

Q8.4.

$$G(s) = \frac{\omega_n^2}{s^2 + 2\omega_n \zeta s + \omega_n^2} = \frac{100^2}{s^2 + 2 \cdot 100 \cdot 0.5 s + 100^2}$$

$$= \frac{10000}{s^2 + 100s + 10000}.$$

$$\mathcal{L}(\text{step response}) = G(s) \cdot \mathcal{L}(\text{step input}) = \frac{10000}{s^2 + 100s + 10000} \cdot \frac{1}{s}$$

$$= \frac{As+B}{s^2 + 100s + 10000} + \frac{C}{s}, \quad 10000 = (As+B)s + (s^2 + 100s + 10000)C$$

$$= s^2(A+C) + s(B+100C) + 10000C$$

$$C=1, A+C=0, A=-1, B+100C=0, B=-100.$$

$$\text{Therefore, } \mathcal{L}(\text{step response}) = -\frac{s+100}{s^2 + 100s + 10000} + \frac{1}{s}$$

$$(\text{step response}) = \mathcal{L}^{-1}\left(-\frac{s+100}{s^2 + 100s + 10000} + \frac{1}{s}\right)$$

$$= -\mathcal{L}^{-1}\left(\frac{s+50}{(s+50)^2 + 7500}\right) - \mathcal{L}^{-1}\left(\frac{50}{(s+50)^2 + 7500}\right) + \mathcal{L}^{-1}\left(\frac{1}{s}\right)$$

$$= -e^{-50t} \cos(\sqrt{7500}t) - \frac{50}{\sqrt{7500}} \cdot e^{-50t} \sin(\sqrt{7500}t) + 1$$

$$= \boxed{-e^{-50t} \left( \cos(\sqrt{7500}t) + \frac{\sqrt{2}}{3} \sin(\sqrt{7500}t) \right) + 1}$$



Q8.5.

$$(a) y(s) = \frac{s+3}{(s^2+2s+2)(s+1)} = \frac{s+3}{(s+1+j)(s+1-j)(s+1)}$$

$$\text{Complex poles: } -\zeta \omega_n \pm j \omega_n \sqrt{1-\zeta^2} = -1 \pm j$$

$$\zeta = \frac{\sqrt{2}}{2}, \omega_n = \sqrt{2}, \zeta \omega_n = 1$$

$$\text{real pole: } -\sigma_r = -1, \sigma_r = 1$$

Since  $\sigma_r$  is not much greater than  $\zeta \omega_n$ , the system cannot be represented as a second order system.

$$(b) y(s) = \frac{1}{(s^2+2s+2)(s+20)} = \frac{1}{(s^2+2\zeta\omega_n s + \omega_n^2)(s+20)}$$

$$\omega_n = \sqrt{2}, \zeta = \frac{\sqrt{2}}{2}, \sigma_r = +20.$$

Since  $\sigma_r \gg \omega_n \zeta$ , we can approximate the system to second-order system.

$$y(s) \approx \frac{1}{20(s^2+2s+2)}, \text{ where } \omega_n = \sqrt{2}, \zeta = \frac{\sqrt{2}}{2}$$

$$PO = e^{\frac{-\zeta\pi}{\sqrt{1-\zeta^2}}} \times 100\% = e^{\frac{-\frac{\sqrt{2}}{2} \cdot \pi}{\sqrt{1-(\frac{\sqrt{2}}{2})^2}}} \times 100\%$$

$$= e^{\frac{-\frac{\sqrt{2}}{2} \cdot \pi}{\frac{\sqrt{2}}{2}}} \times 100\% = e^{-\pi} \times 100\% = \boxed{4.321\%} \text{ (PO)}$$

$$\text{Settling time (2\%)} = \frac{-\ln(0.02\sqrt{1-\zeta^2})}{\zeta\omega_n} = \frac{-\ln(0.02\frac{\sqrt{2}}{2})}{\frac{\sqrt{2}}{2} \cdot \sqrt{2}}$$

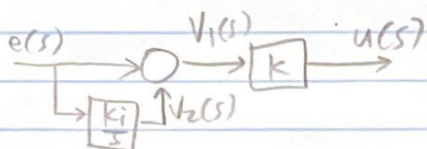
$$= -\ln(0.02\frac{\sqrt{2}}{2}) = \boxed{4.259 \text{ s}} \text{ (Settling time, 2\%)}$$

$$\text{Rise time: } \frac{0.5\pi + \sin^{-1}\zeta}{\omega_n \sqrt{1-\zeta^2}} = \frac{0.5\pi + \sin^{-1}\frac{\sqrt{2}}{2}}{\sqrt{2} \cdot \frac{\sqrt{2}}{2}} = \boxed{2.356 \text{ s}} \text{ (rise time)}$$

$$\text{Peak time: } \frac{\pi}{\omega_n \sqrt{1-\zeta^2}} = \frac{\pi}{\sqrt{2} \cdot \frac{\sqrt{2}}{2}} = \pi = \boxed{3.142 \text{ s}} \text{ (peak time)}$$



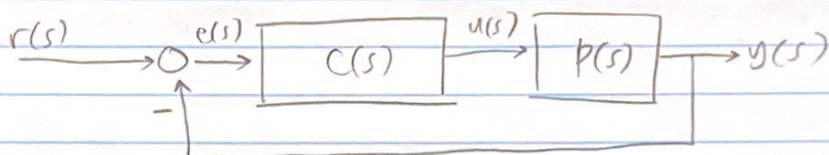
Q8.8.



$$u(s) = k \cdot V_1(s) = k(e(s) + V_2(s)) = k(e(s) + \frac{k_i}{s} e(s))$$

$$= (k + \frac{k \cdot k_i}{s}) e(s), \quad \frac{u(s)}{e(s)} = k + \frac{k \cdot k_i}{s}$$

The diagram can be simplified to



where  $C(s) = k + \frac{k \cdot k_i}{s}$ ,  $P(s) = \frac{b}{s+a}$

(a)  $a_d(s) = (s+3+j2)(s+3-j2) = s^2 + 6s + 13 = s^2 + 2\zeta\omega_n s + \omega_n^2$

$\omega_n = \sqrt{13}$ ,  $\zeta = \frac{6}{2\omega_n} = \frac{6}{2\sqrt{13}} = \frac{3}{\sqrt{13}}$

$\omega_n = \sqrt{13}, \zeta = \frac{3}{\sqrt{13}}$

(b)  $S(s) = \frac{1}{1 + P(s)C(s)} = \frac{1}{1 + \frac{2}{s+2} (k + \frac{k \cdot k_i}{s})} = \frac{1}{1 + \frac{2k}{s+2} + \frac{2k \cdot k_i}{(s+2)s}}$

$$= \frac{(s+2)s}{(s+2)s + 2ks + 2k \cdot k_i} = \frac{s^2 + 2s}{s^2 + (2+2k)s + 2k \cdot k_i}$$

$2+2k=6$ ,  $2k \cdot k_i=13$ ,  $\boxed{k=2, k_i = \frac{13}{4}}$

(c) We should check if for  $C_1$  and  $C_2$ ,  $s^2 + C_1 s + C_2$  always has solution for  $k$  and  $k_i$ , where  $2+2k=C_1$ , and  $2k \cdot k_i=C_2$ .

If  $C_1=2$  and  $C_2 \neq 0$ ,  $k=0$ , and we cannot find  $k_i$ . In other words, if  $a_d(s) = s^2 + 2s + C_2 = (s+1+jm)(s+1-jm)$ , where  $m \in \mathbb{R}$ , there are poles at  $-1 \pm jm$  yet  $(k, k_i)$  cannot be found. Therefore, the statement is false.

(d) If  $a_d(s) = (s+C_1+jC_2)(s+C_1-jC_2)$ , where  $C_1 < 0$ ,  $a_d(s) = s^2 + 2C_1 s + C_1^2 + C_2^2$ . This means  $2+2k=C_1$ ,  $k = \frac{C_1}{2} - 1 > 0$  and  $k_i = \frac{C_1^2 + C_2^2}{2k} > 0$ . This means the gains are big, having risk that  $u(s)$ , the output of the controller, is too big. This can be problematic since there is a limit in the magnitude of the current that can flow without damaging the circuit.