### Q1 What is the true meaning of linearity?

Since linearity enables the computation to easily summing signals, this further makes lots of signal processing and mathematical techniques possible.

### Q2 Derive transfer function from the I/O equation? (FIR)

$$y(n) = b_0 x(n) + b_1 x(n-1) + \dots + b_{L-1} x(n-L+1) \ \therefore H(z) = rac{Y(z)}{X(z)} = b_0 + b_1 z^{-1} + \dots + b_{L-1} z^{-(L-1)} = \sum_{l=0}^{L-1} b_l z^{-l}$$

### Q3 Derive transfer function from the I/O equation? (IIR)

$$y(n) = \sum_{l=0}^{L-1} b_l x(n-l) - \sum_{m=1}^{M} a_m y(n-m) \ dots H(z) = rac{Y(z)}{X(z)} = rac{\sum_{l=0}^{L-1} b_l z^{-l}}{1 + \sum_{m=1}^{M} a_m z^{-m}} = rac{\sum_{l=0}^{L-1} b_l z^{-l}}{\sum_{m=0}^{M} a_m z^{-m}}$$

## Q4 Derive transfer function of MA filter from the I/O equation?

$$egin{aligned} y(n) &= rac{1}{L} igg[ x(n) + x(n-1) + \dots + x(n-L+1) igg] \ &\therefore H(z) &= rac{Y(z)}{X(z)} = rac{1}{L} \sum_{n=0}^{L-1} z^{-n} = rac{1}{L} rac{1-z^{-L}}{1-z^{-1}} = rac{1-z^{-L}}{L imes (1-z^{-1})} \end{aligned}$$

## Q5 Derive recursive I/O equation from the transfer function? (MA)

$$egin{split} y(n) &= y(n-1) + rac{1}{L} \left[ x(n) - x(n-L) 
ight] \ dots \cdot H(z) &= rac{Y(z)}{X(z)} = rac{1-z^{-L}}{L imes (1-z^{-1})} \end{split}$$

https://forums.ni.com/t5/LabVIEW/Transfer-function-of-recursive-moving-average-filter/td-p/3172606

### Q6 Is an MA filter stable?

Since there are always L-1 poles at the origin, one pole at z=1, L zeros spaced on the unit circle, there is no pole located outside the unit circle but one lies on the unit circle. Therefore MA filter is marginally stable.

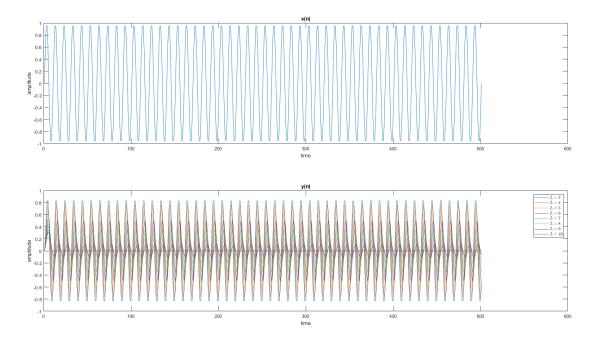
### Q7 Is an MA filter FIR or IIR?

MA filter is FIR filter.

# Q8 When the input to the MA filter is a sinewave, explain it's output waveform in both time and frequency domain.

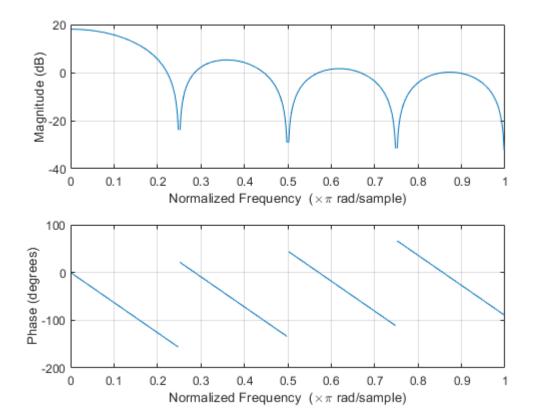
In time domain, the output of MA filter decreases its amplitude as the  ${\it L}$  of MA filter increases.

Code in code1.m



In frequency domain of L=8 MA filter, magnitude have bumps which creates lowest points on  $n\times 0.25\pi$ . It is because those lowest points are where zeros are located on unit

circle.



Q9 We generate 100 samples of sinewave with A=1, f=1000Hz, and sampling rate of 10kHz.

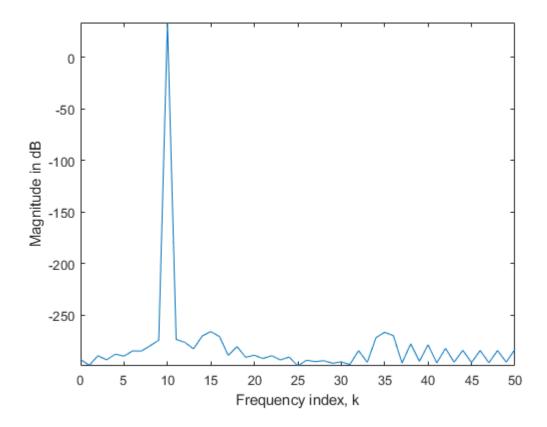
(a) What is the digital frequency  $\omega$  and F?

$$egin{align} \Delta_{\omega} &= rac{2\pi}{N} = rac{2\pi}{100} pprox 0.0628 \ \Delta_{f} &= rac{f_{s}}{N} = rac{1e4}{100} = 100 \ \end{dcases}$$

(b) If we compute DFT with N=100, what value of k corresponding to 1000Hz?

$$k = 0.1$$

(c) Sketch the amplitude spectrum |X(k)|.



### Q10

### (a) Why we see a line at k = 10?

Since the sampling frequency is 10000, and N is 100, which means  $k=rac{f}{f_s} imes N=rac{1000}{10000} imes 100=10$ 

### (b) If we see a line spectrum at k=15, what is the sinewave frequency?

Since k=1.5 imes k',  $f_s=1.5 imes f_s'=15000$ .

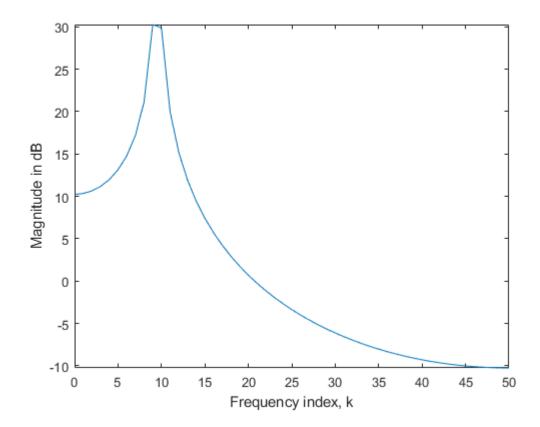
### (c) If f is 950Hz, sketch the amplitude spectrum, what happened?

When f = 950Hz, the center frequency started shifting left.

The reason behind this wider range of frequency is because we use 100 samples to compute FFT, which results in  $\frac{f_s}{samples} = \frac{10000}{100} = 100Hz$  frequency resolution. This results in this wider frequency range. If we use  $f_s = 850$  would results in roughly the s

results in this wider frequency range. If we use  $f_s=850$  would results in roughly the same plot.

#### Code in code3.m



CYEE10828241陳大荃