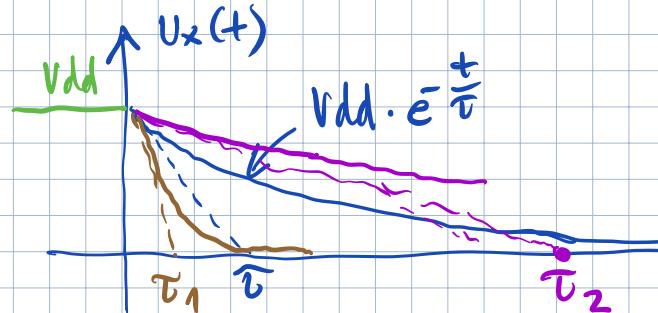
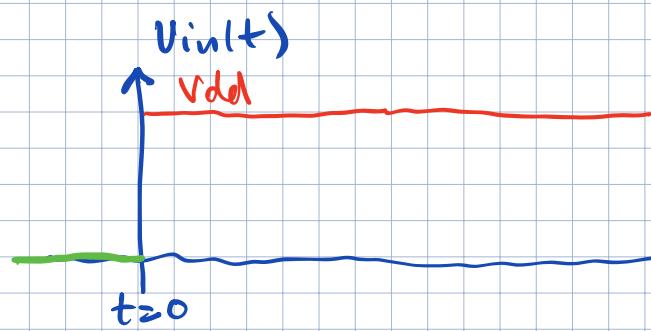
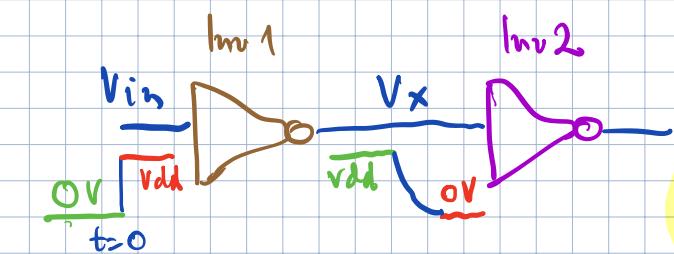


## Lecture 3

- \* Computing : Transistors & Logic
  - \* RC transients finish
  - \* Non-homogeneous diff. eqns.
  - \* constant input
  - \* piece-wise constant input
  - \* continuous input
- ↓  
\* Sensing



$$V_x(t) = V_{dd} e^{-\frac{t}{\tau}}, t \geq 0$$

$$\tau = R_{on,m} \cdot (G_{m2} + G_{p2})$$

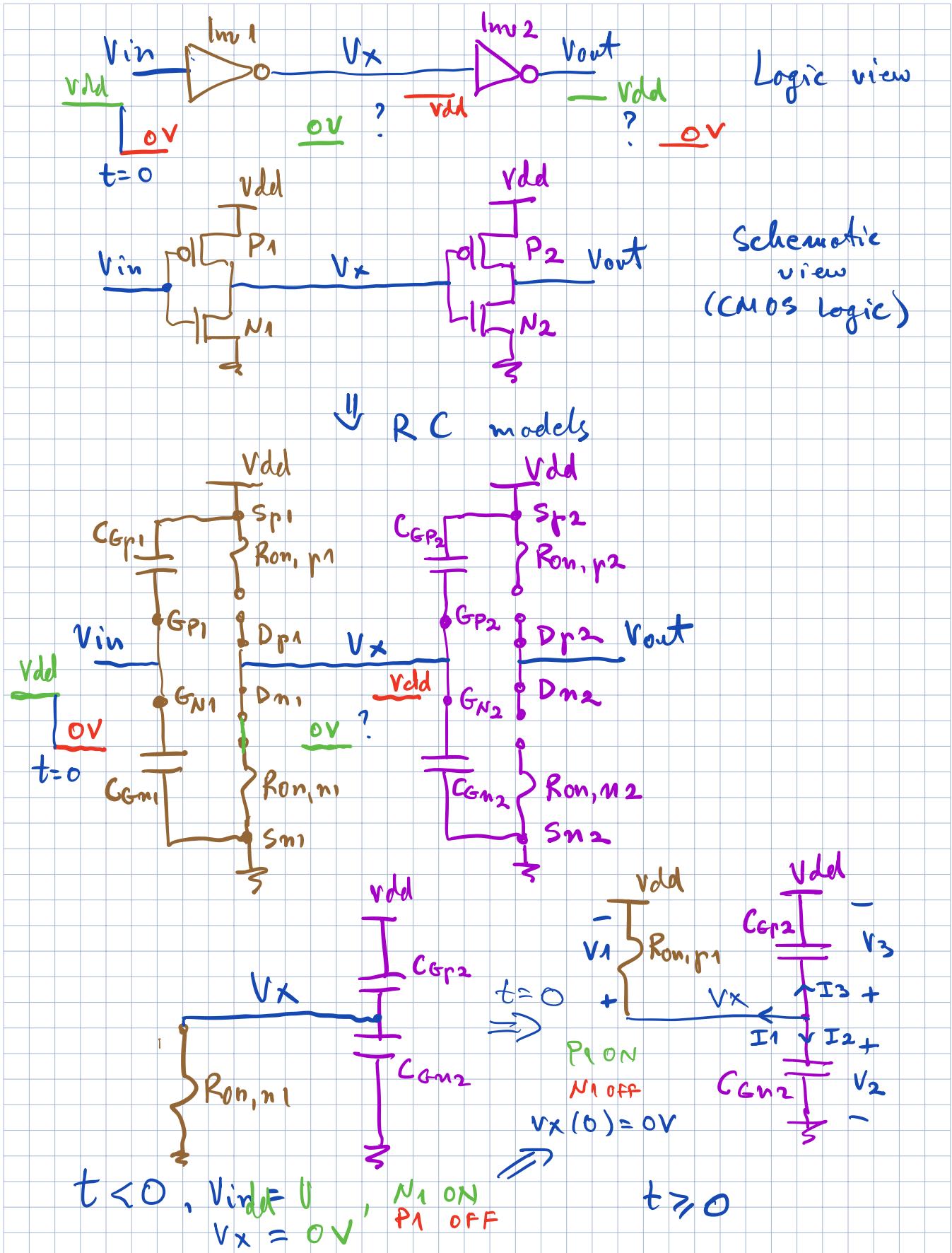
determines the speed  
of transition

What happens for

$$\tau_1 < \tau < \tau_2$$

If you make transistors smaller  $\Rightarrow \tau \downarrow$

Demand scaling  $\Rightarrow$  Moore's law  
(economics law)



$$KCL: I_1 + I_2 + I_3 = 0$$

Elements:

$$V_1 = I_1 \cdot R_{on,p1}$$

$$I_2 = C_{Gn2} \cdot \frac{dV_2}{dt}$$

$$I_3 = C_{Op2} \cdot \frac{dV_3}{dt}$$

Voltages :

$$V_1 = V_x - V_{dd}$$

$$V_2 = V_x$$

$$V_3 = V_x - V_{dd}$$

From KCL & Elements :

$$\underbrace{\frac{V_1}{R_{on,p1}}}_{I_1} + \underbrace{C_{Gn2} \frac{dV_2}{dt}}_{I_2} + \underbrace{C_{Op2} \frac{dV_3}{dt}}_{I_3} = 0$$

From voltages :

$$\frac{V_x - V_{dd}}{R_{on,p1}} + C_{Gn2} \frac{dV_x}{dt} + C_{Op2} \frac{d}{dt}(V_x - V_{dd}) = 0$$

$$\left[ \frac{d}{dt} V_{dd} = 0 \right]$$

$$\frac{V_x - V_{dd}}{R_{on,p1}} + C_{Gn2} \frac{dV_x}{dt} + C_{Op2} \frac{dV_x}{dt} = 0$$

(1)

$$\frac{V_x - V_{dd}}{R_{on,p1}} + ((C_{Gn2} + C_{Op2}) \frac{dV_x}{dt}) = 0$$

$$\frac{dV_x}{dt} = - \underbrace{\frac{V_x}{R_{on,p1} (C_{Gn2} + C_{Op2})}}_{\text{homogeneous term}} + \underbrace{\frac{V_{dd}}{R_{on,p1} (C_{Gn2} + C_{Op2})}}_{\text{non-homogeneous term}}$$

$$\Gamma \text{ Form: } \frac{d}{dt} x(t) = \lambda x(t) \quad (\text{homogeneous})$$

$$\frac{d}{dt} x(t) = \lambda x(t) + a \quad (\text{non-homogeneous})$$

Go back to (1)

$$\frac{V_x - V_{dd}}{R_{on,PN}} + (C_{on2} + C_{op2}) \cdot \frac{d}{dt} (V_x - V_{dd}) = 0$$

Try change of variables to transform the problem into one we know how to solve.

$$\tilde{V}_x = V_x - V_{dd}$$

$$\frac{\tilde{V}_x}{R_{on,PN}} + (C_{on2} + C_{op2}) \frac{d}{dt} \tilde{V}_x = 0$$

$$\frac{d}{dt} \tilde{V}_x = - \frac{\tilde{V}_x}{R_{on,PN} \cdot (C_{on2} + C_{op2})}$$

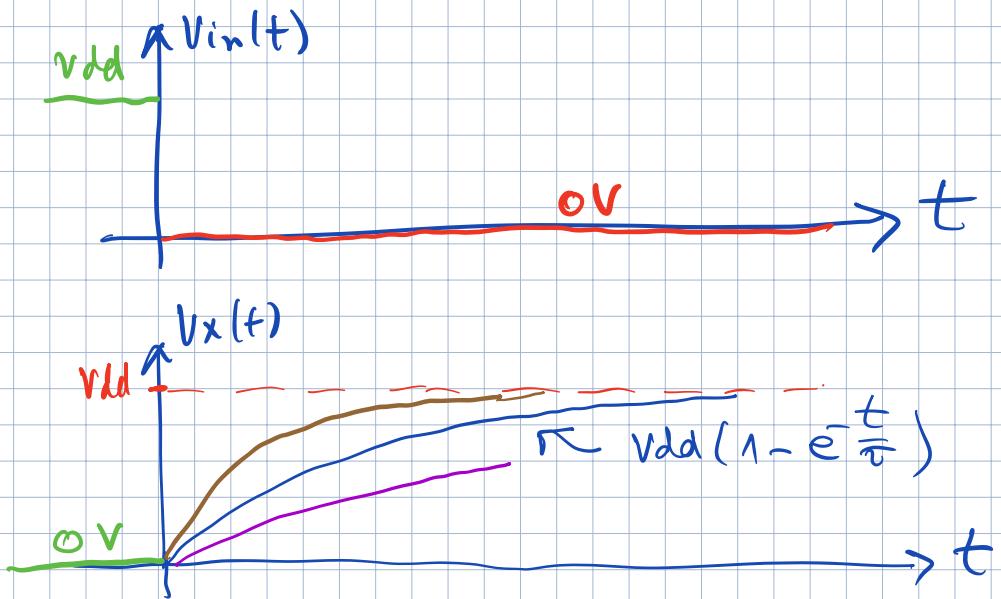
We already  
know how  
to solve  
this  $\Downarrow$   
 $\Leftrightarrow$

$$\tilde{V} = R_{on,PN} (C_{on2} + C_{op2}) \quad \text{homogeneous diff.-eqn.}$$

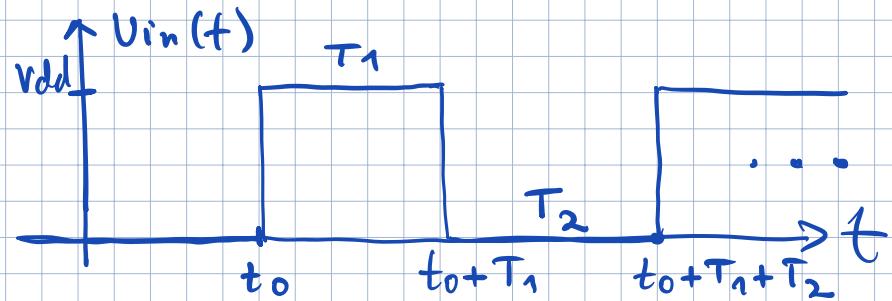
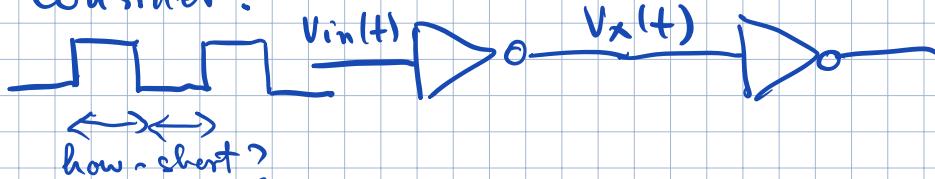
$$\tilde{V}_x(t) = \tilde{V}_x(0) \cdot e^{-\frac{t}{\tilde{V}}} \quad , t \geq 0$$

$$V_x(t) - V_{dd} = (V_x(0) - V_{dd}) \cdot e^{-\frac{t}{\tilde{V}}} \quad V_x(0) = 0V$$

$$V_x(t) = V_{dd} (1 - e^{-\frac{t}{\tilde{V}}}) \quad , t \geq 0$$

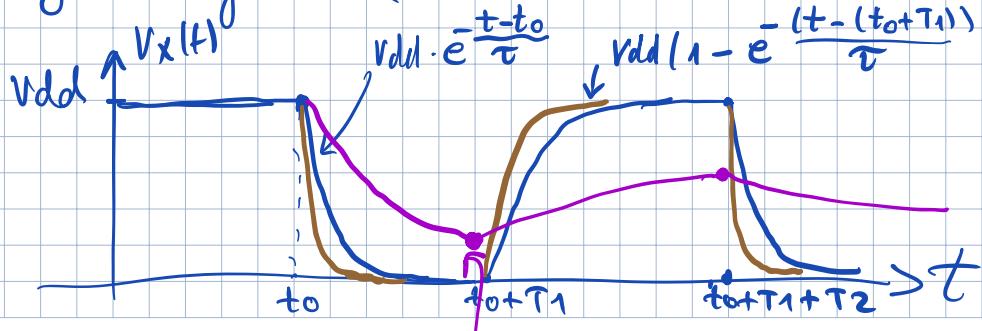


Consider:



Will  $V_x(t)$  be able to follow these changes as

a "logic" signal (i.e. to reach  $OV$  or  $V_{dd}$ )?



$$t_1 < t < t_2$$

} init. condition is no longer  
it is actually  $V_x(t_0 + T_1) =$   
 $= V_{dd} \cdot e^{\frac{t_0 + T_1 - t_0}{T_2}} =$   
 $= V_{dd} \cdot e^{\frac{T_1}{T_2}}$

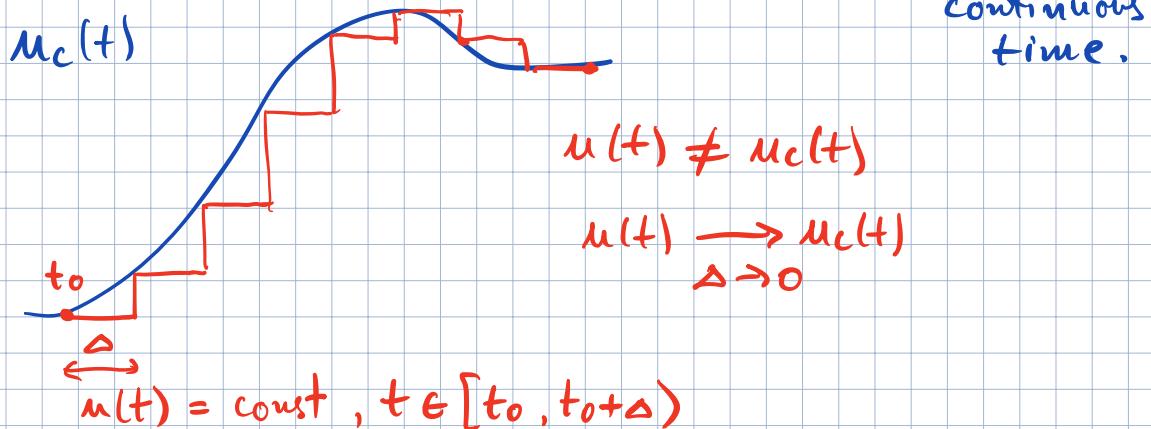
Use previous  $V_x(t)$  solution as an initial condition for the next interval.

Solutions for piece-wise constant input:

Form:  $\frac{d}{dt} x(t) = \lambda x(t) - \lambda u(t) \leftarrow \text{constant}$

or piece-wise constant

What we also want to solve is  $u(t) = u_c(t)$

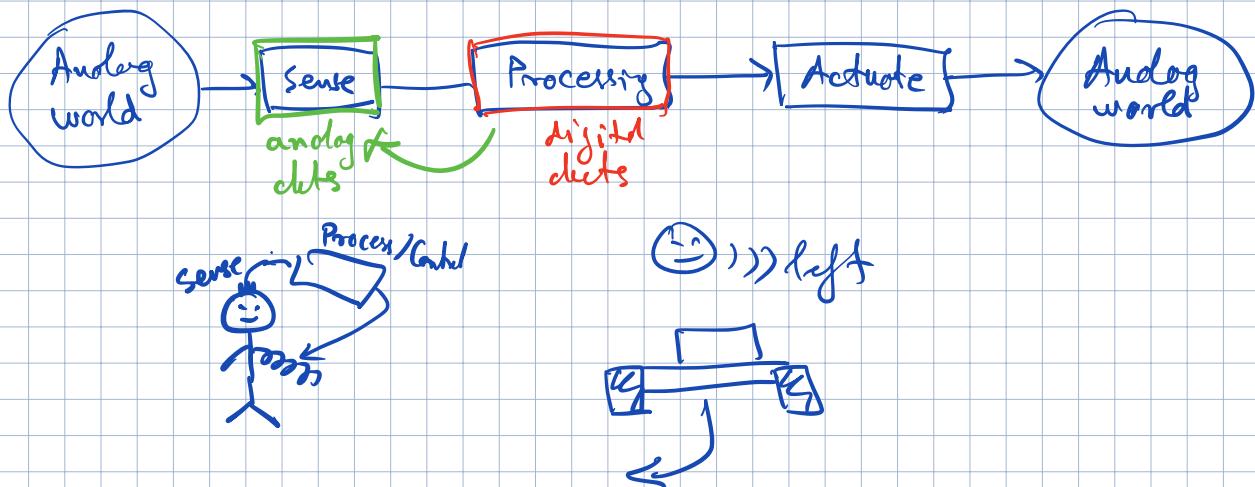


iterate & use the previous solution as initial condition

then let  $\Delta \rightarrow 0$  for  $u(t)$  to approach  $u_c(t)$  in the limit (disc. + hw)

Why do we want the know the response  
to continuous time input?

## EECS 16AB Pipeline



Sensing: brain signals or voice signals

Design sensor ckt/s to pick this up

Signal of interest + ~~interference~~  
(AC power, wifi/cellular,  
others talking in the lot,  
...)  
unwanted

The goal:

Filter / select signals of interest and  
reject interference.

How can a circuit become a filter -  
process the input continuous-time signal?