

Multi-core RISC Processor Design and Implementation (Rev. 2.02)

ELEC5881M - Interim Report

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Abstract

This interim report details the 4-month progress on a project to design, implement, and verify, a multi-core FPGA RISC processor. The project has been split into two stages: firstly to build a functional single-core RISC processor, and then secondly to add multiprocessor principles and functionality to it.

Current multiprocessor and network-on-chip communication methods have been discussed and how they could be included in this multi-core RISC design. To-date, a 16-bit instruction set architecture has been designed featuring common load/store instructions, comparison, and bitwise operations. A single-core processor has been implemented in Verilog and verified using simulations/test benches running various simple software programs.

Future tasks have been planned and will focus on the second stage of the project. Work will start on designing a loosely coupled multiprocessor communication interface and bringing them to the single-core processor.

Revision History

Date	Version	Changes
10/04/2019	2.02	Update future stages.
05/04/2019	2.01	Fix processor RTL diagram.
04/04/2019	2.00	Initial processor RTL diagram.
01/04/2019	1.00	Initial section outline.

Document revisions.

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Chapter 1

Introduction

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This project will detail the design, implementation, and verification, of a new multi-core RISC processor aimed at FPGA devices. This project was chosen due to my interest in processor design, in which I have only previously designed single-core RISC processors and wish to extend this knowledge to gain a basic understanding of multi-core communication, design considerations, and the limitations of parallelism first hand.

I will use this opportunity to further develop my knowledge of FPGA and processor design by implementing, designing, and verifying, a multi-core RISC processor from scratch, including the design of a communication interface between multiple cores.

1.1 Why Multi-core?

Moore's Law states that the number of transistors in a chip will double every 2 years []. CPU designers would utilize the additional transistors to add more pipeline stages in the processor to reduce the propagation delay [] which would allow for higher clock frequencies.

The size of transistors have been decreasing [] and today can be manufactured in sub-10 nanometer range. However, the extremely small transistor size increases electrical leakage and other negative effects resulting in unreliability and potential damage to the transistor []. The high transistor count produces large amounts of heat and requires increasing power to supply the chip. These trade-offs are currently managed by reducing the input voltage, utilising complex cooling techniques, and reducing clock frequency. These factors limit the performance of the chip significantly. These are contributing factors to Moore's Law *slowing* down. The capacity limit of the current-generation planar transistors is approaching and so in order for performance increases to continue, other approaches such as alternate transistor technologies like Multigate transistors [1], software and hardware optimisations, and multi-processor architectures are employed.

This report will focus on the latter: to produce a small multi-core processor that can utilise software-based parallelism to gain performance benefits, compared to a larger single-core design.

1.2 Why RISC?

RISC architectures feature simpler and fewer instructions compared to CISC, which emphasises instructions that perform larger tasks. A single CISC instruction might be performed with multiple RISC instructions. Because of the fewer and simpler instructions, RISC machines rely heavily on software optimisations for performance. RISC instruction sets are based on load/store architectures, where most instructions are either register-to-register or memory reading and writing [?]. This constraint greatly reduces complexity.

RISC architectures are easier to design implement, especially for beginners, due to their simpler instructions that share the same pipeline, compared to CISC where there may be different pipeline for each instruction, which would greatly consume FPGA resources.

1.3 Why FPGA?

Field programmable gate arrays (FPGA) are a great choice for prototyping digital logic designs due to their programmable nature and quick development times.

My previous experience with FPGAs in previous projects will reduce risk and learning times and allow for more time to be spent on adding and extending features (discusses further in section 3.1).

FPGAs, however, may not be suitable for prototyping all register-transistor logic (RTL) projects. Larger RTL projects, such as large commercial processors, may greatly exceed the logic cell resources available in today's high-end FPGA devices and may only be prototyped through silicon fabrication, which can be expensive. This resource limitation will not be problem as the project aims to produce a small and minimal design specifically for learning about multi-core architectures.

Chapter 2

Background

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2.1 Amdahl's Law and Parallelism

In many applications, not restricted to software, there may exist many opportunities for processes or algorithms to be performed in parallel. These algorithms can be split into two parts: a serial part that cannot be parallelised, and a part that can be parallelised. Amdahl's Law defines a formula for calculating the maximum *speedup* of a process with potential parallelism opportunities when ran in parallel with n many processors. Speedup is a term used to describe the potential performance improvements of an algorithm using an enhanced resource (in this case, adding parallel processors) compared to the original algorithm. Amdahl's Law is defined below, where the potential speedup S_p is dependant on the portion of program that can be parallelised p and the number of processing cores n :

$$S_p = \frac{1}{(1 - p) + \frac{p}{n}} \quad (2.1)$$

This formula will be used throughout the project to gauge the performance of the multi-core design running various software algorithms.

2.2 Loosely and Tightly Coupled Processors

Multiprocessor systems can be generalised into two architectures: loosely and tightly coupled, and each architecture has advantages and disadvantages. In loosely coupled systems, each processing node is self-contained – each node has its own dedicated memory and IO modules. Communication between nodes is performed over a *Message Transfer System (MTS)* [?] in a master-slave control architecture.

Scalability in loosely coupled systems is generally easier to implement as each node can simply be appended to the shared MTS interface without large modifications to the rest of the system. Scalability is an important concern in this project as I wish to test the developed solution with a range of processing nodes.

As loosely coupled system's nodes feature their own memory and IO modules, they generally perform better in cases where interaction between nodes is not prominent – each node can store a separate part of the software program in its memory module allowing simultaneous executing of the program.

In scenarios where inter-node communication is prominent however, access to the MTS interface must be scheduled to avoid access conflicts which introduces delays and idle times in the software programs execution, resulting in lower throughput. Figure 2.1 shows a general layout of a loosely coupled multiprocessor system.

Tightly coupled systems feature processing nodes that do not have their own dedicated memory or IO modules – each node is directly connected to a shared memory module using a dedicated port. In scenarios where inter-node communication is prominent, tightly coupled systems are generally better suited as nodes are directly connected to a shared memory and do not need to wait to use a shared bus.

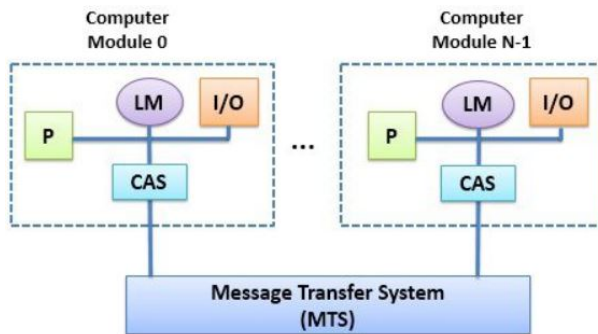


Figure 2.1: A loosely coupled multiprocessor system. Each node features its own memory and IO modules and uses a Message Transfer System to perform inter-node communication. Image source: [?].

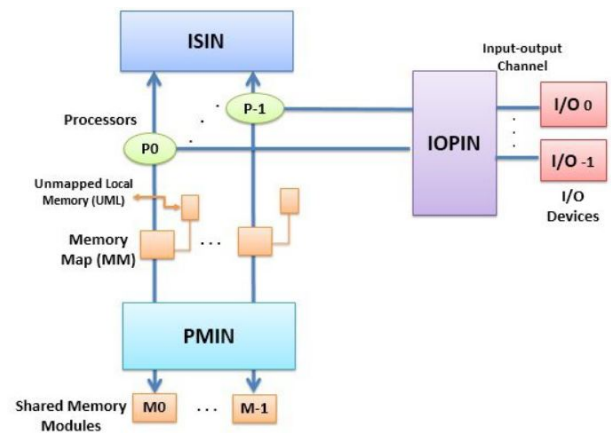


Figure 2.2: A tightly coupled multiprocessor system. Nodes are directly connected to memory and IO modules. Image source: [?].

This project will utilise a loosely coupled architecture due to its easier scalability implementation and my previous experience with the design of single-core processors. Although it will require a scheduler to access the MTS, the experience and knowledge gained from this task will be greatly beneficial for future projects.

2.3 Network-on-chip Architectures

Network-on-chip (NoC) architectures implement on-chip communication mechanisms that are based on network communication principles, such as routing, switching, and massive scalability [?]. NoC's can generally support hundreds to millions of processing cores. Figure 2.3 shows an example 16-core network-on-chip architecture. NoC's can scale to very large sizes while not sacrificing performance because each processor core is able to drive the network rather than needing to wait for a shared bus to become free before doing so.

The greater the number of cores in a network-on-chip design, the greater quality of service (QoS) problems arise. As such, network-on-chip architectures suffer the same problems as networks, such as fairness and throughput [?].

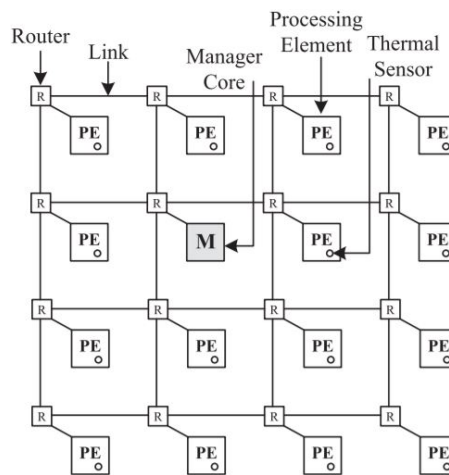


Figure 2.3: A multiprocessor network-on-chip architecture with 16 processing nodes. Nodes are connected in a grid formation with routers and links. Image source: [?].

Chapter 3

Project Overview

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This chapter discusses the the project's requirements, goals, and structure.

3.1 Project Deliverables

The project's deliverables are split into two sections: core deliverables (CD) – each deliverable must be satisfied for the project to be a minimum viable product (MVP), and extended deliverables (ED) – deliverables that are not required for a MVP – features that only improve upon an existing feature.

3.1.1 Core Deliverables (CD)

The project's core deliverables are described below.

CD1 Design a compact 16-bit RISC instruction set architecture.

The instruction set will be the primary interface to control the processor from software. An instruction set will be required to implement the custom multi-core communication interface.

It was decided to design a new instruction set rather than to extend an existing architecture as this will increase my knowledge of the constraints to consider when designing instruction sets and processors.

CD2 Design and implement a Verilog RISC core that implements the ISA in [CD1](#).

The Verilog RISC core will be able to run software program written for the instruction set architecture.

CD3 Design and implement an on-chip interconnect for multi-core processing (2 to 32 cores) using the RISC core from CD2.

The interconnect will be a chief requirement to enable multi-core communication. The interconnect should support up to 32 cores, however FPGA implementation constraints may limit this due to limited resources.

The interconnect will control communication between the cores to enable software parallelism.

CD4 Analyse performance of serial and parallel software algorithms, such as parallel DFT, on the processor.

To evaluate the effectiveness of the developed solution, a serial and parallel implementation of a simple computing algorithm (parallel reduction, sorting) will be ran on the processor and it's performance analysed. Effectiveness will be rated on total algorithm run-time and the speed-up gained by adding more cores.

CD5 Allow the RISC core to be easily compiled to multiple FPGA vendors (Xilinx, Altera).

The developed solution should be generic and portable to allow it to be used across a wide-range of FPGA vendors and devices.

Verilog is a generic implementation-independent hardware-description language and so designing implementation specific modules is recommended.

A key consideration for this requirement is to consider the varying hard IP provided by the FPGA vendors (such as BRAM, ethernet, and PCIe [? ?]). To overcome this problem, the developed Verilog code will conditionally compile where vendor specific requirements are present.

3.1.2 Extended Deliverables (ED)

The project's extended deliverables are described below.

ED1 Design a RISC core with an instructions-per-clock (IPC) rating of at least 1.0 (a single-cycle CPU).

ED2 Design a RISC core with a pipe-lined data path to increase the design's clock speed.

ED3 Design a scalable multi-core interconnect supporting arbitrary (more than 32) RISC core instances (manycore) using Network-on-Chip (NoC) architecture.

ED4 Design a compiler-backend for the PRCO304 [?] compiler to support the ISA from1 CD1. This will make it easier to build complex multi-core software for the processor.

ED5 The RISC core can communicate to peripherals via a memory-mapped addresses using the Wishbone bus.

ED6 Implement various memory-mapped peripherals such as UART, GPIO, LCD, to aid visual representation of the processor during the demonstration viva.

ED7 Store instruction memory in SPI flash.

ED8 Reprogram instruction memory at runtime from host computer.

ED9 Processor external debugger using host-processor link.

3.2 Project Timeline

3.2.1 Project Stages

The project is split up into many stages to aid planning and management of the project. There are 8 unique stage areas: 1. Initial project conception; 2 Basic RISC core development; 3. Extended RISC core development; 4. Multi-core development; 5. Processor quality-of-life (QoL) improvements; 6. Compiler development; 7. Demo preparation, and 8. Final report.

The project stages are shown in Table 3.1.

Stage	Title	Start Date	Days	Core	Applicable Deliverables
1.0	Research	Feb 04	7	x	
1.1	Requirement gathering/review	Feb 11	14	x	
1.1	Processor specification, architecture, ISA	Feb 18	100	x	CD1
1.2	Stage/Time Allocation Planning	Feb 25	7	x	
2.1	Decoder, Register Set, impl & integration	Feb 25	14	x	CD2
2.2	Register set impl & integration	Mar 04	14	x	CD2
2.3	Local memory impl & integration	Mar 11	14	x	CD2
3.1	Memory mapped register layout & impl	Apr 01	21		ED5
3.2	Wishbone peripheral bus connected to MMU	Apr 08	21		ED5
3.3	Pipelined implementation and verification	Apr 15	21		ED2
3.4	Cache memory design & impl	Apr 22	28		ED2
4.1	Multi-core communication interface	TBD	TBD	x	CD3
4.2	Shared-memory controller	TBD	TBD	x	CD3
4.3	Scalable multi-core interface (10s of cores)	TBD	TBD	x	CD3
4.4	Multi-core example program (reduction)	TBD	TBD	x	CD4
5.1	SPI-FPGA interface for OTG programming	TBD	TBD		ED7
5.2	FPGA-PC interfacing	TBD	TBD		ED9
5.3	FPGA-PC debugging (instruction breakpoints)	TBD	TBD		ED9
6.1	Compiler backend for vmicro16	TBD	TBD		ED4
6.2	Compiler support for multi-core codegen	TBD	TBD		ED4
7.1	Wishbone peripherals for demo	TBD	TBD	x	CD4
8.1	Final Report	TBD	TBD	x	

Table 3.1: Project stages throughout the life cycle of the project.

3.2.2 Project Stage Detail

Stages 1.0 through 1.2 – Research and Project Conception

These stages cover initial research of existing problems and solutions in the multiprocessor area. The instruction set architecture is also proposed that later stages will implement.

Stages 2.1 through 2.3 – Processor module Design, Implementation, and Integration

These stages cover the design, implementation, and integration of key processor core modules such as the instruction decoder, register sets and local memory. Integration of all the modules is a challenging task because some modules have both asynchronous and synchronous signals that need to be timed correctly in order for other modules to receive valid data. An example of this is the register set which has asynchronous read ports that are later clocked in the instruction decode stage.

Stages 3.1 through 3.4 – Advanced Processor Implementation

These stages add advanced features to the processor to provide a more functional product. Although these stages are classified as extended, their technical requirement to design and implement is not great and so are have time allocations in the project schedule. The extended features that these stages introduce are: pipelined processor stages – to drastically increase processor performance; provide a memory-mapped peripheral interface through the MMU; provide a Wishbone master interface to the MMU – allowing external peripherals such as GPIO and LCD displays to be utilised in a modular fashion; and to implement a cache memory for each processor core.

Stages 4.1 through 4.4 – Multiprocessor Functionality

These stages are dedicated to adding multiprocessor functionality using a loosely coupled architecture to the processor.

Stages 5.1 through 5.3 – Debugging Features

These stages cover debugging features and are classified as extended due to the large development time required to implement them as well as not being related to multiprocessor systems.

Stages 6.1 through 6.2 – Compiler Backends

These stages cover the implementation of a compiler backend to ease software writing and programming of the processor.

Stage 7.1 – Wishbone Peripherals

Additional Wishbone peripherals, such as SPI and timers will be added to produce a more useful multiprocessor system.

Stage 8.1 – Final Report

This stage is dedicated to the final report write-up. It is expected to be an iterative task that is active throughout the lifespan of the project.

3.2.3 Timeline

The project stages from Table 3.1 are displayed below in a Gantt chart.

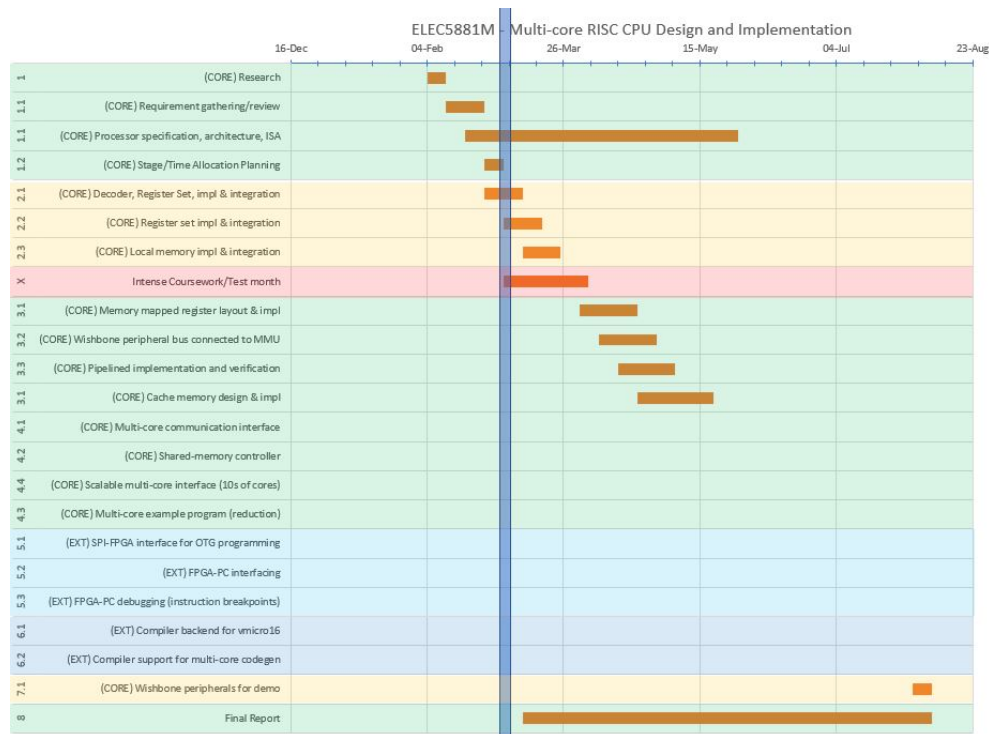


Figure 3.1: Project stages in a Gantt chart.

3.3 Resources

This section describes the hardware and software resources required to fulfil the project.

3.3.1 Hardware Resources

Core deliverable **CD5** requires the designed RISC core to be implemented and demonstrated on multiple FPGA devices. Although my design should synthesise for physical IC implementation, due to high costs and lengthy production times, it is not a primary development target. Due to having past experience with Xilinx FPGAs from my placement work and experience with Altera from university modules it was decided to target the Xilinx Spartan 6 XC6SLX9 and the Altera Cyclone V.

Terasic DE1-SoC Development Board

The Terasic DE1-SoC development board features a large Cyclone V FPGA and many peripherals, such as seven-segment displays, 64 MB SDRAM, ADCs, and buttons and switches, which will aid demonstration of the project. The development board is available through the university so the cost is negligible. Figure 3.2 shows the peripherals (green) available to the FPGA.

Minispartan 6+ FPGA Development Board

The Minispartan 6+ is a hobbyist FPGA development board with fewer peripherals than the DE1-SoC. The board features a Xilinx Spartan 6 XC6LX9 which has far fewer resources than the DE1-

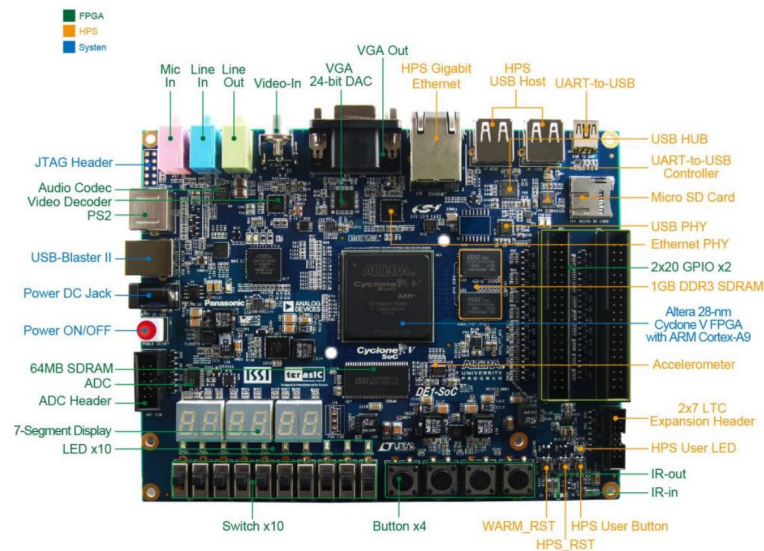


Figure 3.2: Terasic DE1-SoC development board featuring the Altera Cyclone V FPGA and many peripherals. Image source: [2].

SoC's Cyclone V however it's simplicity and my familiarity with Xilinx's software suite will speed up development. The development board is shown in Figure 3.3.

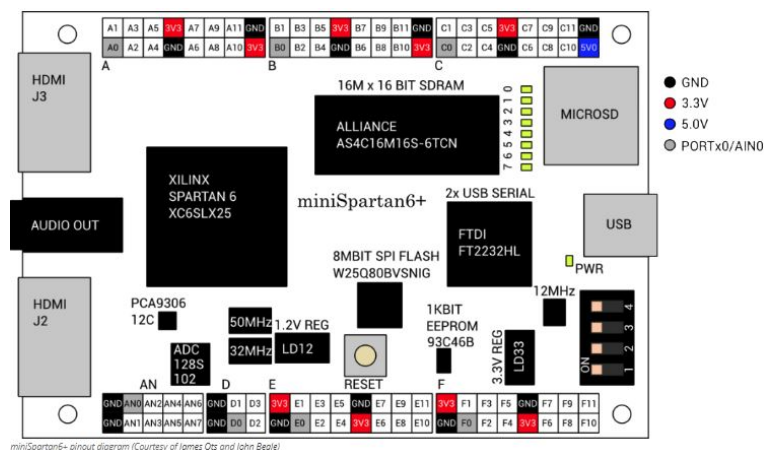


Figure 3.3: Minispartan-6+ development board featuring the Xilinx Spartan 6 XC6SLX9. Note that the XC6SLX9 and XC6SLX25 FPGAs share the same board. Image source: [3].

3.3.2 Software Resources

Intel Quartus

Intel Quartus Prime is a paid-for SoC, CPLD, and FPGA software suite targeting Intel's Stratix, Arria, and Cyclone based FPGAs. The university provides student licences which will be used via VPN.

Xilinx ISE Webpack

Xilinx ISE Webpack is Xilinx's free software suite for FPGA development for Spartan 6 based FPGAs. Due to ISE's intuitive and fast work flow, most of the initial simulation and verification processes will be performed using ISE. This will greatly improve development times.

Verilator

Verilator is an open-source Verilog to C++ transpiler which provides a C++ interface to simulate Verilog modules and read/write values similar to a test bench. Verilator will be used for specific modules within the RISC core such as the ALU and decoder as Verilator is useful when performing exhaustive verification.

3.4 Legal and Ethical Considerations

The RISC core is designed to be used as an academic research and educational tool to aid learning and understanding of RISC and multi-core machines. It should not be used for roles where mission critical or safety is a factor.

The processor does not provide any memory protection features and any software running on the processor has full access to all memory.

The processor does not store/track/predict software instructions. The processor uses pipelining techniques to improve performance which results in future instructions entering the pipeline even if the software's logical sequence does not include these instructions. This could result in security vulnerabilities similar to Intel's Spectre vulnerability [4].

Chapter 4

Current Progress

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This chapter discusses the current progress made towards the project, including designs, implementation, and current results.

4.1 RISC Core

Following the project time line described in section 3.2, the first couple months have been dedicated to the design and implementation of the instruction set architecture and RISC core with stages 1-3. Good progress has been made in both deliverables, the ISA and the RISC core, and the progress is on-time with the initial project time line. The core has been nicknamed *Vmicro16* – short for Verilog microprocessor 16-bit.

4.1.1 Instruction Set Architecture

A 16-bit instruction set architecture (ISA) has been designed using an iterative approach. There currently exists 32 unique instructions covering most generic RISC operations (add, load/store, branch, compare, etc.) and at least 16 opcodes available to be provide multi-core communication and functionality. This number should be adequate to support these features when the work begins on the multi-core project stages (stages 4-7).

Design Goals

Having past experience designing and implementing ISAs for previous projects, I wanted to use that knowledge to design an even more efficient and compact instruction set that could provide much greater functionality. The technical design goals of the ISA are described below:

ISA1 Use a fixed width of 16-bits for all instructions.

This will significantly reduce RTL resources and encourage efficiency by not wasting spare bits. In addition, many SPI flash and RAMs support 16-bit wide data reads which will allow each instruction fetch to only require one clock cycle, thus increasing processor performance.

ISA2 Be able to select at least two registers for common instructions.

This will reduce the number of required instructions to manipulate register data. A disadvantage of using two instead of three register selects is that instructions are always destructive – they always *destroy* existing data in the destination register (e.g. $R0 = \text{ADD } R0 \ R1$) unlike constructive instructions that provide a unique register select for the destination (e.g. $R2 = \text{ADD } R0 \ R1$).

ISA3 Reduce bit-space for frequently used instructions (MOV, MOVI, ADD).

Due to the 16-bit limit, two register selects, and immediate values, the opcode bits are reduced resulting in fewer unique instructions. To overcome this constraint, spare bits in other instructions will be appended to the opcode bits to extend the opcode range. This however, will require a more complex decoder that must first switch the opcode, then switch any spare bits to determine the final opcode. This method will significantly increase the number of unique instructions provided by the instruction set.

ISA4 Provide frequently used actions as options for existing instructions.

In software, frequently used actions include incrementing/decrementing by 1 and performing logical comparisons which usually take more than one instruction on some RISC architectures. As they are common actions, the instruction overhead and time may be significant and can affect performance. To provide a solution to this problem, in addition to using spare bits to extend the opcode range, spare bits will be used to signify a frequently used action action to be performed by the ALU.

As shown in Figure 4.1, frequently used commands such as incrementing/decrementing and logical comparisons are provided by setting spare bits to special values. For example, the instructions ARITH_UADDI and ARITH_SSUBI extend the ARITH_U and ARITH_S opcodes by filling the spare bit, 4. If this bit is not set (0), the instruction allows for a 4-bit immediate value to be added in addition to the two register selects. The 4-bit immediate allows adding a small number to the ALU which is useful in the case of software for loops where an increment/decrement of more than 1 is required.

Another example is the SETC instruction. Inspired by Intel's x86 SETCC, the instructions sets the destination register to zero or one depending on the result of the CMP instruction's flags. Without this instruction, multiple branches would be required to convert the comparison's flags to logical zeros and ones.

ISA5 Provide instructions for performing bitwise manipulations.

RISC processors are commonly used for microprocessing and microcontroller actions which typically includes bit manipulation. The ISA provides bitwise OR, XOR, AND, NOT, and shifting instructions under a single opcode to fill this need.

ISA6 Provide instructions for explicitly performing signed and unsigned arithmetic.

Performing signed and unsigned arithmetic is a key requirement for RISC applications and so it was decided to provide such instructions. Software programmers can easily switch between signed and unsigned arithmetic by setting bit 11 in the ARITH instruction family. Being able to change between signed and unsigned arithmetic instructions by changing a single bit will make the RISC processor's decoder module smaller and less complex.

Without explicit unsigned and signed instructions, extra instructions would be required to perform addition and subtraction. In addition, due to two's complement representation of

signed numbers, the highest immediate operand value would be halved, resulting in more instructions to reach the desired value.

	15-11	10-8	7-5	4-0	rd ra simm5
	15-11	10-8	7-0		rd imm8
	15-11	10-0			nop
	15	14:12	11:0		extended immediate
NOP	00000		X		
LW	00001	Rd	Ra	s5	Rd <= RAM[Ra+s5]
SW	00010	Rd	Ra	s5	RAM[Ra+s5] <= Rd
BIT	00011	Rd	Ra	s5	bitwise operations
BIT_OR	00011	Rd	Ra	00000	Rd <= Rd Ra
BIT_XOR	00011	Rd	Ra	00001	Rd <= Rd ^ Ra
BIT_AND	00011	Rd	Ra	00010	Rd <= Rd & Ra
BIT_NOT	00011	Rd	Ra	00011	Rd <= ~Ra
BIT_LSHFT	00011	Rd	Ra	00100	Rd <= Rd << Ra
BIT_RSHFT	00011	Rd	Ra	00101	Rd <= Rd >> Ra
MOV	00100	Rd	Ra	X	Rd <= Ra
MOVI	00101	Rd		i8	Rd <= i8
ARITH_U	00110	Rd	Ra	s5	unsigned arithmetic
ARITH_UADD	00110	Rd	Ra	11111	Rd <= uRd + uRa
ARITH_USUB	00110	Rd	Ra	10000	Rd <= uRd - uRa
ARITH_UADDI	00110	Rd	Ra	0AAAA	Rd <= uRd + Ra + AAAA
ARITH_S	00111	Rd	Ra	s5	signed arithmetic
ARITH_SADD	00111	Rd	Ra	11111	Rd <= sRd + sRa
ARITH_SSUB	00111	Rd	Ra	10000	Rd <= sRd - sRa
ARITH_SSUBI	00111	Rd	Ra	0AAAA	Rd <= sRd - sRa + AAAA
BR	01000	Rd		i8	conditional branch
BR_U	01000	Rd		0000 0000	Any
BR_E	01000	Rd		0000 0001	Z=1
BR_NE	01000	Rd		0000 0010	Z=0
BR_G	01000	Rd		0000 0011	Z=0 and S=0
BR_GE	01000	Rd		0000 0100	S=0
BR_L	01000	Rd		0000 0101	S != 0
BR_LE	01000	Rd		0000 0110	Z=1 or (S != 0)
BR_S	01000	Rd		0000 0111	S=1
BR_NS	01000	Rd		0000 1000	S=0
CMP	01001	Rd	Ra	X	SZO <= CMP(Rd, Ra)
SETC	01010	Rd	Ra	X	Rd <= Imm8 == SZO ? 1 : 0
MOVI_LARGE	1	Rd	i12		Rd <= i12

Figure 4.1: Initial Vmicro16 16-bit instruction set architecture. Coloured regions represent instruction families (bitwise, branching, arithmetic, etc.).

The ISA table is shown in Figure 4.1. The top 5 bits (15-11) are dedicated to the opcode resulting in 32 unique values. Currently only the bits 14-11 are used (NOP to SETC) leaving the top bit spare. Initially, this bit was reserved to indicate an extended immediate instruction, MOVI12, supporting a large 12-bit immediate value, however later in the design it was decided that the top bit would indicate special instructions dedicated for multi-core operation. This leaves 16 spare unique opcodes for this purpose.

4.1.2 Design and Implementation

The RISC core design is a traditional 5-stage processor (fetch, decode, execute, memory, write-back).

To satisfy [CD5](#), the Verilog code will be self-contained in a single file. This reduces the hierarchical complexity and eases cross-vendor project set-up as only a single file is required to be included. A disadvantage with this single file approach is that some external Verilog verification tools that I plan to use, such as Verilator, do not currently support multiple Verilog modules (due to an unfixed bug) within a single file.

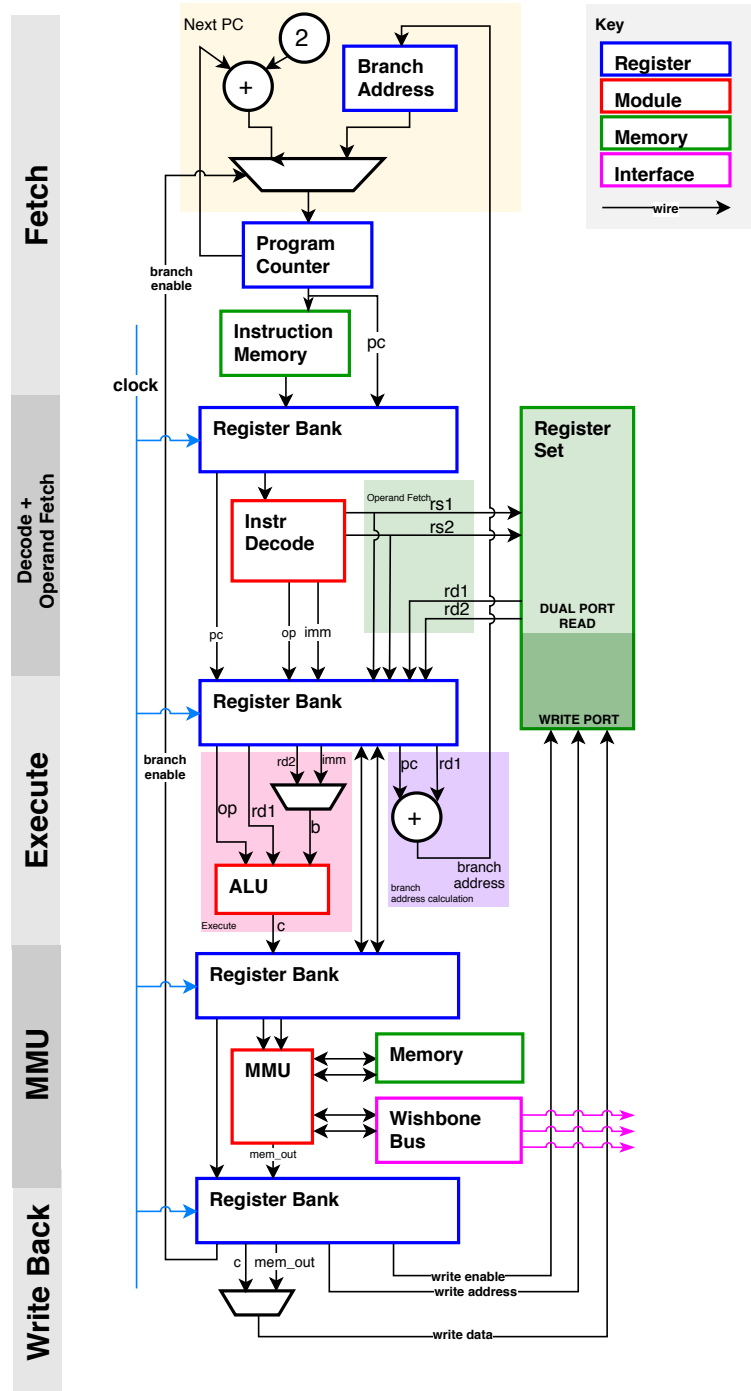


Figure 4.2: Vmicro16 RISC 5-stage RTL diagram showing: instruction pipelining (data passed forward through clocked register banks at each stage); branch address calculation; ALU operand calculation (**rd2** or **imm**); and program counter incrementing.

Instruction and Data Memory

The design uses separate instruction and data memories similar to a Harvard architecture computer. This architecture was chosen due because I find it easier to implement.

Register File

To support design goal [ISA2](#), the register set features a dual-port read and single-port write. This allows instructions to read 2 registers simultaneously for any instruction. The single-port write allows the instruction output to be written to the register file.

Pipelining

The extended deliverable [ED1](#), to provide atleast 1 instructions per clock. Previous processor designs of mine have all required multiple clocks per instruction as it is a lot easier to implement. Modern processors today can output 1 or more instructions per clock through the use of instruction pipelining. This technique increases throughput of the processor by performing each stage in parallel. In this pipeline, instructions still travel through each stage in the same order, the difference is that the fetch stage does not wait for the final stage to complete and so fetches a new instruction every clock cycle, resulting in each stage operating on new data every clock cycle. To extend my knowledge in CPU pipelining, extended deliverable [ED1](#) is proposed.

Instruction pipelining is harder to implement as data and control hazards can occur. Data hazards occur when instructions are dependant on the output of a previous instruction that has not left the pipeline, for example a register dependency. Methods to detect this hazard include checking if the register selects in the decode stage are present in future stages of the pipeline. If this check is true, then the current instruction depends on an instruction in the pipeline, and the processor can either wait until the dependant instruction has left the pipeline (i.e. has been written back to registers) or insert a NOP that will produce a *bubble* in the pipeline allowing the final stage to execute before the dependant instruction continues.

Control hazards occur when conditional or interrupt branching instructions are in the pipeline and their result has not been calculated yet. This results in preceding instructions entering the pipeline when they should not be executed due to the conditional branch. To detect this hazard, for instructions that perform branching or conditional execution, a global flag is set. When the outcome of the conditional check is performed, stages after decode are allowed to commit their results. Fortunately this technique is fairly simple implement.

This project's RISC processor implements these two hazard detectors and solutions to resolve them. The data hazard resolver implements a `valid` signal that is passed forward from stage to stage. This signal is low when a hazard has occured and indicates that receiving stage should not operate on the previous stage's data. Each stage's `valid` signal is dependant on the previous stages `valid` signal. This allows future stages to stall when a hazard is detected in previous stages. A diagram of the implementation of these hazards in the processor is shown in [Figure 4.3](#).

Memory Management Unit

It was decided to use a memory management unit (MMU) to make it easier and extensible to communicate with external peripherals or additional registers. This method would transparently use the

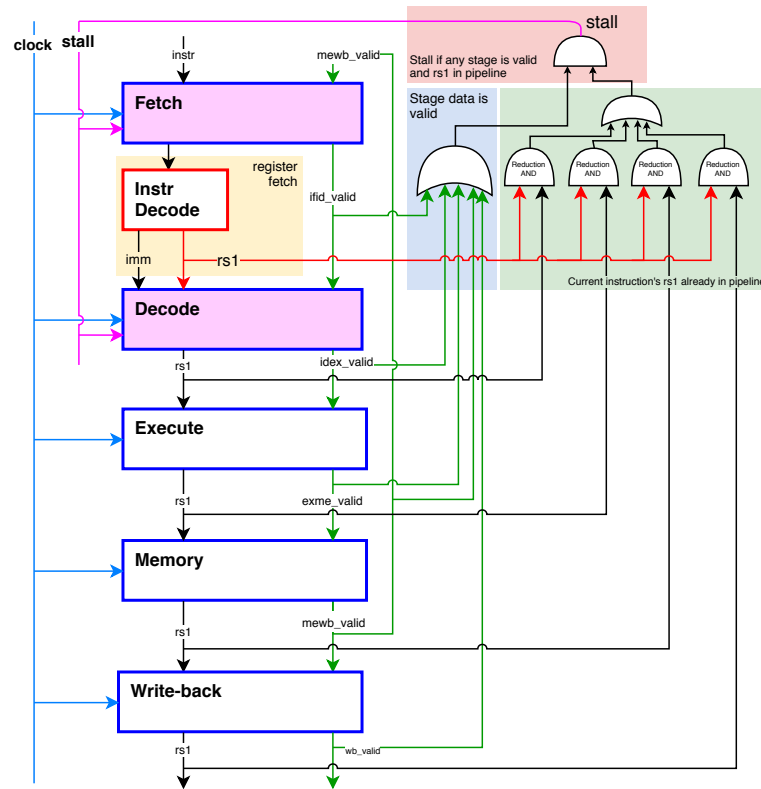


Figure 4.3: Pipeline data hazard detection. The register selects are passed forward through each stage and compared to the IDEX (latest instruction) register selects. If they match, the latest instruction depends on the output of an instruction in the pipeline, the IFID and IDEX stages are stalled to allow the instruction in the pipeline to commit.

existing LW/SW instructions which removes the requirement for a unique instruction for each peripheral.

Proposed Memory Mapped Addresses

The peripheral addresses are currently based on classes. For example, a memory-mapped address may use the upper byte to address a peripheral and the lower byte to address a register/function in that peripheral.

Later in the project, I plan to rewrite the addressing scheme to use a simpler address format which is closer to commonly used peripheral addressing schemes used today. The proposed memory mapped addresses for each system and peripheral are listed below.

Address (16-bit aligned)	Peripheral Name
0x0000	NOP (reads returns 0, writes do nothing)
0x00ZZ	Per-core scratch RAM (ZZ = 8-bit RAM address)
0x0100	Extended Core Registers 1
0x0200	Extended Core Registers 2
0x03ZZ	Wishbone Master controller select (ZZ contains 8-bit wishbone slave address)
0x1XYZ	Master core controller (X = slave select, Y = instruction, Z = data)

Table 4.1: Provisional memory-mapped addresses table.

ALU Design

The Vmicro16's ALU is an asynchronous module that has 3 inputs: data a; data b; and opcode op, and outputs data value c. The ALU is able to operate on both register data (rd1 and rd2) and immediate values. A switch is used to set the b input to either the rd2 or imm value from the previous stage.

Currently, the ALU does not store flags to indicate overflow, equality, or zero values in the module itself. Instead the ALU outputs the result of the CMP, which calculates such flags, to be written back to the register set in the write-back stage. This means that in order to perform a conditional operation, such as a branch, the register containing the CMP flags must be included in the instruction.

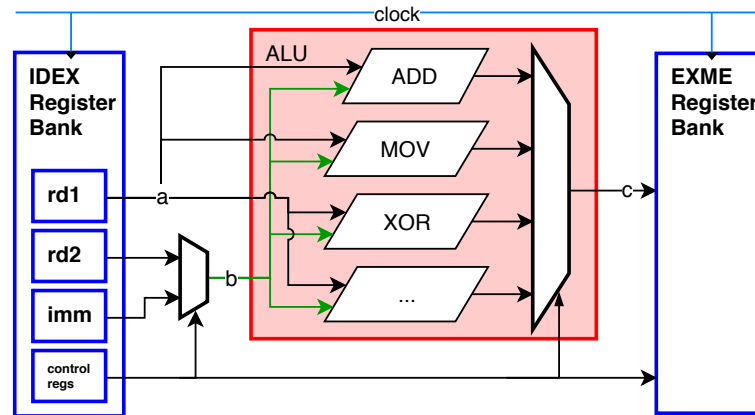


Figure 4.4: Vmicro16 ALU diagram showing clocked inputs from the previous IDEX stage being

The Verilog implementation of the ALU is shown in Figure 4.5. The ALU's asynchronous output is clocked with other registers, such as destination register rs1 and other control signals, in the EXME register bank.

```

322     end
323
324     MMU_STATE_T3: begin
325         // Slave has output a ready signal (finished)
326         M_PENABLE <= 0;
327         M_PADDR   <= 0;
328         M_PWDATA  <= 0;
329         M_PSELx   <= 0;
330         M_PWRITE  <= 0;
331         // Clock the peripheral output into a reg,
332         // to output on the next clock cycle
333         per_out   <= M_PRDATA;
334
335         mmu_state <= MMU_STATE_T1;

```

Figure 4.5: Vmicro16's ALU implementation named vmicro16_alu. vmicro16.v

Decoder Design

Instruction decoding occurs in the between the IFID and IDEX stages. The decoder extracts register selects and operands from the input instruction. The decoder outputs are asynchronous which allows the register selects to be passed to the register set and register data to be read asynchronously. The register selects and register read data is then clocked into the IDEX register bank.


```

224
225 // more luts than below but easier
226 //wire tim0_en = (mmu_addr >= `DEF_MMU_TIMO_S)
227 //      && (mmu_addr <= `DEF_MMU_TIMO_E);
228 //wire sreg_en = (mmu_addr >= `DEF_MMU_SREG_S)
229 //      && (mmu_addr <= `DEF_MMU_SREG_E);
230 //wire intv_en = (mmu_addr >= `DEF_MMU_INTSV_S)
231 //      && (mmu_addr <= `DEF_MMU_INTSV_E);
232 //wire intm_en = (mmu_addr >= `DEF_MMU_INTSM_S)
233 //      && (mmu_addr <= `DEF_MMU_INTSM_E);
234
235 wire tim0_en = ~mmu_addr[12] && ~mmu_addr[9] && ~mmu_addr[7];
236 wire sreg_en = mmu_addr[7] && ~mmu_addr[4] && ~mmu_addr[5];
237 wire intv_en = mmu_addr[8] && ~mmu_addr[3];
238 wire intm_en = mmu_addr[8] && mmu_addr[3];
239
240 wire apb_en   = !(tim0_en, sreg_en, intv_en, intm_en);
241 wire tim0_we  = (tim0_en && mmu_we);
242 wire intv_we  = (intv_en && mmu_we);
243 wire intm_we  = (intm_en && mmu_we);
244
245 // Special register selects

```

Figure 4.6: Vmicro16's decoder module code showing nested bit switches to determine the intended opcode. vmicro16.v

In Figure 4.6, it can be seen that the first 8 opcode cases are represented using the same 15-11 bits, however the VMICR016_OP_BIT instructions require another bit range to be compared to determine the output opcode.

4.1.3 Verification

Currently, the only verification method used is manual inspection of the output waveforms of a test bench. For now, it is easier and faster to spot erroneous states by hand due to the large complexity of the pipeline. Later in the project, automatic test benches will be utilised.

Known Bugs

Known bugs exist within the RISC core however none are critical as they can be easily avoided in software.

BUG1 Stall detection does not consider load/store instructions.

Due to instruction pipelining techniques used by the processor and lack of address checking in the EXME and MEWB stages, LW instructions immediately after SW instructions:

```
SW R0 (R2+16)
```

```
LW R1 (R2+16)
```

will not return the previously stored value. In addition, because of the target address is calculated by the ALU (e.g. R2+16), detecting matching addresses at IFID and IDEX stage is not trivial, and because of this, a hardware fix is not planned for the final version. It is possible to overcome this problem in software by placing at least 5 NOP instructions after each SW.

Chapter 5

Future Work

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5.1.2	Future Work	25

5.1 Project Status

Four months have passed since the start of the project and significant progress has been made to the final deliverable.

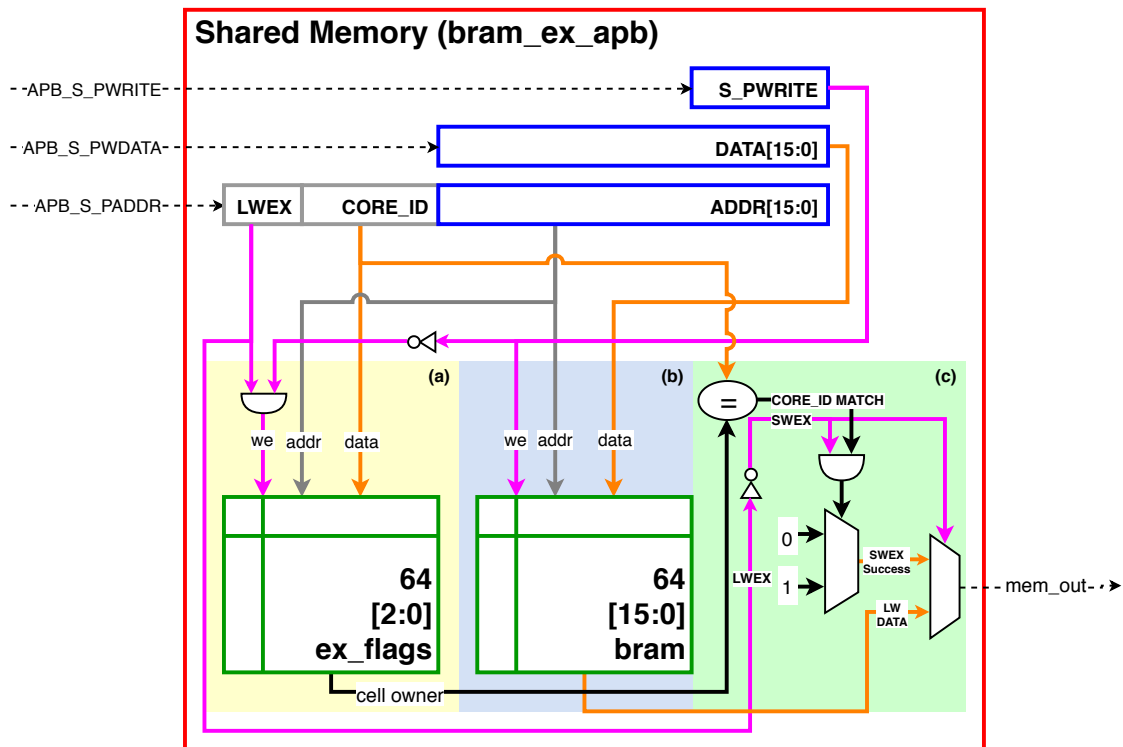


Figure 5.1: Caption for BRAMex

The current active stage is 3.3 *Pipeline Implementation and Verification* where the processor pipeline is being verified against a range of simple software sequences. It is important that this verification is thorough and the output is bug free as future additions to the processor will utilise this foundation.

5.1.1 Updated Project Time Line

The project table described in section 3.2 did not allocate times for stages 4.1 and later. This was due to expected high demand from other modules and exams in this time period and so it was decided to not allocate times that would later not be followed.

Now that this time period is closer, time allocations have been assigned for stages 4, 7, and 8. The state of stage 5's extended deliverables, to implement debugging interfaces, have changed from *Unknown* to *Cancelled* due to expected high workload from other modules in the next month. The cancellation of these stages will not severely affect the final functionality of the deliverable however it will make debugging the processor slightly more difficult. It was decided to remove these extended features to allow for more time to be spent on core functionality.

The updated project status is shown in Table 5.1 and in Figure 5.2.

5.1.2 Future Work

May and early June are reserved for work on other modules and preparation for exams. From mid-June, work will resume on verifying the end of stage 3 and then work will start on stage 4 (focussed on designing and implementing multiprocessor features). After stage 4, software algorithms will be compiled for the ISA and evaluated against Amdahl's Law.

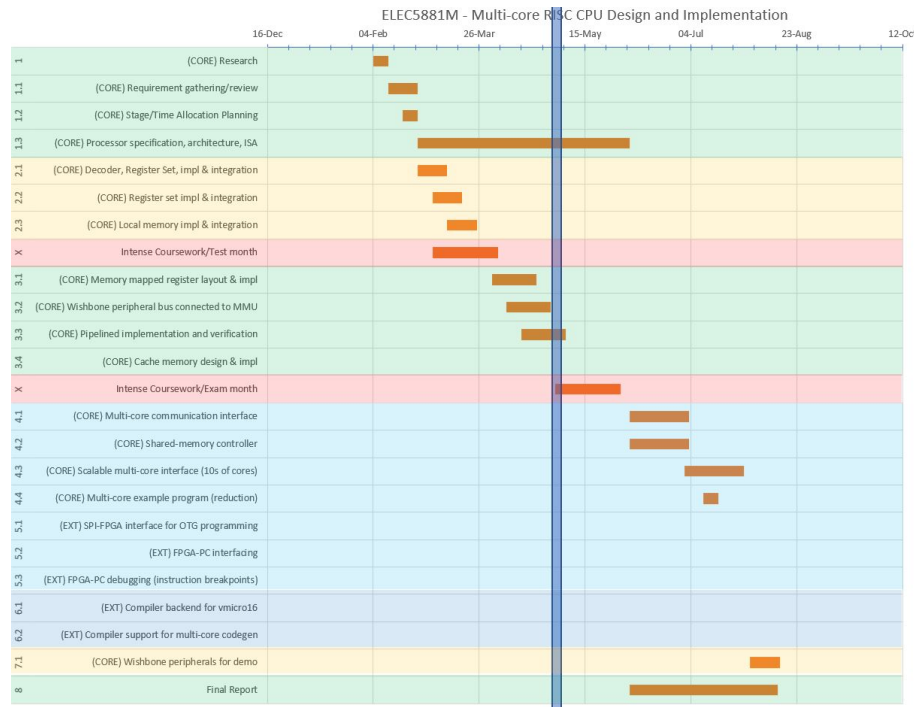


Figure 5.2: Updated project time gantt chart showing time allocations for stage 4.

Stage	Title	Start Date	Core	Status
1.0	Research	Feb 04	x	Completed
1.1	Requirement gathering/review	Feb 11	x	Completed
1.1	Processor specification, architecture, ISA	Feb 18	x	Completed
1.2	Stage/Time Allocation Planning	Feb 25	x	Completed
2.1	Decoder, Register Set, impl & integration	Feb 25	x	Completed
2.2	Register set impl & integration	Mar 04	x	Completed
2.3	Local memory impl & integration	Mar 11	x	Completed
3.1	Memory mapped register layout & impl	Apr 01		On-going
3.2	Wishbone peripheral bus connected to MMU	Apr 08		On-going
3.3	Pipeline implementation and verification	Apr 15		On-going
3.4	Cache memory design & impl	Apr 22		Cancelled
4.1	Multi-core communication interface	Jun 05	x	Planned
4.2	Shared-memory controller	Jun 05	x	Planned
4.3	Scalable multi-core interface (10s of cores)	Jul 01	x	Planned
4.4	Multi-core example program (reduction)	Jul 10	x	Planned
5.1	SPI-FPGA interface for OTG programming	TBD		Cancelled
5.2	FPGA-PC interfacing	TBD		Cancelled
5.3	FPGA-PC debugging (instruction breakpoints)	TBD		Cancelled
6.1	Compiler backend for vmicro16	TBD		Unknown
6.2	Compiler support for multi-core codegen	TBD		Unknown
7.1	Wishbone peripherals for demo	Aug 01	x	Planned
8.1	Final Report	Jun 05	x	Planned

Table 5.1: Updated project stages.

Chapter 6

Conclusion

With the end of Moore's Law looming, processor designers must use other strategies to continue improving performance of processors – multiprocessor and parallelism being a primary strategy. This project sets out to improve my knowledge on multiprocessor communication by designing, implementing, and verifying a multiprocessor – and I believe starting from scratch is the best way to accomplish this learning task.

To date, a compact 16-bit RISC instruction set has been designed and implemented in a Verilog single-core processor. Whilst single-core verification is still on-going, good progress has been made and extended deliverables from stage 3, such as instruction pipelining and memory-mapped peripherals via a Wishbone bus, has been implemented successfully.

Stage 5's extended deliverables and the cache memory have been cancelled but they do not effect the core functionality of the processor. The planned project time-line for future stages is realistic and accomplishing the project's goals appears achievable.

References

- [1] V. Subramanian, "Multiple gate field-effect transistors for future cmos technologies," *IETE Technical review*, vol. 27, no. 6, pp. 446–454, 2010.
- [2] T. Technologies, "Soc platform - cyclone - de1-soc board." [Online]. Available: <https://www.terasic.com.tw/cgi-bin/page/archive.pl?Language=English&No=836>
- [3] *MiniSpartan6+*, Scarab Hardware, 2014. [Online]. Available: <https://www.scarabhardware.com/minispartan6/>
- [4] P. Kocher, D. Genkin, D. Gruss, W. Haas, M. Hamburg, M. Lipp, S. Mangard, T. Prescher, M. Schwarz, and Y. Yarom, "Spectre attacks: Exploiting speculative execution," *arXiv preprint arXiv:1801.01203*, 2018.
- [5] J. Balkind, M. McKeown, Y. Fu, T. Nguyen, Y. Zhou, A. Lavrov, M. Shahradd, A. Fuchs, S. Payne, X. Liang *et al.*, "Openpiton: An open source manycore research framework," in *ACM SIGARCH Computer Architecture News*, vol. 44, no. 2. ACM, 2016, pp. 217–232.
- [6] N. Satish, M. Harris, and M. Garland, "Designing efficient sorting algorithms for manycore gpus," in *2009 IEEE International Symposium on Parallel & Distributed Processing*. IEEE, 2009, pp. 1–10.
- [7] S. Binet, P. Calafiura, S. Snyder, W. Wiedenmann, and F. Winklmeier, "Harnessing multicores: Strategies and implementations in atlas," in *Journal of Physics: Conference Series*, vol. 219, no. 4. IOP Publishing, 2010, p. 042002.

Appendix A - Code Listing

vmicro16.v

The single core RISC processor is defined in this file. It contains many submodules such as the decoder and local memory.

```
1 // This file contains multiple modules.
2 // Verilator likes 1 file for each module
3 /* verilator lint_off DECLFILENAME */
4 /* verilator lint_off UNUSED */
5 /* verilator lint_off BLASEQ */
6 /* verilator lint_off WIDTH */
7
8 // Include Vmicro16 ISA containing definitions for the bits
9 `include "vmicro16_isa.v"
10
11 `include "clog2.v"
12 `include "formal.v"
13
14
15
16 // This module aims to be a SYNCHRONOUS, WRITE_FIRST BLOCK RAM
17 // https://www.xilinx.com/support/documentation/user_guides/ug473.7Series_Memory_Resources.pdf
18 // https://www.xilinx.com/support/documentation/user_guides/ug383.pdf
19 // https://www.xilinx.com/support/documentation/sw_manuals/xilinx2016_4/ug901-vivado-synthesis.pdf
20 module vmicro16_bram # (
21     parameter MEM_WIDTH    = 16,
22     parameter MEM_DEPTH    = 64,
23     parameter CORE_ID      = 0,
24     parameter USE_INITS     = 0,
25     parameter PARAM_DEFAULTS_R0 = 0,
26     parameter PARAM_DEFAULTS_R1 = 0,
27     parameter PARAM_DEFAULTS_R2 = 0,
28     parameter PARAM_DEFAULTS_R3 = 0,
29     parameter NAME          = "BRAM"
30 ) (
31     input clk,
32     input reset,
33
34     input    [^(clog2(MEM_DEPTH)-1:0)] mem_addr,
35     input    [MEM_WIDTH-1:0] mem_in,
36     input                    mem_we,
37     output reg [MEM_WIDTH-1:0] mem_out
38 );
39 // memory vector
40 (* ram_style = "block" *)
41 reg [MEM_WIDTH-1:0] mem [0:MEM_DEPTH-1];
42
43 // not synthesizable
44 integer i;
45 initial begin
46     for (i = 0; i < MEM_DEPTH; i = i + 1) mem[i] = 0;
47     mem[0] = PARAM_DEFAULTS_R0;
48     mem[1] = PARAM_DEFAULTS_R1;
49     mem[2] = PARAM_DEFAULTS_R2;
50     mem[3] = PARAM_DEFAULTS_R3;
51
52     if (USE_INITS) begin
53         `define TEST_SW
54         `ifdef TEST_SW
55             $readmemh("E:\\Projects\\uni\\vmicro16\\sw\\verilog_memh.txt", mem);
56         `endif
57
58         `define TEST_ASM
59         `ifdef TEST_ASM
60             $readmemh("E:\\Projects\\uni\\vmicro16\\sw\\asm.s.hex", mem);
61         `endif
62
63         `//define TEST_COND
64         `ifdef TEST_COND
65             mem[0] = {`VMICRO16_OP_MOVI,    3'h7, 8'hC0}; // lock
66             mem[0] = {`VMICRO16_OP_MOVI,    3'h7, 8'hC0}; // lock
67         `endif
68
69         `//define TEST_CMP
70         `ifdef TEST_CMP
71             mem[0] = {`VMICRO16_OP_MOVI,    3'h0, 8'h0A};
72             mem[1] = {`VMICRO16_OP_MOVI,    3'h1, 8'h0B};
73             mem[2] = {`VMICRO16_OP_CMP,     3'h1, 3'h0, 5'h1};
74         `endif
75
76         `//define TEST_LWEX
77         `ifdef TEST_LWEX
78             mem[0] = {`VMICRO16_OP_MOVI,    3'h0, 8'hC5};
79             mem[1] = {`VMICRO16_OP_SW,      3'h0, 3'h0, 5'h1};
80             mem[2] = {`VMICRO16_OP_LW,      3'h2, 3'h0, 5'h1};
81             mem[3] = {`VMICRO16_OP_LWEX,    3'h2, 3'h0, 5'h1};
```

```

82     mem[4] = {'VMICRO16_OP_SWEX,    3'h3, 3'h0, 5'h1};
83     `endif
84
85     //`define TEST_MULTICORE
86     `ifdef TEST_MULTICORE
87     mem[0] = {'VMICRO16_OP_MOVI,    3'h0, 8'h90};
88     mem[1] = {'VMICRO16_OP_MOVI,    3'h1, 8'h33};
89     mem[2] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
90     mem[3] = {'VMICRO16_OP_MOVI,    3'h0, 8'h80};
91     mem[4] = {'VMICRO16_OP_LW,      3'h2, 3'h0, 5'h0};
92     mem[5] = {'VMICRO16_OP_MOVI,    3'h1, 8'h33};
93     mem[6] = {'VMICRO16_OP_MOVI,    3'h1, 8'h33};
94     mem[7] = {'VMICRO16_OP_MOVI,    3'h1, 8'h33};
95     mem[8] = {'VMICRO16_OP_MOVI,    3'h0, 8'h91};
96     mem[9] = {'VMICRO16_OP_SW,      3'h2, 3'h0, 5'h0};
97     `endif
98
99     //`define TEST_BR
100    `ifdef TEST_BR
101    mem[0] = {'VMICRO16_OP_MOVI,    3'h0, 8'h0};
102    mem[1] = {'VMICRO16_OP_MOVI,    3'h3, 8'h3};
103    mem[2] = {'VMICRO16_OP_MOVI,    3'h1, 8'h2};
104    mem[3] = {'VMICRO16_OP_ARITH_U,  3'h0, 3'h1, 5'b11111};
105    mem[4] = {'VMICRO16_OP_BR,      3'h3, `VMICRO16_OP_BR_U};
106    mem[5] = {'VMICRO16_OP_MOVI,    3'h0, 8'hFF};
107    `endif
108
109    //`define ALL_TEST
110    `ifdef ALL_TEST
111    // Standard all test
112    // REGS0
113    mem[0] = {'VMICRO16_OP_MOVI,    3'h0, 8'h81};
114    mem[1] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0}; // MMU[0x81] = 6
115    mem[2] = {'VMICRO16_OP_SW,      3'h2, 3'h0, 5'h1}; // MMU[0x82] = 6
116    // GP100
117    mem[3] = {'VMICRO16_OP_MOVI,    3'h0, 8'h90};
118    mem[4] = {'VMICRO16_OP_MOVI,    3'h1, 8'hD};
119    mem[5] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
120    mem[6] = {'VMICRO16_OP_LW,      3'h2, 3'h0, 5'h0};
121    // TIMO
122    mem[7] = {'VMICRO16_OP_MOVI,    3'h0, 8'h07};
123    mem[8] = {'VMICRO16_OP_LW,      3'h3, 3'h0, 5'h03};
124    // UART0
125    mem[9] = {'VMICRO16_OP_MOVI,    3'h0, 8'hA0}; // UART0
126    mem[10] = {'VMICRO16_OP_MOVI,    3'h1, 8'h41}; // ascii A
127    mem[11] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
128    mem[12] = {'VMICRO16_OP_MOVI,    3'h1, 8'h42}; // ascii B
129    mem[13] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
130    mem[14] = {'VMICRO16_OP_MOVI,    3'h1, 8'h43}; // ascii C
131    mem[15] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
132    mem[16] = {'VMICRO16_OP_MOVI,    3'h1, 8'h44}; // ascii D
133    mem[17] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
134    mem[18] = {'VMICRO16_OP_MOVI,    3'h1, 8'h45}; // ascii D
135    mem[19] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
136    mem[20] = {'VMICRO16_OP_MOVI,    3'h1, 8'h46}; // ascii E
137    mem[21] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
138    // BRAM0
139    mem[22] = {'VMICRO16_OP_MOVI,    3'h0, 8'hC0};
140    mem[23] = {'VMICRO16_OP_MOVI,    3'h1, 8'hA};
141    mem[24] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h5};
142    mem[25] = {'VMICRO16_OP_LW,      3'h2, 3'h0, 5'h5};
143    // GP101 (SSD 24-bit port)
144    mem[26] = {'VMICRO16_OP_MOVI,    3'h0, 8'h91};
145    mem[27] = {'VMICRO16_OP_MOVI,    3'h1, 8'h12};
146    mem[28] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
147    mem[29] = {'VMICRO16_OP_LW,      3'h2, 3'h0, 5'h0};
148    // GP102
149    mem[30] = {'VMICRO16_OP_MOVI,    3'h0, 8'h92};
150    mem[31] = {'VMICRO16_OP_MOVI,    3'h1, 8'h56};
151    mem[32] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h0};
152    `endif
153
154    //`define TEST_BRAM
155    `ifdef TEST_BRAM
156    // 2 core BRAM0 test
157    mem[0] = {'VMICRO16_OP_MOVI,    3'h0, 8'hC0};
158    mem[1] = {'VMICRO16_OP_MOVI,    3'h1, 8'hA};
159    mem[2] = {'VMICRO16_OP_SW,      3'h1, 3'h0, 5'h5};
160    mem[3] = {'VMICRO16_OP_LW,      3'h2, 3'h0, 5'h5};
161    `endif
162
163 end
164
165 always @ (posedge clk) begin
166     // synchronous WRITE_FIRST (page 13)
167     if (mem_we) begin
168         mem[mem_addr] <= mem_in;
169         $display($time, "\t\t\t%s[%h] <= %h",
170             NAME, mem_addr, mem_in);
171     end else
172         mem_out <= mem[mem_addr];
173     end
174
175     // TODO: Reset impl = every clock while reset is asserted, clear each cell
176     // one at a time, mem[i++] <= 0
177 endmodule
178
179 module vmicro16_core_mmu # (
180     parameter MEM_WIDTH    = 16,
181     parameter MEM_DEPTH    = 64,
182
183     parameter CORE_ID      = 3'h0,
184     parameter CORE_ID_BITS = `clog2(`CORES)
185 ) (
186     input clk,
187     input reset,
188
189     input req,
190     output busy,
191
192     // From core
193     input [MEM_WIDTH-1:0] mmu_addr,
194     input [MEM_WIDTH-1:0] mmu_in,
195     input mmu_we,
196     input mmu_lwex,

```



```

198     input                mmu_swex,
199     output reg [MEM_WIDTH-1:0] mmu_out,
200
201     // interrupts
202     output reg [DATA_WIDTH*DEF_NUM_INT-1:0] ints_vector,
203     output reg [DEF_NUM_INT-1:0] ints_mask,
204
205     // TO APB interconnect
206     output reg [APB_WIDTH-1:0] M_PADDR,
207     output reg                M_PWRITE,
208     output reg                M_PSELx,
209     output reg                M_PENABLE,
210     output reg [MEM_WIDTH-1:0] M_PWDATA,
211     // from interconnect
212     input [MEM_WIDTH-1:0] M_PRDATA,
213     input                M_PREADY
214 );
215 localparam MMU_STATE_T1 = 0;
216 localparam MMU_STATE_T2 = 1;
217 localparam MMU_STATE_T3 = 2;
218 reg [1:0] mmu_state = MMU_STATE_T1;
219
220 reg [MEM_WIDTH-1:0] per_out = 0;
221 wire [MEM_WIDTH-1:0] tim0_out;
222
223 assign busy = req || (mmu_state == MMU_STATE_T2);
224
225 // more luts than below but easier
226 //wire tim0_en = (mmu_addr >= DEF_MMU_TIM0_S)
227 //             && (mmu_addr <= DEF_MMU_TIM0_E);
228 //wire sreg_en = (mmu_addr <= DEF_MMU_SREG_S)
229 //             && (mmu_addr <= DEF_MMU_SREG_E);
230 //wire intv_en = (mmu_addr >= DEF_MMU_INTSV_S)
231 //             && (mmu_addr <= DEF_MMU_INTSV_E);
232 //wire intm_en = (mmu_addr >= DEF_MMU_INTSM_S)
233 //             && (mmu_addr <= DEF_MMU_INTSM_E);
234
235 wire tim0_en = mmu_addr[12] && mmu_addr[9] && mmu_addr[7];
236 wire sreg_en = mmu_addr[7] && mmu_addr[4] && mmu_addr[5];
237 wire intv_en = mmu_addr[8] && mmu_addr[3];
238 wire intm_en = mmu_addr[8] && mmu_addr[3];
239
240 wire apb_en = !((tim0_en, sreg_en, intv_en, intm_en));
241 wire tim0_we = (tim0_en && mmu_we);
242 wire intv_we = (intv_en && mmu_we);
243 wire intm_we = (intm_en && mmu_we);
244
245 // Special register selects
246 localparam SPECIAL_REGS = 8;
247 wire [MEM_WIDTH-1:0] sr_val;
248
249 // Interrupt vector and mask
250 initial ints_vector = 0;
251 initial ints_mask = 0;
252 wire [2:0] intv_addr = mmu_addr[DEF_NUM_INT-1:0];
253 always @(posedge clk)
254     if (intv_we)
255         ints_vector[intv_addr*DATA_WIDTH+:DATA_WIDTH] <= mmu_in;
256
257 always @(posedge clk)
258     if (intm_we)
259         ints_mask <= mmu_in;
260
261
262 always @(ints_vector)
263     $display($time,
264             "\tC%d\t\tints_vector W: | %h %h %h %h | %h %h %h %h |",
265             CORE_ID,
266             ints_vector[0*DATA_WIDTH+:DATA_WIDTH],
267             ints_vector[1*DATA_WIDTH+:DATA_WIDTH],
268             ints_vector[2*DATA_WIDTH+:DATA_WIDTH],
269             ints_vector[3*DATA_WIDTH+:DATA_WIDTH],
270             ints_vector[4*DATA_WIDTH+:DATA_WIDTH],
271             ints_vector[5*DATA_WIDTH+:DATA_WIDTH],
272             ints_vector[6*DATA_WIDTH+:DATA_WIDTH],
273             ints_vector[7*DATA_WIDTH+:DATA_WIDTH]
274     );
275
276 always @(intm_we)
277     $display($time, "\tC%d\t\tintm_we W: %b", CORE_ID, ints_mask);
278
279 // Output port
280 always @(*)
281     if (tim0_en) mmu_out = tim0_out;
282     else if (sreg_en) mmu_out = sr_val;
283     else if (intv_en) mmu_out = ints_vector[mmu_addr[2:0]*DATA_WIDTH+:DATA_WIDTH];
284     else if (intm_en) mmu_out = ints_mask;
285     else mmu_out = per_out;
286
287
288 // APB master to slave interface
289 always @(posedge clk)
290     if (reset) begin
291         mmu_state <= MMU_STATE_T1;
292         M_PENABLE <= 0;
293         M_PADDR <= 0;
294         M_PWDATA <= 0;
295         M_PSELx <= 0;
296         M_PWRITE <= 0;
297     end
298     else
299         casex (mmu_state)
300             MMU_STATE_T1: begin
301                 if (req && apb_en) begin
302                     M_PADDR <= {mmu_lvex,
303                                mmu_swex,
304                                CORE_ID[CORE_ID_BITS-1:0],
305                                mmu_addr[MEM_WIDTH-1:0]};
306
307                     M_PWDATA <= mmu_in;
308                     M_PSELx <= 1;
309                     M_PWRITE <= mmu_we;
310
311                     mmu_state <= MMU_STATE_T2;
312                 end
313             end

```

```

314
315     `ifndef FIX_T3
316         MMU_STATE_T2: begin
317             M_PENABLE <= 1;
318
319             if (M_PREADY == 1'b1) begin
320                 mmu_state <= MMU_STATE_T3;
321             end
322         end
323
324         MMU_STATE_T3: begin
325             // Slave has output a ready signal (finished)
326             M_PENABLE <= 0;
327             M_PADDR <= 0;
328             M_PWDATA <= 0;
329             M_PSELx <= 0;
330             M_PWRITE <= 0;
331             // Clock the peripheral output into a reg,
332             // to output on the next clock cycle
333             per_out <= M_PRDATA;
334
335             mmu_state <= MMU_STATE_T1;
336         end
337     `else
338         // No FIX_T3
339         MMU_STATE_T2: begin
340             if (M_PREADY == 1'b1) begin
341                 M_PENABLE <= 0;
342                 M_PADDR <= 0;
343                 M_PWDATA <= 0;
344                 M_PSELx <= 0;
345                 M_PWRITE <= 0;
346                 // Clock the peripheral output into a reg,
347                 // to output on the next clock cycle
348                 per_out <= M_PRDATA;
349
350                 mmu_state <= MMU_STATE_T1;
351             end else begin
352                 M_PENABLE <= 1;
353             end
354         end
355     `endif
356 endcase
357
358 (* ram_style = "block" *)
359 vmicro16_bram # (
360     .MEM_WIDTH (MEM_WIDTH),
361     .MEM_DEPTH (SPECIAL_REGS),
362     .USE_INITS (0),
363     .PARAM_DEFAULTS_R0 (CORE_ID),
364     .PARAM_DEFAULTS_R1 (CORES),
365     .PARAM_DEFAULTS_R2 (APB_BRAMO_CELLS),
366     .PARAM_DEFAULTS_R3 (SLAVES),
367     .NAME ("ram_sr")
368 ) ram_sr (
369     .clk (clk),
370     .reset (reset),
371     .mem_addr (mmu_addr['clog2(SPECIAL_REGS)-1:0]),
372     .mem_in (0),
373     .mem_we (0),
374     .mem_out (sr_val)
375 );
376
377 // Each M core has a TIM0 scratch memory
378 (* ram_style = "block" *)
379 vmicro16_bram # (
380     .MEM_WIDTH (MEM_WIDTH),
381     .MEM_DEPTH (MEM_DEPTH),
382     .USE_INITS (0),
383     .NAME ("TIM0")
384 ) TIM0 (
385     .clk (clk),
386     .reset (reset),
387     .mem_addr (mmu_addr[7:0]),
388     .mem_in (mmu_in),
389     .mem_we (tim0_we),
390     .mem_out (tim0_out)
391 );
392 endmodule
393
394
395 module vmicro16_regs # (
396     parameter CELL_WIDTH = 16,
397     parameter CELL_DEPTH = 8,
398     parameter CELL_SEL_BITS = 'clog2(CELL_DEPTH),
399     parameter CELL_DEFAULTS = 0,
400     parameter DEBUG_NAME = "",
401     parameter CORE_ID = 0,
402     parameter PARAM_DEFAULTS_R0 = 16'h0000,
403     parameter PARAM_DEFAULTS_R1 = 16'h0000
404 ) (
405     input clk,
406     input reset,
407     // Dual port register reads
408     input [CELL_SEL_BITS-1:0] rs1, // port 1
409     output [CELL_WIDTH-1:0] rd1,
410     //input [CELL_SEL_BITS-1:0] rs2, // port 2
411     //output [CELL_WIDTH-1:0] rd2,
412     // EX/WB final stage write back
413     input we,
414     input [CELL_SEL_BITS-1:0] ws1,
415     input [CELL_WIDTH-1:0] wd
416 );
417
418 (* ram_style = "distributed" *)
419 reg [CELL_WIDTH-1:0] regs [0:CELL_DEPTH-1] /*verilator public_flat*/;
420
421 // Initialise registers with default values
422 // Really only used for special registers used by the soc
423 // TODO: How to do this on reset?
424 integer i;
425 initial
426     if (CELL_DEFAULTS)
427         $readmemh(CELL_DEFAULTS, regs);
428     else begin
429         for(i = 0; i < CELL_DEPTH; i = i + 1)

```

```

430         regs[i] = 0;
431         regs[0] = PARAM_DEFAULTS_R0;
432         regs[1] = PARAM_DEFAULTS_R1;
433         end
434
435     `ifdef ICARUS
436         always @(regs)
437             $display($time, "\tC%02h\t\t| %h %h %h | %h %h %h %h |",
438                     CORE_ID,
439                     regs[0], regs[1], regs[2], regs[3],
440                     regs[4], regs[5], regs[6], regs[7]);
441     `endif
442
443     always @(posedge clk)
444         if (reset) begin
445             for(i = 0; i < CELL_DEPTH; i = i + 1)
446                 regs[i] <= 0;
447             regs[0] <= PARAM_DEFAULTS_R0;
448             regs[1] <= PARAM_DEFAULTS_R1;
449         end
450         else if (we) begin
451             $display($time, "\tC%02h: REGS #s: Writing %h to reg[%d]",
452                     CORE_ID, DEBUG_NAME, wd, ws1);
453
454             // Perform the write
455             regs[ws1] <= wd;
456         end
457
458         // sync writes, async reads
459         assign rd1 = regs[rs1];
460         //assign rd2 = regs[rs2];
461     endmodule
462
463     module vmicro16_dec # (
464         parameter INSTR_WIDTH = 16,
465         parameter INSTR_OP_WIDTH = 5,
466         parameter INSTR_RS_WIDTH = 3,
467         parameter ALU_OP_WIDTH = 5
468     ) (
469         //input clk, // not used yet (all combinational)
470         //input reset, // not used yet (all combinational)
471
472         input [INSTR_WIDTH-1:0] instr,
473
474         output [INSTR_OP_WIDTH-1:0] opcode,
475         output [INSTR_RS_WIDTH-1:0] rd,
476         output [INSTR_RS_WIDTH-1:0] ra,
477         output [3:0] imm4,
478         output [7:0] imm8,
479         output [11:0] imm12,
480         output [4:0] simm5,
481
482         // This can be freely increased without affecting the isa
483         output reg [ALU_OP_WIDTH-1:0] alu_op,
484
485         output reg has_imm4,
486         output reg has_imm8,
487         output reg has_imm12,
488         output reg has_we,
489         output reg has_br,
490         output reg has_mem,
491         output reg has_mem_we,
492         output reg has_cmp,
493
494         output halt,
495         output intr,
496
497         output reg has_lwex,
498         output reg has_swex
499
500         // TODO: Use to identify bad instruction and
501         // raise exceptions
502         //, output is_bad
503     );
504     assign opcode = instr[15:11];
505     assign rd = instr[10:8];
506     assign ra = instr[7:5];
507     assign imm4 = instr[3:0];
508     assign imm8 = instr[7:0];
509     assign imm12 = instr[11:0];
510     assign simm5 = instr[4:0];
511
512     // exme_op
513     always @(*) case (opcode)
514         `VMICRO16_OP_SPCL: casez(instr[11:0])
515             `VMICRO16_OP_SPCL_NOP:
516             `VMICRO16_OP_SPCL_HALT,
517             `VMICRO16_OP_SPCL_INTR: alu_op = `VMICRO16_ALU_NOP;
518             default: alu_op = `VMICRO16_ALU_NOP; endcase
519
520         `VMICRO16_OP_LW: alu_op = `VMICRO16_ALU_LW;
521         `VMICRO16_OP_SW: alu_op = `VMICRO16_ALU_SW;
522         `VMICRO16_OP_LWEX: alu_op = `VMICRO16_ALU_LW;
523         `VMICRO16_OP_SWEX: alu_op = `VMICRO16_ALU_SW;
524
525         `VMICRO16_OP_MOV: alu_op = `VMICRO16_ALU_MOV;
526         `VMICRO16_OP_MOVI: alu_op = `VMICRO16_ALU_MOVI;
527
528         `VMICRO16_OP_BR: alu_op = `VMICRO16_ALU_BR;
529         `VMICRO16_OP_MULT: alu_op = `VMICRO16_ALU_MULT;
530
531         `VMICRO16_OP_CMP: alu_op = `VMICRO16_ALU_CMP;
532         `VMICRO16_OP_SETC: alu_op = `VMICRO16_ALU_SETC;
533
534         `VMICRO16_OP_BIT: casez (simm5)
535             `VMICRO16_OP_BIT_OR: alu_op = `VMICRO16_ALU_BIT_OR;
536             `VMICRO16_OP_BIT_XOR: alu_op = `VMICRO16_ALU_BIT_XOR;
537             `VMICRO16_OP_BIT_AND: alu_op = `VMICRO16_ALU_BIT_AND;
538             `VMICRO16_OP_BIT_NOT: alu_op = `VMICRO16_ALU_BIT_NOT;
539             `VMICRO16_OP_BIT_LSHFT: alu_op = `VMICRO16_ALU_BIT_LSHFT;
540             `VMICRO16_OP_BIT_RSHFT: alu_op = `VMICRO16_ALU_BIT_RSHFT;
541             default: alu_op = `VMICRO16_ALU_BAD; endcase
542
543         `VMICRO16_OP_ARITH_U: casez (simm5)
544             `VMICRO16_OP_ARITH_UADD: alu_op = `VMICRO16_ALU_ARITH_UADD;
545             `VMICRO16_OP_ARITH_USUB: alu_op = `VMICRO16_ALU_ARITH_USUB;

```

```

546         `VMICRO16_OP_ARITH_UADDI: alu_op = `VMICRO16_ALU_ARITH_UADDI;
547     default: alu_op = `VMICRO16_ALU_BAD; endcase
548
549     `VMICRO16_OP_ARITH_S: casez (simm5)
550     `VMICRO16_OP_ARITH_SADD: alu_op = `VMICRO16_ALU_ARITH_SADD;
551     `VMICRO16_OP_ARITH_SSUB: alu_op = `VMICRO16_ALU_ARITH_SSUB;
552     `VMICRO16_OP_ARITH_SSUBI: alu_op = `VMICRO16_ALU_ARITH_SSUBI;
553     default: alu_op = `VMICRO16_ALU_BAD; endcase
554
555     default: begin
556         alu_op = `VMICRO16_ALU_NOP;
557         $display($time, "\tDEC: unknown opcode: %h ... NOPPING", opcode);
558     end
559 endcase
560
561 // Special opcodes
562 //assign nop == ((opcode == `VMICRO16_OP_SPCL) & (~instr[0]));
563 assign halt = ((opcode == `VMICRO16_OP_SPCL) & instr[0]);
564 assign intr = ((opcode == `VMICRO16_OP_SPCL) & instr[1]);
565
566 // Register writes
567 always @(*) case (opcode)
568     `VMICRO16_OP_LWEX,
569     `VMICRO16_OP_SWEX,
570     `VMICRO16_OP_LW,
571     `VMICRO16_OP_MUV,
572     `VMICRO16_OP_MUVI,
573     `VMICRO16_OP_MUVI_L,
574     `VMICRO16_OP_ARITH_U,
575     `VMICRO16_OP_ARITH_S,
576     `VMICRO16_OP_SETC,
577     `VMICRO16_OP_BIT,
578     `VMICRO16_OP_MULT: has_we = 1'b1;
579     default: has_we = 1'b0;
580 endcase
581
582 // Contains 4-bit immediate
583 always @(*)
584     if( ((opcode == `VMICRO16_OP_ARITH_U) && (simm5[4] == 0)) ||
585         ((opcode == `VMICRO16_OP_ARITH_S) && (simm5[4] == 0)) )
586         has_imm4 = 1'b1;
587     else
588         has_imm4 = 1'b0;
589
590 // Contains 8-bit immediate
591 always @(*) case (opcode)
592     `VMICRO16_OP_MUVI,
593     `VMICRO16_OP_BR: has_imm8 = 1'b1;
594     default: has_imm8 = 1'b0;
595 endcase
596
597 // Contains 12-bit immediate
598 //always @(*) case (opcode)
599 // `VMICRO16_OP_MUVI_L: has_imm12 = 1'b1;
600 // default: has_imm12 = 1'b0;
601 //endcase
602
603 // Will branch the pc
604 always @(*) case (opcode)
605     `VMICRO16_OP_BR: has_br = 1'b1;
606     default: has_br = 1'b0;
607 endcase
608
609 // Requires external memory
610 always @(*) case (opcode)
611     `VMICRO16_OP_LW,
612     `VMICRO16_OP_SW,
613     `VMICRO16_OP_LWEX,
614     `VMICRO16_OP_SWEX: has_mem = 1'b1;
615     default: has_mem = 1'b0;
616 endcase
617
618 // Requires external memory write
619 always @(*) case (opcode)
620     `VMICRO16_OP_SW,
621     `VMICRO16_OP_SWEX: has_mem_we = 1'b1;
622     default: has_mem_we = 1'b0;
623 endcase
624
625 // Affects status registers (cmp instructions)
626 always @(*) case (opcode)
627     `VMICRO16_OP_CMP: has_cmp = 1'b1;
628     default: has_cmp = 1'b0;
629 endcase
630
631 // Performs exclusive checks
632 always @(*) case (opcode)
633     `VMICRO16_OP_LWEX: has_lwex = 1'b1;
634     default: has_lwex = 1'b0;
635 endcase
636
637 always @(*) case (opcode)
638     `VMICRO16_OP_SWEX: has_swex = 1'b1;
639     default: has_swex = 1'b0;
640 endcase
641 endmodule
642
643 module vmicro16_alu # (
644     parameter OP_WIDTH = 5,
645     parameter DATA_WIDTH = 16,
646     parameter CORE_ID = 0
647 ) (
648     // input clk, // TODO: make clocked
649
650     input [OP_WIDTH-1:0] op,
651     input [DATA_WIDTH-1:0] a, // rs1/dst
652     input [DATA_WIDTH-1:0] b, // rs2
653     input [3:0] flags,
654     output reg [DATA_WIDTH-1:0] c
655 );
656
657 localparam TOP_BIT = (DATA_WIDTH-1);
658 // 17-bit register
659 reg [DATA_WIDTH:0] cmp_tmp = 0; // {carry, [15:0]}
660 wire r_setc;
661

```

```

662 always @(*) begin
663     cmp_tmp = 0;
664     case (op)
665         // branch/nop, output nothing
666         `VMICRO16_ALU_BR,
667         `VMICRO16_ALU_NOP:    c = {DATA_WIDTH{1'b0}};
668         // load/store addresses (use value in rd2)
669         `VMICRO16_ALU_LW,
670         `VMICRO16_ALU_SW:    c = b;
671         // bitwise operations
672         `VMICRO16_ALU_BIT_OR:    c = a | b;
673         `VMICRO16_ALU_BIT_XOR:    c = a ^ b;
674         `VMICRO16_ALU_BIT_AND:    c = a & b;
675         `VMICRO16_ALU_BIT_NOT:    c = ~(b);
676         `VMICRO16_ALU_BIT_LSHFT:    c = a << b;
677         `VMICRO16_ALU_BIT_RSHFT:    c = a >> b;
678
679         `VMICRO16_ALU_MOV:    c = b;
680         `VMICRO16_ALU_MOVI:    c = b;
681         `VMICRO16_ALU_MOVI_L:    c = b;
682
683         `VMICRO16_ALU_ARITH_UADD:    c = a + b;
684         `VMICRO16_ALU_ARITH_USUB:    c = a - b;
685         // TODO: ALU should have simm5 as input
686         `VMICRO16_ALU_ARITH_UADDI:    c = a + b;
687
688         `ifdef DEF_ALU_HW_MULT
689             `VMICRO16_ALU_MULT:    c = a * b;
690         `endif
691
692         `VMICRO16_ALU_ARITH_SADD:    c = $signed(a) + $signed(b);
693         `VMICRO16_ALU_ARITH_SSUB:    c = $signed(a) - $signed(b);
694         // TODO: ALU should have simm5 as input
695         `VMICRO16_ALU_ARITH_SSUBI:    c = $signed(a) - $signed(b);
696
697         `VMICRO16_ALU_CMP: begin
698             // TODO: Do a-b in 17-bit register
699             // Set zero, overflow, carry, signed bits in result
700             cmp_tmp = a - b;
701             c = 0;
702
703             // N Negative condition code flag
704             // Z Zero condition code flag
705             // C Carry condition code flag
706             // V Overflow condition code flag
707             c[`VMICRO16_SFLAG_N] = cmp_tmp[TOP_BIT];
708             c[`VMICRO16_SFLAG_Z] = (cmp_tmp == 0);
709             c[`VMICRO16_SFLAG_C] = 0; //cmp_tmp[TOP_BIT+1]; // not used
710
711             // Overflow flag
712             // https://stackoverflow.com/questions/30957188/
713             // https://github.com/bendl/prco304/blob/master/prco_core/rtl/prco_alu.v#L50
714             case (cmp_tmp[TOP_BIT+1:TOP_BIT])
715                 2'b01: c[`VMICRO16_SFLAG_V] = 1;
716                 2'b10: c[`VMICRO16_SFLAG_V] = 1;
717                 default: c[`VMICRO16_SFLAG_V] = 0;
718             endcase
719
720             $display($time, "\tC%02h: ALU CMP: %h %h = %h = %b", CORE_ID, a, b, cmp_tmp, c[3:0]);
721         end
722
723         `VMICRO16_ALU_SETC: c = { {15{1'b0}}, r_setc };
724
725         // TODO: Parameterise
726         default: begin
727             $display($time, "\tALU: unknown op: %h", op);
728             c = 0;
729             cmp_tmp = 0;
730         end
731     endcase
732 end
733
734 branch setc_check (
735     .flags      (flags),
736     .cond       (b[7:0]),
737     .en         (r_setc)
738 );
739 endmodule
740
741 // flags = 4 bit r_cmp_flags register
742 // cond = 8 bit VMICRO16_OP_BR_? value. See vmicro16_isa.v
743 module branch (
744     input [3:0] flags,
745     input [7:0] cond,
746     output reg en
747 );
748     always @(*)
749     case (cond)
750         `VMICRO16_OP_BR_U:    en = 1; `VMICRO16_OP_BR_U:    en = 1;
751         `VMICRO16_OP_BR_E:    en = (flags[`VMICRO16_SFLAG_Z] == 1);
752         `VMICRO16_OP_BR_NE:    en = (flags[`VMICRO16_SFLAG_Z] == 0);
753         `VMICRO16_OP_BR_G:    en = (flags[`VMICRO16_SFLAG_Z] == 0) &&
754             (flags[`VMICRO16_SFLAG_N] == flags[`VMICRO16_SFLAG_V]);
755         `VMICRO16_OP_BR_L:    en = (flags[`VMICRO16_SFLAG_Z] != flags[`VMICRO16_SFLAG_N]);
756         `VMICRO16_OP_BR_GE:    en = (flags[`VMICRO16_SFLAG_Z] == flags[`VMICRO16_SFLAG_N]);
757         `VMICRO16_OP_BR_LE:    en = (flags[`VMICRO16_SFLAG_Z] == 1) ||
758             (flags[`VMICRO16_SFLAG_N] != flags[`VMICRO16_SFLAG_V]);
759         default:    en = 0;
760     endcase
761 endmodule
762
763
764
765 module vmicro16_core # (
766     parameter DATA_WIDTH    = 16,
767     parameter MEM_INSTR_DEPTH = 64,
768     parameter MEM_SCRATCH_DEPTH = 64,
769     parameter MEM_WIDTH      = 16,
770
771     parameter CORE_ID        = 3'h0
772 ) (
773     input    clk,
774     input    reset,
775
776     output [7:0] dbug,
777

```

```

778     output        halt,
779
780     // interrupt sources
781     input  [^DEF_NUM_INT-1:0]      ints,
782     input  [^DEF_NUM_INT*DATA_WIDTH-1:0] ints_data,
783     output [^DEF_NUM_INT-1:0]      ints_ack,
784
785     // APB master to slave interface (apb_intercon)
786     output [^APB_WIDTH-1:0]        w_PADDR,
787     output                                w_PWRITE,
788     output                                w_PSELx,
789     output                                w_PENABLE,
790     output [DATA_WIDTH-1:0]        w_PWDATA,
791     input  [DATA_WIDTH-1:0]        w_PRDATA,
792     input                                w_PREADY
793
794 `ifndef DEF_CORE_HAS_INSTR_MEM
795     , // APB master interface to slave instruction memory
796     output reg [^APB_WIDTH-1:0]    w2_PADDR,
797     output reg                                w2_PWRITE,
798     output reg                                w2_PSELx,
799     output reg                                w2_PENABLE,
800     output reg [DATA_WIDTH-1:0]    w2_PWDATA,
801     input  [DATA_WIDTH-1:0]        w2_PRDATA,
802     input                                w2_PREADY
803 `endif
804 );
805
806 localparam STATE_IF = 0;
807 localparam STATE_R1 = 1;
808 localparam STATE_R2 = 2;
809 localparam STATE_ME = 3;
810 localparam STATE_WB = 4;
811 localparam STATE_FE = 5;
812 localparam STATE_IDLE = 6;
813 localparam STATE_HALT = 7;
814 reg [2:0] r_state = STATE_IF;
815
816 reg [DATA_WIDTH-1:0] r_pc      = 16'h0000;
817 reg [DATA_WIDTH-1:0] r_pc_saved = 16'h0000;
818 reg [DATA_WIDTH-1:0] r_instr   = 16'h0000;
819 wire [DATA_WIDTH-1:0] w_mem_instr_out;
820 wire                    w_halt;
821
822 assign debug = {7'h00, w_halt};
823 assign halt = w_halt;
824
825 wire [4:0] r_instr_opcode;
826 wire [4:0] r_instr_alu_op;
827 wire [2:0] r_instr_rsd;
828 wire [2:0] r_instr_rsa;
829 reg [DATA_WIDTH-1:0] r_instr_rdd = 0;
830 reg [DATA_WIDTH-1:0] r_instr_rda = 0;
831 wire [3:0] r_instr_imm4;
832 wire [7:0] r_instr_imm8;
833 wire [4:0] r_instr_simm5;
834 wire r_instr_has_imm4;
835 wire r_instr_has_imm8;
836 wire r_instr_has_we;
837 wire r_instr_has_br;
838 wire r_instr_has_cmp;
839 wire r_instr_has_mem;
840 wire r_instr_has_mem_we;
841 wire r_instr_halt;
842 wire r_instr_has_lwex;
843 wire r_instr_has_swex;
844
845 wire [DATA_WIDTH-1:0] r_alu_out;
846
847 wire [DATA_WIDTH-1:0] r_mem_scratch_addr = $signed(r_alu_out) + $signed(r_instr_simm5);
848 wire [DATA_WIDTH-1:0] r_mem_scratch_in  = r_instr_rdd;
849 wire [DATA_WIDTH-1:0] r_mem_scratch_out;
850 wire r_mem_scratch_we = r_instr_has_mem_we && (r_state == STATE_ME);
851 reg r_mem_scratch_req = 0;
852 wire r_mem_scratch_busy;
853
854 reg [2:0] r_reg_rs1 = 0;
855 wire [DATA_WIDTH-1:0] r_reg_rd1_s;
856 wire [DATA_WIDTH-1:0] r_reg_rd1_i;
857 wire [DATA_WIDTH-1:0] r_reg_rd1 = regs_use_int ? r_reg_rd1_i : r_reg_rd1_s;
858 //wire [15:0] r_reg_rd2;
859 wire [DATA_WIDTH-1:0] r_reg_wd = (r_instr_has_mem) ? r_mem_scratch_out : r_alu_out;
860 wire r_reg_we = r_instr_has_we && (r_state == STATE_WB);
861
862 // branching
863 wire w_intr;
864 wire w_branch_en;
865 wire w_branching = r_instr_has_br && w_branch_en;
866 reg [3:0] r_cmp_flags = 4'h00; // N, Z, C, V
867
868 always @(r_cmp_flags)
869     $display($time, "\tC%02h:\tALU CMP: %b", CORE_ID, r_cmp_flags);
870
871 // 2 cycle register fetch
872 always @(*) begin
873     r_reg_rs1 = 0;
874     if (r_state == STATE_R1)
875         r_reg_rs1 = r_instr_rsd;
876     else if (r_state == STATE_R2)
877         r_reg_rs1 = r_instr_rsa;
878     else
879         r_reg_rs1 = 3'h0;
880 end
881
882 reg regs_use_int = 0;
883 `ifdef DEF_ENABLE_INT
884 wire [^DEF_NUM_INT*DATA_WIDTH-1:0] ints_vector;
885 wire [^DEF_NUM_INT-1:0] ints_mask;
886 wire has_int = ints & ints_mask;
887
888 reg int_pending = 0;
889 reg int_pending_ack = 0;
890 always @(posedge clk)
891     if (int_pending_ack)
892         // We've now branched to the isr
893         int_pending <= 0;
894     else if (has_int)
895         // Notify fsm to switch to the ints_vector at the last stage

```

```

894     int_pending <= 1;
895     else if (w_intr)
896         // Return to Interrupt instruction called,
897         // so we've finished with the interrupt
898         int_pending <= 0;
899     `endif
900
901     // Next program counter logic
902     reg [`DATA_WIDTH-1:0] next_pc = 0;
903     always @(posedge clk)
904         if (reset)
905             r_pc <= 0;
906         else if (r_state == STATE_WB) begin
907             `ifdef DEF_ENABLE_INT
908                 if (int_pending) begin
909                     $display($time, "\tC%02h: Jumping to ISR: %h",
910                         CORE_ID,
911                         ints_vector[0 +: `DATA_WIDTH]);
912                     // TODO: check bounds
913                     // Save state
914                     r_pc_saved <= r_pc + 1;
915                     regs_use_int <= 1;
916                     int_pending_ack <= 1;
917                     // Jump to ISR
918                     r_pc <= ints_vector[0 +: `DATA_WIDTH];
919                 end else if (w_intr) begin
920                     $display($time, "\tC%02h: Returning from ISR: %h",
921                         CORE_ID, r_pc_saved);
922
923                     // Restore state
924                     r_pc <= r_pc_saved;
925                     regs_use_int <= 0;
926                     int_pending_ack <= 0;
927                 end else
928                     `endif
929                 if (w_branching) begin
930                     $display($time, "\tC%02h: branching to %h", CORE_ID, r_instr_rdd);
931                     r_pc <= r_instr_rdd;
932
933                     `ifdef DEF_ENABLE_INT
934                         int_pending_ack <= 0;
935                     `endif
936                 end else if (r_pc < (MEM_INSTR_DEPTH-1)) begin
937                     // normal increment
938                     // pc <= pc + 1
939                     r_pc <= r_pc + 1;
940
941                     `ifdef DEF_ENABLE_INT
942                         int_pending_ack <= 0;
943                     `endif
944                 end
945             end // end r_state == STATE_WB
946         else if (r_state == STATE_HALT) begin
947             `ifdef DEF_ENABLE_INT
948                 // Only an interrupt can return from halt
949                 // duplicate code form STATE_ME!
950                 if (int_pending) begin
951                     $display($time, "\tC%02h: Jumping to ISR: %h", CORE_ID, ints_vector[0 +: `DATA_WIDTH]);
952                     // TODO: check bounds
953                     // Save state
954                     r_pc_saved <= r_pc; // + 1; HALT = stay with same PC
955                     regs_use_int <= 1;
956                     int_pending_ack <= 1;
957                     // Jump to ISR
958                     r_pc <= ints_vector[0 +: `DATA_WIDTH];
959                 end else if (w_intr) begin
960                     $display($time, "\tC%02h: Returning from ISR: %h", CORE_ID, r_pc_saved);
961                     r_pc <= r_pc_saved;
962                     regs_use_int <= 0;
963                     int_pending_ack <= 0;
964                 end
965             `endif
966         end
967
968     `ifdef DEF_CORE_HAS_INSTR_MEM
969         initial w2_PSELx = 0;
970         initial w2_PENABLE = 0;
971         initial w2_PADDR = 0;
972     `endif
973
974     // cpu state machine
975     always @(posedge clk)
976         if (reset) begin
977             r_state <= STATE_IF;
978             r_instr <= 0;
979             r_mem_scratch_req <= 0;
980             r_instr_rdd <= 0;
981             r_instr_rda <= 0;
982         end
983         else begin
984
985             `ifdef DEF_CORE_HAS_INSTR_MEM
986                 if (r_state == STATE_IF) begin
987                     r_instr <= w_mem_instr_out;
988
989                     $display("");
990                     $display($time, "\tC%02h: PC: %h", CORE_ID, r_pc);
991                     $display($time, "\tC%02h: INSTR: %h", CORE_ID, w_mem_instr_out);
992
993                     r_state <= STATE_R1;
994                 end
995             `else
996                 // wait for global instruction rom to give us our instruction
997                 if (r_state == STATE_IF) begin
998                     // wait for ready signal
999                     if (!w2_PREADY) begin
1000                         w2_PSELx <= 1;
1001                         w2_PWRITE <= 0;
1002                         w2_PENABLE <= 1;
1003                         w2_PWDATA <= 0;
1004                         w2_PADDR <= r_pc;
1005                     end else begin
1006                         w2_PSELx <= 0;
1007                         w2_PWRITE <= 0;
1008                         w2_PENABLE <= 0;
1009                         w2_PWDATA <= 0;

```

```

1010
1011         r_instr <= w2_PRDATA;
1012
1013         $display("");
1014         $display($time, "\tC%02h: PC: %h", CORE_ID, r_pc);
1015         $display($time, "\tC%02h: INSTR: %h", CORE_ID, w2_PRDATA);
1016
1017         r_state <= STATE_R1;
1018     end
1019 end
1020 `endif
1021
1022     else if (r_state == STATE_R1) begin
1023         if (w_halt) begin
1024             $display("");
1025             $display("");
1026             $display($time, "\tC%02h: PC: %h HALT", CORE_ID, r_pc);
1027             r_state <= STATE_HALT;
1028         end else begin
1029             // primary operand
1030             r_instr_rdd <= r_reg_rdl;
1031             r_state <= STATE_R2;
1032         end
1033     end
1034     else if (r_state == STATE_R2) begin
1035         // Choose secondary operand (register or immediate)
1036         if (r_instr_has_imm8) r_instr_rda <= r_instr_imm8;
1037         else if (r_instr_has_imm4) r_instr_rda <= r_reg_rdl + r_instr_imm4;
1038         else r_instr_rda <= r_reg_rdl;
1039
1040         if (r_instr_has_mem) begin
1041             r_state <= STATE_ME;
1042             // Pulse req
1043             r_mem_scratch_req <= 1;
1044         end else
1045             r_state <= STATE_WB;
1046     end
1047     else if (r_state == STATE_ME) begin
1048         // Pulse req
1049         r_mem_scratch_req <= 0;
1050         // Wait for MMU to finish
1051         if (!r_mem_scratch_busy)
1052             r_state <= STATE_WB;
1053     end
1054     else if (r_state == STATE_WB) begin
1055         if (r_instr_has_cmp) begin
1056             $display($time, "\tC%02h: CMP: %h", CORE_ID, r_alu_out[3:0]);
1057             r_cmp_flags <= r_alu_out[3:0];
1058         end
1059
1060         r_state <= STATE_FE;
1061     end
1062     else if (r_state == STATE_FE)
1063         r_state <= STATE_IF;
1064     else if (r_state == STATE_HALT) begin
1065         `ifdef DEF_ENABLE_INT
1066             if (int_pending) begin
1067                 r_state <= STATE_FE;
1068             end
1069         `endif
1070     end
1071 end
1072
1073 `ifdef DEF_CORE_HAS_INSTR_MEM
1074 // Instruction ROM
1075 (* rom_style = "distributed" *)
1076 vmicro16_bram # (
1077     .MEM_WIDTH      (DATA_WIDTH),
1078     .MEM_DEPTH      (MEM_INSTR_DEPTH),
1079     .CORE_ID        (CORE_ID),
1080     .USE_INITS      (1),
1081     .NAME            ("INSTR_MEM")
1082 ) mem_instr (
1083     .clk            (clk),
1084     .reset          (reset),
1085     // port 1
1086     .mem_addr       (r_pc),
1087     .mem_in         (0),
1088     .mem_we         (1'b0), // ROM
1089     .mem_out        (w_mem_instr_out)
1090 );
1091 `endif
1092
1093 // MMU
1094 vmicro16_core_mmu # (
1095     .MEM_WIDTH      (DATA_WIDTH),
1096     .MEM_DEPTH      (MEM_SCRATCH_DEPTH),
1097     .CORE_ID        (CORE_ID)
1098 ) mmu (
1099     .clk            (clk),
1100     .reset          (reset),
1101     .req            (r_mem_scratch_req),
1102     .busy           (r_mem_scratch_busy),
1103     // interrupts
1104     .ints_vector    (ints_vector),
1105     .ints_mask      (ints_mask),
1106     // port 1
1107     .mmu_addr       (r_mem_scratch_addr),
1108     .mmu_in         (r_mem_scratch_in),
1109     .mmu_we         (r_mem_scratch_we),
1110     .mmu_lwex       (r_instr_has_lwex),
1111     .mmu_swex       (r_instr_has_swex),
1112     .mmu_out        (r_mem_scratch_out),
1113     // APB master to slave
1114     .M_PADDR        (w_PADDR),
1115     .M_PWRITE       (w_PWRITE),
1116     .M_PSELx        (w_PSELx),
1117     .M_PENABLE      (w_PENABLE),
1118     .M_PWDATA       (w_PWDATA),
1119     .M_PRDATA       (w_PRDATA),
1120     .M_PREADY       (w_PREADY)
1121 );
1122
1123 // Instruction decoder
1124 vmicro16_dec dec (
1125     // input

```



```

1126     .instr          (r_instr),
1127     // output async
1128     .opcode         (),
1129     .rd             (r_instr_rsd),
1130     .ra             (r_instr_rsa),
1131     .imm4           (r_instr_imm4),
1132     .imm8           (r_instr_imm8),
1133     .imm12          (),
1134     .simm5          (r_instr_simm5),
1135     .alu_op         (r_instr_alu_op),
1136     .has_imm4       (r_instr_has_imm4),
1137     .has_imm8       (r_instr_has_imm8),
1138     .has_we         (r_instr_has_we),
1139     .has_br         (r_instr_has_br),
1140     .has_cmp        (r_instr_has_cmp),
1141     .has_mem        (r_instr_has_mem),
1142     .has_mem_we     (r_instr_has_mem_we),
1143     .halt           (w_halt),
1144     .intr           (w_intr),
1145     .has_lwex       (r_instr_has_lwex),
1146     .has_swex       (r_instr_has_swex)
1147 );
1148
1149 // Software registers
1150 vmicro16_regs # (
1151     .CORE_ID      (CORE_ID),
1152     .CELL_WIDTH   (`DATA_WIDTH)
1153 ) regs (
1154     .clk          (clk),
1155     .reset        (reset),
1156     // async port 0
1157     .rs1          (r_reg_rs1),
1158     .rd1          (r_reg_rd1_s),
1159     // async port 1
1160     //.rs2         (),
1161     //.rd2         (),
1162     // write port
1163     .we           (r_reg_we && ~regs_use_int),
1164     .ws1          (r_instr_rsd),
1165     .wd           (r_reg_wd)
1166 );
1167
1168 // Interrupt replacement registers
1169 `ifdef DEF_ENABLE_INT
1170 vmicro16_regs # (
1171     .CORE_ID      (CORE_ID),
1172     .CELL_WIDTH   (`DATA_WIDTH),
1173     .DEBUG_NAME   ("REGSINT")
1174 ) regs_intr (
1175     .clk          (clk),
1176     .reset        (reset),
1177     // async port 0
1178     .rs1          (r_reg_rs1),
1179     .rd1          (r_reg_rd1_i),
1180     // async port 1
1181     //.rs2         (),
1182     //.rd2         (),
1183     // write port
1184     .we           (r_reg_we && regs_use_int),
1185     .ws1          (r_instr_rsd),
1186     .wd           (r_reg_wd)
1187 );
1188 `endif
1189
1190 // ALU
1191 vmicro16_alu # (
1192     .CORE_ID(CORE_ID)
1193 ) alu (
1194     .op           (r_instr_alu_op),
1195     .a            (r_instr_rdd),
1196     .b            (r_instr_rda),
1197     .flags        (r_cmp_flags),
1198     // async output
1199     .c            (r_alu_out)
1200 );
1201
1202 branch branch_check (
1203     .flags        (r_cmp_flags),
1204     .cond         (r_instr_imm8),
1205     .en           (w_branch_en)
1206 );
1207
1208 endmodule

```

vmicro16_soc.v

```

1  //
2  //
3
4  `include "vmicro16_soc_config.v"
5  `include "clog2.v"
6  `include "formal.v"
7
8  module pow_reset # (
9      parameter INIT = 1,
10     parameter N     = 8
11 ) (
12     input    clk,
13     input    reset,
14     output reg resethold
15 );
16     initial resethold = INIT ? (N-1) : 0;
17
18     always @(*)
19         resethold = |hold;
20
21     reg [`clog2(N)-1:0] hold = (N-1);
22     always @(posedge clk)
23         if (reset)
24             hold <= N-1;

```

```

25         else
26             if (hold)
27                 hold <= hold - 1;
28     endmodule
29
30     // Vmicro16 multi-core SoC with various peripherals
31     // and interrupts
32     module vmicro16_soc (
33         input clk,
34         input reset,
35
36         // UART0
37         input          uart_rx,
38         output         uart_tx,
39         //
40         output [`APB_GPIO0_PINS-1:0] gpio0,
41         output [`APB_GPIO1_PINS-1:0] gpio1,
42         output [`APB_GPIO2_PINS-1:0] gpio2,
43         //
44         output          halt,
45         //
46         output          [`CORES-1:0] dbug0,
47         output          [`CORES*8-1:0] dbug1
48     );
49     wire [`CORES-1:0] w_halt;
50     assign halt = &w_halt;
51
52     assign dbug0 = w_halt;
53
54     // Watchdog reset pulse signal.
55     // Passed to pow_reset to generate a longer reset pulse
56     wire wreset;
57     wire prog_prog;
58
59     // soft register reset hold for brams and registers
60     wire soft_reset;
61     `ifndef DEF_GLOBAL_RESET
62         pow_reset # (
63             .INIT      (1),
64             .N          (8)
65         ) por_inst (
66             .clk        (clk),
67             `ifdef DEF_USE_WATCHDOG
68             .reset      (reset | wreset | prog_prog),
69             `else
70             .reset      (reset),
71             `endif
72             .resethold  (soft_reset)
73         );
74     `else
75         assign soft_reset = 0;
76     `endif
77
78     // Peripherals (master to slave)
79     wire [`APB_WIDTH-1:0] M_PADDR;
80     wire                  M_PWRITE;
81     wire [`SLAVES-1:0]    M_PSELx; // not shared
82     wire                  M_PENABLE;
83     wire [`DATA_WIDTH-1:0] M_PWDATA;
84     wire [`SLAVES*DATA_WIDTH-1:0] M_PRDATA; // input to intercon
85     wire [`SLAVES-1:0]    M_PREADY; // input
86
87     // Master apb interfaces
88     wire [`CORES*APB_WIDTH-1:0] w_PADDR;
89     wire [`CORES-1:0]          w_PWRITE;
90     wire [`CORES-1:0]          w_PSELx;
91     wire [`CORES-1:0]          w_PENABLE;
92     wire [`CORES*DATA_WIDTH-1:0] w_PWDATA;
93     wire [`CORES*DATA_WIDTH-1:0] w_PRDATA;
94     wire [`CORES-1:0]          w_PREADY;
95
96     // Interrupts
97     `ifndef DEF_ENABLE_INT
98         wire [`DEF_NUM_INT-1:0] ints;
99         wire [`DEF_NUM_INT*DATA_WIDTH-1:0] ints_data;
100         assign ints[7:1] = 0;
101         assign ints_data[`DEF_NUM_INT*DATA_WIDTH-1:DATA_WIDTH] =
102             {`DEF_NUM_INT*(`DATA_WIDTH-1){1'b0}};
103     `endif
104
105     apb_intercon_s # (
106         .MASTER_PORTS  (`CORES),
107         .SLAVE_PORTS    (`SLAVES),
108         .BUS_WIDTH      (`APB_WIDTH),
109         .DATA_WIDTH     (`DATA_WIDTH),
110         .HAS_PSELX_ADDR (1)
111     ) apb (
112         .clk        (clk),
113         .reset      (soft_reset),
114         // APB master to slave
115         .S_PADDR    (w_PADDR),
116         .S_PWRITE    (w_PWRITE),
117         .S_PSELx     (w_PSELx),
118         .S_PENABLE   (w_PENABLE),
119         .S_PWDATA     (w_PWDATA),
120         .S_PRDATA     (w_PRDATA),
121         .S_PREADY     (w_PREADY),
122         // shared bus
123         .M_PADDR     (M_PADDR),
124         .M_PWRITE     (M_PWRITE),
125         .M_PSELx      (M_PSELx),
126         .M_PENABLE     (M_PENABLE),
127         .M_PWDATA      (M_PWDATA),
128         .M_PRDATA      (M_PRDATA),
129         .M_PREADY      (M_PREADY)
130     );
131
132     `ifndef DEF_USE_WATCHDOG
133         vmicro16_watchdog_apb # (
134             .BUS_WIDTH  (`APB_WIDTH),
135             .NAME        ("WDG0")
136         ) wdog0_apb (
137             .clk        (clk),
138             .reset      (),
139             // apb slave to master interface
140             .S_PADDR    (),

```

```

141     .S_PWRITE      (M_PWRITE),
142     .S_PSELx      (M_PSELx[~APB_PSELX_WD0G0]),
143     .S_PENABLE    (M_PENABLE),
144     .S_PWDATA     (),
145     .S_PRDATA     (),
146     .S_PREADY     (M_PREADY[~APB_PSELX_WD0G0]),
147
148     .wreset       (wreset)
149 );
150 `endif
151
152 vmicro16_gpio_apb # (
153     .BUS_WIDTH    (`APB_WIDTH),
154     .DATA_WIDTH   (`DATA_WIDTH),
155     .PORTS        (`APB_GPIO0_PINS),
156     .NAME         ("GPIO0")
157 ) gpio0_apb (
158     .clk          (clk),
159     .reset        (soft_reset),
160     // apb slave to master interface
161     .S_PADDR      (M_PADDR),
162     .S_PWRITE     (M_PWRITE),
163     .S_PSELx      (M_PSELx[~APB_PSELX_GPIO0]),
164     .S_PENABLE    (M_PENABLE),
165     .S_PWDATA     (M_PWDATA),
166     .S_PRDATA     (M_PRDATA[~APB_PSELX_GPIO0*`DATA_WIDTH +: `DATA_WIDTH]),
167     .S_PREADY     (M_PREADY[~APB_PSELX_GPIO0]),
168     .gpio         (gpio0)
169 );
170
171 // GPIO1 for Seven segment displays (16 pin)
172 vmicro16_gpio_apb # (
173     .BUS_WIDTH    (`APB_WIDTH),
174     .DATA_WIDTH   (`DATA_WIDTH),
175     .PORTS        (`APB_GPIO1_PINS),
176     .NAME         ("GPIO1")
177 ) gpio1_apb (
178     .clk          (clk),
179     .reset        (soft_reset),
180     // apb slave to master interface
181     .S_PADDR      (M_PADDR),
182     .S_PWRITE     (M_PWRITE),
183     .S_PSELx      (M_PSELx[~APB_PSELX_GPIO1]),
184     .S_PENABLE    (M_PENABLE),
185     .S_PWDATA     (M_PWDATA),
186     .S_PRDATA     (M_PRDATA[~APB_PSELX_GPIO1*`DATA_WIDTH +: `DATA_WIDTH]),
187     .S_PREADY     (M_PREADY[~APB_PSELX_GPIO1]),
188     .gpio         (gpio1)
189 );
190
191 // GPIO2 for Seven segment displays (8 pin)
192 vmicro16_gpio_apb # (
193     .BUS_WIDTH    (`APB_WIDTH),
194     .DATA_WIDTH   (`DATA_WIDTH),
195     .PORTS        (`APB_GPIO2_PINS),
196     .NAME         ("GPIO2")
197 ) gpio2_apb (
198     .clk          (clk),
199     .reset        (soft_reset),
200     // apb slave to master interface
201     .S_PADDR      (M_PADDR),
202     .S_PWRITE     (M_PWRITE),
203     .S_PSELx      (M_PSELx[~APB_PSELX_GPIO2]),
204     .S_PENABLE    (M_PENABLE),
205     .S_PWDATA     (M_PWDATA),
206     .S_PRDATA     (M_PRDATA[~APB_PSELX_GPIO2*`DATA_WIDTH +: `DATA_WIDTH]),
207     .S_PREADY     (M_PREADY[~APB_PSELX_GPIO2]),
208     .gpio         (gpio2)
209 );
210
211 apb_uart_tx # (
212     .DATA_WIDTH   (8),
213     .ADDR_EXP     (4) //2**4 = 16 FIFO words
214 ) uart0_apb (
215     .clk          (clk),
216     .reset        (soft_reset),
217     // apb slave to master interface
218     .S_PADDR      (M_PADDR),
219     .S_PWRITE     (M_PWRITE),
220     .S_PSELx      (M_PSELx[~APB_PSELX_UART0]),
221     .S_PENABLE    (M_PENABLE),
222     .S_PWDATA     (M_PWDATA),
223     .S_PRDATA     (M_PRDATA[~APB_PSELX_UART0*`DATA_WIDTH +: `DATA_WIDTH]),
224     .S_PREADY     (M_PREADY[~APB_PSELX_UART0]),
225     // uart wires
226     .tx_wire      (uart_tx),
227     .rx_wire      ()
228 );
229
230 timer_apb timr0 (
231     .clk          (clk),
232     .reset        (soft_reset),
233     // apb slave to master interface
234     .S_PADDR      (M_PADDR),
235     .S_PWRITE     (M_PWRITE),
236     .S_PSELx      (M_PSELx[~APB_PSELX_TIMR0]),
237     .S_PENABLE    (M_PENABLE),
238     .S_PWDATA     (M_PWDATA),
239     .S_PRDATA     (M_PRDATA[~APB_PSELX_TIMR0*`DATA_WIDTH +: `DATA_WIDTH]),
240     .S_PREADY     (M_PREADY[~APB_PSELX_TIMR0])
241     //
242     `ifdef DEF_ENABLE_INT
243     ,.out          (ints [~DEF_INT_TIMR0]),
244     .int_data      (ints_data[~DEF_INT_TIMR0*`DATA_WIDTH +: `DATA_WIDTH])
245     `endif
246 );
247
248 // Shared register set for system-on-chip info
249 // R0 = number of cores
250 vmicro16_regs_apb # (
251     .BUS_WIDTH    (`APB_WIDTH),
252     .DATA_WIDTH   (`DATA_WIDTH),
253     .CELL_DEPTH   (8),
254     .PARAM_DEFAULTS_R0  (`CORES),
255     .PARAM_DEFAULTS_R1  (`SLAVES)
256 ) regs0_apb (

```

```

257     .clk      (clk),
258     .reset    (soft_reset),
259     // apb slave to master interface
260     .S_PADDR  (M_PADDR),
261     .S_PWRITE (M_PWRITE),
262     .S_PSELx  (M_PSELx[APB_PSELX_REGSO]),
263     .S_PENABLE (M_PENABLE),
264     .S_PWDATA  (M_PWDATA),
265     .S_PRDATA  (M_PRDATA[APB_PSELX_REGSO*DATA_WIDTH+: DATA_WIDTH]),
266     .S_PREADY  (M_PREADY[APB_PSELX_REGSO])
267 );
268
269 vmicro16_bram_ex_apb # (
270     .BUS_WIDTH  (`APB_WIDTH),
271     .MEM_WIDTH  (`DATA_WIDTH),
272     .MEM_DEPTH  (`APB_BRAMO_CELLS),
273     .CORE_ID_BITS (`clog2(`CORES))
274 ) bram_apb (
275     .clk      (clk),
276     .reset    (soft_reset),
277     // apb slave to master interface
278     .S_PADDR  (M_PADDR),
279     .S_PWRITE (M_PWRITE),
280     .S_PSELx  (M_PSELx[APB_PSELX_BRAMO]),
281     .S_PENABLE (M_PENABLE),
282     .S_PWDATA  (M_PWDATA),
283     .S_PRDATA  (M_PRDATA[APB_PSELX_BRAMO*DATA_WIDTH+: DATA_WIDTH]),
284     .S_PREADY  (M_PREADY[APB_PSELX_BRAMO])
285 );
286
287 // There must be atleast 1 core
288 `static_assert(`CORES > 0)
289 `static_assert(`DEF_MEM_INSTR_DEPTH > 0)
290 `static_assert(`DEF_MMU_TIMO_CELLS > 0)
291
292 // Single instruction memory
293 `ifndef DEF_CORE_HAS_INSTR_MEM
294 // slave input/outputs from interconnect
295 wire [APB_WIDTH-1:0] instr_m_paddr;
296 wire [1:0] instr_m_pwrite; // not shared
297 wire [1:0] instr_m_pselx;
298 wire [DATA_WIDTH-1:0] instr_m_penable;
299 wire [1*DATA_WIDTH-1:0] instr_m_pwdata; // slave response
300 wire [1:0] instr_m_pready; // slave response
301
302 // Master apb interfaces
303 wire [CORES*APB_WIDTH-1:0] instr_w_paddr;
304 wire [CORES-1:0] instr_w_pwrite;
305 wire [CORES-1:0] instr_w_pselx;
306 wire [CORES-1:0] instr_w_penable;
307 wire [CORES*DATA_WIDTH-1:0] instr_w_pwdata;
308 wire [CORES*DATA_WIDTH-1:0] instr_w_pready;
309
310 `ifdef DEF_USE_REPROG
311 wire [clog2(`DEF_MEM_INSTR_DEPTH)-1:0] prog_addr;
312 wire [DATA_WIDTH-1:0] prog_data;
313 wire prog_we;
314 uart_prog rom_prog (
315     .clk      (clk),
316     .reset    (reset | wdreset),
317     // input stream
318     .uart_rx  (uart_rx),
319     // programmer
320     .addr     (prog_addr),
321     .data     (prog_data),
322     .we       (prog_we),
323     .prog     (prog_prog)
324 );
325 `endif
326
327 `ifndef DEF_USE_REPROG
328 vmicro16_bram_prog_apb
329 `else
330 vmicro16_bram_apb
331 `endif
332 # (
333     .BUS_WIDTH  (`APB_WIDTH),
334     .MEM_WIDTH  (`DATA_WIDTH),
335     .MEM_DEPTH  (`DEF_MEM_INSTR_DEPTH),
336     .USE_INITS  (1),
337     .NAME       ("INSTR_ROM_G")
338 ) instr_rom_apb (
339     .clk      (clk),
340     .reset    (reset),
341     .S_PADDR  (instr_m_paddr),
342     .S_PWRITE (0),
343     .S_PSELx  (instr_m_pselx),
344     .S_PENABLE (instr_m_penable),
345     .S_PWDATA  (0),
346     .S_PRDATA  (instr_m_pready),
347     .S_PREADY  (instr_m_pready)
348
349 `ifdef DEF_USE_REPROG
350     .addr     (prog_addr),
351     .data     (prog_data),
352     .we       (prog_we),
353     .prog     (prog_prog)
354 `endif
355 );
356
357 apb_intercon_s # (
358     .MASTER_PORTS  (`CORES),
359     .SLAVE_PORTS    (1),
360     .BUS_WIDTH      (`APB_WIDTH),
361     .DATA_WIDTH     (`DATA_WIDTH),
362     .HAS_PSELX_ADDR (0)
363 ) apb_instr_intercon (
364     .clk      (clk),
365     .reset    (soft_reset),
366     // APB master from cores
367     // master
368     .S_PADDR  (instr_w_paddr),

```

```

373     .S_PWRITE   (instr_w_PWRITE),
374     .S_PSELx   (instr_w_PSELx),
375     .S_PENABLE (instr_w_PENABLE),
376     .S_PWDATA  (instr_w_PWDATA),
377     .S_PRDATA  (instr_w_PRDATA),
378     .S_PREADY  (instr_w_PREADY),
379     // shared bus slaves
380     // slave outputs
381     .M_PADDR    (instr_M_PADDR),
382     .M_PWRITE   (instr_M_PWRITE),
383     .M_PSELx    (instr_M_PSELx),
384     .M_PENABLE  (instr_M_PENABLE),
385     .M_PWDATA   (instr_M_PWDATA),
386     .M_PRDATA   (instr_M_PRDATA),
387     .M_PREADY   (instr_M_PREADY)
388 );
389 `endif
390
391 genvar i;
392 generate for(i = 0; i < `CORES; i = i + 1) begin : cores
393
394     vmicro16_core # (
395         .CORE_ID      (i),
396         .DATA_WIDTH   (`DATA_WIDTH),
397
398         .MEM_INSTR_DEPTH (`DEF_MEM_INSTR_DEPTH),
399         .MEM_SCRATCH_DEPTH (`DEF_MMU_TIMO_CELLS)
400     ) c1 (
401         .clk      (clk),
402         .reset     (soft_reset),
403
404         // debug
405         .halt      (w_halt[i]),
406
407         // interrupts
408         .ints      (ints),
409         .ints_data  (ints_data),
410
411         // Output master port 1
412         .w_PADDR   (w_PADDR  [`APB_WIDTH*i +: `APB_WIDTH] ),
413         .w_PWRITE  (w_PWRITE [i] ),
414         .w_PSELx   (w_PSELx  [i] ),
415         .w_PENABLE (w_PENABLE [i] ),
416         .w_PWDATA  (w_PWDATA [`DATA_WIDTH*i +: `DATA_WIDTH]),
417         .w_PRDATA  (w_PRDATA  [`DATA_WIDTH*i +: `DATA_WIDTH]),
418         .w_PREADY  (w_PREADY  [i] )
419     )
420 `ifndef DEF_CORE_HAS_INSTR_MEM
421     // APB instruction rom
422     , // Output master port 2
423     .w2_PADDR   (instr_w_PADDR  [`APB_WIDTH*i +: `APB_WIDTH] ),
424     .w2_PWRITE  (instr_w_PWRITE [i] ),
425     .w2_PSELx   (instr_w_PSELx  [i] ),
426     .w2_PENABLE (instr_w_PENABLE [i] ),
427     .w2_PWDATA  (instr_w_PWDATA [`DATA_WIDTH*i +: `DATA_WIDTH]),
428     .w2_PRDATA  (instr_w_PRDATA  [`DATA_WIDTH*i +: `DATA_WIDTH]),
429     .w2_PREADY  (instr_w_PREADY  [i] )
430 `endif
431 );
432 end
433 endgenerate
434
435 ///////////////////////////////////////////////////
436 // Formal Verification
437 ///////////////////////////////////////////////////
438 `ifdef FORMAL
439 wire all_halted = &w_halt;
440 ///////////////////////////////////////////////////
441 // Count number of clocks each core is spending on
442 // bus transactions
443 ///////////////////////////////////////////////////
444 reg [15:0] bus_core_times [0:`CORES-1];
445 reg [15:0] core_work_times [0:`CORES-1];
446 reg [15:0] instr_fetch_times [0:`CORES-1];
447 integer i2;
448 initial
449     for(i2 = 0; i2 < `CORES; i2 = i2 + 1) begin
450         bus_core_times[i2] = 0;
451         core_work_times[i2] = 0;
452     end
453
454 // total bus time
455 generate
456     genvar g2;
457     for (g2 = 0; g2 < `CORES; g2 = g2 + 1) begin : formal_for_times
458         always @(posedge clk) begin
459             if (w_PSELx[g2]) begin
460                 bus_core_times[g2] <= bus_core_times[g2] + 1;
461
462                 // Core working time
463                 `ifndef DEF_CORE_HAS_INSTR_MEM
464                     if (!w_PSELx[g2] && !instr_w_PSELx[g2])
465                         if (!w_PSELx[g2])
466                             if (!w_halt[g2])
467                                 core_work_times[g2] <= core_work_times[g2] + 1;
468                 end
469             end
470         end
471     end
472 endgenerate
473
474 reg [15:0] bus_time_average = 0;
475 reg [15:0] bus_reqs_average = 0;
476 reg [15:0] fetch_time_average = 0;
477 reg [15:0] work_time_average = 0;
478 //
479 always @(all_halted) begin
480     for (i2 = 0; i2 < `CORES; i2 = i2 + 1) begin
481         bus_time_average = bus_time_average + bus_core_times[i2];
482         bus_reqs_average = bus_reqs_average + bus_core_reqs_count[i2];
483         work_time_average = work_time_average + core_work_times[i2];
484         fetch_time_average = fetch_time_average + instr_fetch_times[i2];
485     end
486 end
487
488

```

```

489         bus_time_average = bus_time_average / `CORES;
490         bus_reqs_average = bus_reqs_average / `CORES;
491         work_time_average = work_time_average / `CORES;
492         fetch_time_average = fetch_time_average / `CORES;
493     end
494
495     //////////////////////////////////////////////////
496     // Count number of bus requests per core
497     //////////////////////////////////////////////////
498     // 1 clock delay of w_PSELx
499     reg [`CORES-1:0] bus_core_reqs_last;
500     // rising edges of each
501     wire [`CORES-1:0] bus_core_reqs_real;
502     // storage for counters for each core
503     reg [15:0] bus_core_reqs_count [0:`CORES-1];
504     initial
505         for(i2 = 0; i2 < `CORES; i2 = i2 + 1)
506             bus_core_reqs_count[i2] = 0;
507
508     // 1 clk delay to detect rising edge
509     always @(posedge clk)
510         bus_core_reqs_last <= w_PSELx;
511
512     generate
513         genvar g3;
514         for (g3 = 0; g3 < `CORES; g3 = g3 + 1) begin : formal_for_reqs
515             // Detect new reqs for each core
516             assign bus_core_reqs_real[g3] = w_PSELx[g3] >
517                                     bus_core_reqs_last[g3];
518
519             always @(posedge clk)
520                 if (bus_core_reqs_real[g3])
521                     bus_core_reqs_count[g3] <= bus_core_reqs_count[g3] + 1;
522
523         end
524     endgenerate
525
526     `ifndef DEF_CORE_HAS_INSTR_MEM
527     //////////////////////////////////////////////////
528     // Time waiting for instruction fetches
529     // from global memory
530     //////////////////////////////////////////////////
531     integer i3;
532     initial
533         for(i3 = 0; i3 < `CORES; i3 = i3 + 1)
534             instr_fetch_times[i3] = 0;
535
536     // total bus time
537     // Instruction fetches occur on the w2 master port
538     generate
539         genvar g4;
540         for (g4 = 0; g4 < `CORES; g4 = g4 + 1) begin : formal_for_fetch_times
541             always @(posedge clk)
542                 if (instr_w_PSELx[g4])
543                     instr_fetch_times[g4] <= instr_fetch_times[g4] + 1;
544
545         end
546     endgenerate
547     `endif
548
549     `endif // end FORMAL
550
551 endmodule
552

```

vmicro16_isa.v

```

1 // Vmicro16 multi-core instruction set
2 `include "vmicro16_soc_config.v"
3
4 // TODO: Remove NOP by making a register write/read always 0
5 `define VMICRO16_OP_SPCL      5'b000000
6 `define VMICRO16_OP_LW       5'b000001
7 `define VMICRO16_OP_SW       5'b000010
8 `define VMICRO16_OP_BIT      5'b000011
9 `define VMICRO16_OP_BIT_OR   5'b000000
10 `define VMICRO16_OP_BIT_XOR  5'b000001
11 `define VMICRO16_OP_BIT_AND  5'b000010
12 `define VMICRO16_OP_BIT_NOT  5'b000011
13 `define VMICRO16_OP_BIT_LSHFT 5'b001000
14 `define VMICRO16_OP_BIT_RSHFT 5'b001001
15 `define VMICRO16_OP_MOV      5'b001000
16 `define VMICRO16_OP_MOVI     5'b001001
17 `define VMICRO16_OP_ARITH_U   5'b001100
18 `define VMICRO16_OP_ARITH_UADD 5'b111111
19 `define VMICRO16_OP_ARITH_USUB 5'b100000
20 `define VMICRO16_OP_ARITH_UADDI 5'b0?????
21 `define VMICRO16_OP_ARITH_S   5'b001111
22 `define VMICRO16_OP_ARITH_SADD 5'b111111
23 `define VMICRO16_OP_ARITH_SSUB 5'b100000
24 `define VMICRO16_OP_ARITH_SSUBI 5'b0?????
25 `define VMICRO16_OP_BR       5'b010000
26 `define VMICRO16_OP_CMP      5'b010001
27 `define VMICRO16_OP_SETC     5'b010010
28 `define VMICRO16_OP_MULT     5'b010011
29 `define VMICRO16_OP_LWEX     5'b010101
30 `define VMICRO16_OP_SWEX     5'b010110
31
32 // Special opcodes
33 `define VMICRO16_OP_SPCL_NOP  11'h000
34 `define VMICRO16_OP_SPCL_HALT 11'h001
35 `define VMICRO16_OP_SPCL_INTR 11'h002
36
37 // TODO: wasted upper nibble bits in BR
38 `define VMICRO16_OP_BR_U      8'h00
39 `define VMICRO16_OP_BR_E      8'h01
40 `define VMICRO16_OP_BR_NE     8'h02
41 `define VMICRO16_OP_BR_G      8'h03
42 `define VMICRO16_OP_BR_GE     8'h04
43 `define VMICRO16_OP_BR_L      8'h05

```

```

44  `define VMICRO16_OP_BR_LE      8'h06
45  `define VMICRO16_OP_BR_S      8'h07
46  `define VMICRO16_OP_BR_NS     8'h08
47
48  // flag bit positions
49  `define VMICRO16_SFLAG_N      4'h03
50  `define VMICRO16_SFLAG_Z      4'h02
51  `define VMICRO16_SFLAG_C      4'h01
52  `define VMICRO16_SFLAG_V      4'h00
53
54  // microcode operations
55  `define VMICRO16_ALU_BIT_OR    5'h00
56  `define VMICRO16_ALU_BIT_XOR  5'h01
57  `define VMICRO16_ALU_BIT_AND  5'h02
58  `define VMICRO16_ALU_BIT_NOT  5'h03
59  `define VMICRO16_ALU_BIT_LSHFT 5'h04
60  `define VMICRO16_ALU_BIT_RSHFT 5'h05
61  `define VMICRO16_ALU_LW       5'h06
62  `define VMICRO16_ALU_SW       5'h07
63  `define VMICRO16_ALU_NOP      5'h08
64  `define VMICRO16_ALU_MOV      5'h09
65  `define VMICRO16_ALU_MOVI     5'h0a
66  `define VMICRO16_ALU_MOVI_L   5'h0b
67  `define VMICRO16_ALU_ARITH_UADD 5'h0c
68  `define VMICRO16_ALU_ARITH_USUB 5'h0d
69  `define VMICRO16_ALU_ARITH_SADD 5'h0e
70  `define VMICRO16_ALU_ARITH_SSUB 5'h0f
71  `define VMICRO16_ALU_BR_U      5'h10
72  `define VMICRO16_ALU_BR_E      5'h11
73  `define VMICRO16_ALU_BR_NE     5'h12
74  `define VMICRO16_ALU_BR_G      5'h13
75  `define VMICRO16_ALU_BR_GE     5'h14
76  `define VMICRO16_ALU_BR_L      5'h15
77  `define VMICRO16_ALU_BR_LE     5'h16
78  `define VMICRO16_ALU_BR_S      5'h17
79  `define VMICRO16_ALU_BR_NS     5'h18
80  `define VMICRO16_ALU_CMP       5'h19
81  `define VMICRO16_ALU_SETC      5'h1a
82  `define VMICRO16_ALU_ARITH_UADDI 5'h1b
83  `define VMICRO16_ALU_ARITH_SSUBI 5'h1c
84  `define VMICRO16_ALU_BR        5'h1d
85  `ifdef DEF_ALU_HW_MULT
86  `define VMICRO16_ALU_MULT      5'h1e
87  `endif
88  `define VMICRO16_ALU_BAD       5'h1f

```