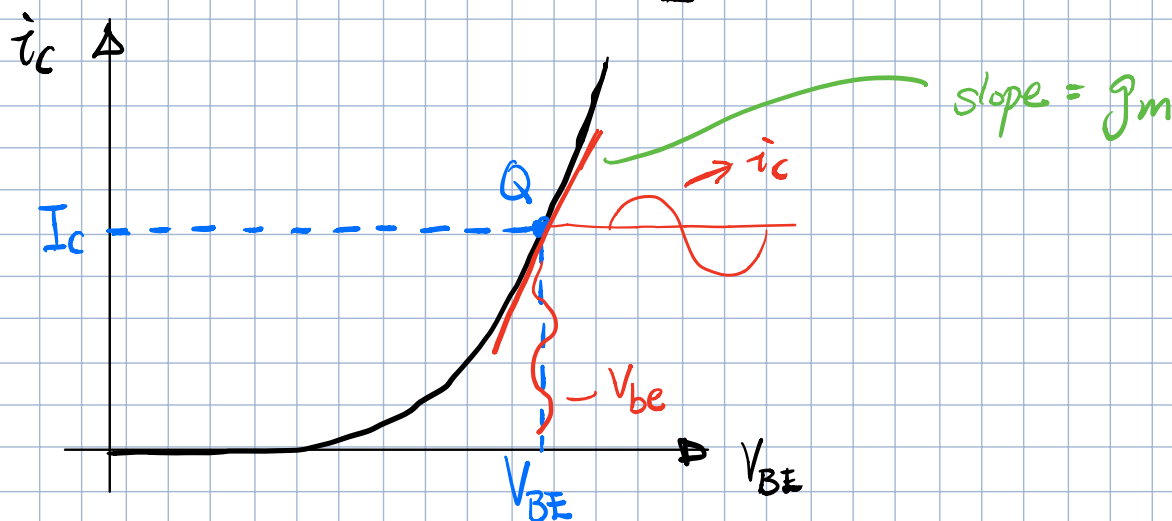
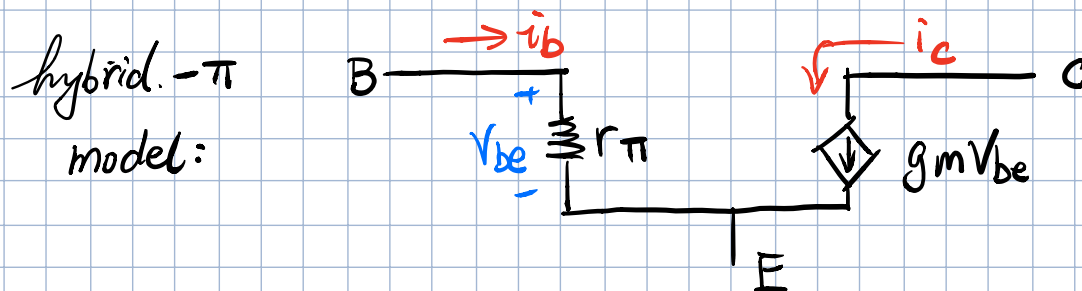


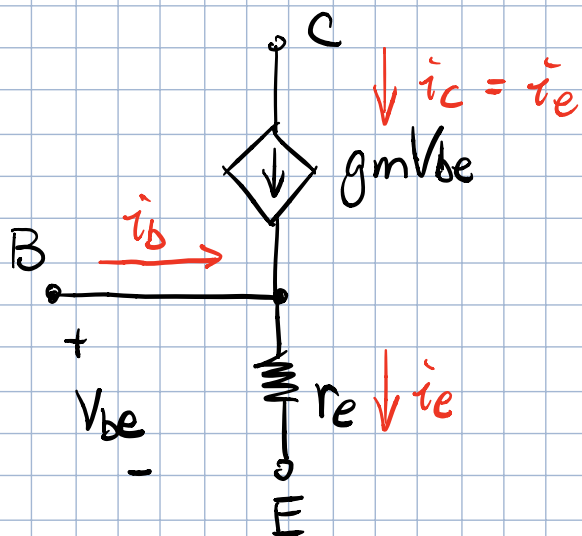
$$i_B = I_B + i_b$$

Dc      ac

$$i_b = \frac{i_c}{\beta} = \frac{g_m V_{be}}{\beta} = \frac{V_{be}}{\frac{\beta}{g_m}} = \frac{V_{be}}{r_\pi}$$



$$i_E = I_E + i_e, \quad i_e = \frac{i_c}{\alpha} = \frac{g_m V_{be}}{\alpha} = \frac{V_{be}}{\frac{\alpha}{g_m}} = \frac{V_{be}}{r_e}$$

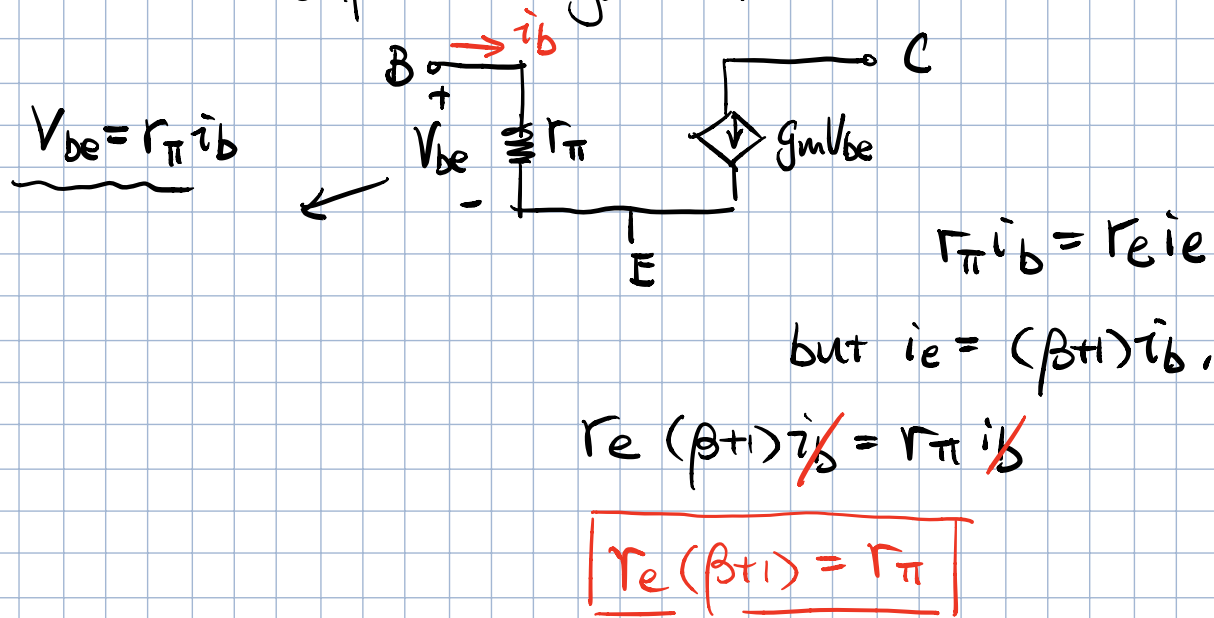


T-model

$$g_m = \frac{I_c}{V_T}$$

$$V_{be} = r_e i_e$$

compare to hybrid- $\pi$ :



## Basic Amplifier (CS/CE)

→ transistors must be biased to operate in **active** region

MOSFET "saturation mode"

BJT "active mode"

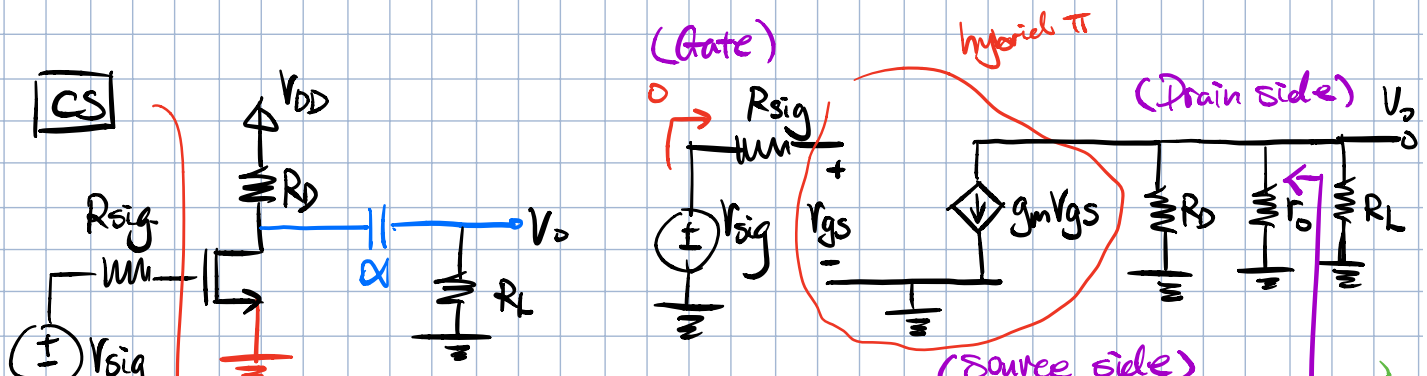
→ keep input signal small

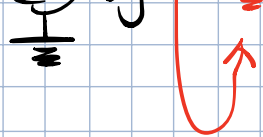
MOSFET:  $V_{gs} \ll 2V_{ov}$

BJT:  $V_{be} \ll V_T$

→ Study voltage gain ( $\frac{V_o}{V_i}$ ), input resistance ( $R_{in}$ ) output resistance ( $R_{out}$ )

## Common Source (CS) / Common Emitter (CE)





In this case,  $V_{gs} = V_{sig}$ . (Drain)  $R_{out}$

Voltage gain =  $V_o/V_{sig}$ ,  $V_o = -g_m V_{gs} \times (R_D // R_L // r_o)$

$= -g_m (R_D // R_L // r_o)$

$= - \frac{(R_D // R_L // r_o)}{\frac{1}{g_m}}$  ← resistance in the drain.

$\frac{1}{g_m}$  ← resistance in the source.

$R_{in} = \infty$  (no current in the gate)

$R_{out} = R_D // r_o$

CS but with a source resistance ( $R_s$ )

(ignore  $r_o$ )

