combat action where a character attempts to grab and restrain an opponent.

**Group Check** A mechanic where the success of the party depends on the number of successful rolls among the group.

**Hit Points (HP)** A measure of a character's health, reduced when taking damage.

**Hit Dice (HD)** Dice used to determine a character's hit points at each level and for healing during short rests.

**Initiative** A roll made at the start of combat to determine the order of turns..

**Inspiration** A DM-awarded bonus that allows a player to gain advantage on a roll.

**Intelligence (INT)** The ability to process problems and wield certain magic. INT affects the number of skill points received.

**Ki** A resource used by monks to perform special abilities.

**Lawful** Abides by a core morality or honor system. Can also be judgmental and close-minded.

**Level** A measure of a character's progression, determining access to new abilities, spells, and increased hit points.

**Long Rest** A period of downtime (usually 8 hours) where characters recover hit points and spell slots.

**Melee** Combat at close range, typically involving hand-to-hand or short-ranged weapons.

**Metagaming** Using out-of-game knowledge within the game, often discouraged as it can break immersion.

**Multiclassing** The practice of taking levels in more than one class, allowing a character to gain abilities from multiple classes.

**Neutral** A balance between Lawful & Chaotic or Good & Evil.

**NPC (Non-Player Character)** Characters controlled by the DM that players interact with, such as villagers, shopkeepers, or enemies.

**Opportunity Attack** A reaction that allows a character to make a melee attack against a creature that moves out of their reach.

**Party** The group of player characters (PCs) adventuring together.

**Perception** A skill representing a character's ability to notice hidden things, typically rolled as a Wisdom check.

**Proficiency Bonus** A bonus added to rolls where a character has proficiency, such as with certain skills, weapons, or saving throws.

**Quiver** A container for holding arrows or bolts, typically used by archers and ranged combatants.

**Ranged Attack** An attack made with a ranged weapon or spell, targeting an enemy at a distance.

**Reaction** An instant response to a trigger, such as casting *Counterspell* or making an opportunity attack.

**Saving Throw** A roll made to resist a spell, trap, or other effect.

**Short Rest** A brief period of downtime (usually 1 hour) where characters can spend Hit Dice to recover hit points.

**Skill Check** A roll made to determine the outcome of an action related to a skill, such as Stealth or Acrobatics.

**Spell Slot** A resource that determines how many spells a character can cast at each level.

**Strength (STR)** The character's physical strength. This effects the potency of melee attacks.

**Turn** A player's time to act during a round of combat, typically consisting of movement, an action, and possibly a bonus action or reaction.



**Vision Types** Various levels of sight in D&D, such as Darkvision, Blindsight, and Truesight.

**Weapon Proficiency** Determines which weapons a character can use effectively, adding their proficiency bonus to attack rolls.

**Wisdom** An ability score representing a player character's insight, perception, and willpower.

**Wisdom (WIS)** Common sense and spirituality.

**XP (Experience Points)** See Experience Points.

# Chapter 1

## Introduction

### 1.1 Background

As a graduate student from the University of Alberta, I am familiar with the challenging task of writing a thesis that adheres to the GPS Minimum Thesis Formatting Requirements. Using a traditional word processor to create a long document filled with equations and figures can be frustrating due to frequent crashes, file corruption, unpredictable formatting changes, and the inability to output a document in the required PDF/A format for submission to GPS.

To overcome these issues, many students turn to  $\text{\LaTeX}$  as an alternative to conventional word processors.

$\text{\LaTeX}$  allows students and researchers to focus separately on the content and the formatting of their documents. Because the writing is independent of the formatting, documents can be written in lightweight text editors or  $\text{\LaTeX}$  editors, which also facilitate the compilation of the documents. These editors can often save work after every keystroke, and due to the plaintext format, they are less prone to file corruption. Moreover,  $\text{\LaTeX}$  ensures a consistent and professional appearance throughout the document.

## 1.2 Objectives

The main objectives of this thesis are:

1. To provide a comprehensive guide on writing a thesis using L<sup>A</sup>T<sub>E</sub>X.
2. To assist students and researchers in mastering the nuances of L<sup>A</sup>T<sub>E</sub>X document preparation.
3. To showcase best practices for structuring and formatting a thesis in L<sup>A</sup>T<sub>E</sub>X.

## 1.3 Scope and Limitations

Although there are existing templates for writing a thesis in L<sup>A</sup>T<sub>E</sub>X for the University of Alberta (see [Table 1.1](#) for a list of available templates), none seem to provide all the necessary information for creating an outstanding thesis. Most templates apply “band-aid” solutions to existing classes, such as `report` or `book`, offering a customized title page and methods for including prefatory pages. However, these templates often fall short by not providing tips and best practices on how to include the various sections and parts that make up a thesis. They also fail to offer a solid foundation for those who are new to L<sup>A</sup>T<sub>E</sub>X. Many of these templates involve extensive patching and fixing, resulting in a large *preamble* section at the beginning of the template that can be confusing to new L<sup>A</sup>T<sub>E</sub>X users and add to the already steep learning curve.

Developer	Last Updated	Link to Template Source
Shivam Garg	May 29, 2023	<a href="https://github.com/svmgrg/ualberta_thesis_template">https://github.com/svmgrg/ualberta_thesis_template</a>
Henry Brausen	Feb 11, 2022	<a href="https://github.com/henrybrausen/thesis_template">https://github.com/henrybrausen/thesis_template</a>
Bernard Llanos	Oct 05, 2019	<a href="https://drive.google.com/file/d/1wKS8fu5e6qiVDRt0VUzEtIW8p7uMyz1T/view?usp=sharing">https://drive.google.com/file/d/1wKS8fu5e6qiVDRt0VUzEtIW8p7uMyz1T/view?usp=sharing</a>
John Bowman	Sep 30, 2019	<a href="https://github.com/vectorgraphics/uofathesis">https://github.com/vectorgraphics/uofathesis</a>
Hongtao Yang & Benjamin Bernard	Sep 28, 2017	<a href="https://github.com/adrs0049/ThesisTemplate">https://github.com/adrs0049/ThesisTemplate</a>
GAME & Hongtao Yang	Feb 03, 2016	<a href="https://www.ualberta.ca/computing-science/media-library/grad/candidacy-template-tex.tex">https://www.ualberta.ca/computing-science/media-library/grad/candidacy-template-tex.tex</a>
Steven Taschuk	Mar 21, 2012	<a href="https://github.com/stebulus/ualberta-math-stat-templates/tree/master/thesis">https://github.com/stebulus/ualberta-math-stat-templates/tree/master/thesis</a>
CMENG	Jul 19, 1999	<a href="https://sites.ualberta.ca/CMENG/research/new-control/stythes.html">https://sites.ualberta.ca/CMENG/research/new-control/stythes.html</a>

Table 1.1: List of Other Available Templates.

This template, document class, and guide aim to address these shortcomings by providing all the necessary information to create a well-structured thesis, along with examples to assist in formatting your thesis written in L<sup>A</sup>T<sub>E</sub>X. To ensure the robustness and ease of maintenance, I developed the class file from the ground up keeping the additional required packages to a minimum. This makes this L<sup>A</sup>T<sub>E</sub>X solution easier to maintain, update, and customize to suit different needs from different areas of the University of Alberta. A key goal with this was to reduce the traditionally steep learning curve associated with L<sup>A</sup>T<sub>E</sub>X to ensure that anyone could create an outstanding thesis.

While the class file (`ualberta.cls`) deserves its own comprehensive documentation, this document will focus on more specifically on the template file (`ualberta.tex`), as well as the following points:

- Installation and basic usage of L<sup>A</sup>T<sub>E</sub>X.
- Document structure and formatting.
- Inclusion of figures and tables.
- Inclusion of plots and graphs.
- Handling mathematical equations.
- Citations and references using BibTeX.
- Use of JabRef—Reference Manager.
- Inclusion of Code and PDF's.
- And more.

This guide does not cover advanced L<sup>A</sup>T<sub>E</sub>X programming or extensive customization of document classes. Instead, the class file `ualberta.cls` provides all the major

document and formatting requirements as provided by GPS, while this document offers references on how to include the various elements that might be required in a thesis. This includes all of the explanations of the packages and macros needed to perform the examples.

## 1.4 Organization of the Thesis

The thesis is organized into several chapters, each addressing a specific aspect of writing a thesis in L<sup>A</sup>T<sub>E</sub>X. The breakdown is as follows:

- ??
- ??
- **Chapter 4:** Figures, Tables, & Plates
- **Chapter 5:** Plots, Charts, & Graphs
- **Chapter 6:** Mathematical Equations
- **Chapter 7:** Citations, References, and Cross-References
- **Chapter 9:** JabRef: Managing Bibliographies Efficiently

Each chapter provides detailed information, examples, and recommendations to help you navigate the thesis writing process within the L<sup>A</sup>T<sub>E</sub>X ecosystem.

## 1.5 Summary

This chapter introduced the background, objectives, scope, and organization of the thesis. The subsequent chapters delve into specific topics, offering practical guidance and examples for mastering the art of writing a thesis in L<sup>A</sup>T<sub>E</sub>X. Through this process,

you will develop an understanding of how to create a thesis and manipulate content within the L<sup>A</sup>T<sub>E</sub>X ecosystem to produce an exceptional document.

## Chapter 2

# Continuous-time Control for Isothermal System

### 2.1 INTRODUCTION

Many chemical, petrochemical, and biochemical unit operation processes are modelled as distributed parameter systems (DPS), ranging from tubular reactors, heat exchangers, and separation columns to processes like digesters in the pulp and paper industry and fluid flow in pipeline networks. When these processes are described using first-principle modelling, they result in a class of partial differential equations (PDEs) to effectively capture diffusion, transport, and reaction phenomena, leading to infinite-dimensional state space representations [1]. This characteristic presents significant challenges, making the control and estimation of DPS inherently more complex than finite-dimensional systems. Two primary methods have emerged for addressing DPS control. One is early lumping, which approximates the infinite-dimensional system with a finite-dimensional model [2, 3]. While this method enables the use of standard regulator design techniques, mismatches between the dynamical properties of the original DPS and the approximate lumped parameter model can occur, negatively affecting the performance of the designed regulator [4]. The second method is late lumping, which directly tackles the infinite-dimensional system before applying



numerical solutions. This approach introduces a challenging yet fertile direction of research, leading to many meaningful contributions that address various aspects of control and estimation of infinite-dimensional systems.

Among notable studies utilizing the lumping method for control of convection-reaction chemical systems resulting in first order hyperbolic PDEs, Christofides explored the robust control of quasi-linear first-order hyperbolic PDEs, providing explicit controller synthesis formulas for uncertainty decoupling and attenuation [5]. Krstic and Smyshlyaev extended boundary feedback stabilization techniques for first-order hyperbolic PDEs using a backstepping method, converting the unstable PDE into a system for finite-time convergence [6]. Relevant applications of reaction-convection systems other than tubular reactors have also been addressed within this field, resulting in regulator/observer design strategies for chemical systems governed by first order hyperbolic PDEs. Xu and Dubljevic addressed the state feedback regulator problem for a countercurrent heat exchanger system, utilizing an infinite-dimensional approach to ensure that the controlled output tracks a reference signal [7]. Xie and Dubljevic developed a discrete-time output regulator for gas pipeline networks, emphasizing the transformation of continuous-time models into discrete-time systems while preserving essential continuous-time properties [8]. This work was further extended by Zhang *et al.*, who proposed a tracking model predictive control and moving horizon estimation design for pipeline systems, addressing the challenges of state and parameter estimation in an infinite-dimensional chemical system governed by first order hyperbolic PDEs [9]. For a similar convection-reaction system, Zhang *et al.* proposed a model predictive control strategy, incorporating a Luenberger observer to achieve output constrained regulation in a system modelled by nonlinear coupled hyperbolic PDEs [10].

Additionally, diffusion-convection-reaction systems resulting in parabolic PDEs are

also addressed in several works. For example, Christofides addressed order reduction methods for diffusion-convection-reaction type of reactors [11]. Djurjic *et al.* utilized modal decomposition to capture dominant modes of a DPS to construct a reduced order finite dimensional system, which enables the design of a low dimensional controller for a diffusion-convection-reaction type reactor described by second order parabolic PDEs [12]. Cassol *et al.* designed and compared the performance of a full-state and output feedback controller for a diffusion-convection heat exchanger system [13]. In Khatibi *et al.*'s work, an axial dispersion tubular reactor equipped with recycle stream is considered as a second order parabolic DPS, with a predictive controller being utilized to optimally control the reactor. Although the presence of recycle is common in industrial reactor designs, their study has thus far been one of the few contributions in this field addressing a diffusion-convection-reaction system equipped with a recycle stream [14].

Moreover, continuous-time optimal control design is a well-developed concept for distributed parameter systems, particularly when the system generator is either a self-adjoint operator or can be transformed into one through a proper linear transformation [15]. However, there are distributed parameter systems that do not possess this property. Instead, the system generator belongs to the class of Riesz-spectral operators. Rather than an orthonormal basis for the function-space, these generators introduce a bi-orthonormal set of eigenfunctions as the basis. Optimal controller design for these systems was initially addressed by Curtain and Zwart [16]. Since then, significant work has been done in this field; for instance, continuous-time optimal control design for a cracking catalytic reactor, another convection-reaction system governed by first-order hyperbolic PDEs, has been achieved by solving an operator Riccati equation (ORE)[17]. This work has been further extended to time-varying PDEs of the same class[17]. The same approach has been applied to develop a full-

state feedback[18] and output feedback[19] linear quadratic (LQ) optimal regulator for a boundary-controlled convection-reaction system, utilizing the properties of a Riesz-spectral generator for the system.

On top of those dynamic systems that are distributed in space, delay systems are another example of distributed parameter systems [16]. Although delay is commonly represented in the form of delay differential equations (DDEs), it can also be modelled as a transport partial differential equation (PDE), which offers advantages in more complex scenarios or when employing alternative norms on infinite-dimensional states. This approach allows for a smoother transition to problems involving more intricate PDE dynamics while maintaining notational consistency [20]. Input/output delay with relevant applications in chemical engineering has been addressed previously in the field of control theory for DPS. For example, time-delayed boundary observation is considered while addressing an output feedback regulator for a tubular reactor [21]. However, the notion of state-delay (as opposed to delayed-inputs or delayed-measurements) seems to be less addressed in this field compared to other relevant fields like signal processing, self-driving cars, or network control theory (NCT). This is probably due to the fact that not many applications in the field of distributed parameter chemical engineering systems can be described by state delays in the first place. Cassol *et al.*'s work is one of the few instances that addressed a delayed-state distributed parameter chemical engineering system, where they designed a full-state and output feedback regulator for a system of heat exchangers [13]. The notion of state-delay comes from the time it takes for a stream to leave one pass of the heat exchanger and enter the next pass. As stated previously, not much work is published addressing chemical reactors equipped with recycle as distributed parameter systems. Even in Khatibi *et al.*'s work, the recycle is assumed to be instantaneous [14]; a simplifying assumption that does not resonate well with reality. In fact, taking the time

it takes for the recycle stream to re-enter the reactor input can be another instance for the rare concept of a delayed state DPS in the field of chemical engineering. In another attempt, Qi *et al.* addressed the challenge of state delay imposed by a recycle stream in a system modelled by interconnected first-order hyperbolic partial integro differential equations (PIDEs), introducing a transport PDE to account for the in-domain recycle delay [22]. However, the diffusion term was not addressed, leaving a gap in the literature regarding diffusion-convection-reaction systems with a recycle stream imposing state delay.

The present work focuses on the control of an axial tubular reactor equipped with a recycle stream—a configuration common in industrial processes such as catalytic reactors, polymerization units, and biochemical fermenters—but one that is inadequately addressed in the literature. Unlike previous studies that assumed instantaneous recycle, this work incorporates the time delay associated with the recycle stream re-entering the reactor, presenting a rare example of state-delay in the field of chemical engineering DPS. The model comprises a second-order parabolic PDE to capture the diffusion-convection-reaction nature of the reactor, coupled with a first-order hyperbolic PDE to account for the delay. The boundary conditions are chosen as Danckwerts boundary conditions, which are particularly suitable for this type of reactor. The system results in a non-self-adjoint operator. However, by utilizing the bi-orthogonal theorem, given that the generator is Riesz-spectral, a full-state feedback optimal LQ regulator is developed, followed by an output feedback regulator. The control feedback is derived by solving an operator Riccati equation (ORE) in order to implement a late lumping approach. Actuation and observation are applied at the boundaries, making it a boundary-actuated system involving finite-dimensional dynamics for an infinite-dimensional DPS. These contributions are presented in the following order: In **Section 2.2**, the system analysis is first addressed by modelling

the delay infinite-dimensional system (DPS) and transforming it into a system of coupled PDEs using the delay-transport approach. The system's characteristics are explored by examining the system generator and its eigenvalues, followed by analyzing the adjoint operator and its spectrum, which allows introducing the bi-orthogonal basis for the infinite-dimensional system. Consequently in **Section 2.3.1**, the design strategy for an optimal full-state regulator is developed by formulating the infinite-time horizon LQ control problem, converting the ORE into matrix Riccati equations (MRE), and calculating the feedback gain. Practical limitations of the full-state feedback mechanism are then addressed in **Section 2.3.2** by introducing a Luenberger observer for state reconstruction, followed by the design of an output feedback regulator. Finally, numerical simulations are illustrated in **Section 2.4** to showcase the results of the developed theoretical concepts, demonstrating closed-loop responses of the system equipped with both full-state feedback regulator and output feedback compensator in various settings.

## 2.2 OPEN-LOOP SYSTEM

### 2.2.1 System model

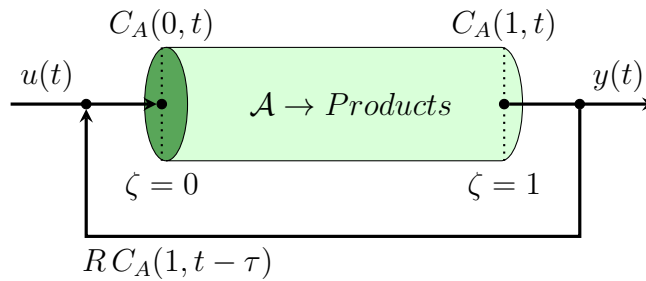


Figure 2.1: Axial tubular reactor with recycle stream.

The chemical process illustrated in Figure 2.1 represents an axial dispersion tubular reactor, which incorporates diffusion, convection, and a chemical reaction where

reactant  $A$  is converted into products [23]. The reactor is equipped with a recycle mechanism, allowing a fraction of the product stream to re-enter the reactor to ensure the consumption of any unreacted substrate. By applying first-principle modelling through relevant mass balance relations on an infinitesimally small section of the reactor, the dynamics of the reactant concentration can be described by the PDE given in Equation (2.1), belonging to the class of second order parabolic PDEs commonly used to characterize diffusion-convection-reaction systems [24] in chemical engineering.

$$\dot{C}_A(\zeta, t) = D\partial_{\zeta\zeta}C_A(\zeta, t) - v\partial_{\zeta}C_A(\zeta, t) + r(C_A) \quad (2.1)$$

Here,  $C_A(\zeta, t)$  denotes the concentration of reactant  $A$  along the reactor. The physical parameters  $D$  and  $v$  correspond to the diffusion coefficient and flow velocity along the reactor, respectively. It is worth noting that the system properties are assumed to be constant against changes in temperature and pressure. The spatial and temporal coordinates of the system are represented by  $\zeta$  and  $t$ , where  $\zeta \in [0, 1]$  and  $t \in [0, \infty)$ . In addition,  $r(C_A)$  is the reaction rate by which the reactant is consumed. Considering the reaction term in general can be non-linear, the model is further linearized around its steady-state, followed by replacing the reactant concentration  $C_A$  with its deviations from the steady-state concentration  $C_{A,ss}$ . The result is given in Equation (3.1).

$$\dot{c}(\zeta, t) = D\partial_{\zeta\zeta}c(\zeta, t) - v\partial_{\zeta}c(\zeta, t) - k_r c(\zeta, t) \quad (2.2)$$

where  $c(\zeta, t) \equiv C_A(\zeta, t) - C_{A,ss}(\zeta)$  is the deviation from the steady-state concentration and the linearized reaction coefficient is defined as  $k_r \equiv \left. \frac{\partial r(C_A)}{\partial C_A} \right|_{C_{A,ss}}$  in the vicinity of the steady-state. The system output is assumed to be the deviation of the reactant concentration from the steady-state measured at the reactor outlet, while the control input is set to be equal to the deviation of the reactant concentration

from the steady-state, applied at the reactor inlet after being mixed with the delayed state resulting from the recycled portion of the flow occurring  $\tau$  time units ago. Incorporating input, output, and state delay in addition to the assumption of Danckwerts boundary condition will result in Equation (3.3) that describe the boundary conditions of the system.

$$\left\{ \begin{array}{l} D\partial_{\zeta}c(0,t) - vc(0,t) = -v[Rc(1,t-\tau) + (1-R)u(t)] \\ \partial_{\zeta}c(1,t) = 0 \\ y(t) = c(1,t) \end{array} \right. \quad (2.3)$$

Here, parameters  $R$  and  $\tau$  correspond to the recycle ratio and the residence time along the recycle stream, respectively. Accounting for deviations from perfect mixing and piston flow and assuming negligible transport lags in connecting lines [25], the Danckwerts boundary conditions have become an inseparable part of modelling axial tubular reactors in the field of chemical engineering process control and dynamics. While capturing physical significance, Danckwerts boundary conditions maintain generality without unnecessarily simplifying the model as they belong to the general class of Robin boundary conditions.

### 2.2.2 PDE representation of the delay term

One effective method for addressing delay in systems is to represent the delay using an alternative transport partial differential equation (PDE). This approach is particularly advantageous when the problem already involves similar forms of PDEs, as is the case in the current study. To specifically address the delay in the system under consideration, the state variable  $c(\zeta, t)$  is expanded into a vector of functions  $x(\zeta, t) \equiv [x_1(\zeta, t), x_2(\zeta, t)]^T$ , where  $x_1(\zeta, t)$  is the same as  $c(\zeta, t)$ , while  $x_2(\zeta, t)$  is introduced as a new state variable to account for the concentration along the recycle stream. The delay is thus modelled as a pure transport process, as if the first state

$x_1(\zeta, t)$  is being transported from the reactor outlet to the inlet, experiencing a delay of  $\tau$  time units while in the recycle stream. As a result, Equations 3.1 and 3.3 may be re-formulated as follows:

$$\partial_t \begin{bmatrix} x_1(\zeta, t) \\ x_2(\zeta, t) \end{bmatrix} = \begin{bmatrix} D\partial_{\zeta\zeta} - v\partial_{\zeta} + k_r & 0 \\ 0 & \frac{1}{\tau}\partial_{\zeta} \end{bmatrix} \begin{bmatrix} x_1(\zeta, t) \\ x_2(\zeta, t) \end{bmatrix} \quad (2.4)$$

$$\begin{cases} D\partial_{\zeta}x_1(0, t) - vx_1(0, t) = -v[Rx_2(0, t) + (1 - R)u(t)] \\ \partial_{\zeta}x_1(1, t) = 0 \\ x_1(1, t) = x_2(1, t) \\ y(t) = x_1(1, t) \end{cases} \quad (2.5)$$

With all state variables now expressed explicitly at a specific time instance  $t$ —in contrast to the previous representation where states at  $t$  were directly involved with states at  $(t - \tau)$ —the open-loop system can be described in the standard state-space form of an infinite-dimensional linear time-invariant (LTI) system as  $\dot{x} = \mathfrak{A}x$ . Here,  $\mathfrak{A}$  is a linear operator  $\mathcal{L}(X)$  acting on a Hilbert space  $X : L^2[0, 1] \times L^2[0, 1]$  and  $x(\zeta, t)$ , as defined previously, is the vector of functions describing the states of the system. The operator  $\mathfrak{A}$  and its domain are defined in detail as shown in Equation (2.6):

$$\begin{aligned} \mathfrak{A} &\equiv \begin{bmatrix} D\partial_{\zeta\zeta} - v\partial_{\zeta} + k_r & 0 \\ 0 & \frac{1}{\tau}\partial_{\zeta} \end{bmatrix} \\ \mathcal{D}(\mathfrak{A}) &= \left\{ x = [x_1, x_2]^T \in X : x(\zeta), \partial_{\zeta}x(\zeta), \partial_{\zeta\zeta}x(\zeta) \quad \text{a.c.}, \right. \\ &\quad D\partial_{\zeta}x_1(0) - vx_1(0) = -v[Rx_2(0) + (1 - R)u], \\ &\quad \left. \partial_{\zeta}x_1(1) = 0, x_1(1) = x_2(1) \right\} \end{aligned} \quad (2.6)$$

### 2.2.3 Adjoint operator

The adjoint operator  $\mathfrak{A}^*$  plays a critical role in analyzing the spectral properties of the system. It is obtained in Equation (2.7):



$$\begin{aligned}
 \langle \mathfrak{A}\phi, \psi \rangle &= \langle \phi, \mathfrak{A}^*\psi \rangle \Rightarrow \\
 \mathfrak{A}^* &= \begin{bmatrix} D\partial_{\zeta\zeta} + v\partial_{\zeta} + k_r & 0 \\ 0 & -\frac{1}{\tau}\partial_{\zeta} \end{bmatrix} \\
 \mathcal{D}(\mathfrak{A}^*) &= \left\{ y = [y_1, y_2]^T \in Y : y(\zeta), \partial_{\zeta}y(\zeta), \partial_{\zeta\zeta}y(\zeta) \quad \text{a.c.}, \right. \\
 &\quad D\partial_{\zeta}y_1(1) + vy_1(1) = \frac{1}{\tau}y_2(1) \\
 &\quad Rvy_1(0) = \frac{1}{\tau}y_2(0) \\
 &\quad \left. \partial_{\zeta}y_1(0) = 0 \right\}
 \end{aligned} \tag{2.7}$$

where  $\phi_i(\zeta) = [\phi_{i,1}(\zeta), \phi_{i,2}(\zeta)]^T$  and  $\psi_i(\zeta) = [\psi_{i,1}(\zeta), \psi_{i,2}(\zeta)]^T$  are the eigenfunction of  $\mathfrak{A}$  and  $\mathfrak{A}^*$ , respectively. Given that  $\mathfrak{A}$  is not self-adjoint (i.e.,  $\mathfrak{A} \neq \mathfrak{A}^*$ ), their combined eigenmodes may still form a bi-orthonormal basis, typical of a Riesz-spectral operator [16]. Therefore their spectral properties must be determined by solving their characteristic equations.

### 2.2.4 Eigenvalue problem

The eigenvalue problem for  $\mathfrak{A}$  is formulated as follows:

$$\mathfrak{A}\phi_i(\zeta) = \lambda_i\phi_i(\zeta) \tag{2.8}$$

where  $\lambda_i \in \mathbb{C}$  is the  $i^{\text{th}}$  eigenvalue. To obtain the characteristic equation, the system of PDEs shall be reduced to the ODE system in Equation (2.9)  $\forall i \geq 0$ :

$$\partial_{\zeta} \begin{bmatrix} \phi_1 \\ \partial_{\zeta}\phi_1 \\ \phi_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ \frac{\lambda - k_r}{D} & \frac{v}{D} & 0 \\ 0 & 0 & \tau\lambda \end{bmatrix} \begin{bmatrix} \phi_1 \\ \partial_{\zeta}\phi_1 \\ \phi_2 \end{bmatrix} \tag{2.9}$$

which is in the form of  $\tilde{\phi}_{\zeta} = \tilde{\mathfrak{A}}\tilde{\phi}$ , with the solution stated in Equation (2.10):

$$\begin{bmatrix} \phi_1 \\ \partial_\zeta \phi_1 \\ \phi_2 \end{bmatrix}_{\zeta=1} = \begin{bmatrix} \Lambda_{1,1} & \Lambda_{1,2} & \Lambda_{1,3} \\ \Lambda_{2,1} & \Lambda_{2,2} & \Lambda_{2,3} \\ \Lambda_{3,1} & \Lambda_{3,2} & \Lambda_{3,3} \end{bmatrix} \begin{bmatrix} \phi_1 \\ \partial_\zeta \phi_1 \\ \phi_2 \end{bmatrix}_{\zeta=0} \quad (2.10)$$

where the  $3 \times 3$  matrix  $\Lambda_{(m,n)}$  is defined as  $\Lambda \equiv e^{\tilde{\mathfrak{A}}(\zeta-0)} \Big|_{\zeta=1}$ . By applying the boundary conditions to Equation (2.10), the algebraic system of equations in Equation (2.11) is obtained:

$$\begin{bmatrix} -v & D & Rv \\ \Lambda_{2,1} & \Lambda_{2,2} & \Lambda_{2,3} \\ (\Lambda_{1,1} - \Lambda_{3,1}) & (\Lambda_{1,2} - \Lambda_{3,2}) & (\Lambda_{1,3} - \Lambda_{3,3}) \end{bmatrix} \begin{bmatrix} \phi_1 \\ \partial_\zeta \phi_1 \\ \phi_2 \end{bmatrix}_{\zeta=0} = \tilde{\Lambda} \tilde{\phi} \Big|_{\zeta=0} = 0 \quad (2.11)$$

where  $\tilde{\Lambda}$  is defined as the square matrix shown in Equation (2.11). Equation (2.11) suggests that the matrix  $\tilde{\Lambda}$  must be rank-deficient for appropriate values of  $\lambda_i$ . Attempts to analytically solve the characteristic equation  $\det(\tilde{\Lambda}) = 0$  have failed; therefore, it is solved numerically using the parameters in Table 2.1. The resulting eigenvalue distribution is depicted in Figure 2.2 in the complex plane.

Following the same procedure for  $\mathfrak{A}^*$  shows that the eigenvalues of  $\mathfrak{A}$  match the ones of its adjoint, confirming that  $\mathfrak{A}$  and  $\mathfrak{A}^*$  form a bi-orthogonal basis according to Equation (2.12):

$$\begin{aligned} \langle \mathfrak{A}\phi_i, \psi_j \rangle &= \langle \lambda_i \phi_i, \psi_j \rangle = \lambda_i \langle \phi_i, \psi_j \rangle \\ \text{L.H.S.} &= \langle \phi_i, \mathfrak{A}^* \psi_j \rangle = \langle \phi_i, \lambda_j^* \psi_j \rangle = \overline{\lambda_j^*} \langle \phi_i, \psi_j \rangle \\ \lambda_i &= \overline{\lambda_i^*} \Rightarrow \langle \phi_i, \psi_j \rangle = \delta_{ij} \end{aligned} \quad (2.12)$$

The eigenfunctions  $\{\phi_i(\zeta), \psi_i(\zeta)\}$  (for  $\mathfrak{A}$  and  $\mathfrak{A}^*$ , respectively) may be obtained following the calculation of eigenvalues. The first 3 eigenfunctions are plotted in Figure 2.3.

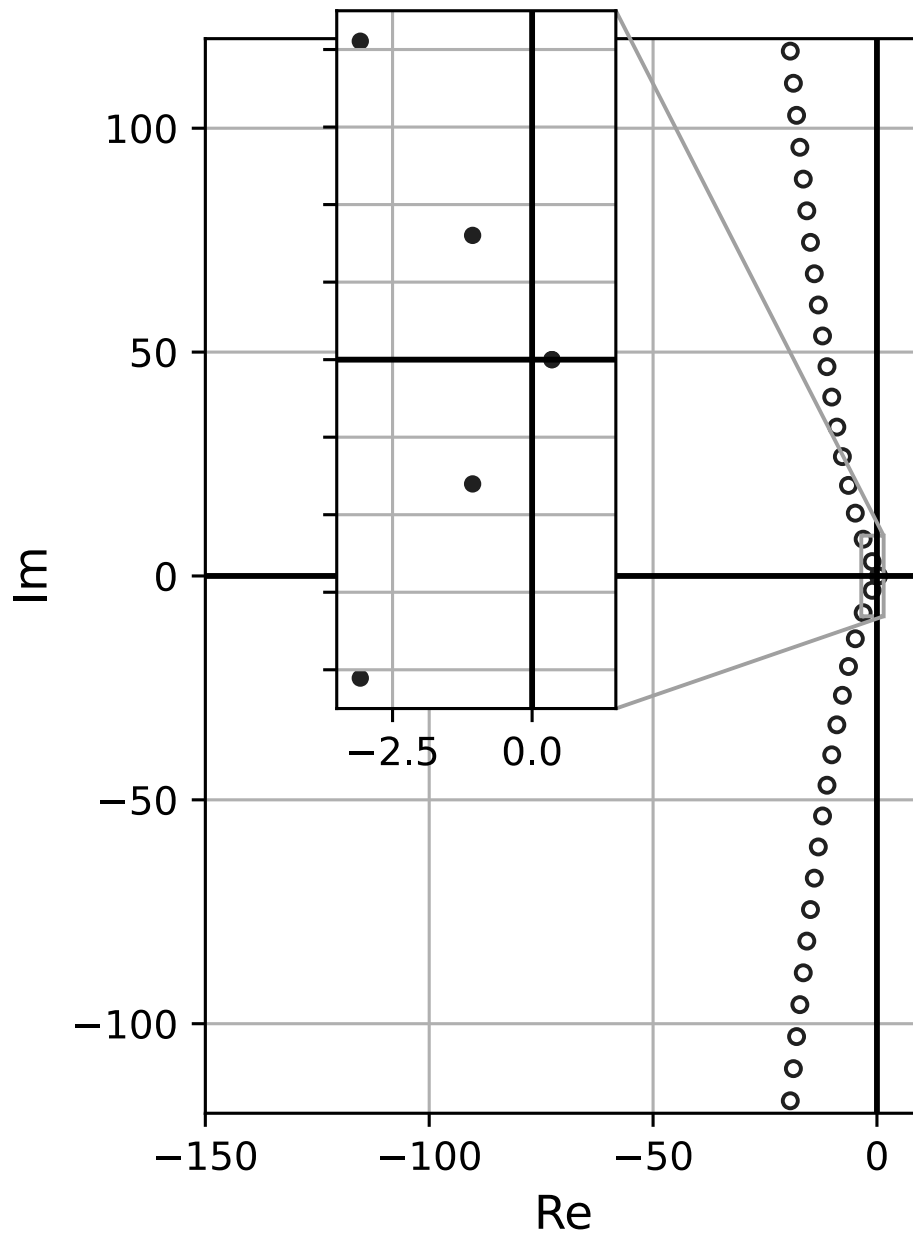


Figure 2.2: Eigenvalues of operator  $\mathfrak{A}$  obtained by solving Equation (2.11).

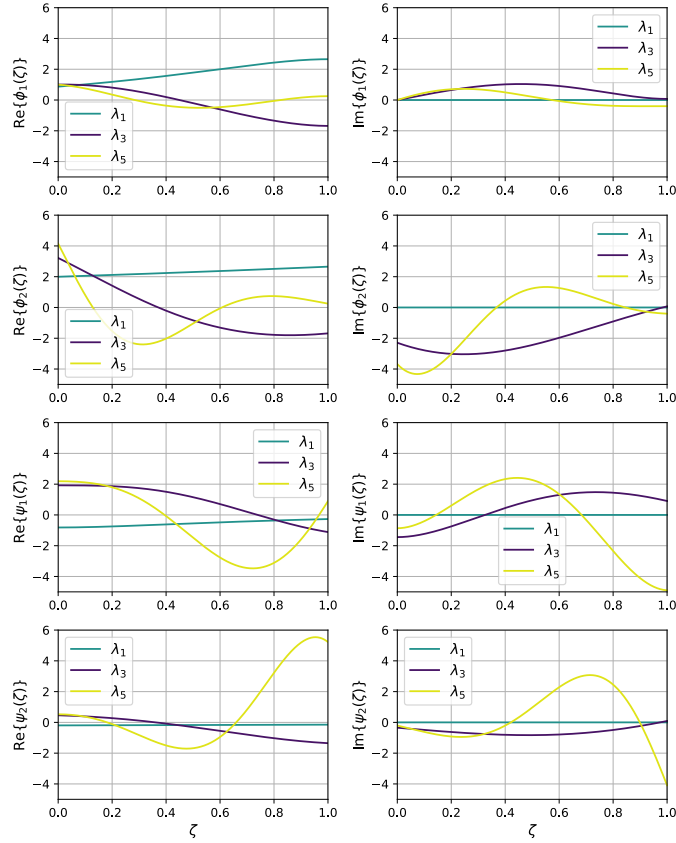

 Figure 2.3: First few eigenmodes of  $\mathfrak{A}$  and  $\mathfrak{A}^*$ .

Table 2.1: Physical parameters for the system.

Parameter	Symbol	Value	Unit
Diffusivity	$D$	$2 \times 10^{-5}$	$m^2/s$
Velocity	$v$	0.01	$m/s$
Reaction constant	$k_r$	-1.5	$s^{-1}$
Recycle residence time	$\tau$	80	$s$
Recycle ratio	$R$	0.3	—

The parameters of the system are carefully chosen to highlight all its key characteristics simultaneously—namely, significant diffusion, convection, and reaction occurring within the reactor—while also ensuring that the delay term and recycle ratio

have a pronounced effect on system dynamics. Additionally, the parameters are deliberately selected to introduce instability into the system, emphasizing the proposed control strategy's ability to stabilize an inherently unstable system. While no isothermal reactor can truly exhibit exponential instability due to the finite availability of reactants, such systems can still become unstable near the steady state. In this context, deviations from the steady state may cause the system to transition toward a different steady state, thereby altering the underlying dynamics and invalidating the original model used for system design and control optimization.

It has been observed that for the linearized system to have an unstable steady state, the reaction coefficient,  $k_r$ , must be negative. Although rare, this scenario can arise in certain reaction mechanisms where the reaction rate decreases as the reactant concentration increases, such as autocatalytic reactions, enzyme-catalyzed reactions, or reactions involving inhibitory effects. This instability can be qualitatively understood as follows: a negative reaction coefficient causes a decline in the reaction rate as the reactant accumulates, leading to further reactant accumulation and thus, driving the system away from its steady state. Quantitative confirmation of this behaviour can be achieved through eigenvalue analysis, where the presence of at least one eigenvalue with a positive real part indicates the fact that the open-loop linearized system is exponentially unstable.

## 2.3 LINEAR QUADRATIC REGULATOR DESIGN

### 2.3.1 Full-state feedback regulator

The bi-orthogonal basis generated by the Riesz-spectral operator  $\mathfrak{A}$  in the LTI system  $\Sigma(\mathfrak{A}, \mathfrak{B}, \mathfrak{C}, -)$  provides the foundation for solving the operator Riccati equation (ORE), a crucial step in the design of a linear quadratic regulator (LQR). The objective is to determine an offline feedback control law that drives the system's states from

an arbitrary initial condition to zero, thereby maintaining the system at its steady state. This is achieved within an optimal control framework, minimizing the infinite-time cost function defined in Equation 2.13. In this context,  $\mathfrak{Q}$  and  $\mathfrak{R}$  are self-adjoint coercive operators that penalize state deviations and control actions, respectively.

$$J(x_0, u) = \int_{t=0}^{\infty} \langle x(s), \mathfrak{Q}x(s) \rangle + \langle u(s), \mathfrak{R}u(s) \rangle ds \quad (2.13)$$

### 2.3.1.1 Operator Riccati equation

The LQR problem is solved by finding the unique positive semi-definite operator  $\mathbf{\Pi}$ , which satisfies the ORE presented in Equation 2.14. This operator is then used to compute the feedback gain that ensures optimal control of the system.

$$\langle \mathfrak{A}^* \mathbf{\Pi} x, y \rangle + \langle \mathbf{\Pi} \mathfrak{A} x, y \rangle - \langle \mathbf{\Pi} \mathfrak{B} \mathfrak{R}^{-1} \mathfrak{B}^* \mathbf{\Pi} x, y \rangle + \langle \mathfrak{Q} x, y \rangle = 0 \quad (2.14)$$

Given that the solution to the ORE is unique for any set of functions in the domain of operator  $\mathfrak{A}$ , we can arbitrarily set  $x = \phi_m$  and  $y = \phi_n$ , that is, the eigenfunctions of  $\mathfrak{A}$ . Applying this choice, and noting that  $\mathbf{\Pi}$  is self-adjoint, leads to the simplified Equation 2.15.

$$\langle \mathbf{\Pi} \phi_m, \mathfrak{A} \phi_n \rangle + \langle \mathfrak{A} \phi_m, \mathbf{\Pi} \phi_n \rangle - \mathfrak{R}^{-1} \langle \mathfrak{B}^* \mathbf{\Pi} \phi_m, \mathfrak{B}^* \mathbf{\Pi} \phi_n \rangle + \langle \mathfrak{Q} \phi_m, \phi_n \rangle = 0 \quad (2.15)$$

To ensure that the domain and range of  $\mathbf{\Pi}$  match those of  $\mathfrak{A}$  and  $\mathfrak{A}^*$ , respectively,  $\mathbf{\Pi}$  can be expressed as an infinite series, as shown in Equation 2.16. The coefficients  $p_{i,j}$  can be interpreted as elements of an infinite-dimensional matrix  $\tilde{P}$ , which represents the operator  $\mathbf{\Pi}$ . This forms the first step in converting the ORE to the corresponding matrix Riccati equation (MRE).

$$\mathbf{\Pi}x \equiv \sum_{i=1}^{\infty} \sum_{j=1}^{\infty} p_{i,j} \langle x, \psi_j \rangle \psi_i \quad \forall i, j : \quad p_{i,j} \in \mathbb{C} \quad (2.16)$$

### 2.3.1.2 Obtaining $\mathfrak{B}$ and $\mathfrak{B}^*$

Before further simplifying the ORE, it is essential to define the operators  $\mathfrak{B}$  and  $\mathfrak{B}^*$ . Given the boundary-control nature of the system as seen in Equation 3.3,  $\mathfrak{B}$  is defined to properly project the control input  $u \in \mathbb{R}^1$  onto the state space  $X : L^2[0, 1] \times L^2[0, 1]$ , as outlined in Equation 2.17.

$$\mathfrak{B}u \equiv v(1 - R) \begin{bmatrix} \delta(\zeta) \\ 0 \end{bmatrix} \cdot u \quad (2.17)$$

where  $\delta(\zeta)$  denotes the Dirac delta function. The adjoint operator  $\mathfrak{B}^*$  is obtained by leveraging the properties of  $\mathfrak{A}$  and  $\mathfrak{A}^*$ , that is, their expressions as well as their domains (as shown in Equations 3.5 and 3.6), after applying integration by parts to the result of the inner products, as summarized in Equation 2.18.

$$\begin{aligned} \langle \mathfrak{A}x + \mathfrak{B}u, y \rangle &= \langle \mathfrak{A}x, y \rangle + \langle \mathfrak{B}u, y \rangle = \langle x, \mathfrak{A}^*y \rangle + \langle u, \mathfrak{B}^*y \rangle \\ \langle u, \mathfrak{B}^*y \rangle &= \langle \mathfrak{A}x + \mathfrak{B}u, y \rangle - \langle x, \mathfrak{A}^*y \rangle \Rightarrow \dots \\ \Rightarrow \mathfrak{B}^*(\cdot) &= \left[ v(1 - R) \int_0^1 \delta(\zeta)(\cdot) d\zeta \quad , \quad 0 \right] \end{aligned} \quad (2.18)$$

### 2.3.1.3 Matrix Riccati equation

Using the expression for  $\mathbf{\Pi}$  in Equation 2.16, along with the derived  $\mathfrak{B}^*$  from Equation 2.18, and the eigenvalue problem  $\mathfrak{A}\phi_i = \lambda_i\phi_i$ , the ORE can be reformulated as the MRE shown in Equation 2.19. Here,  $\gamma_i \equiv v(1 - R) \psi_1^{(i)} \Big|_{\zeta=0}$ , and  $q_{m,n} = \langle \mathfrak{Q}\phi_m, \phi_n \rangle$ .

$$p_{n,m}(\lambda_m + \overline{\lambda_n}) - \Re^{-1} \left\langle \sum_{i=1}^{\infty} p_{i,m} \gamma_i, \sum_{i=1}^{\infty} p_{i,n} \gamma_i \right\rangle + q_{m,n} = 0 \quad (2.19)$$

Due to the infinite-dimensional nature of  $\tilde{P}$ , a numerical solution is impractical. This challenge is addressed by selecting the first  $N$  eigenmodes of the system as its dominant modes. This translates to truncating the infinite sums in the MRE and reducing the infinite-dimensional system to a finite set of nonlinear algebraic equations that can be solved to obtain an equivalent  $N \times N$  matrix  $P$ , that is, a truncated approximation of matrix  $\tilde{P}$ . The optimal full-state feedback gain is then calculated using Equation 2.20, ensuring closed-loop stability.

$$\begin{aligned}
 u(t) &= -\langle K(\zeta), x(\zeta, t) \rangle = -\mathfrak{B}^* \mathbf{P} x(\zeta, t) \\
 &= -\sum_{i=1}^N \sum_{j=1}^N p_{i,j} \langle x(\zeta, t), \psi_j(\zeta) \rangle \gamma_i \\
 &= -\sum_{i=1}^N \sum_{j=1}^N p_{i,j} \gamma_i \int_0^1 x(\zeta, t) \cdot \overline{\psi_j}(\zeta) d\zeta \\
 &= -\int_0^1 \sum_{i=1}^N \sum_{j=1}^N p_{i,j} \gamma_i \overline{\psi_j}(\zeta) \cdot x(\zeta, t) d\zeta \\
 \Rightarrow K(\zeta) &\equiv \sum_{i=1}^N \sum_{j=1}^N p_{i,j} \gamma_i \overline{\psi_j}(\zeta)
 \end{aligned} \tag{2.20}$$

The computed gain is a function of space and is calculated offline. The control action at any given time instance is the inner product of this gain with the current state of the system, thus justifying the term “full-state” feedback. The dynamics of the resulting closed-loop full-state feedback system may be described by the state-space representation shown in Equation 2.21.

$$\begin{aligned}
 \dot{x}(\zeta, t) &= \mathfrak{A}x(\zeta, t) + \mathfrak{B}u(t) \\
 &= (\mathfrak{A} - \mathfrak{B}K)x(\zeta, t) \\
 &= \mathfrak{A}_{reg} x(\zeta, t)
 \end{aligned} \tag{2.21}$$

By selecting  $\mathfrak{Q} = 0.05$  as a constant function over  $\zeta = [0, 1]$ , and  $\mathfrak{R} = 50$ , the full-state feedback gain is obtained and represented in Figure 2.4. The obtained gains



are used to design the optimal full-state feedback regulator to stabilize the control system. A block diagram representation of the full-state feedback control system is shown in Figure 3.3.

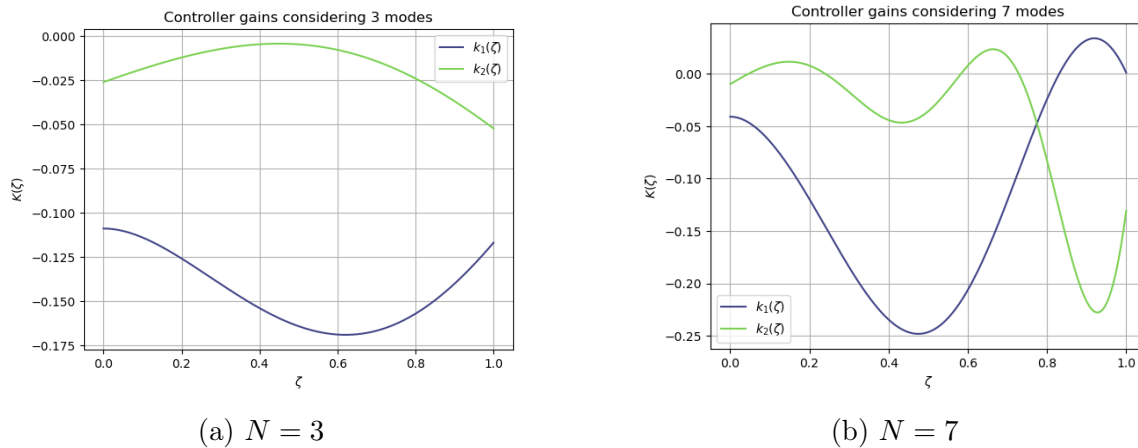


Figure 2.4: Full-state feedback gain  $K(\zeta)$  utilizing the first  $N$  modes of the system given by Equation (2.20).

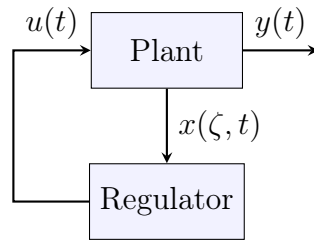


Figure 2.5: Block diagram representation of the optimal full-state feedback control system.

### 2.3.2 Output feedback compensator

Thus far, the optimal regulator is designed under the assumption that it has full access to the system's states. However, this assumption is not feasible in realistic applications. To address this, an observer is introduced to estimate and reconstruct the states by measuring the system's output in real time, and providing the regulator with the reconstructed states to further stabilize the system. The output, in this

context, is taken as the concentration at the reactor outlet, as defined in Equation 3.3. This leads to the definition of the output operator  $\mathfrak{C}$  in the linear time-invariant (LTI) system  $\Sigma(\mathfrak{A}, \mathfrak{B}, \mathfrak{C}, -)$ , which is subsequently used to determine the observer gain,  $L(\zeta)$ . The formulation is shown in Equation 2.22:

$$\mathfrak{C} \equiv \left[ \int_0^1 \delta(\zeta - 1)(\cdot) d\zeta \quad , \quad 0 \right] \quad (2.22)$$

where  $\delta(\zeta)$  denotes the Dirac delta function. Regarding the choice of observer, Luenberger-based observers are well-suited for infinite-dimensional systems when the system parameters are perfectly known [26]. Among the various methods to compute the gain for this class of observers, pole-placement is a solid, straightforward, and reliable approach for state reconstruction. To ensure that the state reconstruction dynamics converge more quickly than the regulation dynamics, the poles of the observer-based controller are placed to the left of the poles of the full-state feedback controller. This practice is common in the design of observer-based controllers for infinite-dimensional systems [15]. The observer gain in Figure 2.6 is obtained by limiting the eigenmodes of the observer-based controller to have real parts that are at least 3 times more negative than the real part of the dominant eigenmodes of the full-state feedback system. This is done for the case where the first 7 modes of the system are considered for designing the controller. The first few eigenvalues of the observer-based controller and the full-state feedback controller, along with the eigenvalues of the open-loop system are shown in Figure 2.7 to demonstrate the pole placement strategy explained above. It can be confirmed that both control systems have eigenvalues with negative real parts, ensuring stability. Note that the eigenvalues of both control systems are identical after the 7th mode, as the real parts of these eigenvalues are already sufficiently negative.

The dynamics of the augmented observer-controller system are described by the

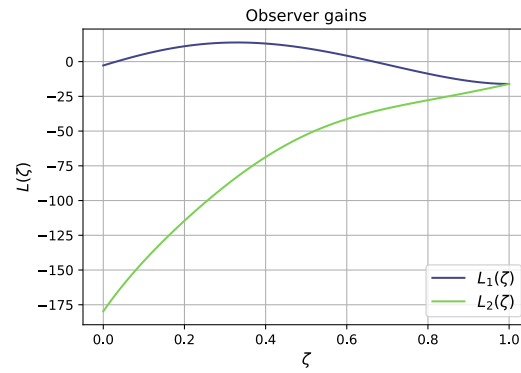


Figure 2.6: Observer gain  $L(\zeta)$ .

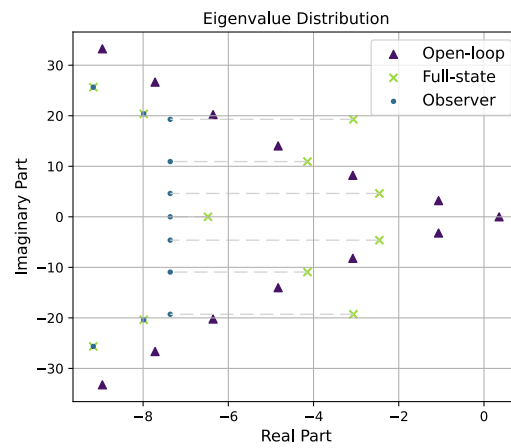


Figure 2.7: Eigenvalues of the observer-based controller, full-state feedback controller, and open-loop system.

state-space representation shown in Equation ??, where  $\hat{x}(\zeta, t)$  and  $e(\zeta, t)$  refer to the estimated state and the state estimation error, respectively. A block diagram representation of the output feedback control system is also shown in Figure 2.8.

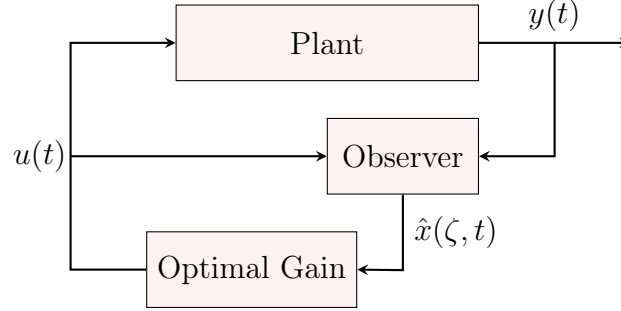


Figure 2.8: Block diagram representation of the observer-based output feedback control system.

## 2.4 RESULTS AND DISCUSSION

In this section, the obtained control strategies are applied to a finite-difference method (FDM) representation of the system to evaluate its dynamic response. The system is discretized in space using a uniform grid with 100 equidistributed points, resulting in a system of ordinary differential equations (ODEs) with respect to time. This spatial discretization is introduced solely at the evaluation stage to numerically approximate the system's behaviour under the influence of the optimal control input and is not involved in the design of the control law. The control law is derived directly in the infinite-dimensional space, fully capturing the continuous nature of the original system.

To solve the resulting ODEs, an adaptive Runge–Kutta method of order 5(4), commonly referred to as RK45, is employed. This method dynamically adjusts time steps to balance accuracy and computational efficiency, ensuring a reliable numerical solution while evaluating the system at specific points as required [27, 28]. The imple-

mentation of this method is facilitated using the `solve_ivp` function from Python's SciPy library [29], which provides a robust framework for handling time integration of ODEs.

Employing the outlined approach to evaluate the dynamic response of the systems under consideration, a comparative analysis is conducted between two identical systems with full-state access, differing only in the number of eigenmodes employed to compute the optimal full-state feedback gains. Subsequently, the performance of the proposed observer-based controller is assessed, with particular attention given to the dynamics of state reconstruction errors. Finally, a sensitivity analysis is performed to examine the impact of key parameters on the model behaviour and the effectiveness of the control strategy. Across all simulations presented,  $\mathfrak{Q} = 0.05 \cdot \mathfrak{I}$  and  $\mathfrak{R} = 50$  are used as the deviation penalty and control effort weight operators, respectively, where  $\mathfrak{I}$  denotes the identity operator matching the size of  $\mathfrak{A}$ .

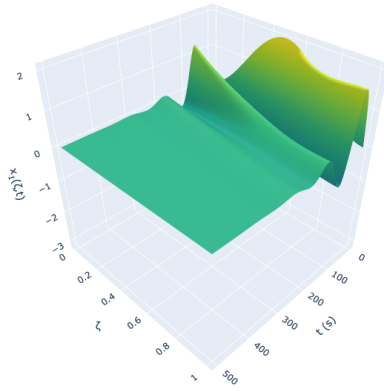
### 2.4.1 Full-state feedback regulator FDM representation

Initially, the input response of the system provided by the full-state feedback is explored using the mentioned FDM setup. Two configurations are compared where the optimal feedback gain is obtained using different numbers of eigenmodes: one with  $N = 3$  and another with  $N = 7$ , according to Figure 2.4. The state profile versus time and space is illustrated for both cases in Figure 2.9.

In order to offer a clearer representation of the state trajectory in time, spatial cross-sectional plots are provided in Figure 2.10 for the  $N = 7$  case at different lengths of the domain. The delay-imposing state, that is, the concentration along the recycle stream  $x_2(\zeta, t)$ , is provided only in Figure 2.10 for the sake of conciseness.

Both optimal feedback gains are able to successfully stabilize the system within finite time horizon. However, the case where more eigenmodes are considered in the

Reactor concentration profile


 (a)  $N = 3$ 

Reactor concentration profile

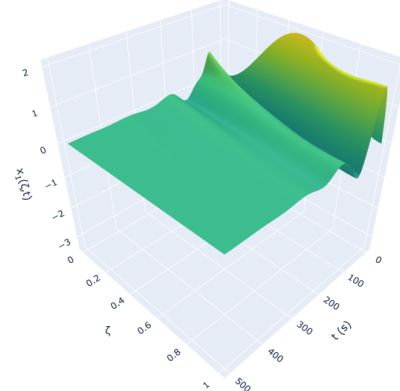
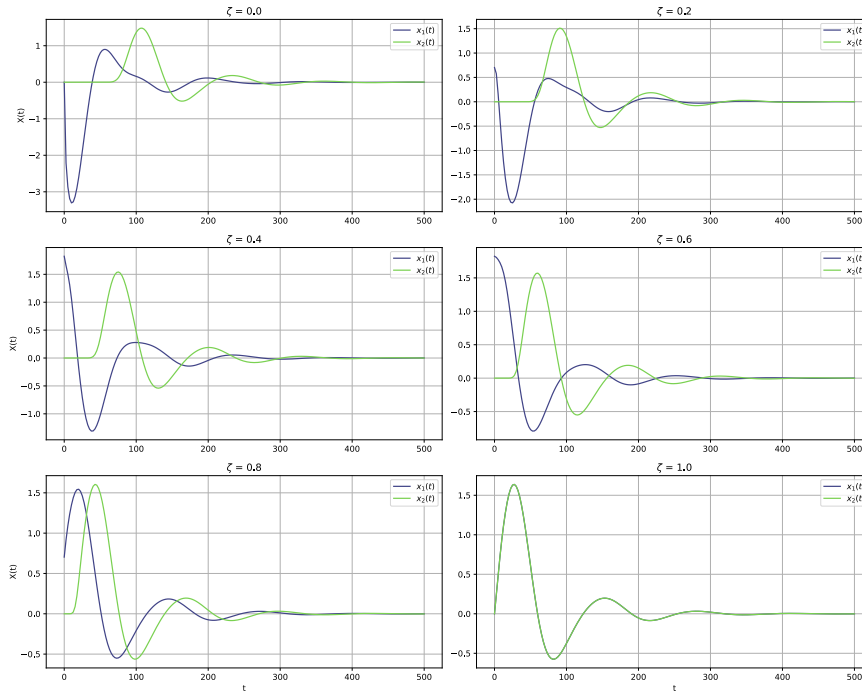

 (b)  $N = 7$ 

Figure 2.9: Input response of the system under full-state feedback control given by Equation (2.21), utilizing the feedback gain obtained in Figure 2.4.


 Figure 2.10: 2D cross-section plots of the full-state feedback input response at various  $\zeta$  positions, utilizing the feedback gain obtained in Figure 2.4b.

controller design shows better performance, as the higher dimensional controller is able to stabilize the system quicker with lower cost function values in general.

### 2.4.2 Observer-based regulator FDM representation

Omitting the need to have full access to system states, the observer-based regulator is evaluated using the same FDM representation. The states reconstruction is done by applying the observer gain obtained in Figure 2.6 to the system output. The estimated states are now used with the previously obtained optimal feedback gain with  $N = 7$  eigenmodes to calculate the input. Similar to the previous case, the state profile  $x_1(\zeta, t)$  is illustrated in Figure 2.11, as well as cross-sectional plots for both states in Figure 2.12 for better visualization of state trajectories in time.

In the next step, the state estimation error dynamics of the observer are plotted in Figures 2.13 and 2.14 to demonstrate the performance of the observer. The error dynamics are calculated as the squared difference between the true state and the estimated state at each grid point and time instance.

While the performance of the observer-based controller is slightly more sluggish compared to that of the full-state feedback regulator, it successfully stabilizes the system within a finite time horizon using only output measurements instead of full state information. In the absence of uncertainty in the system model, the observer gain can theoretically be designed so that the state estimation error converges to zero very fast compared to full-state feedback regulator dynamics. In practice, however, the observer gain is constrained by factors such as noise in the system output and plant-model mismatches. Despite these challenges, the proposed observer design mechanism achieves system stabilization with reasonable performance.

Estimated state profile

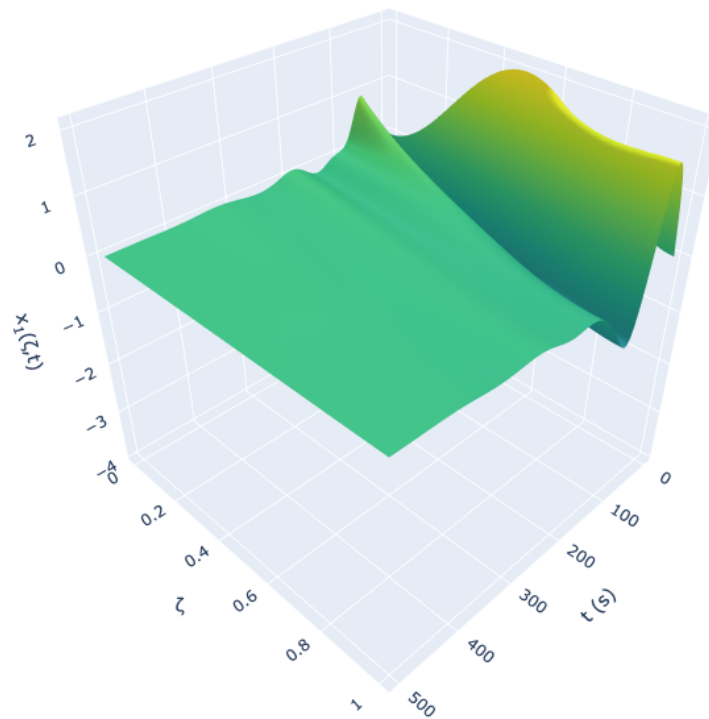


Figure 2.11: Input response of the system under observer-based output feedback control given by Equation (??), utilizing the observer gain obtained in Figure 2.6 and the feedback gain obtained in Figure 2.4b.



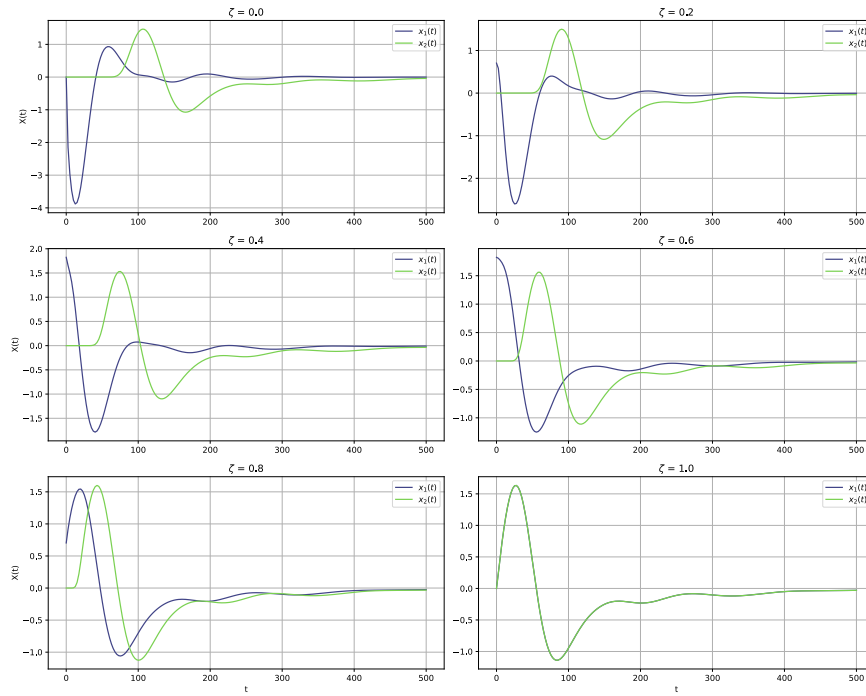


Figure 2.12: 2D cross-section plots of the input response at various  $\zeta$  positions, utilizing the observer gain obtained in Figure 2.6 and the feedback gain obtained in Figure 2.4b.

State reconstruction error profile

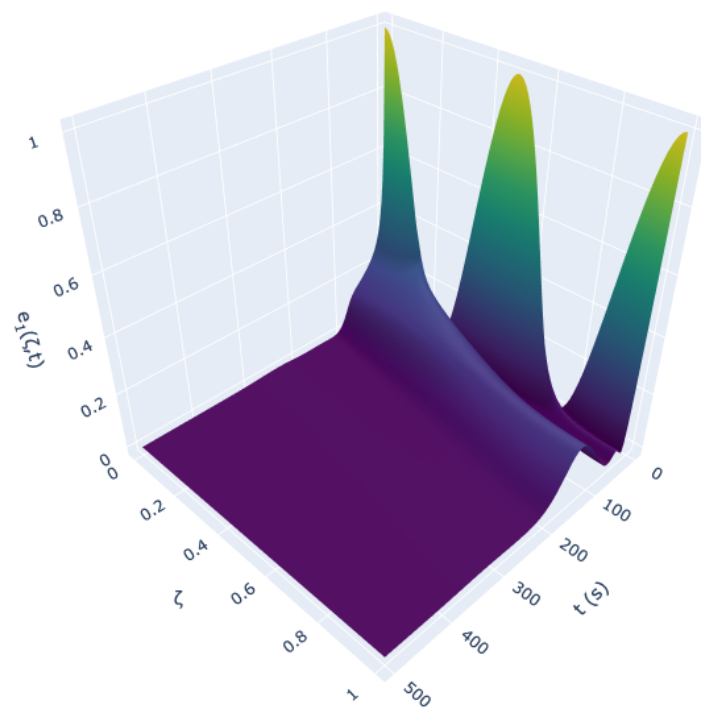


Figure 2.13: Error dynamics of the observer-based regulator utilizing the observer gain obtained in Figure 2.6 and the feedback gain obtained in Figure 2.4b.

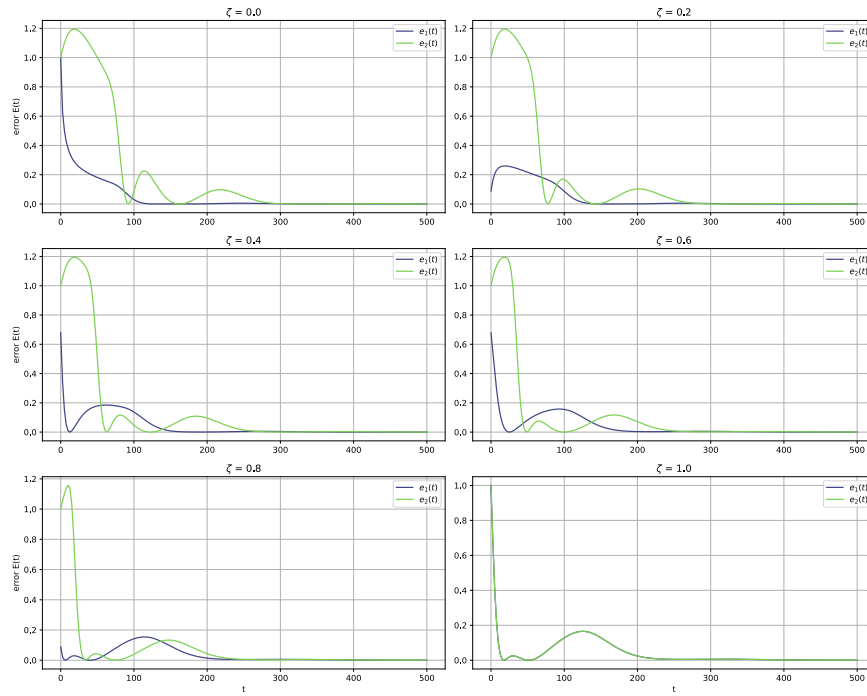


Figure 2.14: 2D cross-section plots of the error dynamics of the observer-based regulator at various  $\zeta$  positions, utilizing the observer gain obtained in Figure 2.6 and the feedback gain obtained in Figure 2.4b.

### 2.4.3 Parameter sensitivity analysis

Followed by showcasing the ability of the proposed controller to stabilize an unstable system using merely output measurements, a brief parameter sensitivity analysis of the model dynamics and controller performance is conducted at the end of this section. The effects of varying the recycle ratio  $R$  and mass transfer Peclet number (i.e., the ratio of convection to diffusion,  $Pe = v/D$ ) were explored during initial simulations. In addition, as it is central to this work, the effect of varying time delay  $\tau$  on the response of the system under the original controller design is investigated in more detail.

Regarding the effect of recycle ratio on system dynamics and controller performance, it was observed that as the recycle ratio approaches unity, the open-loop system exhibits behaviour similar to that of a well-mixed reactor, with concentration profiles flattening. This also influences the controller performance as it becomes more challenging to affect the system dynamics with the control input as the controller's action becomes diluted at the reactor inlet due to mixing with the recycle stream. For changes in mass Peclet number and its effect on the system dynamics, it was observed that a decrease in the Peclet number causes the eigenvalues of the system generator to shift closer to the real axis of the complex plane. This implies that greater diffusion relative to convection dampens the oscillatory behaviour of the system that is originally imposed as the result of the delayed recycle stream.

Finally, the effect of varying time delay  $\tau$  on the system dynamics is also investigated, as it is a key parameter in the model and controller design within this work. Input responses of several systems with different time delays are compared in Figure 2.15, where the input for all cases is calculated assuming  $\tau = 80$  s, which results in the same output feedback gain as the one obtained in Section 2.4.2.

It can be seen that as long as the actual time delay of a system is less than the

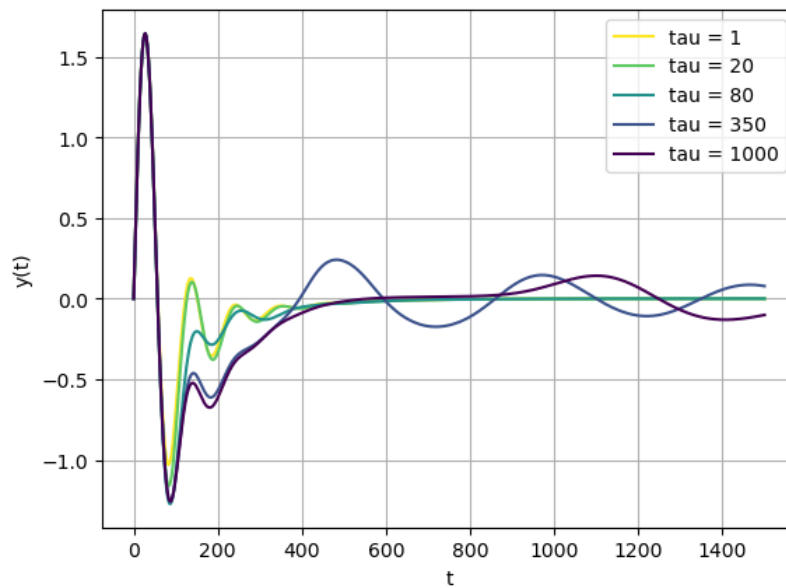


Figure 2.15: Measured output of the systems with different time delays  $\tau$ , under observer-based output feedback control utilizing the observer gain obtained in Figure 2.6 and the feedback gain obtained in Figure 2.4b, where  $\tau = 80$  s.

assumed delay used in the controller design, the controller is still able to stabilize the system within a finite time horizon, although transient deviations from the desired behaviour are observed. However, as the actual delay increases, the response of the system starts to deviate significantly from the desired behaviour, especially after a certain threshold close to the actual recycle delay of the system. This shows the importance of including the time delay in the controller design to ensure the stability of the system within an optimal framework.

Although further parameter sensitivity analysis is possible as it naturally raises readers' curiosity, expanding the analysis to include more detailed investigations would risk exceeding the scope of this work, which is to offer a novel modeling and control framework for a certain class of distributed parameter systems in chemical engineering. Nonetheless, the proposed modelling and control strategy is able to effectively stabilize the system over a broad range of parameter sets, including but not limited to variations in the imposed time delay, Peclet number, and recycle ratio, ensuring the practicality of the proposed framework across different system configurations.

## 2.5 CONCLUSION

The control of an axial tubular reactor equipped with recycle stream is addressed as a significant class of distributed parameter systems in chemical engineering industries. The notion of time delay introduced by the recycle process has not been adequately addressed in the literature despite being a common and intrinsic feature of such systems; introducing a rare example of state-delay in this field. By converting the notion of delay into an equivalent transport PDE, the DPS is formulated as a system of coupled parabolic and hyperbolic PDEs. The infinite-dimensional system is assumed to be boundary controlled, with the control input acting on the reactor

inlet. Particularly suited for the class of axial tubular reactors, Danckwerts boundary conditions are considered. A continuous-time linear quadratic optimal regulator is then developed to stabilize the system.

To address the infinite-dimensional nature of the system, a late lumping approach is employed, ensuring that the infinite-dimensional characteristics of the system are preserved in the control design. The system's Riesz-spectral properties are utilized to derive the full-state feedback regulator by solving the ORE, utilizing dominant modes of the system to obtain low-dimensional feedback gains. Recognizing practical limitations of the full-state feedback strategy, an observer-based regulator is also introduced to reconstruct the system states using boundary measurements, addressing the challenge of limited state access in real-world applications.

The proposed framework may be extended to more complex diffusion-convection reactor configurations, such as non-isothermal reactors. More complex control strategies may also be considered for this framework, such as model-predictive control (MPC) strategy, enabling constraints to be incorporated into the control design. The proposed observer-based control strategy may also be extended to handle measurement noise as well as plant-model mismatches, which are common in real-world applications. Another interesting aspect to explore is the briefly described impact that picking different numbers of eigenmodes may have on the optimality of controller design. This could be further investigated to provide a more comprehensive understanding of the effects of this choice on the controller's performance.

In summary, this research introduces a comprehensive optimal control strategy for a novel yet practically significant class of distributed parameter systems, that is, axial tubular reactors with delayed recycle streams. A late-lumping approach is employed to address the infinite-dimensional nature of the system, leveraging the Riesz-spectral properties of the system generator to derive an optimal feedback law utilizing both

full-state and estimated states of the system, setting the stage for future advancements in this area of research.

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## Chapter 3

# Discrete-time Control for Isothermal System

### 3.1 Introduction

Many chemical and petrochemical processes, such as reactions in tubular reactors, heat transfer in exchangers, and separations in columns, involve states distributed in space and time. These systems, known as distributed parameter systems (DPS), are often modeled using partial differential equations (PDEs) to describe distributed state dynamics. Due to their infinite-dimensional nature, the control and estimation of DPSs are inherently more challenging compared to the well-established control theories for finite-dimensional systems [1], making this field an active area of research. Two primary methods, “Early Lumping” and “Late Lumping,” have been proposed to address DPS control in the literature. The first, “Early Lumping,” reduces the infinite-dimensional system to a finite-dimensional one through spatial discretization during the modeling phase [2]. While this enables standard control strategies, it often compromises model accuracy due to mismatches between the original and reduced-order systems [4]. In contrast, “Late Lumping” preserves the infinite-dimensional system until the final numerical implementation stage, resulting in more accurate but computationally complex control strategies [1].

State reconstruction for DPSs has also been addressed using discrete-time Luenberger observers without spatial discretization, a key feature consistent with the late lumping paradigm [26, 30–32]. Numerous studies have employed Late Lumping approaches to control infinite-dimensional systems in the field of chemical engineering. These efforts primarily focus on convection-reaction systems governed by first-order hyperbolic PDEs and diffusion-convection-reaction systems governed by second-order parabolic PDEs. For example, robust control of first-order hyperbolic PDEs was explored in [33], where a plug flow reactor system was stabilized under distributed input. Similarly, boundary feedback stabilization using the backstepping method was proposed in [6] for such systems. State feedback regulator design for a countercurrent heat exchanger, another example of a chemical engineering DPS, was addressed in [7]. Introducing the effects of dispersion in tubular reactors, robust control of second-order parabolic PDEs was studied in [5]. Modal decomposition methods for designing low-dimensional predictive controllers for diffusion-convection-reaction systems have also been applied in [34], while observer-based model predictive control (MPC) was developed in [14] for axial dispersion tubular reactors, considering recycle stream effects.

Delay systems represent another class of infinite-dimensional systems studied extensively [16]. Commonly modeled using delay differential equations (DDEs), delays can alternatively be described using transport PDEs, offering advantages in complex scenarios [20]. In chemical engineering DPS control, input/output delays have been widely addressed, as industrial processes often feature both measurement and actuation delays. Such delays are typically handled by modeling them as transportation lag blocks, resulting in cascade PDE systems [18, 21, 35]. State delays, though less common, have been investigated in specific applications, such as heat exchangers with stream delays between passes [13], and plug flow tubular reactors with recycle

delays [22]; with the effect of dispersion not being addressed in any of these works. Even in [14], where the effect of recycle is studied for an axial dispersion tubular reactor, the recycle is assumed to be instantaneous, leaving a gap in the literature regarding state delays in diffusion-convection-reaction systems with recycle streams.

In this work, an axial dispersion reactor with recycle is modeled as a diffusion-convection-reaction DPS. The reactor dynamics are described by a second-order parabolic PDE, coupled with a first-order hyperbolic transport PDE to account for the recycle stream's state delay. A Late Lumping approach is employed, obtaining the system's resolvent in a closed operator form without spatial discretization. To implement MPC as a digital controller, the system is discretized using the Cayley-Tustin method, a Crank-Nicolson-type discretization that conserves the continuous system's characteristics, avoiding the need for model reduction. Numerical simulations demonstrate that the proposed controller stabilizes an unstable system optimally under input constraints. A discrete-time infinite-dimensional Luenberger observer is designed to reconstruct unmeasured states, enabling output feedback MPC. Simulations show that the proposed controller successfully stabilizes the otherwise unstable system under input constraints.

## 3.2 Mathematical Modeling of the Reactor System

### 3.2.1 Model representation

The chemical process depicted in Fig. 3.1 illustrates a chemical reaction within an axial dispersion tubular reactor [23] where reactant  $A$  is converted into products. The reactor features a recycle mechanism, allowing a portion of the product stream to re-enter the reactor, ensuring the consumption of any unreacted substrate. The dynamics of the reactant concentration can be described by the second-order parabolic PDE given by (3.1), a common class of equations used to characterize diffusion-convection-

reaction systems [24]. The resulting PDE that describes the reactor model is obtained by utilizing first-principle modeling through relevant mass balance relations on an infinitesimally thin disk element along the longitudinal axis of the reactor.

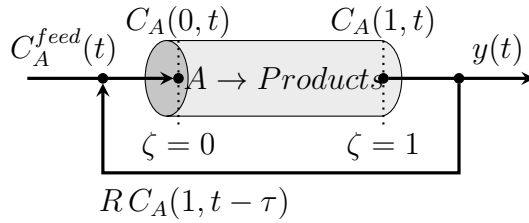


Figure 3.1: Axial tubular reactor with recycle stream.

$$\dot{C}_A(\zeta, t) = D\partial_{\zeta\zeta}C_A(\zeta, t) - v\partial_{\zeta}C_A(\zeta, t) - r(C_A) \quad (3.1)$$

Here,  $C_A(\zeta, t)$  is the concentration of reactant  $A$  along the reactor. The physical parameters  $D$  and  $v$  represent the diffusion coefficient and flow velocity along the reactor, respectively. Physical parameters are assumed to be constant, hence changes in temperature or pressure will not affect the reactor model. The coordinate system in space and time is represented by  $\zeta$  and  $t$ , where  $\zeta \in [0, 1]$  and  $t \in [0, \infty)$ . In addition,  $r(C_A)$  is the reaction rate of the reactant in general, which is often a non-linear function of  $C_A$ . Therefore, the model is further linearized around its steady-state, followed by introducing the deviation variable  $c(\zeta, t) = C_A(\zeta, t) - C_{A,ss}(\zeta)$ , where  $C_{A,ss}(\zeta)$  is the steady-state concentration of the reactant. The linearized model is then given by (3.2).

$$\dot{c}(\zeta, t) = D\partial_{\zeta\zeta}c(\zeta, t) - v\partial_{\zeta}c(\zeta, t) - k_r c(\zeta, t) \quad (3.2)$$

Here,  $k_r \equiv \left. \frac{\partial r(C_A)}{\partial C_A} \right|_{C_{A,ss}}$  is the linearized reaction rate coefficient in the vicinity of the steady-state. The system input is defined as  $u(t) \equiv C_A^{feed} - C_{A,ss}^{feed}$ , representing the deviation of the concentration of the reactant being fed into the reactor from

its steady-state value. The output of the system is also considered as the deviation of the concentration of the reactant being measured at the reactor outlet from its steady-state value, denoted as  $y(t)$ .

To accurately represent the behavior of the given axial dispersion tubular reactor, Dankwerts boundary conditions are applied; as they effectively capture deviations from ideal mixing and piston flow while assuming negligible transport lags in connecting lines [25]. The inlet boundary condition is modified to reflect the mixing of the input stream with the delayed state, i.e. the recycled reactant concentration coming from the reactor outlet, occurring  $\tau$  seconds earlier. These boundary conditions are therefore summarized in (3.3), with  $R$  and  $\tau$  denoting the recycle ratio and the residence time in the recycle stream, respectively. The system output will consequently be defined as  $y(t) = x_1(1, t)$ .

$$\begin{cases} D\partial_{\zeta}c(0, t) - vc(0, t) = -v[Rc(1, t - \tau) + (1 - R)u(t)] \\ \partial_{\zeta}c(1, t) = 0 \\ y(t) = c(1, t) \end{cases} \quad (3.3)$$

In the case where the problem involves similar forms of PDEs, an effective general practice to address delays in systems is to reformulate the problem such that the notion of delay is replaced with an alternative transport PDE. Therefore, a new state variable  $\underline{x}(\zeta, t) \equiv [x_1(\zeta, t), x_2(\zeta, t)]^T$  is defined as a vector of functions, where  $x_1(\zeta, t)$  represents the concentration within the reactor—analogueous to  $c(\zeta, t)$ —and  $x_2(\zeta, t)$  is introduced as a new state variable to account for the concentration along the recycle stream. The delay is thus modeled as a pure transport process, wherein the first state  $x_1(\zeta, t)$  is transported from the reactor outlet to the inlet, experiencing a delay of  $\tau$  time units while in the recycle stream. This makes all state variables expressed explicitly at a specific time instance  $t$ , resulting in the standard state-space form for

a given infinite-dimensional linear time-invariant (LTI) system given in (3.4).

$$\begin{aligned}\dot{\underline{x}}(\zeta, t) &= \mathfrak{A}\underline{x}(\zeta, t) + \mathfrak{B}u(t) \\ y(t) &= \mathfrak{C}\underline{x}(\zeta, t) + \mathfrak{D}u(t)\end{aligned}\tag{3.4}$$

Here,  $\mathfrak{A}$  is a linear operator  $\mathcal{L}(X)$  acting on a Hilbert space  $X : L^2[0, 1] \times L^2[0, 1]$  and  $\underline{x}(\zeta, t)$ , as defined previously, is the vector of functions describing the states of the system. Input operator  $\mathfrak{B}$  is a linear operator that maps the scalar input from input-space onto the state space. Output operator  $\mathfrak{C}$  on the other hand, is a linear operator that maps the infinite-dimensional state space onto the finite-dimensional output space, resulting in a scalar output. The operator  $\mathfrak{D}$  is the direct transmission operator, which is set to zero in this case as there is no direct transmission from the input to the output in the continuous-time system. The operators ( $\mathfrak{A}$ ,  $\mathfrak{B}$ ,  $\mathfrak{C}$ , and  $\mathfrak{D}$ ) are shown in (3.5) for the infinite-dimensional LTI system.

$$\begin{aligned}\mathfrak{A} &\equiv \begin{bmatrix} D\partial_{\zeta\zeta} - v\partial_{\zeta} + k_r & 0 \\ 0 & \frac{1}{\tau}\partial_{\zeta} \end{bmatrix} \\ D(\mathfrak{A}) &= \left\{ \underline{x}(\zeta) = [x_1(\zeta), x_2(\zeta)]^T \in X : \right. \\ &\quad \underline{x}(\zeta), \partial_{\zeta}\underline{x}(\zeta), \partial_{\zeta\zeta}\underline{x}(\zeta) \quad \text{a.c.}, \\ &\quad D\partial_{\zeta}x_1(0) - vx_1(0) = -vRx_2(0), \\ &\quad \left. \partial_{\zeta}x_1(1) = 0, x_1(1) = x_2(1) \right\} \\ \mathfrak{B} &\equiv \begin{bmatrix} \delta(\zeta) \\ 0 \end{bmatrix} v(1 - R) \\ \mathfrak{C} &\equiv \left[ \int_0^1 \delta(\zeta - 1)(\cdot) d\zeta \quad 0 \right] \\ \mathfrak{D} &= 0\end{aligned}\tag{3.5}$$

where  $\delta(\zeta)$  is dirac delta function. This will enable the derivation of the system's spectrum using the eigenvalue problem. The characteristics equation of the system is obtained by solving the equation  $\det(\mathfrak{A} - \lambda_i I) = 0$  for  $\lambda_i$ , where  $\lambda_i \in \mathbb{C}$  is the  $i^{\text{th}}$



eigenvalue of the system and  $I$  is the identity operator. Attempts to analytically solve this equation have failed; therefore, it is solved numerically using the parameters in Table I. These parameters are carefully chosen to reflect key characteristics of the system, i.e. diffusion, convection, reaction, and delayed recycle.

Similar to [36], a negative reaction coefficient ( $k_r$ ) is used to induce instability for analysis, a condition uncommon for isothermal reactors but possible in specific cases like autocatalytic or inhibitory reactions. Figure 2 depicts the resulting eigenvalue distribution in the complex plane, confirming instability of the linearized model near its steady state.

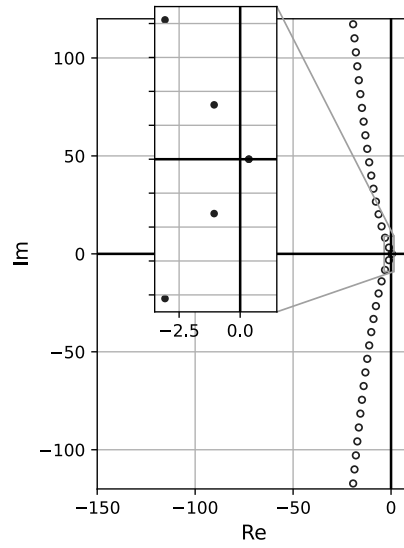


Figure 3.2: Eigenvalues of operator  $\mathfrak{A}$ .

### 3.2.2 Adjoint system

Next step is to obtain the adjoint system operators  $\mathfrak{A}^*$  and  $\mathfrak{B}^*$ . Utilizing the relation  $\langle \mathfrak{A}\underline{x} + \mathfrak{B}u, \underline{y} \rangle = \langle \underline{x}, \mathfrak{A}^*\underline{y} \rangle + \langle u, \mathfrak{B}^*\underline{y} \rangle$ , the adjoint operators  $\mathfrak{A}^*$  and  $\mathfrak{B}^*$  are obtained as shown in (3.6) and (3.7), respectively.

Table 3.1: Physical Parameters for the System

Parameter	Symbol	Value	Unit
Diffusivity	$D$	$2 \times 10^{-5}$	$m^2/s$
Velocity	$v$	0.01	$m/s$
Reaction Constant	$k_r$	-1.5	$s^{-1}$
Recycle Residence Time	$\tau$	80	$s$
Recycle Ratio	$R$	0.3	—

$$\begin{aligned}
\mathfrak{A}^* &= \begin{bmatrix} D\partial_{\zeta\zeta} + v\partial_{\zeta} - k_r & 0 \\ 0 & -\frac{1}{\tau}\partial_{\zeta} \end{bmatrix} \\
D(\mathfrak{A}^*) &= \left\{ \underline{y} = [y_1, y_2]^T \in Y : \right. \\
&\quad \underline{y}(\zeta), \partial_{\zeta}\underline{y}(\zeta), \partial_{\zeta\zeta}\underline{y}(\zeta) \quad \text{a.c.}, \\
&\quad D\partial_{\zeta}y_1(1) + vy_1(1) = \frac{1}{\tau}y_2(1), \\
&\quad Rvy_1(0) = \frac{1}{\tau}y_2(0), \partial_{\zeta}y_1(0) = 0 \left. \right\}
\end{aligned} \tag{3.6}$$

$$\mathfrak{B}^*(\cdot) = \left[ v(1-R) \int_0^1 \delta(\zeta)(\cdot) d\zeta \quad , \quad 0 \right] \tag{3.7}$$

Once the adjoint operators are determined, the eigenfunctions  $\{\underline{\phi}_i(\zeta), \underline{\psi}_i(\zeta)\}$  (for  $\mathfrak{A}$  and  $\mathfrak{A}^*$ , respectively) may be obtained and properly scaled following the calculation of eigenvalues. The set of scaled eigenfunctions will then form a bi-orthonormal basis for the Hilbert space  $X$ ; which will be later used in the controller design. It is important to note that the system is not self adjoint, as the obtained adjoint operator and its domain are not the same as the original operator and its domain.

### 3.2.3 Resolvent operator

One must obtain the resolvent operator of the system  $\mathfrak{R}(s, \mathfrak{A}) = (sI - \mathfrak{A})^{-1}$  prior to constructing the discrete-time representation of the system. One way to obtain

it is by utilizing the modal characteristics of the system, resulting in an infinite-sum representation of the operator. While being a common practice in the literature, truncating the infinite-sum representation for numerical implementation may lead to a loss of accuracy. Another way to express the resolvent operator is by treating it as an operator that maps either the initial condition of the system  $\underline{x}(\zeta, 0)$  or the input  $u(t)$ , to the Laplace transform of the state of the system  $\underline{X}(\zeta, s)$ . This approach, although more computationally intensive, results in a closed form expression for the resolvent operator, preserving the infinite-dimensional nature of the system. In (3.8), Laplace transform is applied to the LTI representation of the system for both zero-input response and zero-state response to obtain a general expression for the resolvent operator.

$$\begin{aligned} \dot{\underline{x}}(\zeta, t) &= \mathfrak{A}\underline{x}(\zeta, t) + \mathfrak{B}u(t) \xrightarrow{\mathcal{L}} \\ s\underline{X}(\zeta, s) - \underline{x}(\zeta, 0) &= \mathfrak{A}\underline{X}(\zeta, s) + \mathfrak{B}U(s) \\ \begin{cases} \xrightarrow{u=0} & \underline{X}(\zeta, s) = (sI - \mathfrak{A})^{-1}\underline{x}(\zeta, 0) = \mathfrak{R}(s, \mathfrak{A})\underline{x}(\zeta, 0) \\ \xrightarrow{\underline{x}(0, \zeta)} & \underline{X}(\zeta, s) = (sI - \mathfrak{A})^{-1}\mathfrak{B}U(s) = \mathfrak{R}(s, \mathfrak{A})\mathfrak{B}U(s) \end{cases} \end{aligned} \quad (3.8)$$

The goal is to obtain the solution for  $\underline{X}(\zeta, s)$  and compare it with the general expression obtained in (3.8) to get the closed form expression for the resolvent operator. First step is to apply Laplace transform to the original system of PDEs in (3.5). The second order derivative term is decomposed to two first order PDEs, constructing a new  $3 \times 3$  system of first order ODEs with respect to  $\zeta$  after Laplace transformation, as shown in (3.9).

$$\begin{aligned}
 \overbrace{\partial_\zeta \begin{bmatrix} X_1(\zeta, s) \\ \partial_\zeta X_1(\zeta, s) \\ X_2(\zeta, s) \end{bmatrix}}^{\underline{\tilde{X}}(\zeta, s)} &= \overbrace{\begin{bmatrix} 0 & 1 & 0 \\ \frac{s+k_r}{D} & \frac{v}{D} & 0 \\ 0 & 0 & s\tau \end{bmatrix}}^{P(s)} \begin{bmatrix} X_1(\zeta, s) \\ \partial_\zeta X_1(\zeta, s) \\ X_2(\zeta, s) \end{bmatrix} \\
 &+ \underbrace{\begin{bmatrix} 0 \\ -\frac{x_1(\zeta, 0)}{D} + v(1-R)\delta(\zeta)U(s) \\ -\tau x_2(\zeta, 0) \end{bmatrix}}_{Z(\zeta, s)} \\
 &\Rightarrow \partial_\zeta \underline{\tilde{X}}(\zeta, s) = P(s)\underline{\tilde{X}}(\zeta, s) + Z(\zeta, s)
 \end{aligned} \tag{3.9}$$

with solution given by (3.10).

$$\underline{\tilde{X}}(\zeta, s) = \underbrace{e^{P(s)\zeta}}_{T(\zeta, s)} \underline{\tilde{X}}(0, s) + \int_0^\zeta \underbrace{e^{P(s)(\zeta-\eta)}}_{F(\zeta, \eta)} Z(\eta, s) d\eta \tag{3.10}$$

Since the boundary conditions are not homogeneous,  $\underline{\tilde{X}}(0, s)$  needs to be obtained by solving the system of algebraic equations given in (3.11); which is the result of applying Danckwerts boundary conditions to the Laplace transformed system of PDEs at  $\zeta = 1$ .

$$\begin{aligned}
 &\overbrace{\begin{bmatrix} -v & D & Rv \\ T_{11}(1, s) & T_{12}(1, s) & -T_{33}(1, s) \\ T_{21}(1, s) & T_{22}(1, s) & 0 \end{bmatrix}}^{M^{-1}(s)} \underline{\tilde{X}}(0, s) = \\
 &\underbrace{\int_0^1 \begin{bmatrix} 0 \\ F_{33}(1, \eta)Z_3(\eta, s) - F_{12}(1, \eta)Z_2(\eta, s) \\ -F_{22}(1, \eta)Z_2(\eta, s) \end{bmatrix} d\eta}_{\underline{b}(s)} \\
 &\Rightarrow \underline{\tilde{X}}(0, s) = M(s)\underline{b}(s)
 \end{aligned} \tag{3.11}$$

Having access to  $\underline{\tilde{X}}(0, s)$ , the solution for  $\underline{X}(\zeta, s)$  can be explicitly derived. The

resolvent operator for zero-input and zero-state cases are therefore obtained in a closed form as shown in (3.12) and (3.13), respectively.

$$\begin{aligned}
 U(s) = 0 \Rightarrow \mathfrak{R}(s, \mathfrak{A})\underline{(\cdot)} &= \begin{bmatrix} \mathfrak{R}_{11} & \mathfrak{R}_{12} \\ \mathfrak{R}_{21} & \mathfrak{R}_{22} \end{bmatrix} \begin{bmatrix} (\cdot)_1 \\ (\cdot)_2 \end{bmatrix} \Rightarrow \\
 \mathfrak{R}_{11} &= \sum_{j=1}^2 \frac{T_{1j}(\zeta)}{D} \int_0^1 [M_{j2}F_{12}(1, \eta) + M_{j3}F_{22}(1, \eta)] (\cdot)_1 d\eta \\
 &\quad - \frac{1}{D} \int_0^\zeta F_{12}(\zeta, \eta) (\cdot)_1 d\eta \\
 \mathfrak{R}_{12} &= \sum_{j=1}^2 -\tau T_{1j}(\zeta) \int_0^1 M_{j2}F_{33}(1, \eta) (\cdot)_2 d\eta \\
 \mathfrak{R}_{21} &= \frac{T_{33}(\zeta)}{D} \int_0^1 [M_{32}F_{12}(1, \eta) + M_{33}F_{22}(1, \eta)] (\cdot)_1 d\eta \\
 \mathfrak{R}_{22} &= -\tau T_{33}(\zeta) \int_0^1 M_{32}F_{33}(1, \eta) (\cdot)_2 d\eta \\
 &\quad - \tau \int_0^\zeta F_{33}(\zeta, \eta) (\cdot)_2 d\eta
 \end{aligned} \tag{3.12}$$

$$\begin{aligned}
 \underline{x}(\zeta, 0) = 0 \Rightarrow \mathfrak{R}(s, \mathfrak{A})\mathfrak{B}(\cdot) &= \begin{bmatrix} \mathfrak{R}_1\mathfrak{B} \\ \mathfrak{R}_2\mathfrak{B} \end{bmatrix} (\cdot) \Rightarrow \\
 \mathfrak{R}_1\mathfrak{B} &= -v(1 - R) \left[ \sum_{j=1}^2 T_{1j}(\zeta) (M_{j2}T_{12}(1) + M_{j3}T_{22}(1)) \right. \\
 &\quad \left. - T_{12}(\zeta) \right] (\cdot) \\
 \mathfrak{R}_2\mathfrak{B} &= -v(1 - R) [T_{33}(\zeta) (M_{32}T_{12}(1) + M_{33}T_{22}(1))] (\cdot)
 \end{aligned} \tag{3.13}$$

Since the system generator  $\mathfrak{A}$  is not self-adjoint, the resolvent operator for the adjoint system shall also be obtained. This is done in a similar manner as the original system, resulting in a closed-form expression for the adjoint resolvent operator  $\mathfrak{R}^*(s, \mathfrak{A}^*)$ . To avoid redundancy, the derivation of the resolvent operator for the adjoint system is not included in this manuscript.

### 3.2.4 Cayley–Tustin Time Discretization

To implement the system on digital controllers, it is necessary to transition to a discrete-time framework while preserving critical properties such as stability and controllability. The Cayley–Tustin time-discretization method achieves this by mapping the continuous-time system to the discrete domain [37, 38]. This Crank–Nicolson type of discretization is also known as the lowest order symplectic integrator in Gauss quadrature-based Runge–Kutta methods [39]. Considering  $\Delta t$  as the sampling time, and assuming a piecewise constant input within time intervals (zero-order hold), the discrete-time representation  $\underline{x}(\zeta, k) = \mathfrak{A}_d \underline{x}(\zeta, k-1) + \mathfrak{B}_d u(k)$  is obtained, with discrete-time operators  $\mathfrak{A}_d$ ,  $\mathfrak{B}_d$ ,  $\mathfrak{C}_d$ , and  $\mathfrak{D}_d$  defined in (3.14), where  $\alpha = 2/\Delta t$ .

$$\begin{bmatrix} \mathfrak{A}_d & \mathfrak{B}_d \\ \mathfrak{C}_d & \mathfrak{D}_d \end{bmatrix} = \begin{bmatrix} -I + 2\alpha \Re(\alpha, \mathfrak{A}) & \sqrt{2\alpha} \Re(\alpha, \mathfrak{A}) \mathfrak{B} \\ \sqrt{2\alpha} \mathfrak{C} \Re(\alpha, \mathfrak{A}) & \mathfrak{C} \Re(\alpha, \mathfrak{A}) \mathfrak{B} \end{bmatrix} \quad (3.14)$$

As required for systems with nonself-adjoint generators, the adjoint discrete-time operators  $\mathfrak{A}_d^*$  and  $\mathfrak{B}_d^*$  are also obtained in a similar manner, as shown in (3.15).

$$\begin{bmatrix} \mathfrak{A}_d^* & \mathfrak{B}_d^* \end{bmatrix} = \begin{bmatrix} -I + 2\alpha \Re^*(\alpha, \mathfrak{A}^*) & \sqrt{2\alpha} \mathfrak{B}^* \Re^*(\alpha, \mathfrak{A}^*) \end{bmatrix} \quad (3.15)$$

## 3.3 Estimation and Control

### 3.3.1 Model predictive control design, full-state availability

The proposed full-state feedback model predictive control strategy, as shown in Fig. 3.3, is developed in this section with the goal of stabilizing the given unstable infinite-dimensional system within an optimal framework while satisfying input constraints. An infinite-time open-loop objective function sets the foundation of the controller design in the discrete-time setting at each sampling instant  $k$ , which consists of a weighted sum of state deviations and actuation costs for all future time instances,

subject to the system dynamics and input constraints, as shown in (3.16).

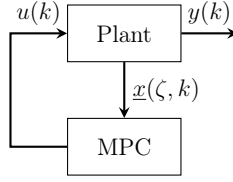


Figure 3.3: Proposed full-state feedback model predictive control system.

$$\begin{aligned}
 \min_U \quad & \sum_{l=0}^{\infty} \langle \underline{x}(\zeta, k+l|k), \mathfrak{Q} \underline{x}(\zeta, k+l|k) \rangle \\
 & + \langle u(k+l+1|k), \mathfrak{F} u(k+l+1|k) \rangle
 \end{aligned} \tag{3.16}$$

$$\text{s.t.} \quad \underline{x}(\zeta, k+l|k) = \mathfrak{A}_d \underline{x}(\zeta, k+l-1|k) + \mathfrak{B}_d u(k+l|k)$$

$$u^{\min} \leq u(k+l|k) \leq u^{\max}$$

where  $\mathfrak{Q}$  and  $\mathfrak{F}$  are positive definite operators of appropriate dimensions, responsible for penalizing state deviations and actuation costs, respectively. The notation  $(k+l|k)$  indicates the future time states or input instance  $k+l$  obtained at time  $k$ . The infinite-time optimization problem may be reduced to a finite-time setup by assigning zero-input beyond a certain control horizon  $N$ , resulting in the optimization problem in (3.17).

$$\begin{aligned}
 \min_U \quad & \sum_{l=0}^{N-1} \langle \underline{x}(\zeta, k+l|k), \mathfrak{Q}\underline{x}(\zeta, k+l|k) \rangle \\
 & + \langle u(k+l+1|k), \mathfrak{F}u(k+l+1|k) \rangle \\
 & + \langle \underline{x}(\zeta, k+N|k), \mathfrak{P}\underline{x}(\zeta, k+N|k) \rangle
 \end{aligned} \tag{3.17}$$

$$\begin{aligned}
 \text{s.t.} \quad & \underline{x}(\zeta, k+l|k) = \mathfrak{A}_d \underline{x}(\zeta, k+l-1|k) + \mathfrak{B}_d u(k+l|k) \\
 & u^{min} \leq u(k+l|k) \leq u^{max} \\
 & \langle \underline{x}(\zeta, k+N|k), \underline{\phi}_u(\zeta) \rangle = 0
 \end{aligned}$$

Obtained as the solution to the discrete-time Lyapunov equation,  $\mathfrak{P}$  is the terminal cost operator as shown in (3.18); which can be proven to be positive definite only if the terminal state  $\underline{x}(\zeta, k+N|k)$  is in a stable subspace. Therefore, an equality constraint is introduced to guarantee that the resulting quadratic optimization problem is convex. The terminal constraint is enforced by setting the projection of the terminal state onto the unstable subspace of the system to zero [14, 16, 38]. Here,  $\underline{\phi}_u(\zeta)$  is the set of unstable eigenfunctions of the system, for all eigenvalues where  $\text{Re}(\lambda_u) \geq 0$ .

$$\mathfrak{P}(\cdot) = \sum_{m=0}^{\infty} \sum_{n=0}^{\infty} -\frac{\langle \underline{\phi}_m, \mathfrak{Q}\underline{\psi}_n \rangle}{\lambda_m + \bar{\lambda}_n} \langle (\cdot), \underline{\psi}_n \rangle \underline{\phi}_m \tag{3.18}$$

One may further process the optimization problem in (3.17) to obtain a standard format for quadratic programming (QP) solvers by substituting the future states in terms of the current state and the sequence of future inputs using system dynamics expression. The resulting QP problem is given in (3.19). The optimal input sequence  $U$  is then obtained by solving the QP problem at each sampling instant  $k$ . To implement a receding horizon control strategy, only the first input of the optimal sequence  $u(k+1|k)$  is applied to the system, and the optimization problem is solved again at the next sampling instant  $k+1$ .



$$\begin{aligned}
 \min_U J &= U^T \langle I, H \rangle U + 2U^T \langle I, P\underline{x}(\zeta, k|k) \rangle \\
 \text{s.t.} \quad &U^{min} \leq U \leq U^{max} \\
 &T_u \underline{x}(\zeta, k|k) + S_u U = 0
 \end{aligned}$$

with  $H =$

$$\begin{bmatrix}
 \mathfrak{B}_d^* \mathfrak{P} \mathfrak{B}_d + \mathfrak{F} & \mathfrak{B}_d^* \mathfrak{A}_d^* \mathfrak{P} \mathfrak{B}_d & \dots & \mathfrak{B}_d^* \mathfrak{A}_d^{*N-1} \mathfrak{P} \mathfrak{B}_d \\
 \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d \mathfrak{B}_d & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{B}_d + \mathfrak{F} & \dots & \mathfrak{B}_d^* \mathfrak{A}_d^{*N-2} \mathfrak{P} \mathfrak{B}_d \\
 \vdots & \vdots & \ddots & \vdots \\
 \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^{N-1} \mathfrak{B}_d & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^{N-2} \mathfrak{B}_d & \dots & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{B}_d + \mathfrak{F}
 \end{bmatrix} \quad (3.19)$$

$$P = \begin{bmatrix} \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^2 & \dots & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^N \end{bmatrix}^T$$

$$T_u(\cdot) = \begin{bmatrix} \langle \mathfrak{A}_d^N(\cdot), \underline{\phi}_u \rangle \end{bmatrix}$$

$$S_u = \begin{bmatrix} \langle \mathfrak{A}_d^{N-1} \mathfrak{B}_d, \underline{\phi}_u \rangle & \langle \mathfrak{A}_d^{N-2} \mathfrak{B}_d, \underline{\phi}_u \rangle & \dots & \langle \mathfrak{B}_d, \underline{\phi}_u \rangle \end{bmatrix}$$

$$U = \begin{bmatrix} u(k+1|k) & u(k+2|k) & \dots & u(k+N|k) \end{bmatrix}^T$$

### 3.3.2 Continuous-Time Observer Design

For the purpose of state reconstruction of a diffusion-convection-reaction system, where the feedforward term  $\mathfrak{D}$  is generally absent, the continuous-time observer dynamics are given by (??). where  $\hat{\underline{x}}(\zeta, t)$  is the reconstructed state of the original system and  $\mathfrak{L}_c$  is the continuous-time observer gain. By subtracting the observer dynamics from the original system dynamics, the error dynamics  $e(\zeta, t)$  are obtained as shown in (3.20).

$$\dot{e}(\zeta, t) = (\mathfrak{A} - \mathfrak{L}_c \mathfrak{C})e(\zeta, t) \equiv \mathfrak{A}_o e(\zeta, t) \quad (3.20)$$

The goal is to design the observer gain  $\mathfrak{L}_c$  such that the error dynamics are exponentially stable, i.e.  $\max\{\text{Re}(\lambda_o)\} < 0$  where  $\{\lambda_o\}$  is the set of eigenvalues of the error dynamics operator  $\mathfrak{A}_o$ . Three different forms of the observer gain are considered as spatial functions  $\mathfrak{L}_c = f(\zeta, l_{obs})$  with the effect of the scalar coefficient  $l_{obs}$  on

$\max\{\text{Re}(\lambda_o)\}$  shown in Fig. 3.4.

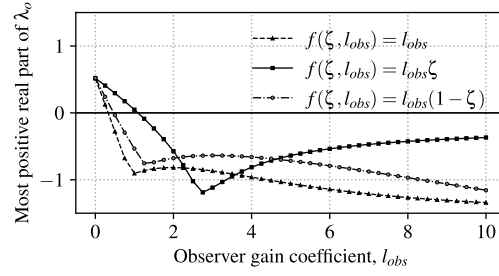


Figure 3.4: The effect of various observer gains  $\mathfrak{L}_c = f(\zeta, l_{obs})$  on the eigenvalues of state reconstruction error dynamics  $\lambda_o$ .

### 3.3.3 Discrete-Time Observer Design

Once an appropriate continuous-time observer gain is determined, the discrete-time observer gain  $\mathfrak{L}_d$  may be obtained using the same Cayley-Tustin time discretization approach, as shown in (3.21).

$$\begin{aligned}\hat{\underline{x}}(\zeta, k) &= \mathfrak{A}_d \hat{\underline{x}}(\zeta, k-1) + \mathfrak{B}_d u(k) + \mathfrak{L}_d [y(k) - \hat{y}(k)] \\ \hat{y}(k) &= \mathfrak{C}_{d,o} \hat{\underline{x}}(\zeta, k-1) + \mathfrak{D}_{d,o} u(k) + \mathfrak{M}_{d,o} y(k)\end{aligned}\tag{3.21}$$

with  $\mathfrak{A}_d$  and  $\mathfrak{B}_d$  defined in (3.14), and  $\mathfrak{C}_{d,o}$ ,  $\mathfrak{D}_{d,o}$ ,  $\mathfrak{M}_{d,o}$ , and  $\mathfrak{L}_d$  are given in (3.22).

$$\begin{aligned}\mathfrak{C}_{d,o}(\cdot) &= \sqrt{2\alpha} [I + \mathfrak{C}(\alpha I - \mathfrak{A})\mathfrak{L}_c]^{-1} \mathfrak{C}\mathfrak{R}(\alpha, \mathfrak{A})(\cdot) \\ \mathfrak{D}_{d,o} &= [I + \mathfrak{C}(\alpha I - \mathfrak{A})\mathfrak{L}_c]^{-1} \mathfrak{C}\mathfrak{R}(\alpha, \mathfrak{A})\mathfrak{B} \\ \mathfrak{M}_{d,o} &= [I + \mathfrak{C}\mathfrak{R}(\alpha, \mathfrak{A})\mathfrak{L}_c]^{-1} \mathfrak{C}\mathfrak{R}(\alpha, \mathfrak{A})\mathfrak{L}_c \\ \mathfrak{L}_d &= \sqrt{2\alpha}\mathfrak{R}(\alpha, \mathfrak{A})\mathfrak{L}_c\end{aligned}\tag{3.22}$$

It can be shown that using this approach, the discrete-time error dynamics will be stable if the continuous-time observer gain  $\mathfrak{L}_c$  is chosen such that  $\mathfrak{A}_o$  is stable. It is also worth noting that the proposed methodology skips the need for model reduction associated with the discrete-time Luenberger observer, with no spatial approximation required as well [Ali2015Ali2015Review, 14, 30–32].

### 3.3.4 Model predictive control design, output feedback implementation

To enable real-time implementation under limited state access, the discrete-time model predictive controller is now augmented with the obtained discrete-time Luenberger observer. The reconstructed state  $\hat{\underline{x}}(\zeta, k)$  is substituted for the full state in the MPC formulation, yielding an observer-based output-feedback controller, as illustrated in Fig. 3.5.

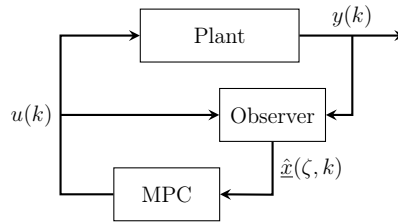


Figure 3.5: Block diagram representation of the observer-based MPC.

The cost function and terminal condition remain unchanged, but the predicted state trajectory is now driven by the estimated state:

$$\begin{aligned}
 \min_U \quad & \sum_{l=0}^{N-1} \langle \hat{\underline{x}}(\zeta, k+l|k), \mathbf{Q} \hat{\underline{x}}(\zeta, k+l|k) \rangle \\
 & + \langle u(k+l+1|k), \mathbf{R} u(k+l+1|k) \rangle \\
 & + \langle \hat{\underline{x}}(\zeta, k+N|k), \mathbf{P} \hat{\underline{x}}(\zeta, k+N|k) \rangle
 \end{aligned} \tag{3.23}$$

$$\begin{aligned}
 \text{s.t.} \quad & \hat{\underline{x}}(\zeta, k+l|k) = \mathbf{A}_d \hat{\underline{x}}(\zeta, k+l-1|k) + \mathbf{B}_d u(k+l|k) \\
 & u^{min} \leq u(k+l|k) \leq u^{max} \\
 & \langle \hat{\underline{x}}(\zeta, k+N|k), \phi_u(\zeta) \rangle = 0
 \end{aligned}$$

The observer provides  $\hat{\underline{x}}(\zeta, k)$  at each time step by processing most recent output and control input. This reconstructed state initializes the prediction horizon and closes the loop in the absence of full-state access. The resulting control law inherits

all properties of the full-state MPC while enabling output feedback implementation. The QP formulation follows analogously by substituting  $\underline{x}$  with  $\hat{\underline{x}}$  in (3.19), and is detailed in (3.24).

$$\begin{aligned}
 \min_U J &= U^\top \langle I, H \rangle U + 2U^\top \langle I, P \hat{\underline{x}}(\zeta, k|k) \rangle \\
 \text{s.t.} \quad &U^{\min} \leq U \leq U^{\max} \\
 &T_u \hat{\underline{x}}(\zeta, k|k) + S_u U = 0
 \end{aligned}$$

with  $H =$

$$\begin{bmatrix}
 \mathfrak{B}_d^* \mathfrak{P} \mathfrak{B}_d + \mathfrak{F} & \mathfrak{B}_d^* \mathfrak{A}_d^* \mathfrak{P} \mathfrak{B}_d & \cdots & \mathfrak{B}_d^* \mathfrak{A}_d^{*N-1} \mathfrak{P} \mathfrak{B}_d \\
 \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d \mathfrak{B}_d & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{B}_d + \mathfrak{F} & \cdots & \mathfrak{B}_d^* \mathfrak{A}_d^{*N-2} \mathfrak{P} \mathfrak{B}_d \\
 \vdots & \vdots & \ddots & \vdots \\
 \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^{N-1} \mathfrak{B}_d & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^{N-2} \mathfrak{B}_d & \cdots & \mathfrak{B}_d^* \mathfrak{P} \mathfrak{B}_d + \mathfrak{F}
 \end{bmatrix} \quad (3.24)$$

$$P = \left[ \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d \quad \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^2 \quad \cdots \quad \mathfrak{B}_d^* \mathfrak{P} \mathfrak{A}_d^N \right]^\top$$

$$T_u(\cdot) = \left[ \langle \mathfrak{A}_d^N(\cdot), \underline{\phi}_u \rangle \right]$$

$$S_u = \left[ \langle \mathfrak{A}_d^{N-1} \mathfrak{B}_d, \underline{\phi}_u \rangle \quad \langle \mathfrak{A}_d^{N-2} \mathfrak{B}_d, \underline{\phi}_u \rangle \quad \cdots \quad \langle \mathfrak{B}_d, \underline{\phi}_u \rangle \right]$$

$$U = \left[ u(k+1|k) \quad u(k+2|k) \quad \cdots \quad u(k+N|k) \right]^\top$$

### 3.4 Simulation Results

This section presents numerical simulations of the closed-loop system under both full-state feedback and output-feedback model predictive control schemes. The reactor model and all physical parameters follow those in Table 3.1, and the same control settings are used throughout: initial condition  $c(\zeta, 0) = \sin^2(\pi\zeta)$ , empty recycle stream, state and input penalty weights  $Q = 0.04I$ ,  $F = 27$ , sampling time  $\Delta t = 20$  s, control horizon  $N = 9$ , and input constraints  $0 \leq u(t) \leq 0.15$ . The control horizon corresponds to 180 s, which exceeds the recycle delay of 80 s. The subsections below compare the controller performance under full-state and output-feedback implemen-

tations.

### 3.4.1 Full-State Feedback MPC Performance

As the eigenvalue distribution obtained in Fig. 3.2 suggests, the open-loop system is unstable due to the presence of an eigenvalue with positive real part. The zero-input response of the system is shown in Fig. 3.6 where the initial condition for the reactor is set to  $c(\zeta, 0) = \sin^2(\pi\zeta)$ . The recycle stream is assumed to be empty at the beginning of the simulation.

An infinite-dimensional MPC is designed and applied to the unstable system. The closed-loop response of the system is shown in Fig. 3.7 and the control input as well as the measured output is shown in Fig. 3.8. It may be confirmed that the MPC successfully stabilizes the unstable system while satisfying the input constraints.

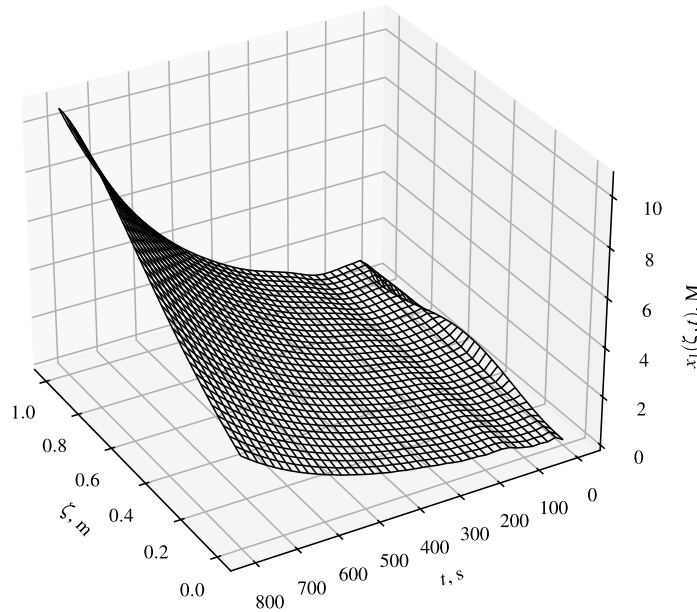


Figure 3.6: Open-loop concentration profile along the reactor.

One interesting aspect of considering a recycle stream is the oscillatory behavior

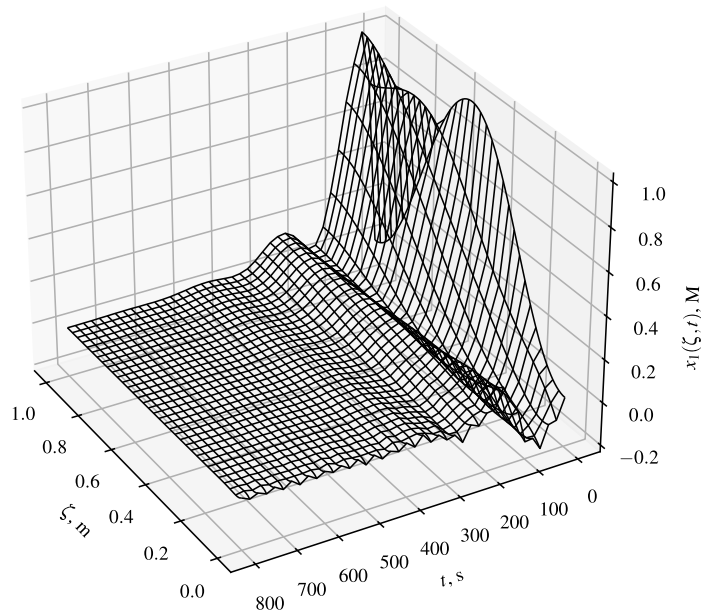


Figure 3.7: Stabilized reactor concentration profile under the proposed full-state MPC.

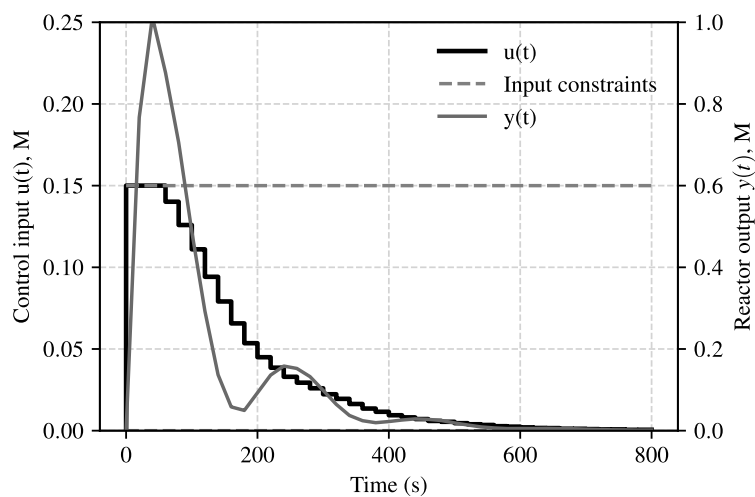


Figure 3.8: Input profile and reactor output under full-state MPC, subject to constraints.

of the system dynamics. While axial dispersion reactors show no oscillation in the absence of recycle, the nature of recycle streams can introduce such behavior. The choice of control horizon is another key factor. A short control horizon relative to the residence time of the recycle stream can lead to oscillatory input profiles due to the presence of delayed recycle stream. In this example, the control horizon, i.e., 180 s, is set to be considerably longer than the recycle delay, which is 80 s; resulting in a non-oscillatory input profile.

### 3.4.2 Observer-Based Output Feedback MPC

To evaluate the performance of the output-feedback controller, numerical simulations are conducted under the same conditions as in the full-state feedback case. This subsection presents the closed-loop behavior of the system when using the discrete-time Luenberger observer to reconstruct the states based on output measurements.

The eigenvalue distribution shown previously in Fig. 3.2 confirms that the open-loop system is unstable due to the presence of an eigenvalue with a positive real part. The observer gain is selected as a constant function  $L_c = 1$ , and the estimated state is initialized to zero across the domain.

The closed-loop reactor response under the proposed output-feedback controller is shown in Fig. 3.9, and the corresponding control input and measured output are shown in Fig. 3.10. The evolution of the state estimation error is depicted in Fig. 3.11. These results confirm that the observer-based MPC successfully stabilizes the unstable system while adhering to input constraints, using only output measurements.

An important aspect of the proposed observer-based controller is the relative speed of the observer error convergence compared to the system dynamics. As seen in Fig. 3.11, the observer error dynamics decay significantly faster than the closed-loop reactor response, helping prevent oscillations that may arise from poor state

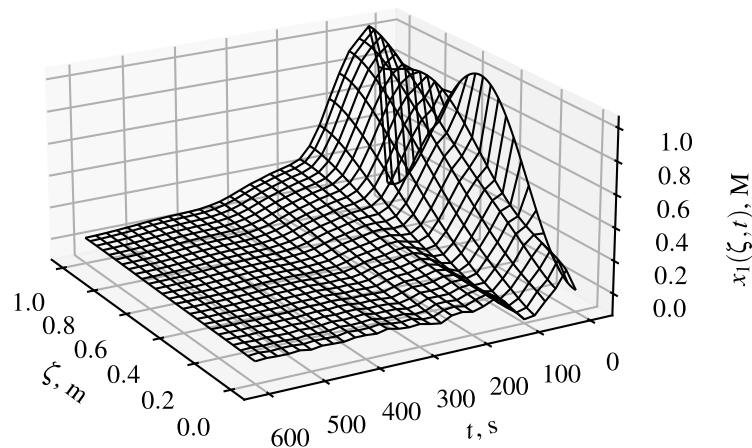


Figure 3.9: Stabilized reactor concentration profile under the proposed observer-based MPC.

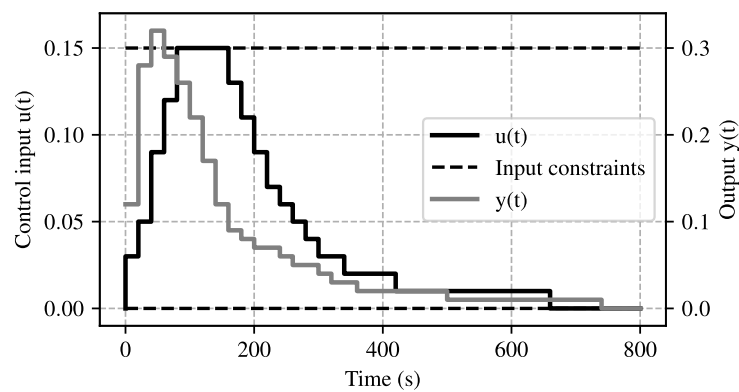


Figure 3.10: Input profile and reactor output under observer-based MPC.

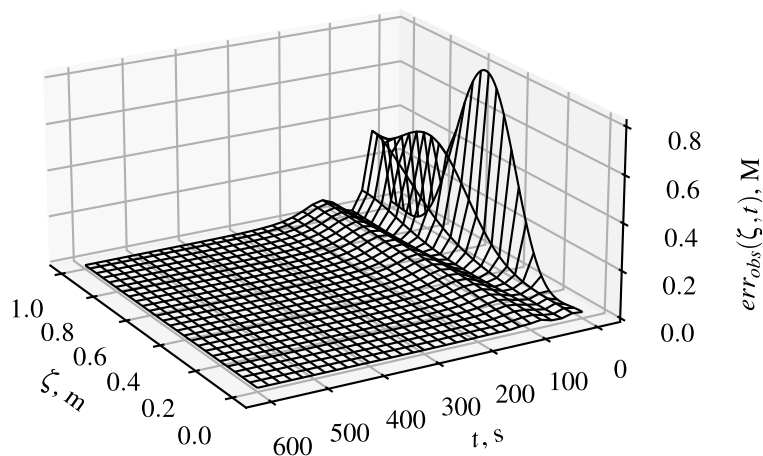


Figure 3.11: State reconstruction error profile along the reactor.



reconstruction.

Oscillatory behavior induced by the recycle stream is discussed at the end of the previous subsection. Since all simulation settings are shared, the same rationale applies here; the long control horizon relative to the recycle delay ensures a smooth input profile and stable closed-loop response.

### 3.5 Conclusion

In this work, model predictive control of an axial dispersion tubular reactor equipped with recycle is addressed, while considering the delay imposed by the recycle stream. This setup is common in industry but has received limited attention in the chemical engineering distributed parameter systems literature. The diffusion-convection-reaction dynamics of the reactor is modeled by a second-order parabolic PDE, while a notion of state delay is introduced to account for the delay imposed by the recycle stream. The state delay is addressed as a separate transport PDE, resulting in a boundary-controlled system governed by a coupled set of parabolic and hyperbolic PDEs under Danckwerts boundary conditions. Utilizing a late-lumping approach, the resolvent operator is obtained in a closed form in order to preserve the infinite-dimensional nature of the system without requiring spatial discretization. To implement MPC as a digital controller, the Cayley–Tustin transformation is used. This Crank–Nicolson type of discretization is chosen as it maintains important properties of the system such as stability and controllability when mapping the continuous-time system to a discrete-time one. Numerical simulations demonstrate the effectiveness of the proposed controller in stabilizing an unstable system while satisfying input constraints under full-state feedback. Recognizing, however, that full-state information is often unavailable in practical implementations of distributed parameter systems, this work is further extended by designing and integrating a discrete-time Luenberger

observer to reconstruct the state from output measurements alone, without any spatial approximation. A family of observer gains is examined, and spectral analysis is performed to select gains that ensure the state reconstruction error converges faster than the closed-loop system dynamics. This guarantees accurate state estimates during transients and prevents performance degradation due to estimation delay. The proposed approach can be further extended to incorporate the effects of temperature as well as disturbance rejection or set-point tracking in future work.

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- [38] Q. Xu and S. Dubljevic, “Linear model predictive control for transport-reaction processes,” *AIChE Journal*, vol. 63, no. 7, pp. 2644–2659, 2017, ISSN: 0001-1541. DOI: [10.1002/aic.15592](https://doi.org/10.1002/aic.15592).
- [39] E. Hairer, M. Hochbruck, A. Iserles, and C. Lubich, “Geometric numerical integration,” *Oberwolfach Reports*, vol. 3, no. 1, pp. 805–882, 2006, ISSN: 1660-8933. DOI: [10.4171/owr/2011/16](https://doi.org/10.4171/owr/2011/16).

# Chapter 4

## Figures, Tables, & Plates

### 4.1 Introduction

Figures and Tables play a crucial role in conveying information effectively in academic documents. This chapter will delve into the intricacies of incorporating figures and tables in your  $\text{\LaTeX}$  document, exploring various features and advanced techniques to enhance the visual appeal and clarity of your content. While there are many types of figures that one might have in their document, this Chapter more specifically focuses on the inclusion of graphic figures (pictures), as well as how to lay out multi figures (sub-figures).

**⚠ WARNING** Throughout this Chapter there will be code listings that include lines that show the required packages. These lines should be included in your  $\text{\LaTeX}$  preamble, not in the body of your document. To make this easy I have actually included a document specifically to add your packages too. This can be found in the `includePackages.tex` file found in the `00_LaTeX_Files` folder.

💡 `./00_LaTeX_Files/includePackages.tex` includes all the packages required for creating all the examples in this document, some of these will not be necessary for your own thesis, however, I have included comments on all packages for what they are used for. This allows you to make a decision on if you will need them as well. If you are unsure, you may just comment out the line so as to not forget the packages that were originally included.

## 4.2 Inserting Figures

In  $\text{\LaTeX}$ , figures are included using the `graphicx` package. The `\includegraphics` command is used to insert an image. Let's consider an example:

Listing 4.1: A Basic Example of Including a Figure.

```
\usepackage{graphicx}

\begin{figure}[htb]
  \centering
  \includegraphics[width=0.7\linewidth]{example-image}
  \caption{Example Figure}
  \label{fig:example}
\end{figure}
```

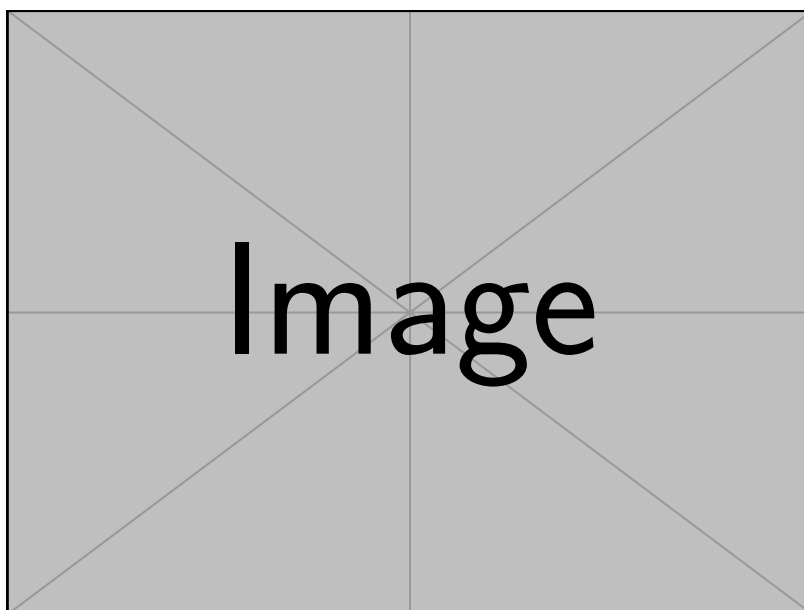


Figure 4.1: This is an example of a single figure similar to that produced by Listing 4.1.

In this example, the `figure` environment is used to contain the image. The `\centering` command ensures the image is centred horizontally. The `width` parameter is used to control the size of the image; in this case, it has been set to `0.45\cmd {linewidth}` which will make it fill a space that is 0.45 tiems the width of

the current line. The `\caption` and `\label` commands provide a caption and label for referencing, respectively.

Figures can be formatted to meet specific requirements. The `\subfigure` command from the `subcaption` package can be used for side-by-side figures:

```
\usepackage{subcaption}

\begin{figure}[htb]
  \begin{subfigure}{0.45\linewidth}
    \centering
    \includegraphics[width=\linewidth]{example-image-a}
    \caption{Subfigure A} % Leave blank for just letter
    \label{subfig:a}
  \end{subfigure}
  \hfill
  \begin{subfigure}{0.45\linewidth}
    \centering
    \includegraphics[width=\linewidth]{example-image-b}
    \caption{Subfigure B} % Leave blank for just letter
    \label{subfig:b}
  \end{subfigure}
  \caption{Example with Subfigures}
  \label{fig:subfigures}
\end{figure}
```

This example uses the `subfigure` environment to create subfigures within a larger figure (as show in [Figure 4.2](#). The `\hfill` command adds horizontal space between the subfigures.

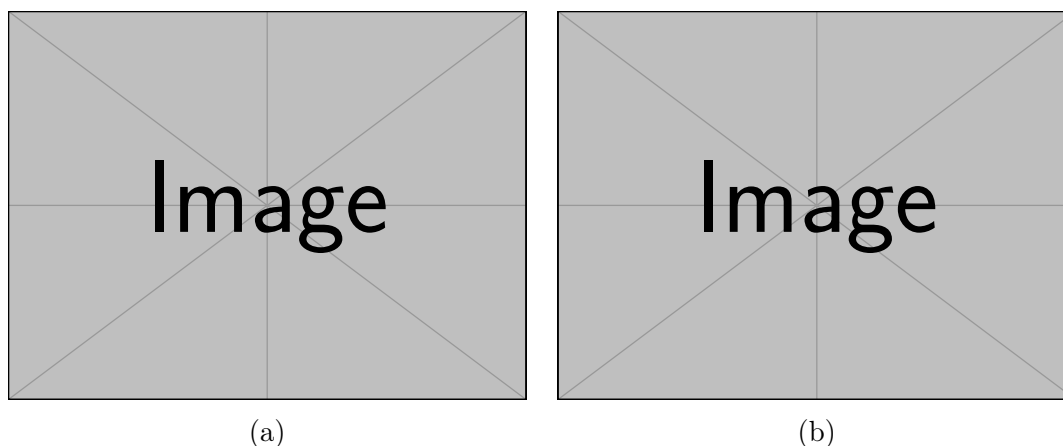


Figure 4.2: This is an example of a double image figure similar to that produced by [Section 4.2](#).

While this section provided a few examples on how to make some figures and subfigures, I would strongly recommend checking out some of the more complex examples of figures shown in [Appendix A](#). This will provide the code and examples for how to create more intricate subfigures and layouts.

### 4.3 Tables and Tabularx

Tables in  $\text{\LaTeX}$  are created using the `tabular` environment. The `tabularx` package is particularly useful when you want the table to automatically adjust its width. Let's define some custom column types for convenience:

```
\usepackage{tabularx}
\newcolumntype{C}{>\centering\arraybackslash}X}
\newcolumntype{L}{>\raggedright\arraybackslash}X}
\newcolumntype{R}{>\raggedleft\arraybackslash}X}
```

Now, let's create a table using `tabularx`:

In this example, the `tabularx` environment is used, and the custom column types `C`, `L`, and `R` are applied to the columns. This ensures the content is centered, left-justified, and right-justified, respectively.



```

\begin{table}[htb]
  \centering
  \begin{tabularx}{\linewidth}{|C|L|R|}
    \hline
    \textbf{Centered} & \textbf{Left Justified} & \textbf{Right Justified} \\
    \hline
    Content & More content & Additional content \\
    \hline
  \end{tabularx}
  \caption{Example Table with Tabularx}
  \label{tab:example}
\end{table}

```

## 4.4 Advanced Table Features

To create professional-looking tables, the `booktabs` package can be employed. It provides commands for better spacing and styling of tables:

```

\usepackage{booktabs}

\begin{table}[htb]
  \centering
  \begin{tabular}{ccc}
    \toprule
    \textbf{Header 1} & \textbf{Header 2} & \textbf{Header 3} \\
    \midrule
    Content 1 & Content 2 & Content 3 \\
    Content 4 & Content 5 & Content 6 \\
    \bottomrule
  \end{tabular}
  \caption{Example Table with Booktabs}
  \label{tab:booktabs_example}
\end{table}

```

The `\toprule`, `\midrule`, and `\bottomrule` commands create horizontal rules with appropriate spacing.

## 4.5 Additional Packages for Enhanced Table Functionality

Several other packages can be employed to enhance table functionality:

The `longtable` package allows tables to span multiple pages, which is useful for large datasets. The `multirow` and `multicolumn` packages provide commands for cells that span multiple rows or columns, respectively. The `makecell` package enables more complex table layouts. Each of these packages comes with its set of commands and options. Let's briefly explore the usage of `longtable`, `multirow`, and `multicolumn`:

In these examples, the `longtable` environment is used for tables that span multiple pages. The `\multirow` command is employed to create cells that span multiple rows, while `\multicolumn` is used for cells that span multiple columns.

## 4.6 Conclusion

This Chapter provided a comprehensive overview of including figures and tables in your  $\text{\LaTeX}$  document. From basic insertion of figures to advanced table formatting using packages like `tabularx`, `booktabs`, and others, you now have a solid foundation to include tables and figures in your thesis. While this provides a lot of details on how to add figures that have already been pre-generated, one might want to generate figures on the spot potentially even using data generated from other programs (such as MatLab<sup>®</sup>, Python, *etc.*). For this [Chapter 5](#) provides in-depth workings of the `pgfplots` package and how to generate consistent and professional looking plots and graphs.

```

\usepackage{longtable}
\usepackage{multirow}
\usepackage{multicolumn}

% Example Longtable
\begin{longtable}{|c|c|}
  \caption{Longtable Example} \label{tab:longtable} \\
  \hline
  \textbf{Header 1} & \textbf{Header 2} \\
  \hline
  \endfirsthead
  \hline
  \textbf{Header 1} & \textbf{Header 2} \\
  \hline
  \endhead
  Content 1 & Content 2 \\
  Content 3 & Content 4 \\
  \hline
\end{longtable}

% Example Multirow and Multicolumn
\begin{table}[htb]
  \centering
  \begin{tabular}{|c|c|c|}
    \hline
    \multirow{2}{*}{\textbf{Multirow-Col1}} & \multicolumn{2}{|c|}{\textbf{Multicolumn-Col2-3}} \\
    \cline{2-3}
    & \textbf{Column 2} & \textbf{Column 3} \\
    \hline
    Content 1 & Content 2 & Content 3 \\
    \hline
  \end{tabular}
  \caption{Example Table with Multirow and Multicolumn}
  \label{tab:multirow_multicolumn}
\end{table}

```

# Chapter 5

## Plots, Charts, & Graphs

Throughout this chapter we will be exploring some of the different ways of displaying your data in your thesis. Mainly this will be accomplished with the `pgfplots` package. In the following sections, there will be a few examples of how to generate different plots. For more information on how to create plots, [here](#) is the manual for `pgfplots` (the package used to generate the information for TikZ to create the plots). For an extensive list of examples please refer to [this](#)

### 5.1 Line Plots

A simple line plot can be effectively created using the `axis` environment from the `pgfplots` package in L<sup>A</sup>T<sub>E</sub>X. The `pgfplots` package is a powerful tool for creating high-quality plots directly within L<sup>A</sup>T<sub>E</sub>X documents. It builds upon the `TikZ` package and provides a comprehensive set of options for plotting and customizing graphs.

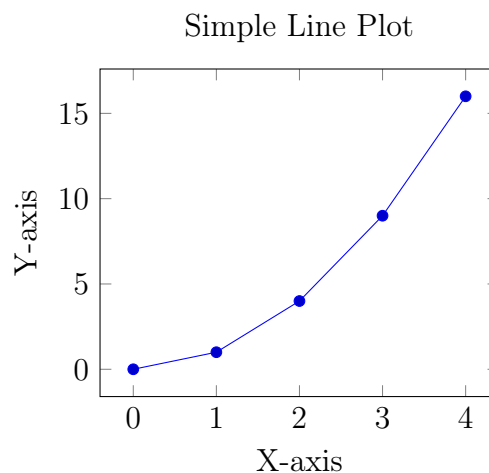
The following code (see [Figure 5.1a](#)) can be used to create the figure shown in [Figure 5.1b](#). Expanding on this example, we can add a second plot by adding the following code below the closing bracket and semi-colon (`};`) of the `\addplot` command: This will result in the addition of the second red line shown in [Figure 5.2](#).

```

\begin{figure}[htbp]
  \centering
  \begin{tikzpicture}
    \begin{axis}[
      title={Simple Line Plot},
      xlabel={X-axis},
      ylabel={Y-axis},
    ]
      \addplot coordinates {
        (0,0)
        (1,1)
        (2,4)
        (3,9)
        (4,16)
      };
    \end{axis}
  \end{tikzpicture}
  \caption{A simple line plot.}
  \label{fig:line-plot}
\end{figure}

```

(a)



(b)

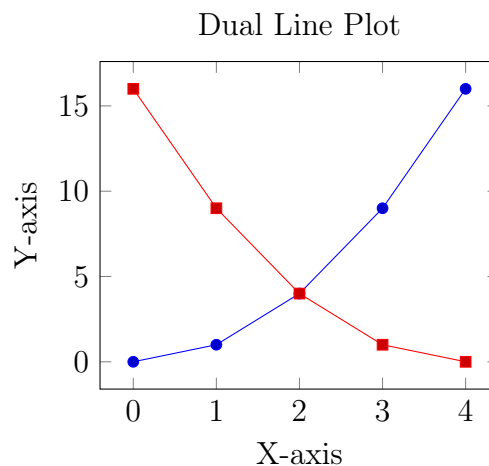
Figure 5.1: A simple line plot (b) and the code to generate the plot (a)

```

\begin{tikzpicture}
  \begin{axis}[
    title={Dual Line Plot},
    xlabel={X-axis},
    ylabel={Y-axis},
  ]
    \addplot coordinates {
      (0,0)
      (1,1)
      (2,4)
      (3,9)
      (4,16)
    };
    \addplot coordinates {
      (0,16)
      (1,9)
      (2,4)
      (3,1)
      (4,0)
    };
  \end{axis}
\end{tikzpicture}

```

(a)



(b)

Figure 5.2: A simple line plot with two sets of data.

```
\addplot coordinates {
    (0,16)
    (1,9)
    (2,4)
    (3,1)
    (4,0)
};
```

## 5.2 Customizing Plots

This section provides some ways to increase the readability and customization of the plots we generate.

### 5.2.1 Adding a Legend

Legends can be added to plots for better readability. To add a legend to your plot you can use the code in [Section 5.2.1](#) to generate the plot shown in [Figure 5.3](#). Each plot is individually added to the legend by adding a `\addlegendentry{YOUR LEGEND ENTRY HERE}` command following the `\addplot` command.

**Note:** *The position of the legend can be specified by using the optional parameter `legend pos=` followed by a set of compass coordinates.*

```
\begin{figure}[H]
  \centering
  \begin{tikzpicture}
    \begin{axis}[
      title={Plot with Added Legend},
      xlabel={X-axis},
      ylabel={Y-axis},
      legend pos=north west,
    ]
```

```

\addplot coordinates {
    (0,0)
    (1,1)
    (2,4)
    (3,9)
    (4,16)
};
\addlegendentry{\(y = x^2\)}
\addplot coordinates {
    (0,16)
    (1,9)
    (2,4)
    (3,1)
    (4,0)
};
\addlegendentry{\(y = 16 - x^2\)}
\end{axis}
\end{tikzpicture}
\caption{A customized plot with a legend.}
\label{fig:legend-plot}
\end{figure}

```

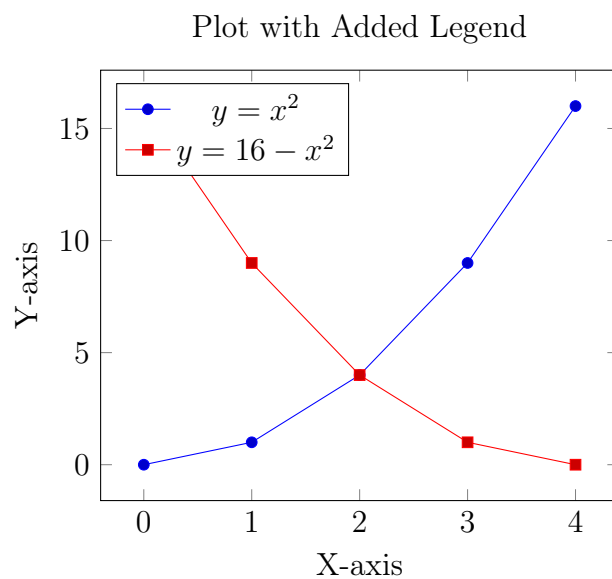


Figure 5.3: A customized plot with a legend.

### 5.2.2 Adding Grid Lines

To add gridlines to your plot

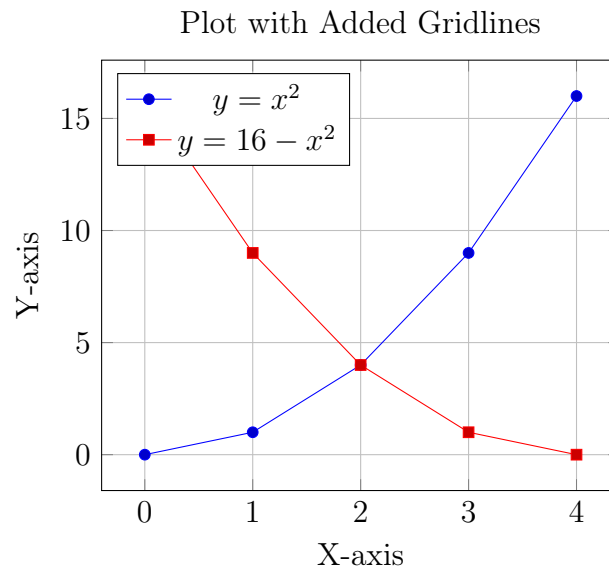


Figure 5.4: A customized plot with added gridlines.

### 5.2.3 Changing Colors and Line Styles

Colors and line styles can be easily modified:



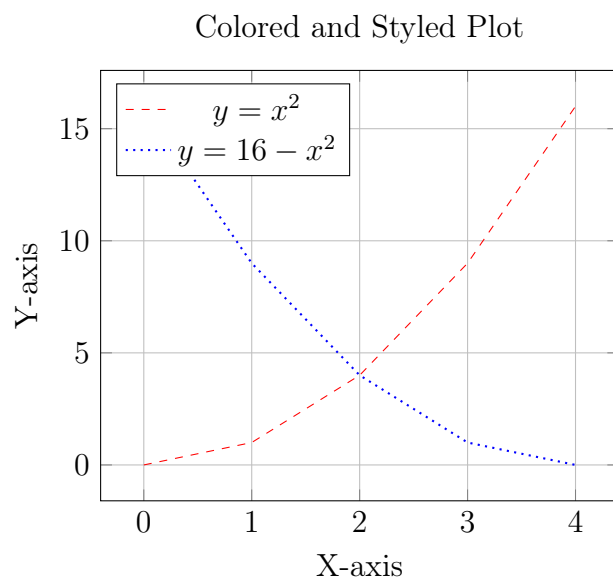


Figure 5.5: A plot with customized colors and line styles

## 5.3 Advanced Plot Types

### 5.3.1 Equations

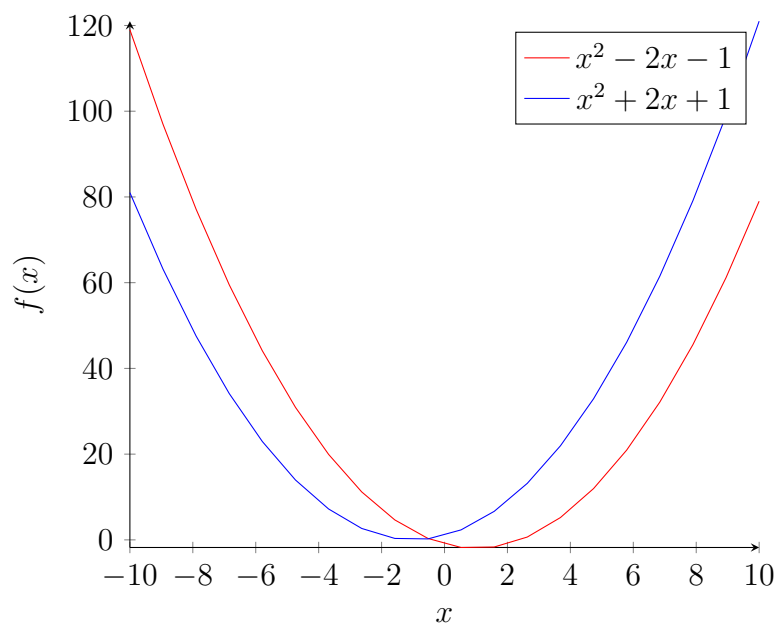


Figure 5.6: Plot of two parabola.

### 5.3.2 Scatter Plot with External Data

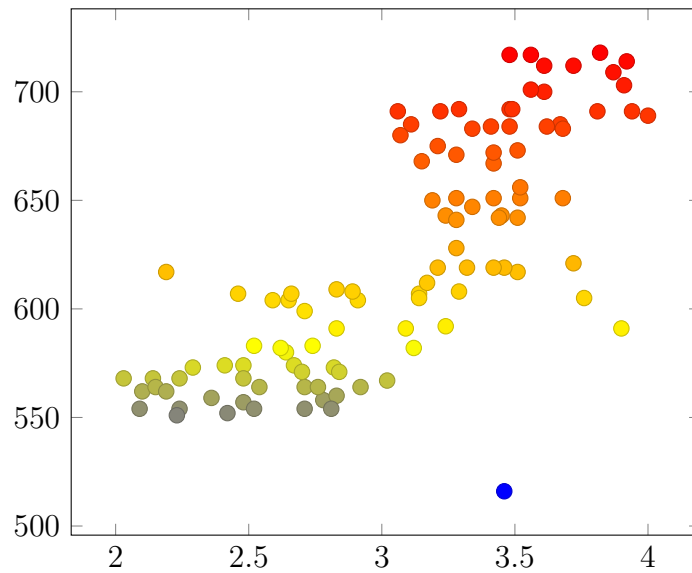


Figure 5.7: Example of a Scatter Plot.

### 5.3.3 Bar Plot

Bar plots are useful for categorical data. Here's how to create one:

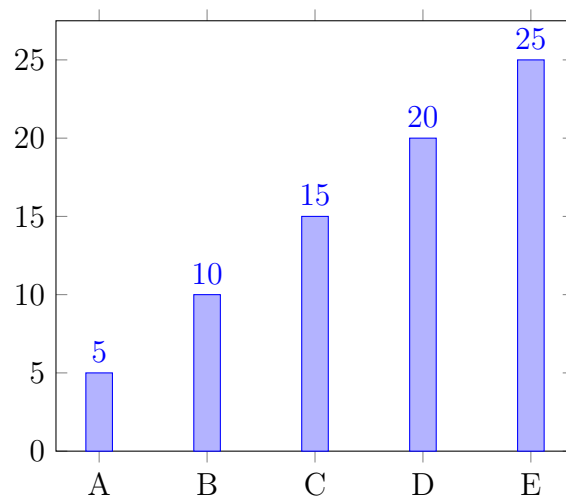


Figure 5.8: A bar plot

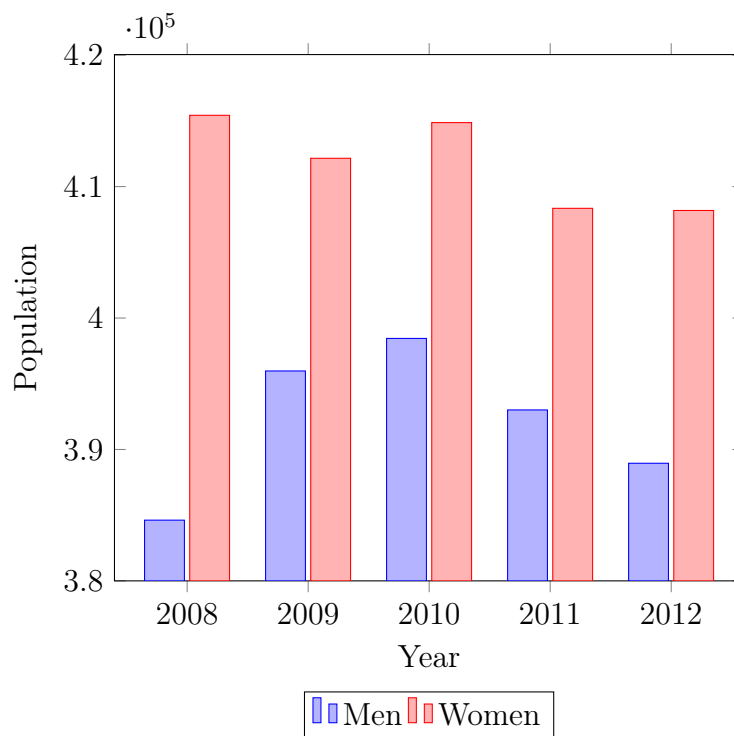


Figure 5.9: Example of a Bar Graph.

### 5.3.4 Pie Chart

Pie charts are less common in  $\text{\LaTeX}$ , as well as, a lot of other media due to the poor representation of the data. However, if you are still inclined to use them the remainder of this section will show how the package `pgf-pie` can be used to create consistent graphs that look good.

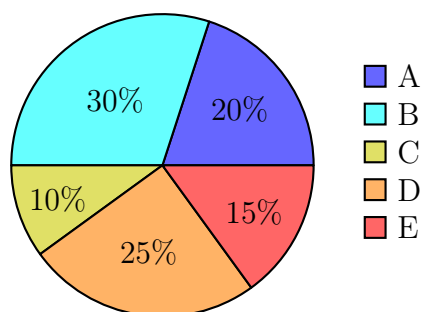


Figure 5.10: A basic pie chart.

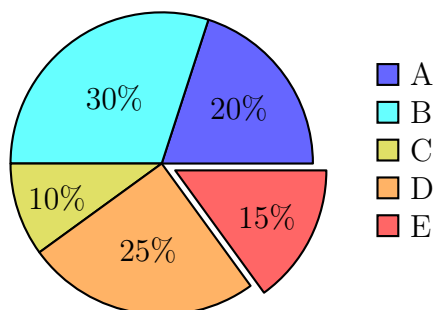


Figure 5.11: A pie chart with an “Exploded” slice.

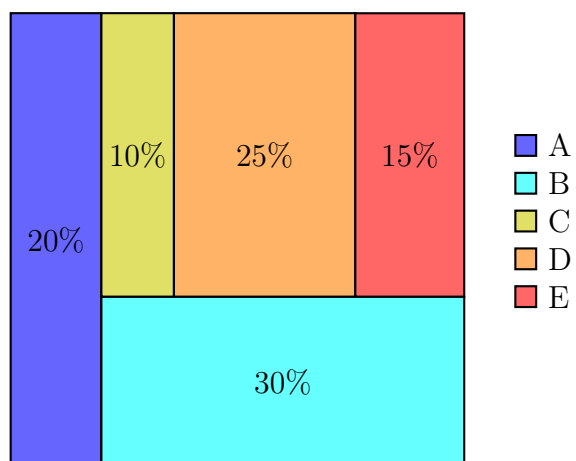


Figure 5.12: A “square” pie chart.

### 5.3.5 3D Plot

3D plots can be created for more complex data visualization:

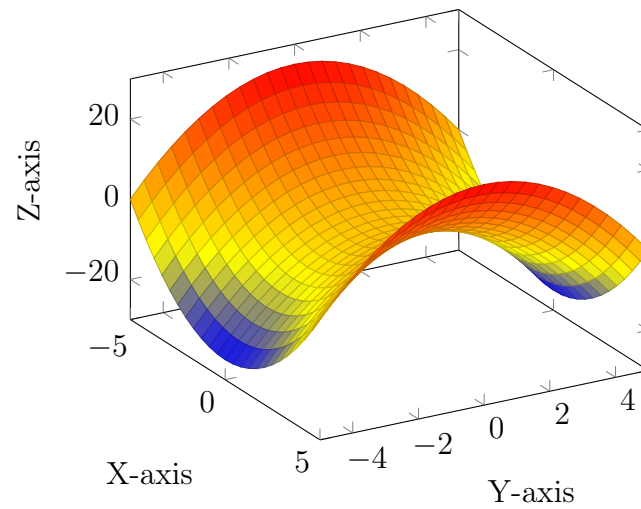


Figure 5.13: A 3D surface plot

Example using the mesh parameter

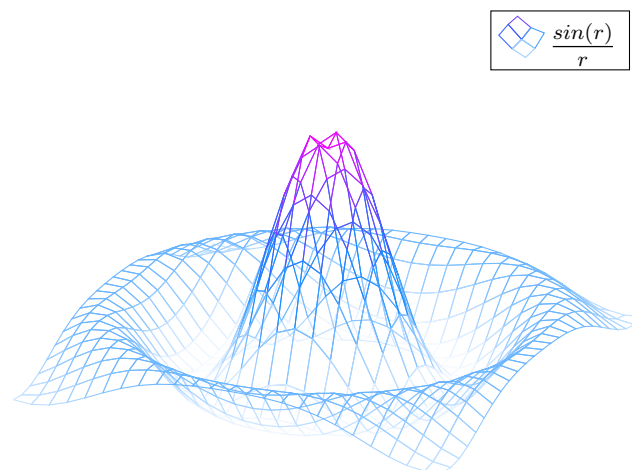


Figure 5.14: Example of a 3D Plot

### 5.3.6 Polar Plot

Polar plots are useful for circular data:

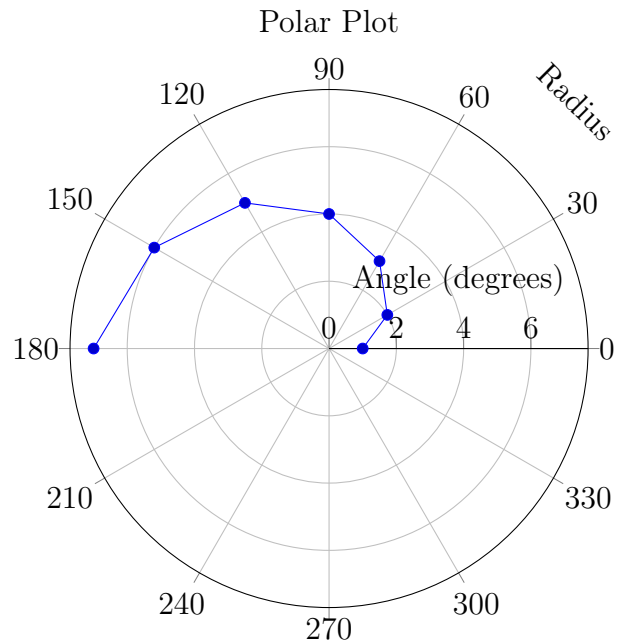


Figure 5.15: A polar plot

### 5.3.7 Box Plot

Box plots are used to visualize the distribution of data:

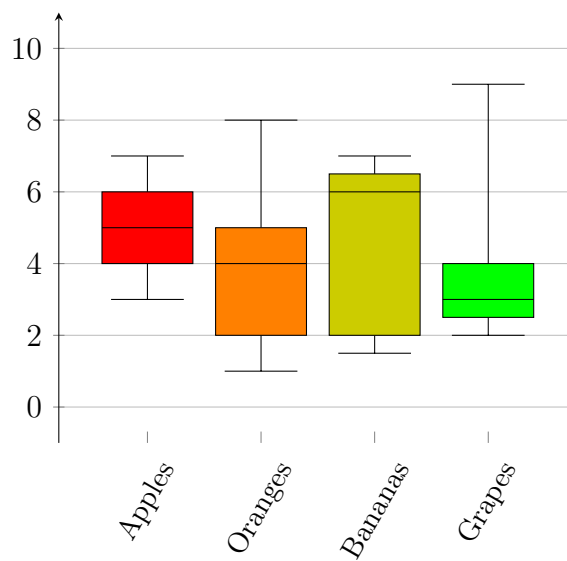


Figure 5.16: A box plot

## 5.4 Conclusion

The `pgfplots` package is an incredibly versatile tool for creating a wide range of plots and graphs in L<sup>A</sup>T<sub>E</sub>X. This chapter has provided examples of various plot types and customization options, showcasing the power and flexibility of `pgfplots`. By leveraging these capabilities, you can create high-quality, publication-ready figures for your thesis.

## Chapter 6

# Mathematical Equations

There are many ways to include formulas in your thesis. This section will provide some different ways of adding them (inline and standalone), as well as provide some ways of referencing the equations.

To start the simplest way to add an equation is using the built-in L<sup>A</sup>T<sub>E</sub>X math mode. To enter and exit math mode one just needs to use the `\(` and `\)` symbols around an equation. While there also exists `$(Equation)$` to add math, it is not recommended due to potential compatibility issues. Additionally, this, `\(<Equation>\)`, method is capable of being redefined to add further customization. An example of using math mode to get an inline equation is by using the following command:

```
\(\vec{F}_{d}=\frac{1}{2}A\ C_{d}\ \vec{V}^2\)
```

The above command has the effect of creating the following output:  $\vec{F}_d = \frac{1}{2} A C_d \vec{V}^2$ . Sometimes it can be quite beneficial to separate what would be an inline equation to be on its own line. For this, we have two different ways of doing it. The first way will produce an equation that has no reference:

$$E = m c^2$$

$$E = m c^2$$



The second will produce an equation with a reference. For this, there are two main ways of creating the reference, the first one, see [Equation \(6.1\)](#), creates a numbered reference; the other one, see [Equation \(Constant pi\)](#), creates a reference with a ‘tag’. The difference between the two is the inclusion of a `\tag{<text>}` command that will replace the regular number with `<text>` and the changing from the `equation` environment to the `equation*` environment. If you do not want the brackets around the tag, as shown in [Equation \(Constant pi\)](#), use the starred version of the command: `\tag*{<text>}`. This will not remove the braces in the reference for the equation, but will remove them from appearing next to the equation definition.

**NOTICE** Using the `\tag` command in conjunction with the `\label` command can create a L<sup>A</sup>T<sub>E</sub>X warning when used in the non-star `equation` environment. This warning can be safely ignored, however, the better way to deal with this is to make sure one is using the star version, `equation*`.

$$\pi = 3.14... \tag{6.1}$$

$$\pi = 3.1415... \tag{Constant pi}$$

$$\pi = 3.1415... \tag{Constant pi}$$

If you have multiple equations that you want arranged very neatly, use the `align` environment and you can assign individual equations numbers as shown in [Equations \(6.2\)](#) to [\(6.4\)](#). Note that it is the `&` symbol that determines what will be aligned. Further note that spaces in “math mode” are ignored and need to be specified using the space commands in

$$Equation1 = 1 \tag{6.2}$$

$$Equation2 = 2 + 2 \tag{6.3}$$

$$Equation3 = 3 + 3 + 3 \tag{6.4}$$

It may be very important in a math heavy thesis to be able to show your equations,

or even data in a readable way. For this, we will explore some of the ways to create specific data.

## 6.1 Vector, Sets, Piecewise Functions, Matrix Math, and More

$$f(x) = \begin{cases} x^{2*\ln x}, & \text{if } x < 3 \\ -\frac{x}{2}, & \text{if } 3 \leq x \leq 4 \\ x, & \text{if } 4 < x \end{cases} \quad (6.5)$$

Vectors and Matrices are used in many fields of math and science and provide a convenient way to represent 2-Dimensional arrays of numbers.

$$x \in \{1, 2, 3, 4, 5, 6, 7\} \quad (6.6)$$

$$V_1 = \left( \begin{array}{cccc} a, & b, & c, & d \end{array} \right) \quad (6.7)$$

$$V_2 = \left( \begin{array}{c} a \\ b \\ c \\ d \end{array} \right) \quad (6.8)$$

$$M = \left[ \begin{array}{cccc} a & b & c & d \\ e & f & g & h \\ i & j & k & l \\ m & n & o & p \end{array} \right] \quad (6.9)$$

$$p(x) = 3x^6 + 14x^5y + 590x^4y^2 + 19x^3y^3$$

$$- 12x^2y^4 - 12xy^5 + 2y^6 - a^3b^3$$

$$ralign = lalign \quad (6.10)$$

$$ralign = lalign \quad (6.11)$$

$x = y$	$w = z$	$a = b + c$
$2x = -y$	$3w = \frac{1}{2}z$	$a = b$
$-4 + 5x = 2 + y$	$w + 2 = -1 + w$	$ab = cb$

$m$	Modulo	$m > 0$
$a$	Multiplier	$0 < a < m$
$c$	Constant	$0 \leq c < m$
$x_0$	Initial Value	$0 \leq x_0 < m$

$m$	Modulo	$m > 0$
$a$	Multiplier	$0 < a < m$
$c$	Constant	$0 \leq c < m$
$x_0$	Initial Value	$0 \leq x_0 < m$

$eqn1 = fdsjalk$

(6.12)

$eqn2 = fdsa * 243$

(6.13)

$eqn$

$= eqn \quad (6.14)$

## 6.2 Functions

<code>\sin</code>	sin	<code>\cos</code>	cos	<code>\tan</code>	tan
<code>\arcsin</code>	arcsin	<code>\arccos</code>	arccos	<code>\arctan</code>	arctan
<code>\sec</code>	cos	<code>\csc</code>	csc	<code>\cot</code>	cot
<code>\sinh</code>	sinh	<code>\cosh</code>	cosh	<code>\tanh</code>	tanh
<code>\ln</code>	ln	<code>\log</code>	log	<code>\exp</code>	exp

## 6.3

## 6.4 Vector, Sets, Piecewise Functions, Matrix Math, and More

# Chapter 7

## Citations, References, and Cross-References

This section will be showing off some of the different ways to include “citations” and “cross-references” within your document. Note that **cross-references** in  $\text{\LaTeX}$  utilize `\ref` as a command, while one might think that this is short for reference this is not the case citation/references utilize the `\cite` command.

### 7.1 Cross-References

In  $\text{\LaTeX}$ , references will “reference” a `\label{Reference:Label}` command. This section has the following command to define the Chapter:

```
\chapter{Citations/References, and Cross-References}\label{ch:citref}
```

By using `\ref{ch:citref}`, this allows you to insert a cross-reference that look like this 7. Now this by itself is not the most useful, to make it a bit better we should keep track of what we are cross-referencing, in this case a **Chapter**, and add this label in front of the cross-reference (`Chapter~\ref{sec:citref}`) and this will display like this: Chapter 7.

**Note:** *To ensure the reference is not split we need add a non-breaking space (~) to prevent  $\text{\LaTeX}$  from accidentally adding a line-break between the label we added and the actual cross-reference.*

While using the `\ref` command, you might ask “*Why does L<sup>A</sup>T<sub>E</sub>X not just know what it is that I am referencing and insert that automatically in front of the reference?*” The answer is to provide more flexibility to the user. However, that being said, individuals have created a number of packages that work to enhance the workflow of adding these cross-references. Some of these are provided by the `hyperref` and `cleveref` packages. To include these packages add the following lines to the **bottom** of your preamble (order matters, `cleveref` needs to be after `hyperref`; and `hyperref` should be one of the last packages loaded):

```
\usepackage{hyperref}
\usepackage[nameinlink]{cleveref}
```

With these packages loaded we can now use the commands listed in [Table 7.1](#).<sup>1</sup>

---

<sup>1</sup>Note that because the floats are added where they are in the text this causes them to insert large amounts of white space because it only fits on the following page.

Table 7.1: Built-in, hyperref, and cleveref commands and outputs

Command	Output
built-in	
<code>\ref{}</code>	7.1
<code>\pageref{}</code>	94
hyperref	
<code>\autoref{}</code>	Table 7.1
cleveref	
<code>\cref{}</code>	table 7.1
<code>\Cref{}</code>	Table 7.1
<code>\cref*{}</code>	table 7.1
<code>\Cref*{}</code>	Table 7.1
<code>\cpageref{}</code>	page 94
<code>\Cpageref{}</code>	Page 94
<code>\namecref{}</code>	table
<code>\nameCref{}</code>	Table

Further, the `cleveref` also includes features that allows for the auto sorting and combining of references:

```
\Cref{fig:doubleImage,fig:singleImage,fig:
tripleImage1,fig:quadImage}
```

Noting that there are **NO** spaces between the labels; this will produce: **Figures 4.1, 4.2, A.2 and A.4**. Allowing one to quickly and efficiently keep references up-to-date and consistent in their style. More examples of the use of the `cleveref` cross-referencing is found through the rest of this Chapter.

## 7.2 Citations/References

## 7.3 Citation Managers

### 7.3.1 JabRef

Information on the use of this has been moved to it's own Chapter

## 7.4 This is old Material

This section will be showing off some of the different ways to include “citations” and “cross-references” within your document. Note that **cross-references** in  $\text{\LaTeX}$  utilize `\ref` as a command, while one might think that this is short for reference this is not the case citation/references utilize the `\cite{}` commands.

### 7.4.1 Cross-References

In  $\text{\LaTeX}$ , references will “reference” a `\label{Reference:Label}` command. This section has the following command to define the the section:

```
\subsection{Cross-References}\label{subsec:cross-reference}
```

By using `\ref{subsec:cross-reference}`, this allows you to insert a reference that look like this: 7.4.1. Now this by itself is not the most useful, to make it a bit better we should keep track of what we are referencing, in this case a **Section**, and add this label in front of the reference (`Section~\ref{subsec:cross-reference}`) and this will display like this: Section 7.4.1. Note to ensure the reference is not split we add a non-breaking space (`~`) to prevent  $\text{\LaTeX}$  from adding a line break.

While using the `ref` command, you might ask “*Why does  $\text{\LaTeX}$  not just know what it is that I am referencing and insert that automatically in front of the reference?*” The answer is to provide more flexibility to the user. However, that being said,



individuals have created a number of packages that work to enhance the workflow of adding these cross-references. Some of these are provided by the `hyperref` and `cleveref` packages. To include these packages add the following lines to the bottom of your preamble (order matters, `cleveref` needs to be after `hyperref` and `hyperref` should be one of the last packages loaded):

```
\usepackage{hyperref}
\usepackage[nameinlink]{cleveref}
```

With these packages installed we can now use the commands in Table 7.1.<sup>2</sup>

Table 7.2: Built-in, `hyperref`, and `cleveref` commands and outputs

Command	Output
built-in	
<code>\ref{}</code>	7.2
<code>\pageref{}</code>	96
hyperref	
<code>\autoref{}</code>	Table 7.2
cleveref	
<code>\cref{}</code>	table 7.2
<code>\Cref{}</code>	Table 7.2
<code>\cref*{}</code>	table 7.2
<code>\Cref*{}</code>	Table 7.2
<code>\cpageref{}</code>	page 96
<code>\Cpageref{}</code>	Page 96
<code>\namecref{}</code>	table
<code>\nameCref{}</code>	Table

<sup>2</sup>Note that because the floats are added where they are in the text this causes them to insert large amounts of white space because it only fits on the following page.

Further, the `cleveref` also includes features that allows for the auto sorting and combining of references:

```
\Cref{fig:doubleImage,fig:singleImage,fig:
tripleImage1,fig:quadImage}
```

Noting that there are **NO** spaces between the labels; this will produce: **Figures 4.1, 4.2, A.2 and A.4**. Allowing one to quickly and efficiently keep references up-to-date and consistent in their style. More examples of the use of the `cleveref` cross-referencing is found through the rest of this document.

## 7.4.2 Citations

Citations are a lot easier than dealing with the cross-referencing. There are no additional packages required for citations, the built-in ones are feature-rich enough. Now, while there are no additional packages required to make citations in your document, there are in fact a few programs that should help you manage all of your citations/references. These programs can include Mendeley, JabRef, or Zotero; a comparison of the softwares can be found in **Table 7.3**, and more information of the use of JabRef can be found in **Chapter 9**. Single citations can be included with the `\cite{citationKey}`

Table 7.3: Comparison of Reference Softwares

Software	Developer	Version	Cost	License
JabRef	The JabRef Team	5.11	Free	MIT
Mendeley	Elsevier	2.99.0	Free up to 2 GB	Proprietary
Zotero	CDS	6.0.27	Free up to 300 MB	AGPL

command, the one at the end of this sentence is created with the `\cite{TEST}` command[**TEST**]. Multiple citations can be included in a single cite command by adding commas in between the citation keys. The citation at the end of this sentence

shows how to create more than one citation and how they are grouped together, it is created with the `\cite{testone,cite2,cite3,cite4,cite5}` command[**testone, cite2, cite3, cite4, cite5**]. Finally this sentence shows how a gap in the citations is handled, this is created with the `\cite{testone,cite2,cite3,cite5}` command[**testone, cite2, cite3, cite5**].

# Chapter 8

## Submitting Your Thesis

So you have seemingly gotten to the end of the writing, and you may be already taking the steps to set up your last review with your supervisor, set up your thesis defence, or even submit your thesis to GPS... but now what do you do?

Quick answer is a lot, long answer will be discussed throughout this Chapter.

There are a number of steps that you will want to take to make sure that you are submitting the best version of your work. This includes checking for some of the more obvious and less obvious pitfalls that writing a Thesis in L<sup>A</sup>T<sub>E</sub>X or really any software poses.

# Chapter 9

## JabRef: Managing Bibliographies Efficiently

### 9.1 Introduction

JabRef stands as a powerful tool for researchers and academics engaged in scholarly writing. JabRef offers a robust solution for bibliography management, including a number of features to ensure that you are not only able to organize your references but keep track of progress, and notes on each reference. This chapter aims to provide an review of JabRef, including its myriad features that I find particularly useful, and to guide you through its implementation in writing your thesis.

### 9.2 Key Features of JabRef

JabRef, with its versatile features, emerges as an indispensable tool for bibliography/reference management. Delving deeper into its functionalities reveals a wealth of tools designed to streamline the often cumbersome process of handling references.

#### 9.2.1 BibTeX Compatibility

JabRef's commitment to the BibTeX format is a testament to its roots in the L<sup>A</sup>T<sub>E</sub>X ecosystem. This compatibility ensures a seamless integration between the reference

management process and the L<sup>A</sup>T<sub>E</sub>X document preparation workflow. Users can easily export and import BibTeX files, facilitating collaboration and compatibility across various platforms.

### 9.2.2 Reference Import

The capability to import references directly from online databases and journal websites significantly accelerates the reference collection process. JabRef supports various import formats, allowing users to effortlessly populate their databases with accurate and structured reference information. This feature is particularly valuable for researchers dealing with large/extensive bibliographies.

### 9.2.3 Customizable Entry Types

The flexibility offered by customizable entry types allows users to categorize references based on the nature of the source. Whether it's a book, article, conference proceeding, or any other reference type, JabRef accommodates diverse sources, ensuring a well-organized and easily navigable bibliography.

### 9.2.4 Search and Filter

The ability to efficiently search and filter references is a hallmark of JabRef's usability. Researchers dealing with extensive databases will appreciate the quick and precise retrieval of references based on author names, titles, keywords, or any other criteria. This feature is crucial for maintaining order in a rapidly growing bibliography.

### 9.2.5 Grouping

JabRef's grouping functionality provides a systematic approach to organizing references. Users can create custom groups to categorize references based on themes, projects, or any other criteria. This feature is especially useful for large research

projects where a systematic organization of references is essential for maintaining clarity and coherence.

### 9.2.6 Integration with L<sup>A</sup>T<sub>E</sub>X

The seamless integration of JabRef with L<sup>A</sup>T<sub>E</sub>X editors fortifies the synergy between bibliography management and document preparation. This integration minimizes the manual effort required for citation insertion and ensures consistency between the bibliography and the in-text citations. Users can easily copy citation keys from JabRef and paste them directly into their L<sup>A</sup>T<sub>E</sub>X documents.

## 9.3 Getting Started with JabRef

Now that we’ve outlined the key features of JabRef, let’s embark on a comprehensive guide on how to get started with JabRef. This step-by-step walkthrough will cover everything from installation to creating a new bibliography and populating it with references.

### 9.3.1 Installation

The initial step in utilizing JabRef is to install the software on your system. For all users, regardless of OS, the easiest way to download JabRef is to visit their website: <https://www.jabref.org/>. Once there select ‘Download’ from the navigation bar, and press the “Download JabRef” button. This will take you to the FossHub page where you can select the appropriate version for your OS and download and install it.

### 9.3.2 Creating a New Bibliography

Once JabRef is successfully installed, launch the application. When the program loads you will be faced with a window that looks like the one shown in [Figure 9.1](#). Now that the program is open, to create a new bibliography:

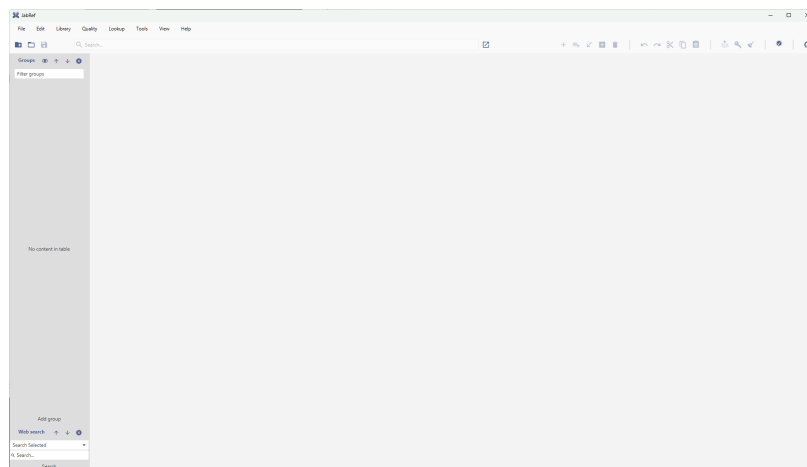


Figure 9.1: JabRef Main Window.

1. Click on ‘File → New Library’.

To save the database:

1. Click on ‘File → Save Library’.
2. Choose an appropriate and location.
3. Click ‘Save’.

Congratulations! You’ve initiated your bibliography using JabRef. Now that we have this created, the next step is to add references to the database.

### 9.3.3 Adding References

JabRef offers multiple avenues for adding references to your database. Some of the methods are generally more useful than others but we will go over a few that you are likely to use:

#### 9.3.3.1 Web Search

JabRef’s integrated web search (see [Figure 9.2](#)) feature simplifies the process of importing references from online sources. This is by-far the easiest way to enter a



reference.

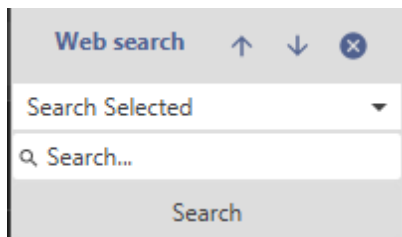


Figure 9.2: JabRef Web Search Tool.

1. Click on ‘Web Search’.
2. Search for the desired reference using the integrated search feature.
3. Select the reference all the references you wish to import, as shown in [Figure 9.3](#).
4. Click ‘Import entries’ to import the selected entries.
5. The references are added to your library.

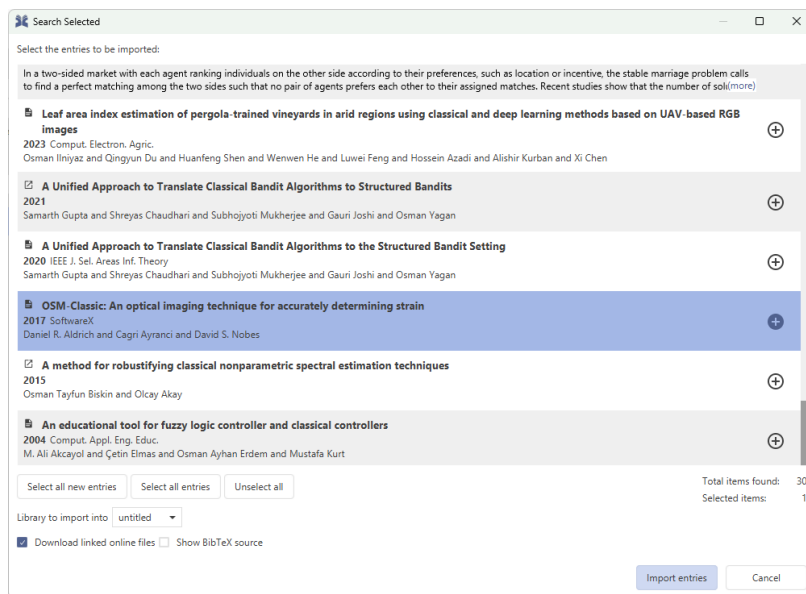


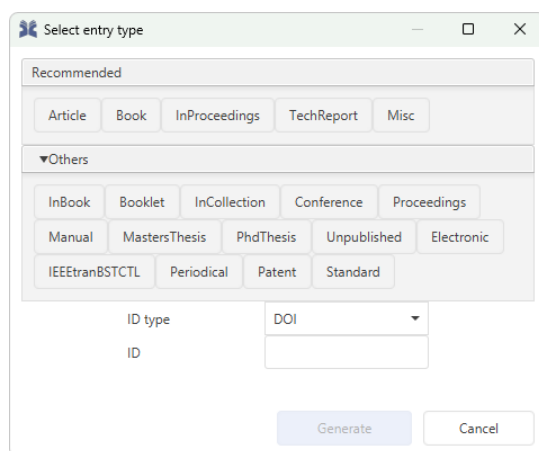
Figure 9.3: Example Web Search Results for ‘OSM-Classic’.

**Note:** The Web Search tool by default uses a general search, however, a specific database can be chosen as well using the drop down arrow next to “Search Selected”.

### 9.3.3.2 Manual Entry

Manually adding a reference can be done in a ‘manual’ and ‘automatic’ way. When adding a new entry you will be faced with the option to select an entry type or to enter an ID (DOI, ArXiv, ISBN, *etc.*). If you enter an ID, the information for the reference will be automatically pulled from the internet. Otherwise to manually enter all the information for a reference:

1. Click Library → ‘New entry’ or use the shortcut ‘Ctrl + N’ and this will show the following window.



2. Choose the entry type (*e.g.*, article, book, inproceedings).
3. Fill in the required fields like author, title, journal, *etc.*

By following these steps, you can efficiently populate your JabRef database with the necessary references.

### 9.3.4 Organizing References

Effectively organizing references is essential for a streamlined bibliography. JabRef's grouping feature allows you to categorize references based on your preferences:

1. On the left panel, select 'Add Group'.
2. Give the group an appropriate name.
3. Optionally you can add a Description, Icon, Colour, *etc.*

To assign a reference to a group:

1. Select the Reference(s) from the centre list.
2. Drag them to the group on the left of the screen.

**Note:** *Groups can even be nested into groups to provide more levels of organization.*

Organizing references into groups enhances accessibility and facilitates a more systematic approach to bibliography management.

## 9.4 Exploring Advanced Features of JabRef

JabRef's capabilities extend beyond the basics covered in the previous sections. In this section, we'll explore some of the advanced features that enhance the efficiency and effectiveness of JabRef as a reference manager.

### 9.4.1 Quality Assurance: Checking and Correcting Entries

Ensuring the accuracy and completeness of references is crucial. JabRef provides tools for quality assurance, allowing users to check and correct entries.

To check for duplicate entries:

1. Click on 'Quality → Find duplicates'.

2. JabRef will identify and display duplicate entries.

To correct entries:

1. Click on ‘Quality → Cleanup entries’.
2. JabRef will provide some useful option to ensure conformity within the different references. This includes renaming Linked PDF’s to match the standard of ‘CitationKey - Title’.

These quality assurance features contribute to maintaining a clean and error-free bibliography.

### 9.4.2 Managing PDFs and File Links

JabRef facilitates the management of associated PDF’s and file links, offering a consolidated approach to reference and document management.

To link a PDF or file:

1. Open the entry editor for a reference.
2. Click on ‘General’ and use the ‘PDF’ or ‘File’ field to link the document.

This integration helps to streamline the retrieval of PDF’s or other associated documents directly from JabRef. Further, this allows JabRef to keep track of the comments and highlights in a single place (see [Figure 9.4](#)). These annotations can be found by selecting the entry, and selecting the ‘File annotations’ tab.

### 9.4.3 Additional Information

JabRef keeps track of a lot of information and can even help you with your research. Some additional information JabRef provides includes:

- Citation information

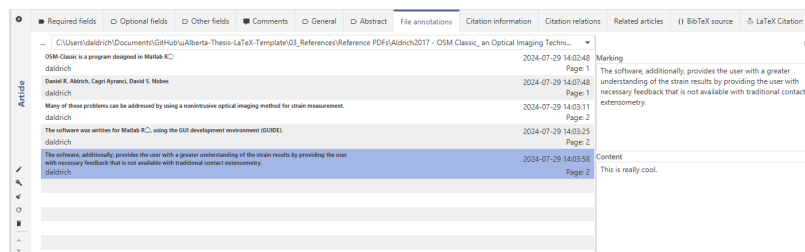


Figure 9.4: Showcase of the file annotations in JabRef.

- Citation relationships (what the reference cites and who has cited the reference).  
This further lets you open the links to the reference's source, or even add these references directly to your library.
- If one right clicks an entry you are provided the following options:
  - Rank - rank the reference with one to five stars.
  - Toggle Relevance - add a marker to show this is a relevant source.
  - Priority - rank items as low, medium, or high priority.
  - Read Status - set the status to read or skimmed.

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# Appendix A

## Additional Example Figures

Each of the following pages will provide an example of a different figure configuration. In addition to the examples the code that generates the figure will be provided and explanations of what the different parts of the code do will be included. From all of the included information in this Appendix it should be possible to even develop your own figures that potentially suit your needs best.

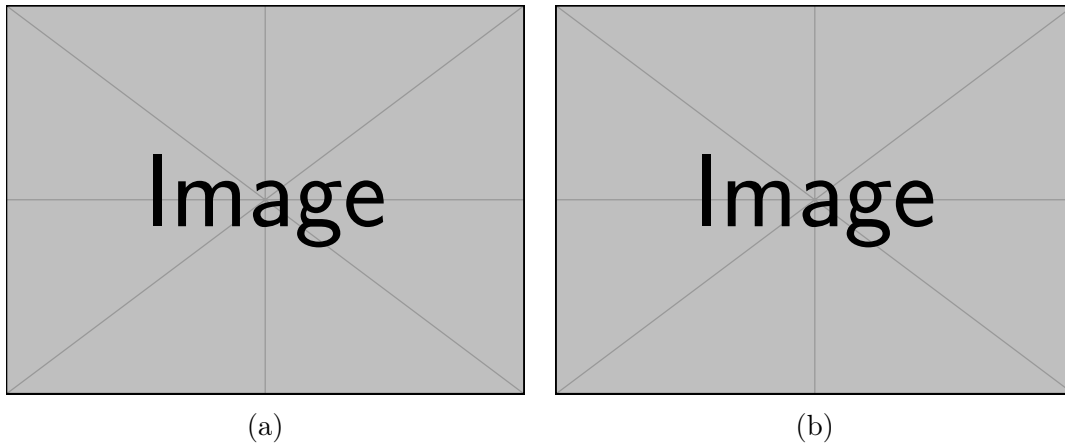


Figure A.1: This is an example of a double image figure.

```
\begin{figure}[H]
\centering
\begin{subfigure}{0.45\linewidth}
\includegraphics[width=\linewidth]{example-image}
\caption{} % Leave blank for just letter
\label{fig:doubleImage2:a}
\end{subfigure}
~
\begin{subfigure}{0.45\linewidth}
\includegraphics[width=\linewidth]{example-image}
\caption{} % Leave blank for just letter
\label{fig:doubleImage2:b}
\end{subfigure}
\caption{This is an example of a double image figure.}
\label{fig:doubleImage2}
\end{figure}
```

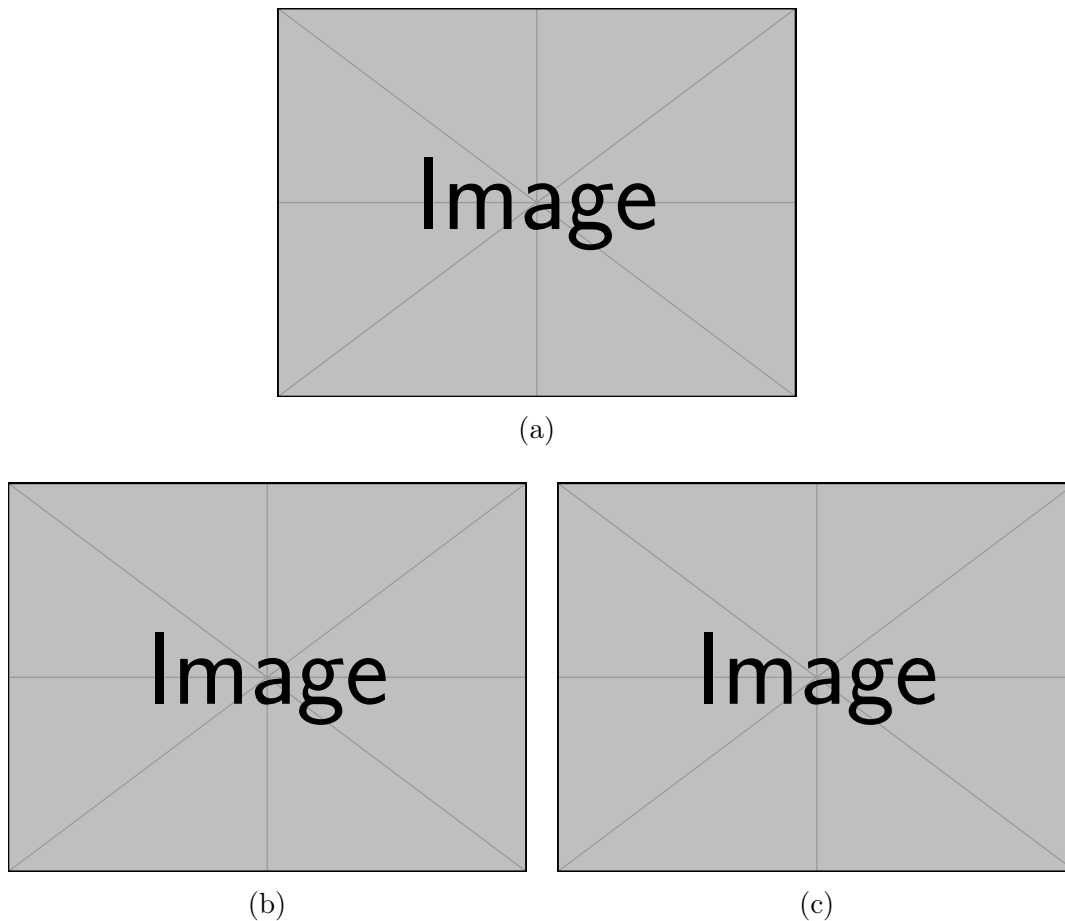
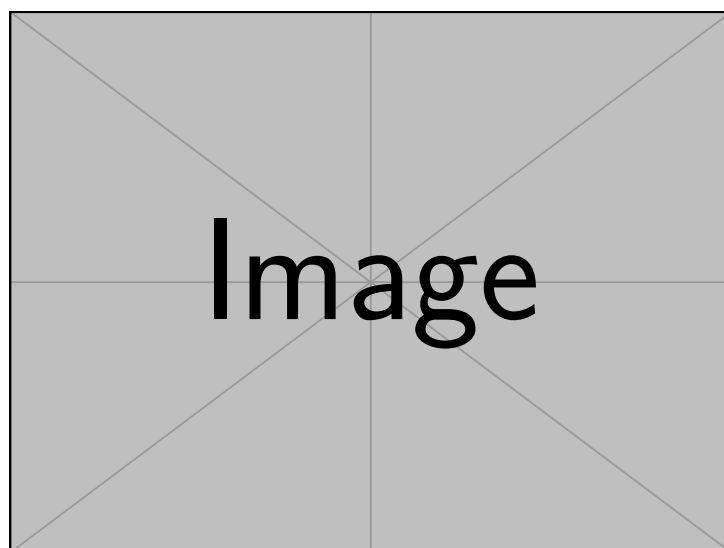


Figure A.2: This is an example of a triple image figure.

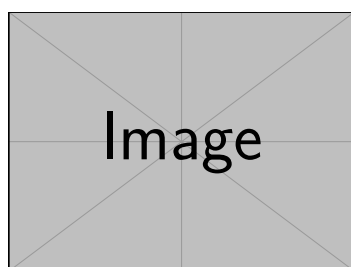
```

\begin{figure}[H]
\centering
\hspace*{\fill}% Adds space to left of top image (prevents two images from going to top)
\begin{subfigure}{0.45\linewidth}
\includegraphics[width=\linewidth]{example-image}
\caption{} % Leave blank for just letter
\label{fig:tripleImage1:a}
\end{subfigure}
\hspace*{\fill}% Adds space to right of top image (prevents two images from going to top)
\par\vspace{1em}% Adds space between upper and lower images
\begin{subfigure}{0.45\linewidth}
\includegraphics[width=\linewidth]{example-image}
\caption{} % Leave blank for just letter
\label{fig:tripleImage1:b}
\end{subfigure}
~ % Adds space between the two lower figures
\begin{subfigure}{0.45\linewidth}
\includegraphics[width=\linewidth]{example-image}
\caption{} % Leave blank for just letter
\label{fig:tripleImage1:c}
\end{subfigure}
\caption{This is an example of a triple image figure.}
\label{fig:tripleImage1}
\end{figure}

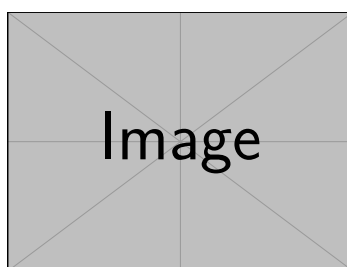
```



(a)



(b)



(c)

Figure A.3: This is a second example of a triple image figure.

```

\begin{figure}[H]
  \centering
  \hspace*{\fill}% Adds space to left of top image (prevents two images from going to top)
  \begin{subfigure}{0.90\linewidth+1em}% 0.9 = 0.45 + 0.45, and 1em is the width of ~
    \includegraphics[width=\linewidth]{example-image}
    \caption{} % Leave blank for just letter
    \label{fig:tripleImage2:a}
  \end{subfigure}
  \hspace*{\fill}% Adds space to right of top image (prevents two images from going to top)
  \par\vspace{1em}% Adds space between upper and lower images
  \begin{subfigure}{0.45\linewidth}
    \includegraphics[width=\linewidth]{example-image}
    \caption{} % Leave blank for just letter
    \label{fig:tripleImage2:b}
  \end{subfigure}
  ~ % Adds space between the two lower figures
  \begin{subfigure}{0.45\linewidth}
    \includegraphics[width=\linewidth]{example-image}
    \caption{} % Leave blank for just letter
    \label{fig:tripleImage2:c}
  \end{subfigure}
  \caption{This is a second example of a triple image figure.}
  \label{fig:tripleImage2}
\end{figure}

```

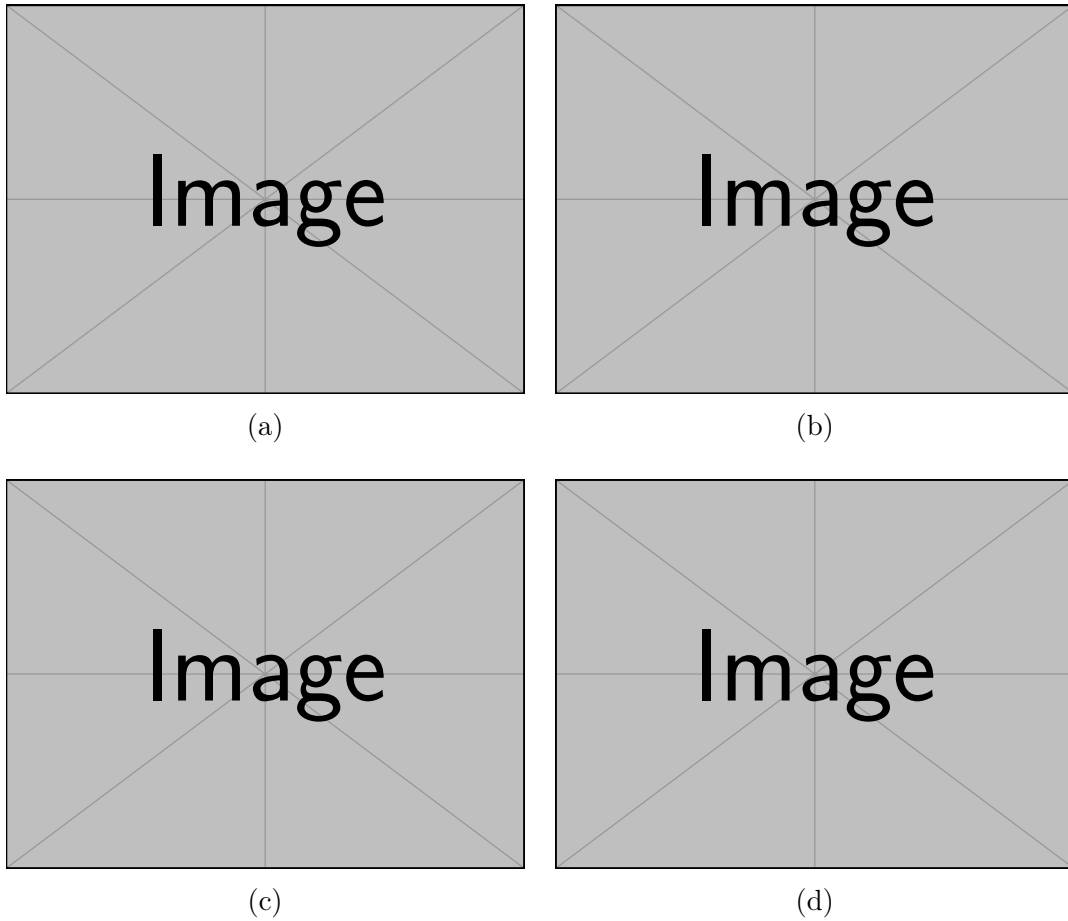


Figure A.4: This is an example of a quad image figure.

```

\begin{figure}[H]
  \centering
  \begin{subfigure}{0.45\linewidth}
    \includegraphics[width=\linewidth]{example-image}
    \caption{} % Leave blank for just letter
    \label{fig:quadImage:a}
  \end{subfigure}
  ~ % Adds space between the two top figures
  \begin{subfigure}{0.45\linewidth}
    \includegraphics[width=\linewidth]{example-image}
    \caption{} % Leave blank for just letter
    \label{fig:quadImage:b}
  \end{subfigure}
  \par\vspace{1em} % Adds space between upper and lower images
  \begin{subfigure}{0.45\linewidth}
    \includegraphics[width=\linewidth]{example-image}
    \caption{} % Leave blank for just letter
    \label{fig:quadImage:c}
  \end{subfigure}
  ~ % Adds space between the two lower figures
  \begin{subfigure}{0.45\linewidth}
    \includegraphics[width=\linewidth]{example-image}
    \caption{} % Leave blank for just letter
    \label{fig:quadImage:d}
  \end{subfigure}
  \caption{This is an example of a quad image figure.}
  \label{fig:quadImage}
\end{figure}

```

# Appendix B

## Additional Example Tables

### B.1 Section 1

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## B.2 Section 2

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# Appendix C

## Including Code Listings

This appendix provides guidelines for including code listings in your thesis. Code listings are often used to demonstrate algorithms, data processing scripts, or other relevant programming content. Proper formatting ensures that code is both readable and aesthetically pleasing.

### C.1 Using the `listings` Package

The `listings` package is a powerful tool for displaying code in LaTeX. It supports syntax highlighting for a wide variety of programming languages and offers many customization options.

#### C.1.1 Basic Usage

To include a simple code listing, you can use the following command:

```
\begin{lstlisting}[language=Python]
    # Your code here
    print("Hello, world!")
\end{lstlisting}
```

The `language` option specifies the programming language, which enables syntax highlighting. Replace `Python` with the appropriate language for your code.

## C.1.2 Customizing Listings

The `listings` package allows for extensive customization. You can adjust the appearance of your code by setting options such as `frame`, `backgroundcolor`, `keywordstyle`, and more.

Here is an example of how to customize your code listing:

```
\lstset{
    language=Python,
    frame=single,
    backgroundcolor=\color{gray!10},
    keywordstyle=\color{blue}\bfseries,
    commentstyle=\color{green},
    stringstyle=\color{red},
    basicstyle=\ttfamily,
    breaklines=true
}
```

This configuration adds a single-line frame around the code, sets a light gray background, and defines styles for keywords, comments, and strings.

Listing C.1: This is a caption for the inserted code

```
function [outputs] = functionName(inputs)
%{
    This is a Comment Block
    That
    can
    span
    multiple
    lines.
%}

% This is a regular comment
a = 1 + 2 * sin(angle);
b = 'This is a String';
```

Listing C.2: This is a caption for the inserted code

```
#include <iostream>
using namespace std;
/* This function adds two integer values
 * and returns the result
 */

int sum(int num1, int num2){
    int num3 = num1 + num2; return num3;
}

void main() {
```

```
//Calling the function
cout << 'The sum is:' << sum(1,99);
}
```

## C.2 Advanced Features

### C.2.1 Including External Files

The `listings` package allows you to include code from external files. This is particularly useful if you have long code files that you want to reference directly.

```
\lstinputlisting[language=Python]{path/to/your/code.py}
```

Replace `path/to/your/code.py` with the actual path to your file. Or you can use the `\addmedia{./99_Inclusions/}` and `\addcode{Code/}` commands to define the location for code files, then you can just use `\insertcode{filename.ext}` command instead of the full path. You can customize the display in the same way as inline listings.

### C.2.2 Handling Special Characters

If your code contains special characters (*e.g.*, #, %, \$), you may need to escape them or use the `literate` option to ensure proper display.

```
\lstset{
  literate={~} {$\sim$}{1}
}
```

This command, for example, replaces the tilde symbol with the appropriate LaTeX command.

## C.3 Line Breaks in Long Code Lines

To automatically break long lines of code, use the `breaklines=true` option as shown in the earlier examples. This prevents code from running off the page and maintains readability.

## C.4 Conclusion

Including well-formatted code listings in your thesis can enhance the clarity of your work and demonstrate your technical skills. By following the guidelines in this appendix, you can ensure that your code is presented professionally.

# Appendix D

## Including PDFs

### WARNING

While it is possible to have horizontal pages with the page numbers centered on the bottom long edge, I *DO NOT* recommend it. This is because, while it looks okay in a digital format, this is not suitable for printing... this would print page numbers on the side of the page rather than consistently on the bottom or in the heading.

The package that is used to include PDF's is `pdfpages`. This provides the main command `\includepdf` that can be used to include a PDF. Further, this package provides another command, `\includepdfset`, that can be used in the start of the document to pre-set some of the default values. This class file detects the presence of this package and invokes this command as follows:

```
\includepdfset{pages=-,scale=0.85,pagecommand=\  
  thispagestyle{STYLE}}
```

Where `pages=-` defaults to including all pages from the PDF and `scale=0.85` scales the inserted PDF to 85% of its original size, so that the documents fit within the page's margins, headers, and footers.

Further, the option `pagecommand=\cmd {thispagestyle}\mopt {STYLE}` sets the pagestyle for the PDF pages, where, `STYLE` is replaced with the current pagestyle<sup>1</sup>

---

<sup>1</sup>To change the pagestyle of this one can add or remove the class option `fancyheaders`.

that is in use.

## D.1 How to Insert a Portrait PDF

To insert a portrait-oriented PDF into your LaTeX document, you can use the `pdfpages` package, which provides a convenient way to include external PDF files. The following code snippet demonstrates how to include a portrait PDF with the specified options:

```
\includepdf{./99_Inclusions/PDFs/examplePDF}
```

This is an Example PDF that is Portrait



This is the second page

## D.2 How to Insert a Landscape PDF

Inserting a landscape-oriented PDF is similarly straightforward using the `pdfpages` package. The code snippet below demonstrates how to include a landscape PDF:

```
\includepdf[landscape=true]{./99_Inclusions/PDFs/  
  landscapePDF}
```

Here, `landscape` sets the orientation to landscape. This configuration ensures that your landscape PDF is correctly oriented and properly sized within your document.

This is an Example PDF that is Landscape

This is the second page

# Appendix E

## Math Lettering

Table E.1: Math Mode Greek Letters

Command	Output	Command	Output	Command	Output
<code>\alpha</code>	$\alpha$	<code>\beta</code>	$\beta$	<code>\gamma</code>	$\gamma$
<code>\delta</code>	$\delta$	<code>\epsilon</code>	$\epsilon$	<code>\zeta</code>	$\zeta$
<code>\eta</code>	$\eta$	<code>\theta</code>	$\theta$	<code>\iota</code>	$\iota$
<code>\kappa</code>	$\kappa$	<code>\lambda</code>	$\lambda$	<code>\mu</code>	$\mu$
<code>\nu</code>	$\nu$	<code>\xi</code>	$\xi$	<code>o</code>	$o$
<code>\pi</code>	$\pi$	<code>\rho</code>	$\rho$	<code>\sigma</code>	$\sigma$
<code>\tau</code>	$\tau$	<code>\upsilon</code>	$\upsilon$	<code>\phi</code>	$\phi$
<code>\chi</code>	$\chi$	<code>\psi</code>	$\psi$	<code>\omega</code>	$\omega$
A	$A$	B	$B$	<code>\Gamma</code>	$\Gamma$
<code>\Delta</code>	$\Delta$	E	$E$	Z	$Z$
H	$H$	<code>\Theta</code>	$\Theta$	I	$I$
K	$K$	<code>\Lambda</code>	$\Lambda$	M	$M$
N	$N$	<code>\Xi</code>	$\Xi$	O	$O$
<code>\Pi</code>	$\Pi$	P	$P$	<code>\Sigma</code>	$\Sigma$
T	$T$	<code>\Upsilon</code>	$\Upsilon$	<code>\Phi</code>	$\Phi$
X	$X$	<code>\Psi</code>	$\Psi$	<code>\Omega</code>	$\Omega$

Table E.2: Blackboard Bold Letters

Command	Output	Command	Output	Command	Output
<code>\mathbb\mopt{A}</code>	A	<code>\mathbb\mopt{B}</code>	B	<code>\mathbb\mopt{C}</code>	C
<code>\mathbb\mopt{D}</code>	D	<code>\mathbb\mopt{E}</code>	E	<code>\mathbb\mopt{F}</code>	F
<code>\mathbb\mopt{G}</code>	G	<code>\mathbb\mopt{H}</code>	H	<code>\mathbb\mopt{I}</code>	I
<code>\mathbb\mopt{J}</code>	J	<code>\mathbb\mopt{K}</code>	K	<code>\mathbb\mopt{L}</code>	L
<code>\mathbb\mopt{M}</code>	M	<code>\mathbb\mopt{N}</code>	N	<code>\mathbb\mopt{O}</code>	O
<code>\mathbb\mopt{P}</code>	P	<code>\mathbb\mopt{Q}</code>	Q	<code>\mathbb\mopt{R}</code>	R
<code>\mathbb\mopt{S}</code>	S	<code>\mathbb\mopt{T}</code>	T	<code>\mathbb\mopt{U}</code>	U
<code>\mathbb\mopt{V}</code>	V	<code>\mathbb\mopt{W}</code>	W	<code>\mathbb\mopt{X}</code>	X
<code>\mathbb\mopt{Y}</code>	Y	<code>\mathbb\mopt{Z}</code>	Z		

Table E.3: Calligraphic Letters

Command	Output	Command	Output	Command	Output
<code>\mathcal\mopt{A}</code>	$\mathcal{A}$	<code>\mathcal\mopt{B}</code>	$\mathcal{B}$	<code>\mathcal\mopt{C}</code>	$\mathcal{C}$
<code>\mathcal\mopt{D}</code>	$\mathcal{D}$	<code>\mathcal\mopt{E}</code>	$\mathcal{E}$	<code>\mathcal\mopt{F}</code>	$\mathcal{F}$
<code>\mathcal\mopt{G}</code>	$\mathcal{G}$	<code>\mathcal\mopt{H}</code>	$\mathcal{H}$	<code>\mathcal\mopt{I}</code>	$\mathcal{I}$
<code>\mathcal\mopt{J}</code>	$\mathcal{J}$	<code>\mathcal\mopt{K}</code>	$\mathcal{K}$	<code>\mathcal\mopt{L}</code>	$\mathcal{L}$
<code>\mathcal\mopt{M}</code>	$\mathcal{M}$	<code>\mathcal\mopt{N}</code>	$\mathcal{N}$	<code>\mathcal\mopt{O}</code>	$\mathcal{O}$
<code>\mathcal\mopt{P}</code>	$\mathcal{P}$	<code>\mathcal\mopt{Q}</code>	$\mathcal{Q}$	<code>\mathcal\mopt{R}</code>	$\mathcal{R}$
<code>\mathcal\mopt{S}</code>	$\mathcal{S}$	<code>\mathcal\mopt{T}</code>	$\mathcal{T}$	<code>\mathcal\mopt{U}</code>	$\mathcal{U}$
<code>\mathcal\mopt{V}</code>	$\mathcal{V}$	<code>\mathcal\mopt{W}</code>	$\mathcal{W}$	<code>\mathcal\mopt{X}</code>	$\mathcal{X}$
<code>\mathcal\mopt{Y}</code>	$\mathcal{Y}$	<code>\mathcal\mopt{Z}</code>	$\mathcal{Z}$		



Table E.4: Fraktur Letters

Command	Output	Command	Output	Command	Output
<code>\mathfrak\mopt{a}</code>	$\mathfrak{a}$	<code>\mathfrak\mopt{b}</code>	$\mathfrak{b}$	<code>\mathfrak\mopt{c}</code>	$\mathfrak{c}$
<code>\mathfrak\mopt{d}</code>	$\mathfrak{d}$	<code>\mathfrak\mopt{e}</code>	$\mathfrak{e}$	<code>\mathfrak\mopt{f}</code>	$\mathfrak{f}$
<code>\mathfrak\mopt{g}</code>	$\mathfrak{g}$	<code>\mathfrak\mopt{h}</code>	$\mathfrak{h}$	<code>\mathfrak\mopt{i}</code>	$\mathfrak{i}$
<code>\mathfrak\mopt{j}</code>	$\mathfrak{j}$	<code>\mathfrak\mopt{k}</code>	$\mathfrak{k}$	<code>\mathfrak\mopt{l}</code>	$\mathfrak{l}$
<code>\mathfrak\mopt{m}</code>	$\mathfrak{m}$	<code>\mathfrak\mopt{n}</code>	$\mathfrak{n}$	<code>\mathfrak\mopt{o}</code>	$\mathfrak{o}$
<code>\mathfrak\mopt{p}</code>	$\mathfrak{p}$	<code>\mathfrak\mopt{q}</code>	$\mathfrak{q}$	<code>\mathfrak\mopt{r}</code>	$\mathfrak{r}$
<code>\mathfrak\mopt{s}</code>	$\mathfrak{s}$	<code>\mathfrak\mopt{t}</code>	$\mathfrak{t}$	<code>\mathfrak\mopt{u}</code>	$\mathfrak{u}$
<code>\mathfrak\mopt{v}</code>	$\mathfrak{v}$	<code>\mathfrak\mopt{w}</code>	$\mathfrak{w}$	<code>\mathfrak\mopt{x}</code>	$\mathfrak{x}$
<code>\mathfrak\mopt{y}</code>	$\mathfrak{y}$	<code>\mathfrak\mopt{z}</code>	$\mathfrak{z}$		
<code>\mathfrak\mopt{A}</code>	$\mathfrak{A}$	<code>\mathfrak\mopt{B}</code>	$\mathfrak{B}$	<code>\mathfrak\mopt{C}</code>	$\mathfrak{C}$
<code>\mathfrak\mopt{D}</code>	$\mathfrak{D}$	<code>\mathfrak\mopt{E}</code>	$\mathfrak{E}$	<code>\mathfrak\mopt{F}</code>	$\mathfrak{F}$
<code>\mathfrak\mopt{G}</code>	$\mathfrak{G}$	<code>\mathfrak\mopt{H}</code>	$\mathfrak{H}$	<code>\mathfrak\mopt{I}</code>	$\mathfrak{I}$
<code>\mathfrak\mopt{J}</code>	$\mathfrak{J}$	<code>\mathfrak\mopt{K}</code>	$\mathfrak{K}$	<code>\mathfrak\mopt{L}</code>	$\mathfrak{L}$
<code>\mathfrak\mopt{M}</code>	$\mathfrak{M}$	<code>\mathfrak\mopt{N}</code>	$\mathfrak{N}$	<code>\mathfrak\mopt{O}</code>	$\mathfrak{O}$
<code>\mathfrak\mopt{P}</code>	$\mathfrak{P}$	<code>\mathfrak\mopt{Q}</code>	$\mathfrak{Q}$	<code>\mathfrak\mopt{R}</code>	$\mathfrak{R}$
<code>\mathfrak\mopt{S}</code>	$\mathfrak{S}$	<code>\mathfrak\mopt{T}</code>	$\mathfrak{T}$	<code>\mathfrak\mopt{U}</code>	$\mathfrak{U}$
<code>\mathfrak\mopt{V}</code>	$\mathfrak{V}$	<code>\mathfrak\mopt{W}</code>	$\mathfrak{W}$	<code>\mathfrak\mopt{X}</code>	$\mathfrak{X}$
<code>\mathfrak\mopt{Y}</code>	$\mathfrak{Y}$	<code>\mathfrak\mopt{Z}</code>	$\mathfrak{Z}$		