

Phase Locked Loop

Ben Lorenzetti

Start Date: April 2, 2015

Submission Date: April 20, 2015

Contents

1	Objective	1
2	Principles of Operation	2
3	Theory	3
3.1	Texas Instruments LM656	3
3.2	Voltage Controlled Oscillator	4
3.3	Signal Multiplier	4
3.4	Frequency Independent Amplifier	4
3.5	Low-Pass Filter	4
4	Design	4
4.1	Voltage Controlled Oscillator	4
4.1.1	LM565 vs SPICE Model	4
4.1.2	Free Running Frequency f_0	4
4.1.3	Oscillator Sensitivity K_O	4
4.2	Signal Multiplier	4
4.3	Frequency Independent Amplifier	4
4.4	Low-Pass Filter	4
4.5	Phase Detector Sensitivity K_D	4
4.6	Loop Gain $K_O K_D$	4
5	Results	4
5.1	Free Running Frequency f_0	4
5.2	Oscillator Sensitivity K_O	4
5.3	FSK Demodulator	4
6	Conclusions	4
7	Appendices	4
7.1	LM565 Schematic Diagram	5

1 Objective

To measure the characteristics of a phase locked loop and to investigate and model its behavior in an FSK (frequency-shift keying) demodulator.

2 Principles of Operation

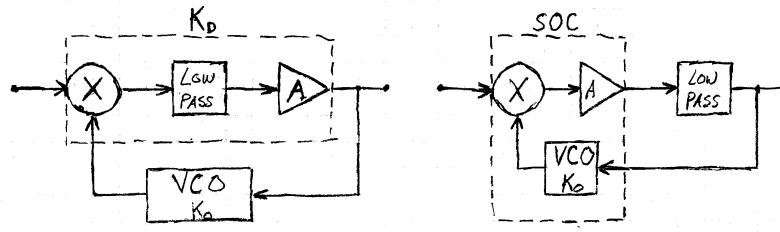


Figure 1: PLL Demodulator Block Diagram (left); Rearrangement for SOC Integration (right).

3 Theory

3.1 Texas Instruments LM656

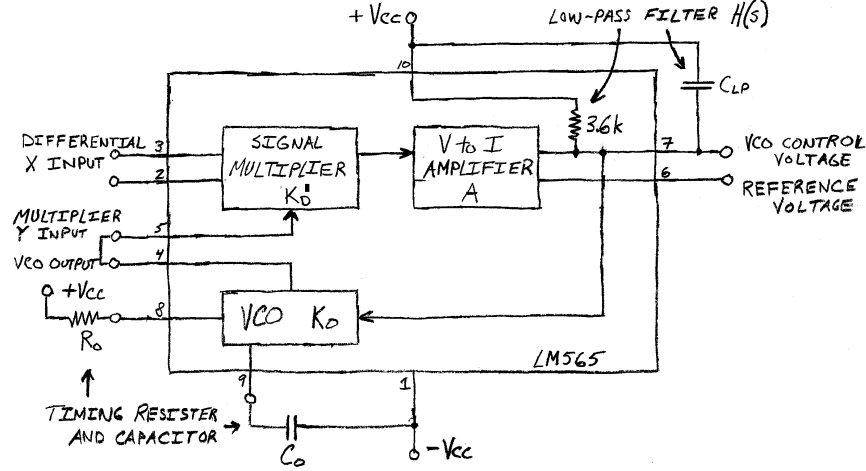


Figure 2: Rough Block Diagram for the LM565 Phase Locked Loop

As discussed in the section 2, most of a PLL demodulator can be integrated onto a single silicon wafer—called a system-on-a-chip (SOC). The only parts that cannot be fabricated on silicon are the capacitors, so the PLL demodulator has been redesigned with this in mind. A company called Signetics was the first to put a PLL on an SOC, but they are out of business so we have a similar chip from Texas Instruments, the LM565.

The LM565 makes it easy for the circuit designer to incorporate frequency-shift keying into their product; you just have to slap on a few capacitors to have a working demodulator. The only design that must be done is selecting a free-running frequency f_0 near the transmission frequency and building a low-pass filter with bandwidth greater than the frequency shift but less than twice the transmission frequency. Use the following equation from the LM565 datasheet to select f_0 .

$$f_0 = \frac{0.3}{R_0 C_0} \quad (1)$$

For more information about how the LM565 works, see the schematic diagram in section 7.1.

3.2 Voltage Controlled Oscillator

3.3 Signal Multiplier

3.4 Frequency Independent Amplifier

3.5 Low-Pass Filter

4 Design and Simulation

4.1 Voltage Controlled Oscillator

4.1.1 LM565 vs SPICE Model

For the timing resistor R_0 , I decided to use $3k\Omega$ because it was available with 1% tolerance. Using equation 1, the timing capacitors for 1 kHz, 10 kHz, and 100 kHz operating ranges should be

$$R_0 = 3k\Omega$$

$$C_{0(1kHz)} = 0.1\mu F \quad | \quad C_{0(10kHz)} = 0.01\mu F \quad | \quad C_{0(100kHz)} = 1,000pF$$

4.1.2 Free Running Frequency f_0

4.1.3 Oscillator Sensitivity K_O

4.2 Signal Multiplier

4.3 Frequency Independent Amplifier

4.4 Low-Pass Filter

4.5 Phase Detector Sensitivity K_D

4.6 Loop Gain $K_O K_D$

5 Results

5.1 Free Running Frequency f_0

5.2 Oscillator Sensitivity K_O

5.3 FSK Demodulator

6 Conclusions

7 Appendices

7.1 LM565 Schematic Diagram

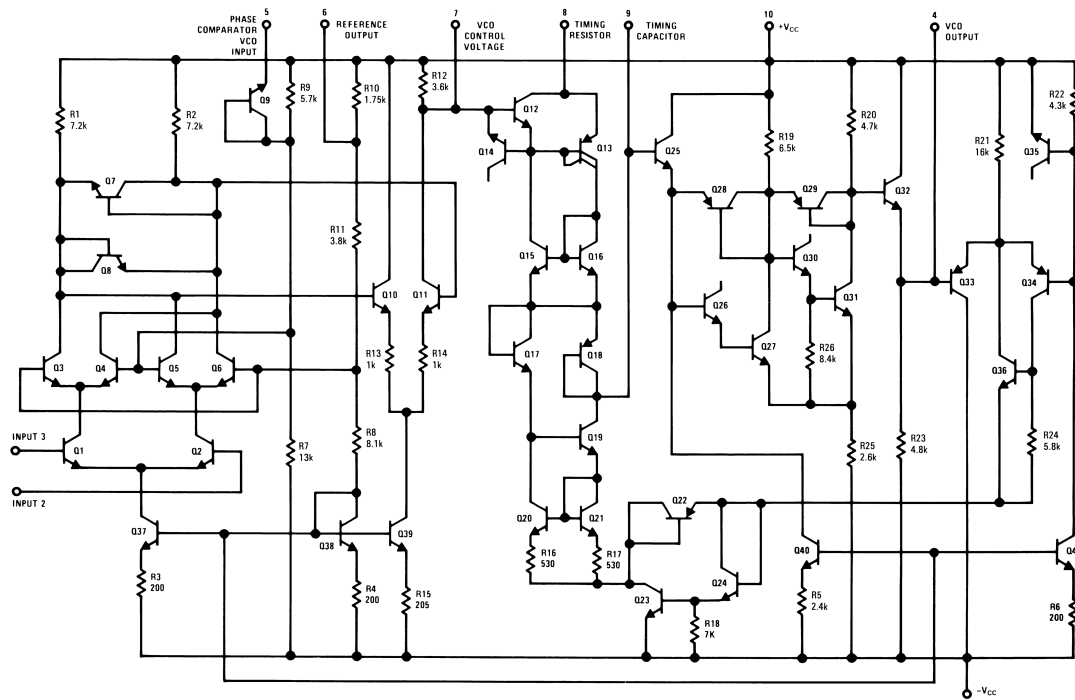


Figure 3: LM565 Schematic Diagram; taken from page 6 of the TI datasheet