

6. Design via Root Locus

Design of cascade compensators using root locus

PI/Lag Compensator

PD/Lead Compensator

PID/Lag-Lead Compensator

Implementation of Controllers

Notch Filter

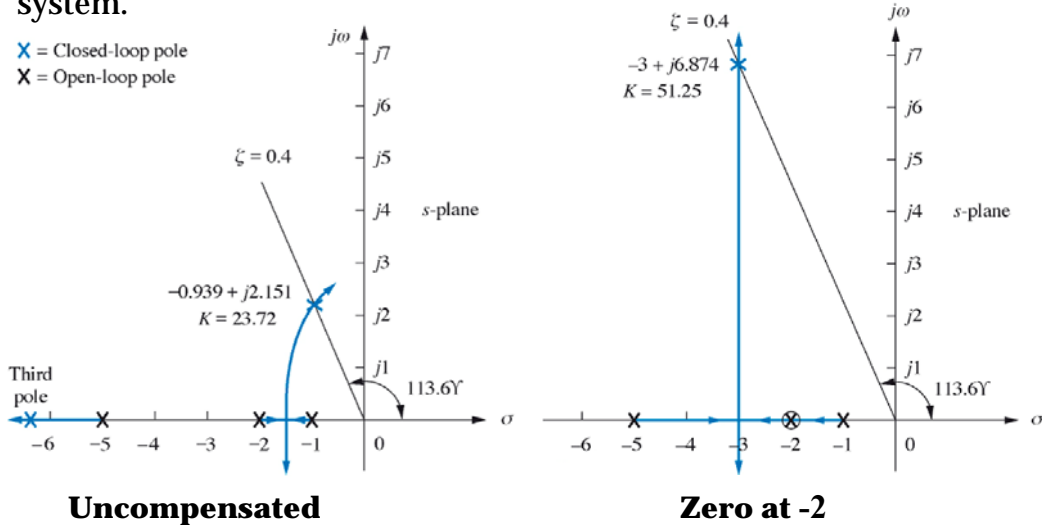
Improving Transient Response

- ☐ One method used to improve the transient response is **PD** control/compensation. It is also called _____ **compensation**.
- ☐ The result of adding differentiation is the addition of a zero to the forward-path transfer function.
- ☐ Typically, the objective is to design a response that has a desirable overshoot and a shorter settling time than the uncompensated system.
- ☐ One major problem with PD compensation is that PD control amplifies high frequency noise.
- ☐ **Note:** The transient response of a system can be selected by choosing an appropriate closed-loop pole location (design point) on the s-plane. If this **point is on the root locus**, then all that is required is a simple gain adjustment. If the **pole is not on the root locus**, then _____ are added to produce a new open loop function whose root locus goes through the design point.

- In PD control, _____ is added to the forward path such that the transfer function of the compensator becomes

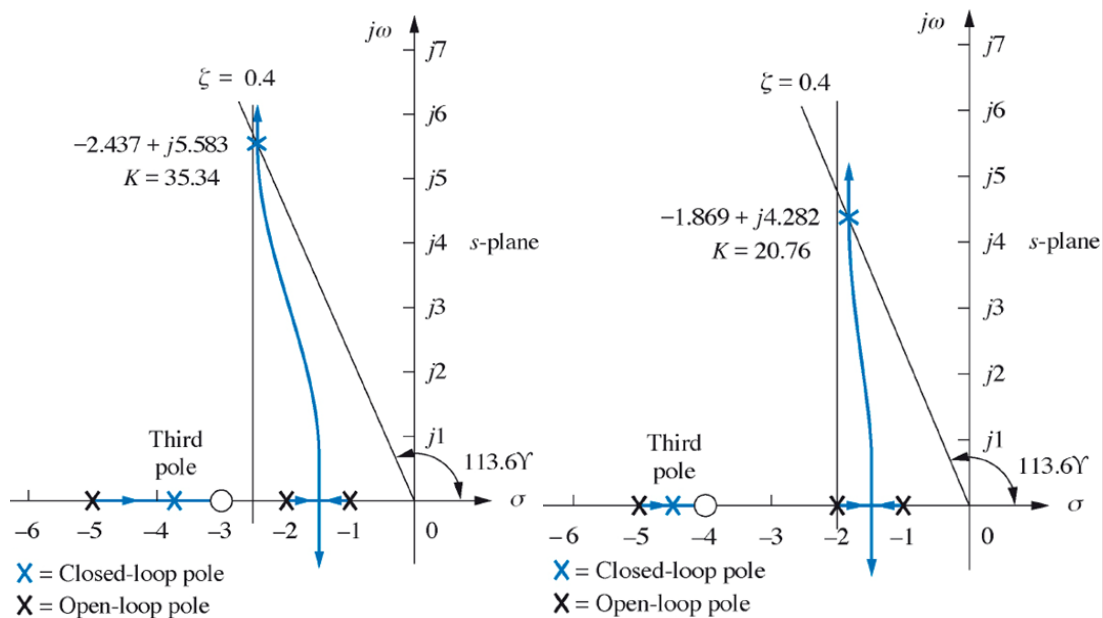
$$G_c(s) = s + z_c$$

- Example: Ideal derivative compensation speeds up the response of a system.



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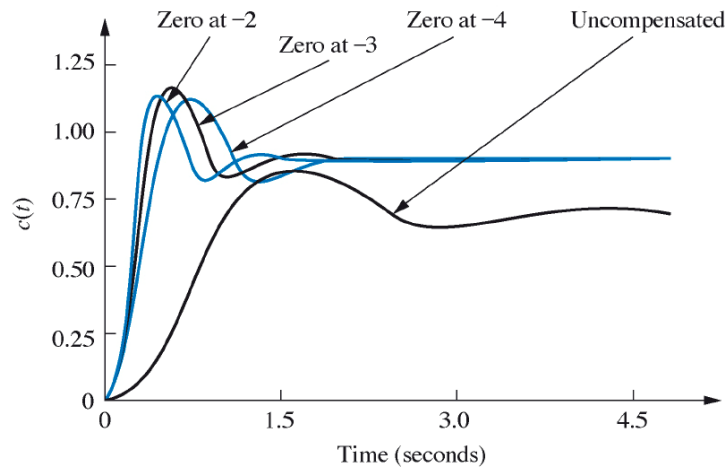
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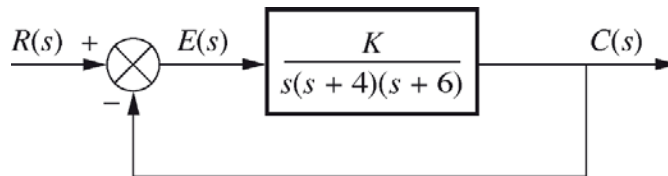
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- The time response of the four cases:



- All compensated cases have dominant poles with the _____, so the overshoots are almost the same.
- Dominant closed-loop poles with the more negative real parts have the shorter settling times.

- Given the system, design an ideal derivative compensator to yield a 16% overshoot, with a threefold reduction in settling time.



1. Transients

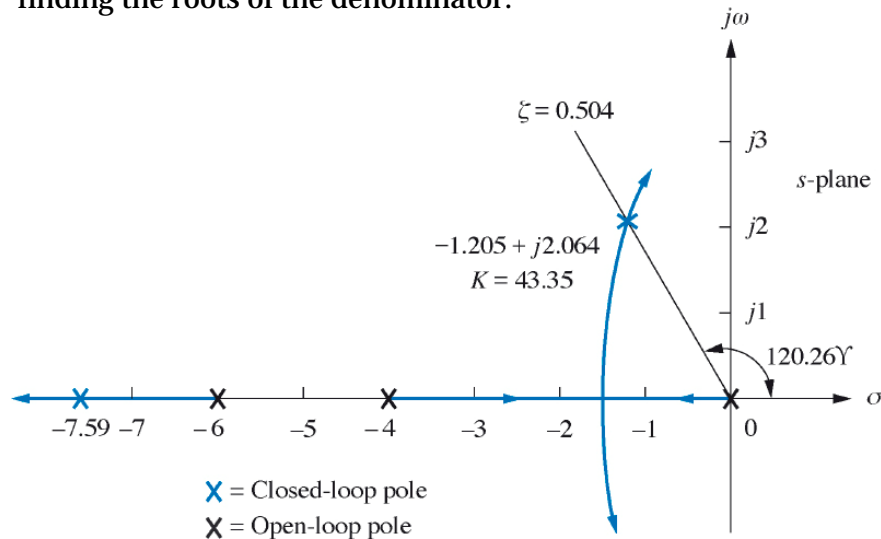
- 1) Plot the root locus of the uncompensated system

$$M_p = 0.16 \rightarrow \zeta = 0.504$$

- 2) Assume 2nd order response

$$t_s = \frac{4}{\zeta \omega_n} = \underline{\hspace{2cm}}$$

- 3) Assumptions is justified because at $K = 43.35$, the third pole is at 7.59 (6 times farther from the $j\omega$ axis than the dominant poles). This can be found by Matlab or getting the closed loop TF and plugging in K , then finding the roots of the denominator.



2. Calculate the value of the design point (or desired pole).
- This is the point that we want our new root locus to go through.
 - The real part of the pole can be found from the settling time given in the problem statement.

$$t_s = 1.107 \Rightarrow \sigma = \underline{\hspace{2cm}}$$

- In general, $\sigma \pm j\omega_d$ is the pole. Since we know σ and we know the angle θ from the $\zeta = 0.504$ line,

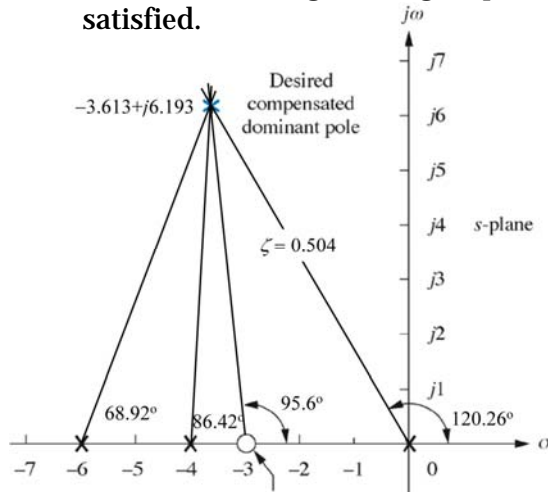
$$\tan \theta = \frac{\sigma}{\omega_d} \Rightarrow \omega_d = \underline{\hspace{2cm}}$$

- So the NEW desired pole is $\underline{\hspace{2cm}}$

3. Calculate the value of K_p/K_D , the zero in the $K_D \left(s + \frac{K_p}{K_D} \right)$.

- We want to add a zero to our system to force the Root Locus through this NEW design point.

- To get the actual value of the zero or (K_p/K_D) we use the fact that for the root locus to go through a point, the angle criterion must be satisfied.



$$\theta_1 = 120.265^\circ$$

$$\theta_2 = \tan^{-1}\left(\frac{6.192}{4 - 3.613}\right) = 86.42^\circ$$

$$\theta_3 = \tan^{-1}\left(\frac{6.192}{6-3.613}\right) = 68.92^\circ$$

$$\begin{aligned}\angle \text{Poles} &= 120.26 + 68.917 + 86.423 \\ &= 275.61^\circ\end{aligned}$$

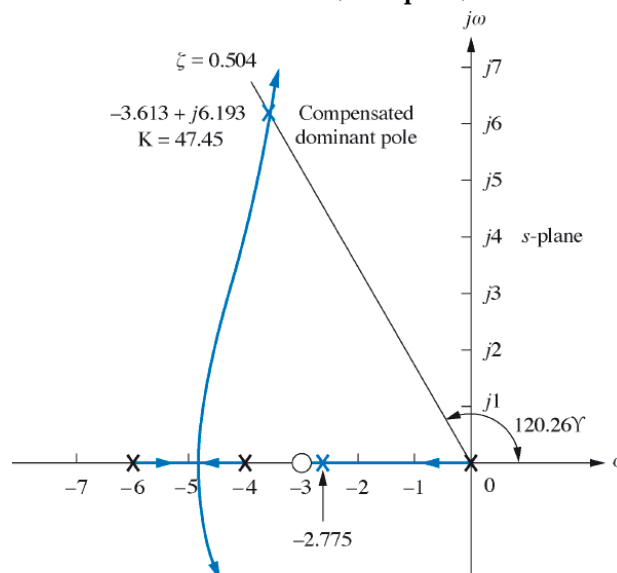
$$\begin{aligned}\angle KG(s) &= \angle \text{Zeros} - \angle \text{Poles} \\ &= (2k+1)180^\circ\end{aligned}$$

- The zero when added to -275.61° must give some odd multiple of $\pm 180^\circ$

$$\angle \text{Zeros} - 275.61^\circ =$$

$$\angle \text{Zeros} = \underline{\hspace{2cm}}$$

- To find the exact value of the zero, use the small right triangle formed between the zero, the pole, and the **real part (s)** of desired point.



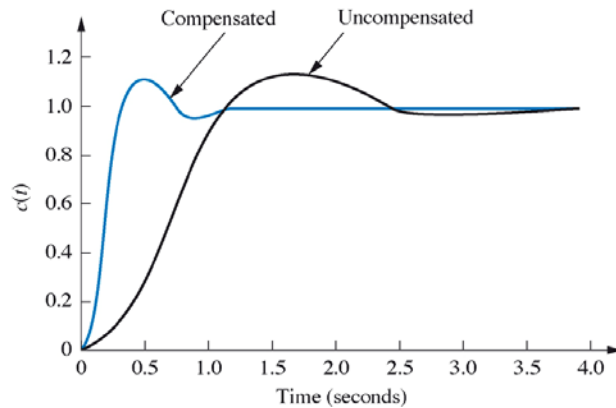
X = Closed-loop pole

X = Open-loop pole

$$\tan(180^\circ - 95.6^\circ) = \underline{\hspace{2cm}}$$

$$\sigma =$$

4. Assuming a second-order system behavior, get transient quantities ζ , rise time, settling time, and the location of the third pole.
 - For uncompensated system: the second-order approximation is accurate since the third pole is at least five times the real part of the dominant, second-order pair.
 - For compensated system: the second-order approximation may be invalid because the CL third pole is at $-2.775 \rightarrow$ Simulation required!
5. Simulation - the system meets the basic requirements!
 - The gain K is calculated by $K = \frac{1}{|G(s_o)|}$
 - $G(s)$ includes plant and controller



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	Uncompensated	Simulation	Compensated	Simulation
Plant and compensator	$\frac{K}{s(s+4)(s+6)}$		$\frac{K(s+3.006)}{s(s+4)(s+6)}$	
Dominant poles	$-1.205 \pm j2.064$		$-3.613 \pm j6.193$	
K	43.35		47.45	
ζ	0.504		0.504	
ω_n	2.39		7.17	
% overshoot	16	14.8	16	11.8
t_s	3.320	3.6	1.107	1.2
t_p	1.522	1.7	0.507	0.5
K_v	1.806		5.94	
$e(\infty)$	0.554		0.168	
Third pole	-7.591		-2.775	
Zero	None		-3.006	
Comments	2 nd order approx OK		Pole-zero not canceling	

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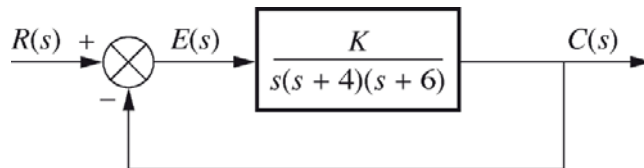
- A passive lead compensator approximates an active ideal derivative compensator (PD controller).

$$G_c(s) = \frac{s + z}{s + p}, \text{ where } z < p$$

- Advantages: (1) Passive network → no additional power supplies required. (2) noise due to differentiation is reduced.
- Disadvantage: the additional pole does not reduce the number of branches of the root locus that cross the imaginary axis into the RHP.
- To design a lead compensator, select either a compensator pole or zero arbitrarily (selection of zero first is recommended). Then use the angle criterion to find the remaining zero or pole.
- There exist more than one possible solution. The differences are in the values of static error constants, the gain required to reach the design point, the difficulty in justifying a second-order approximation and the transient response.

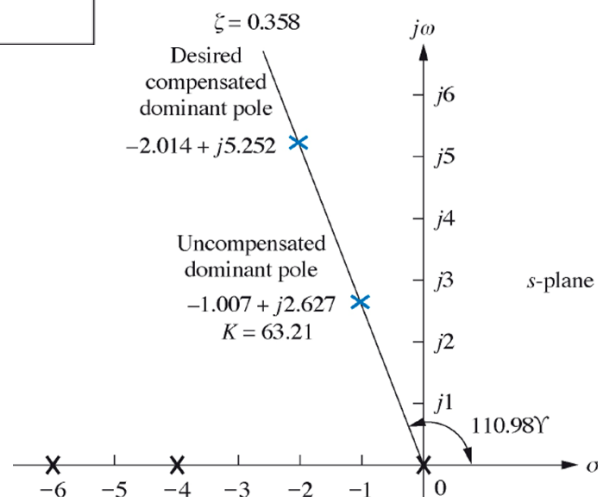
Example: Lead Compensation (1/6)

- Given the system, design three lead compensators that will reduce the settling time by a factor of 2 while maintaining 30% overshoot.



1. Transient response

- 1) Plot the root locus of the uncompensated system
 $M_p = 0.30 \rightarrow$ _____
- 2) Assume 2nd order response
 $t_s =$ _____



2. Find the design point

- For a two fold reduction in settling time:

$$t_s = \frac{3.972}{2} = 1.986 \Rightarrow \sigma = \frac{4}{t_s} = \frac{4}{1.986} = 2.014$$

- The imaginary part:

$$\theta = \sin^{-1}(.358) = 20.98^\circ \Rightarrow \tan \theta = \frac{\sigma}{\omega_d} =$$

$$\Rightarrow \omega_d =$$

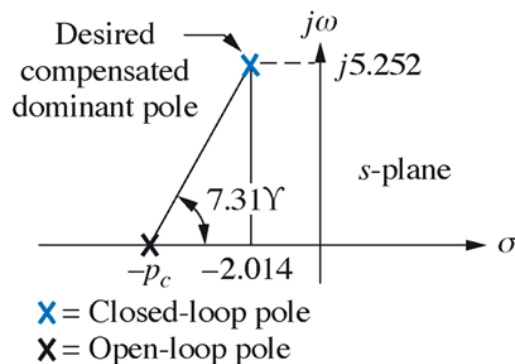
- The design point:

$$s = \sigma + j\omega_d =$$

3. Arbitrarily choose the value of the zero

- Arbitrarily choose: $z_c = -5$

4. Use Angle criterion to calculate the pole



$$\psi_c - \theta_1 - \theta_2 - \theta_3 - \theta_c = (2k+1)180^\circ$$

$$\psi_c = \tan^{-1}\left(\frac{5.253}{5-2.014}\right) = 60.38^\circ$$

$$\theta_1 = \tan^{-1}\left(\frac{5.253}{4-2.014}\right) = 69.29^\circ$$

$$\theta_2 = \tan^{-1}\left(\frac{5.253}{6-2.014}\right) = 52.81^\circ$$

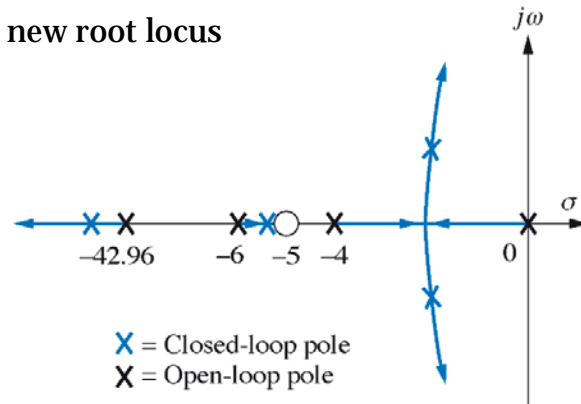
$$\theta_3 = 110.98^\circ$$

$$(2k+1)180^\circ = \Rightarrow \theta_c =$$

$$\tan(7.31^\circ) = \Rightarrow p_c = G_c(s) =$$

Lead compensator

5. Sketch the new root locus



6. Calculate the gain K

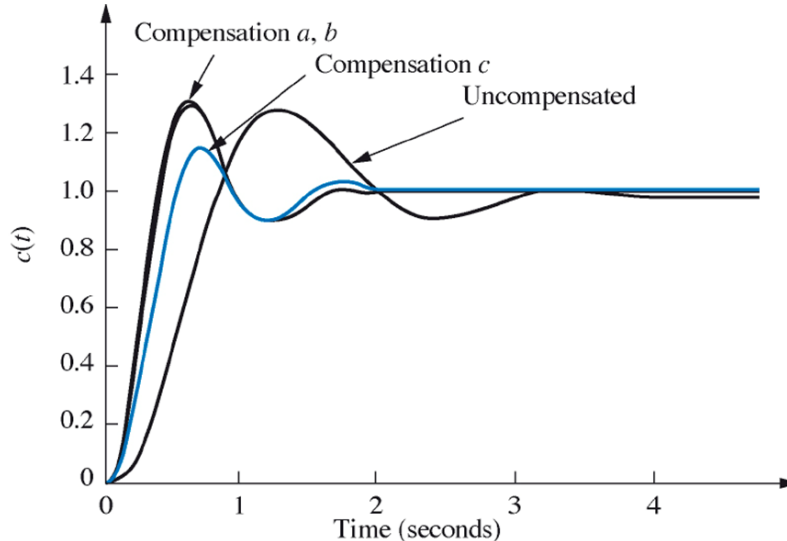
$$K = \frac{1}{|G(s_o)|}, \text{ where } s_o \text{ is the design point.}$$

7. Verify the second-order approximation

- Higher order poles are much farther to the left of the dominant poles.
- Closed loop poles and zeros cancel, so second order approximation is valid.

	Uncompensated	Compensation a	Compensation b	Compensation c
Plant and compensator	$\frac{K}{s(s+4)(s+6)}$	$\frac{K(s+5)}{s(s+4)(s+6)(s+42.96)}$	$\frac{K(s+4)}{s(s+4)(s+6)(s+20.09)}$	$\frac{K(s+2)}{s(s+4)(s+6)(s+8.971)}$
Dominant poles	$-1.007 \pm j2.627$	$-2.014 \pm j5.252$	$-2.014 \pm j5.252$	$-2.014 \pm j5.252$
K	63.21	1423	698.1	345.6
ζ	0.358	0.358	0.358	0.358
ω_n	2.813	5.625	5.625	5.625
% overshoot	30 (28)	30 (30.7)	30 (28.2)	30 (14.5)
t_s	3.972 (4)	1.986 (2)	1.986 (2)	1.986 (1.7)
t_p	1.196 (1.3)	0.598 (0.6)	0.598 (0.6)	0.598 (0.7)
K_v	2.634	6.90	5.791	3.21
$e(\infty)$	0.380	0.145	0.173	0.3112
Third pole	-7.986	-43.8, -5.134	-22.06	-13.3, -1.642
Zero	None	-5	None	-2
Comments	2 nd order approx OK	2 nd order approx OK	2 nd order approx OK	No pole-zero cancellation

8. Notice differences in the following:
- 1) The position of the arbitrarily selected zero
 - 2) The amount of required gain, K .
 - 3) The position of the third and fourth poles and their relative effect upon the second-order approximation.



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Lead Compensation – Less Analytical Method

1. Place the zero in the vicinity of the closed loop natural frequency (ω_n). This is determined by settling time or rise time requirements.
2. Place the pole at a distance between 5 () and 20 times () the value of the zero location.
3. Then use steps 5, 6 and 7 for the analytical method.
4. Note: this method will NOT force the root locus to go through a specific point, but it can place the dominant poles in a general region where the response is acceptable.
5. Use trial and error to improve results if necessary