

Welcome to EECS 16A!

Designing Information Devices and Systems I

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Fall 2022

Lecture 12A
Correlation and Tri-Lateration



Good morning!

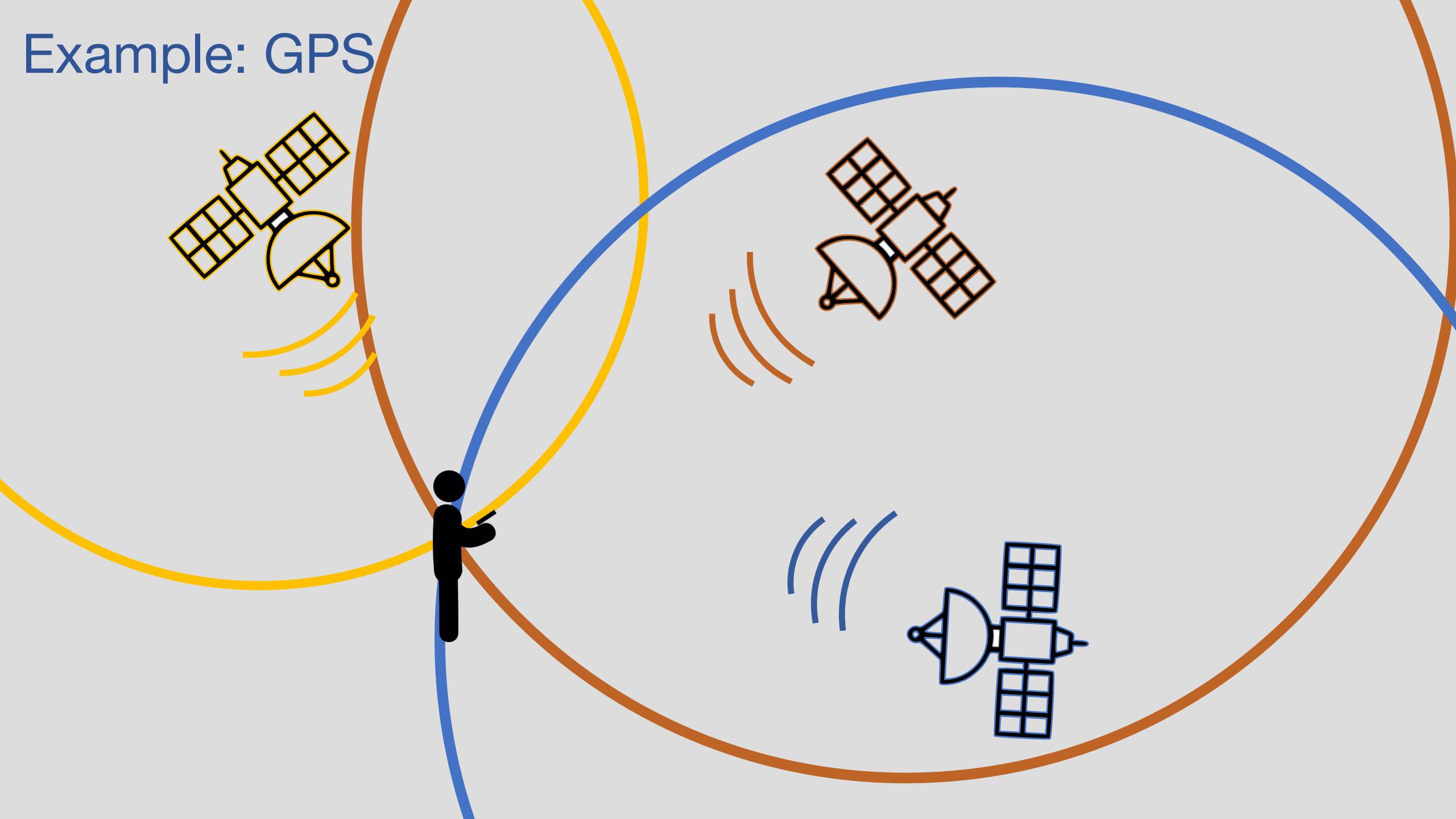
Last time:

- Talked about GPS
 - Known position of satellites
 - Each satellite has its own signature
- Talked about inner product
 - Measure of similarity between vectors
 - When zero – orthogonal vectors
- Talked about using inner products for classification

Today:

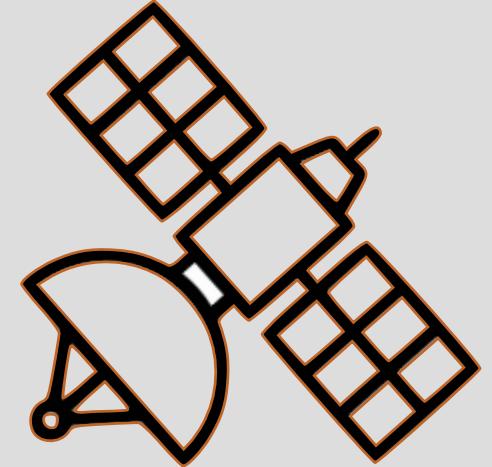
- Computing delay with cross-correlation
- Finding position with multi-lateration

Example: GPS



GPS

- 24 satellites
 - Known position
 - Time synchronized
 - 8 usually visible
- Problem:
 - Classify which satellite is transmitting
 - Estimate distance to GPS
 - Estimate position from noisy data
- Tools:
 - Inner product
 - Cross correlation
 - Least Squares



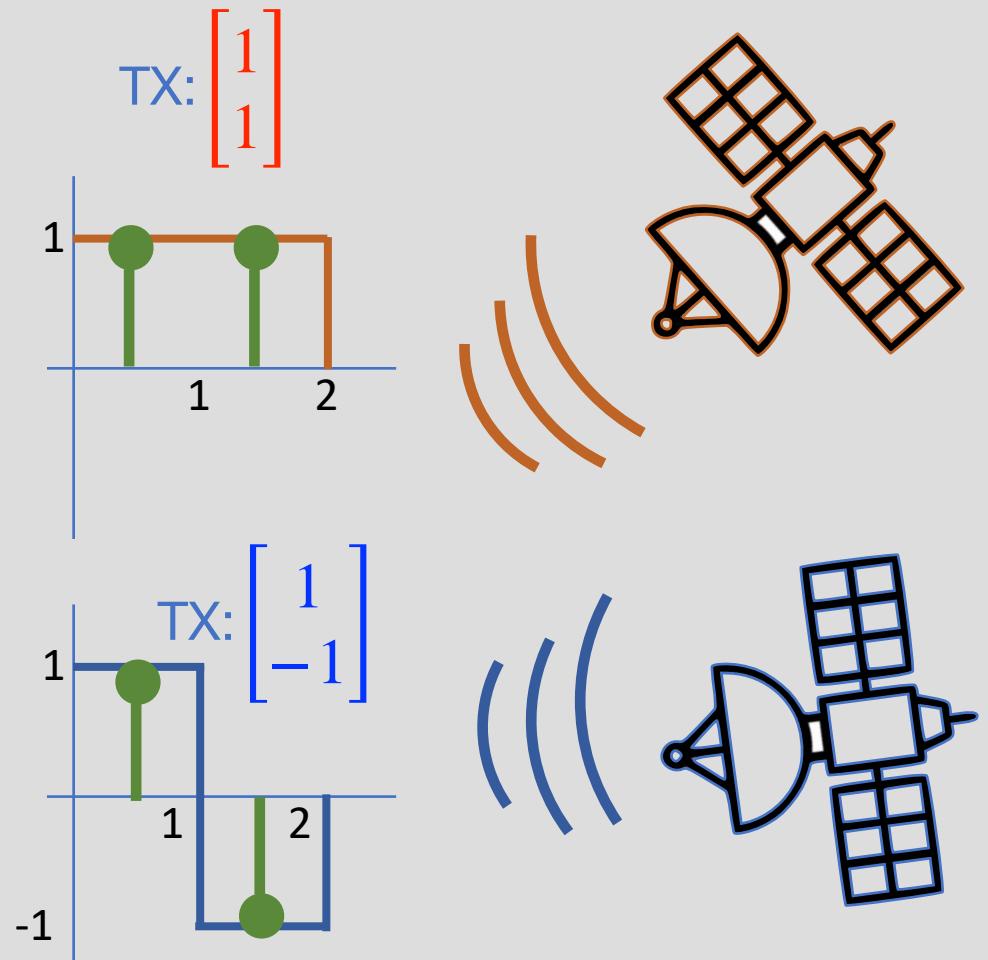
Localization

- Satellites transmit a unique code
 - Radio signal
- Signal is received and digitized by a receiver



Two problems:

1. Interference
2. Timing (Next!)



Interference

Possibility 1: Both sats are in TX

$$\vec{r} = \vec{s}_1 + \vec{s}_2 + \vec{n}$$

Possibility 2: Only S1 is in Tx

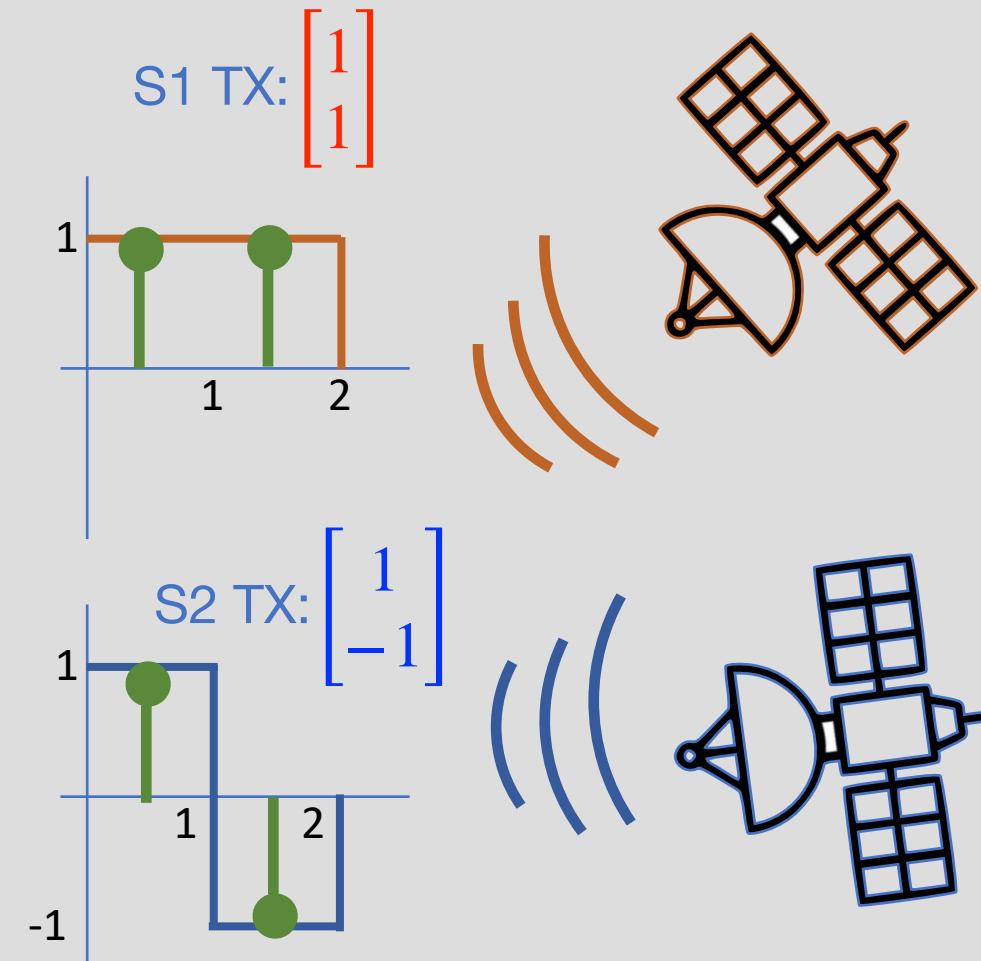
$$\vec{r} = \vec{s}_1 + \vec{n}$$

Possibility 3: Only S2 is in Tx

$$\vec{r} = \vec{s}_2 + \vec{n}$$

Possibility 4: None is in Tx

$$\vec{r} = \vec{n}$$



Interference

Possibility 1: Both sats are in TX

$$\vec{r} = \vec{s}_1 + \vec{s}_2 + \vec{n}$$

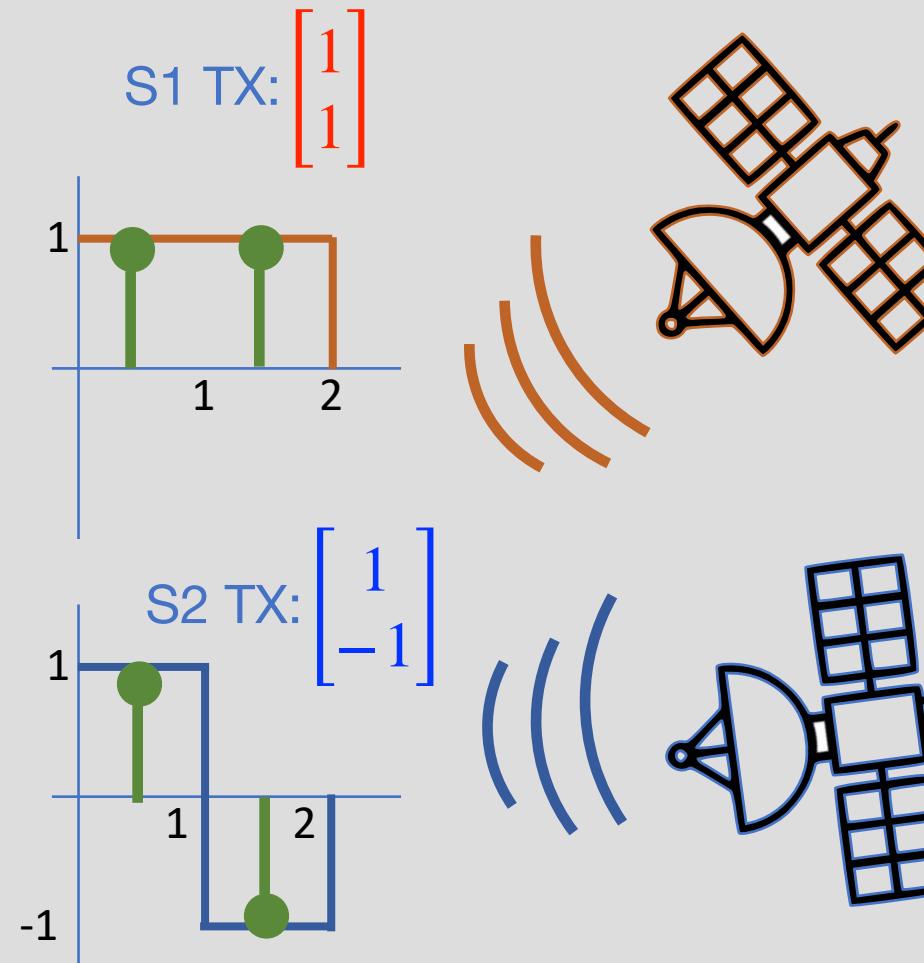
$$\begin{aligned}\langle \vec{r}, \vec{s}_1 \rangle &= \langle \vec{s}_1 + \vec{s}_2 + \vec{n}, \vec{s}_1 \rangle \\ &= \langle \vec{s}_1, \vec{s}_1 \rangle + \langle \vec{s}_2, \vec{s}_1 \rangle + \langle \vec{n}, \vec{s}_1 \rangle\end{aligned}$$

Desired Interference

Q: How to design codes that don't interfere?

A: Make them orthogonal!

$$\langle \vec{s}_2, \vec{s}_1 \rangle = 0$$

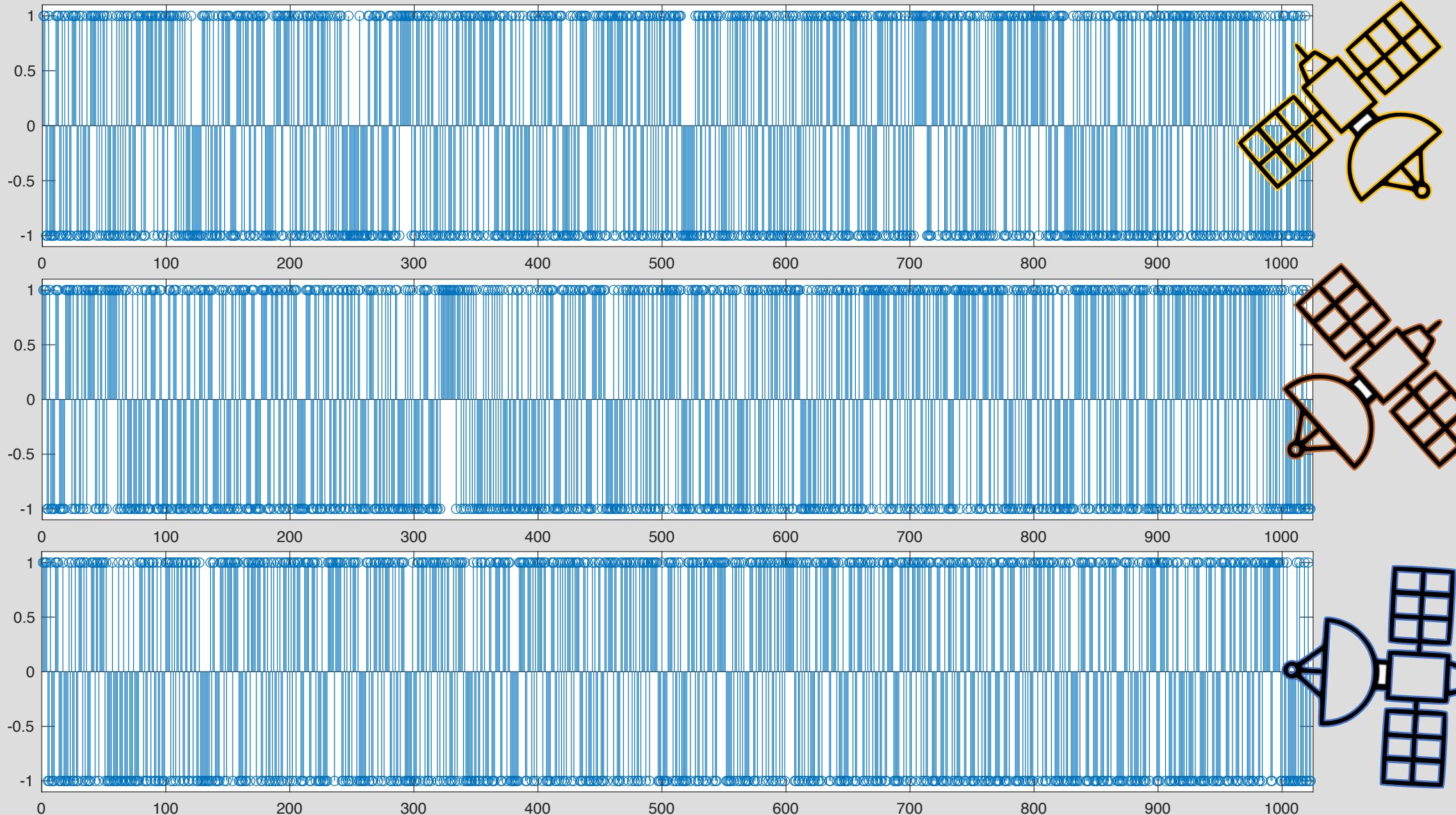


Example

$$\vec{F}_1 = [2.01, -0.03]^T \quad \vec{n}_L = [1.01, -1.03]$$

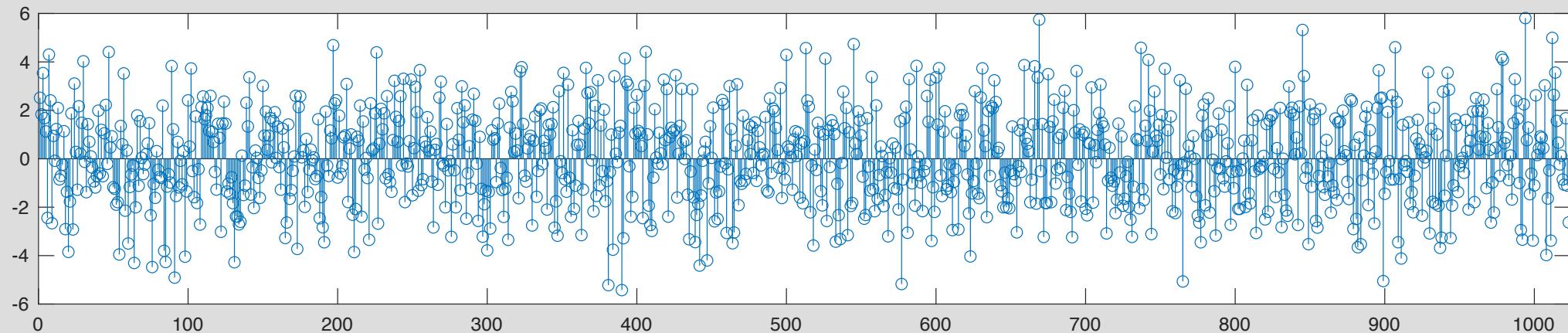
$$\begin{aligned}\vec{s}_1 &= [1, 1]^T \\ \vec{s}_2 &= [1, -1]^T \\ \vec{n} &= [0.01, -0.03]^T\end{aligned}$$

GPS Gold Codes



Example:

$$\vec{r} =$$



$$\langle \vec{r}, \vec{s}_i \rangle = \vec{r}^T \vec{s}_i$$

$$\vec{r}^T$$

$$\vec{r}^T \vec{s}_1 \quad \vec{r}^T \vec{s}_2 \quad \dots \quad \vec{r}^T \vec{s}_{24}$$



$$\vec{s}_1$$

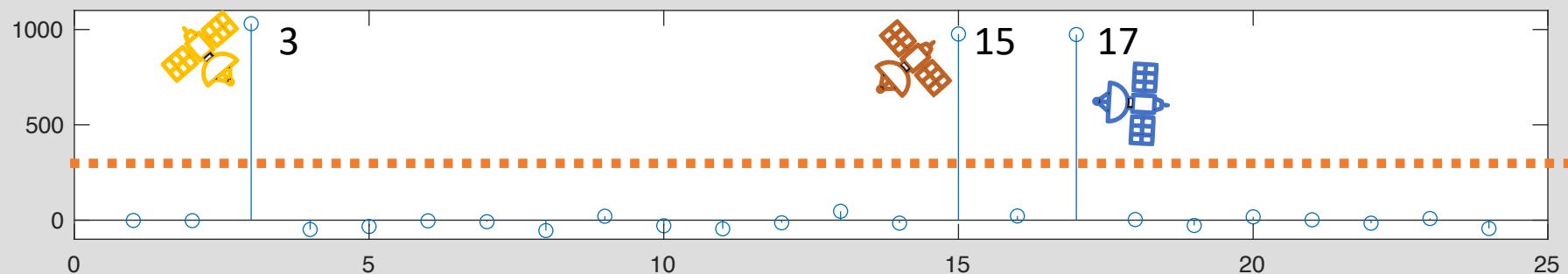
$$\vec{s}_2$$

$$\vec{s}_3$$

...

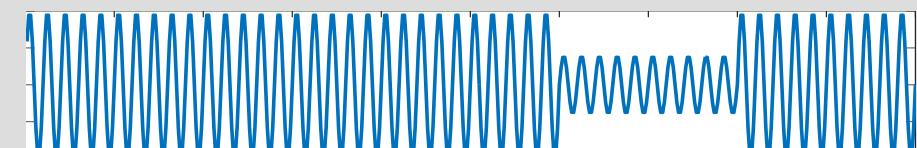
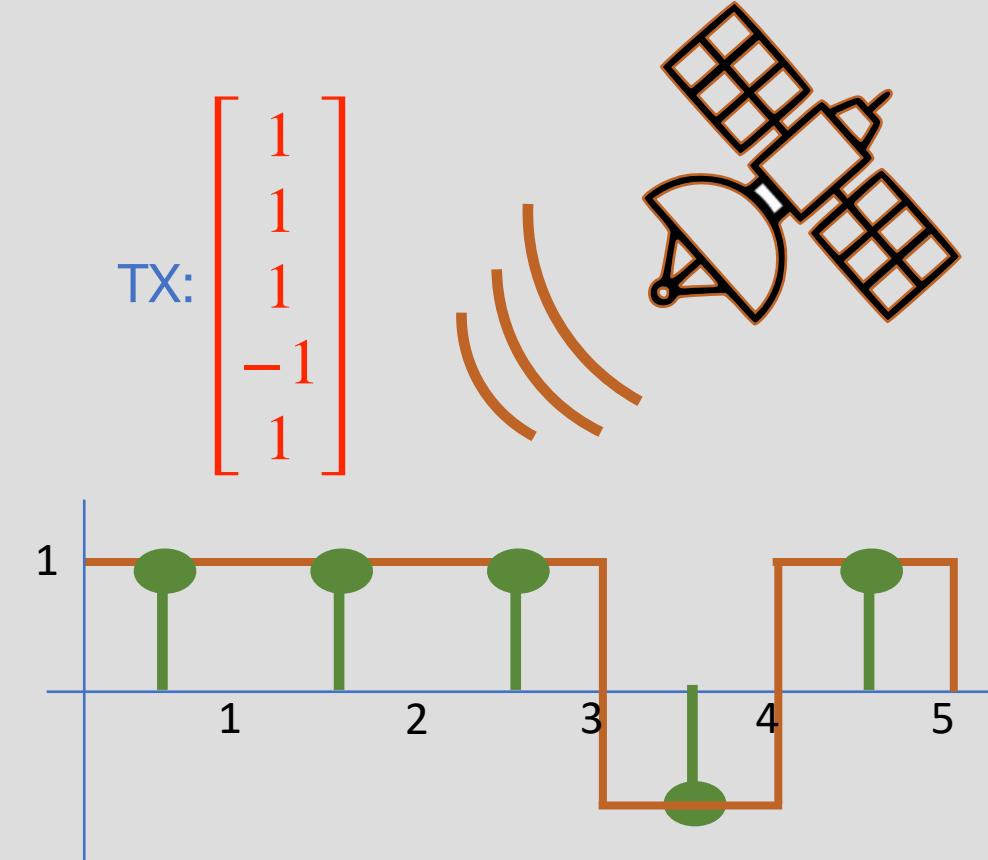
$$\vec{s}_{24}$$

=

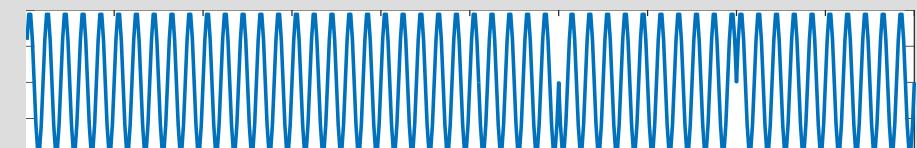


Timing....

- Satellites transmit a (modulated) unique code
 - Radio signal
- Signal is received (demodulated) and digitized by a receiver



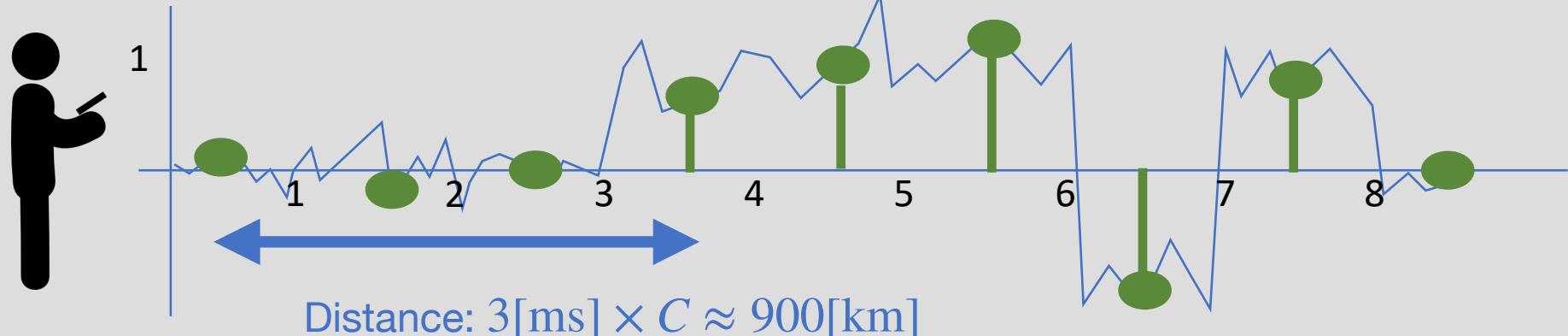
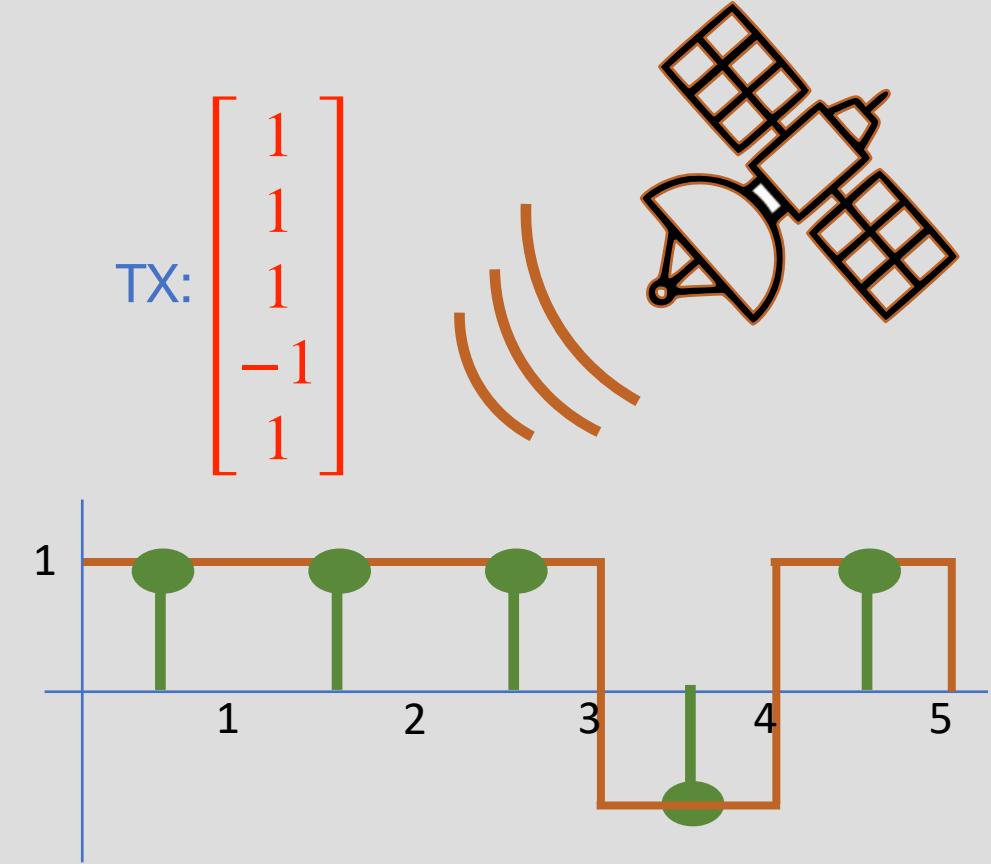
Amplitude Modulation (AM)



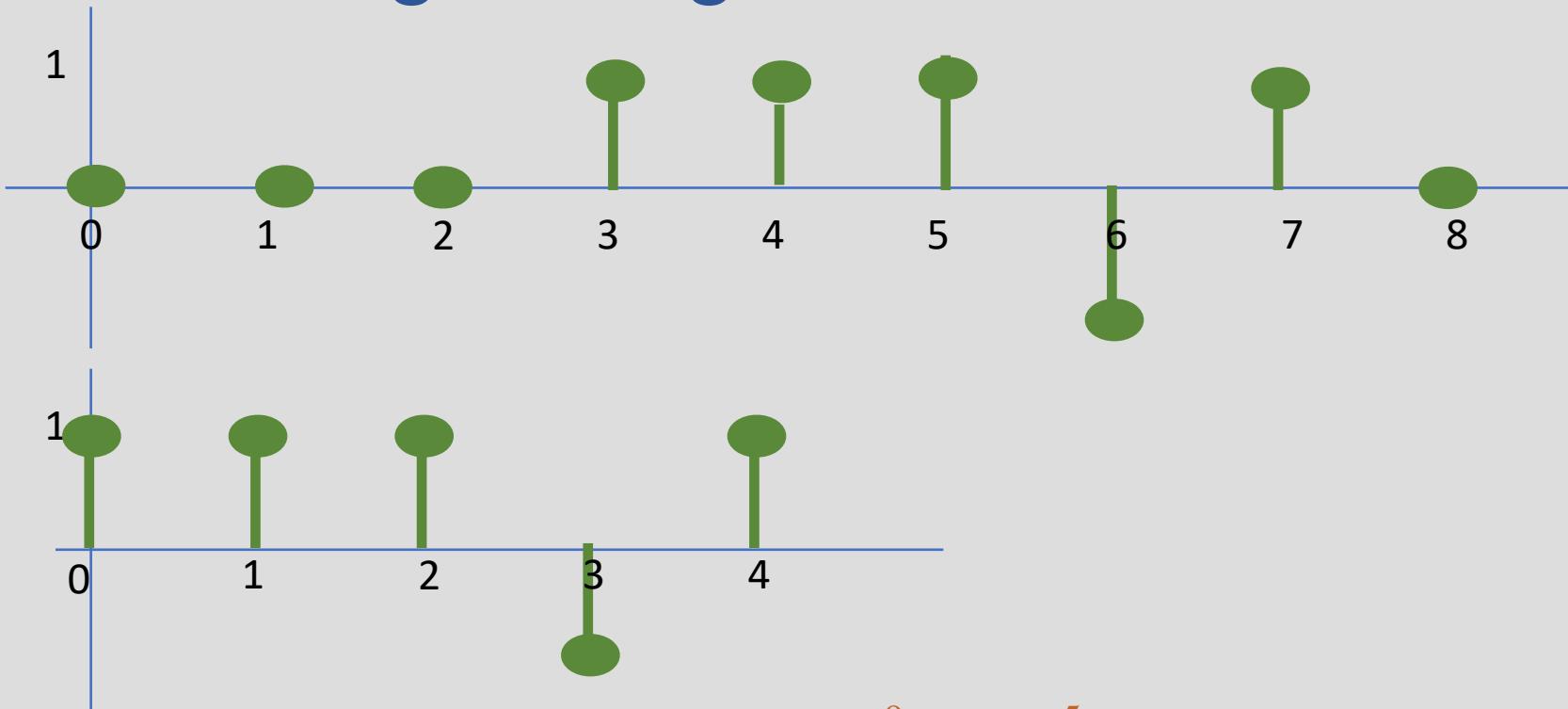
Phase Modulation (PM)

Timing....

- Satellites transmit a unique code
 - Radio signal
- Signal is received and digitized by a receiver



“Pattern Matching” of Signals



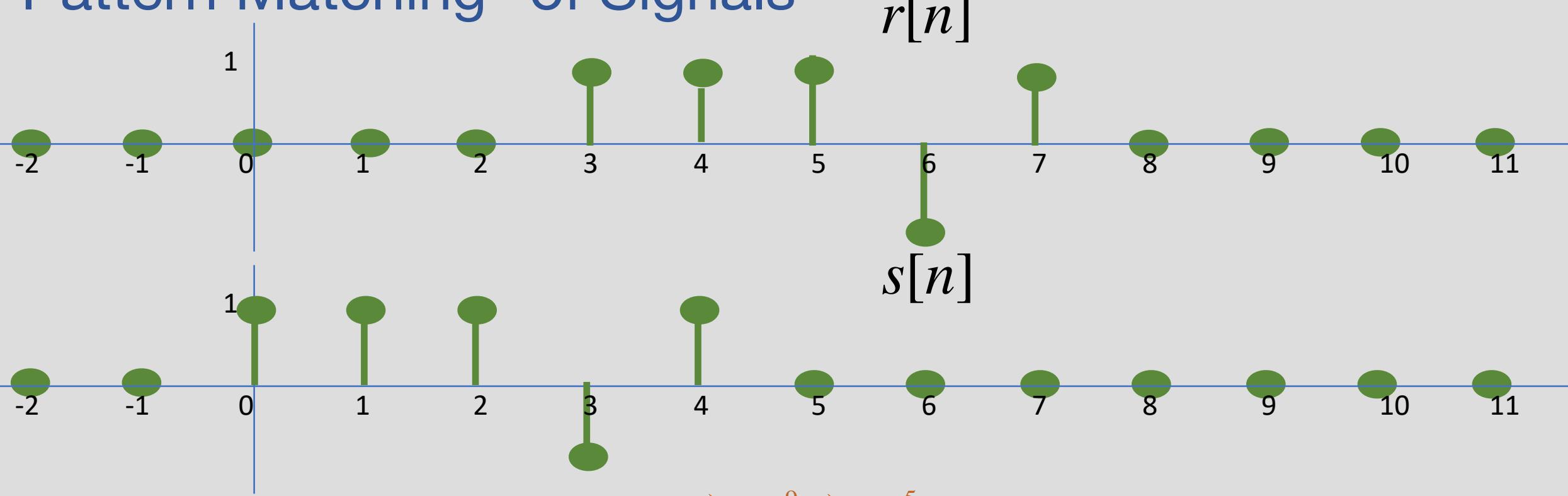
Problem: vectors (signals) not the same length... $\vec{r} \in \mathbb{R}^9$, $\vec{s} \in \mathbb{R}^5$

Solution: Define infinite signals $r[n]$, $s[n]$ by zero-padding

$$\vec{r} = [r_0 \ r_1 \ r_2 \ \dots \ r_8]^T \Rightarrow r[n] = \begin{cases} r_n & 0 \leq n \leq 8 \\ 0 & \text{elsewhere} \end{cases}$$

$$\vec{s} = [s_0 \ s_1 \ s_2 \ \dots \ s_4]^T \Rightarrow s[n] = \begin{cases} s_n & 0 \leq n \leq 4 \\ 0 & \text{elsewhere} \end{cases}$$

“Pattern Matching” of Signals



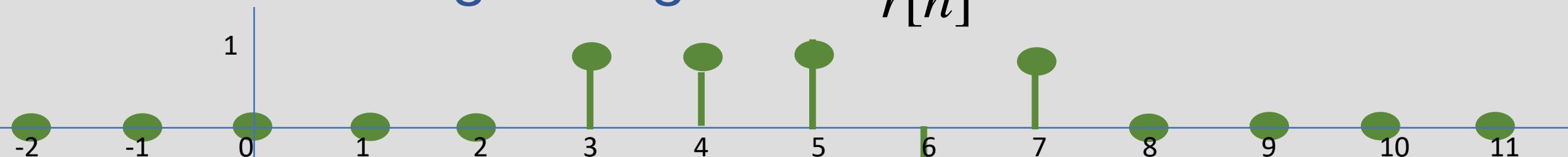
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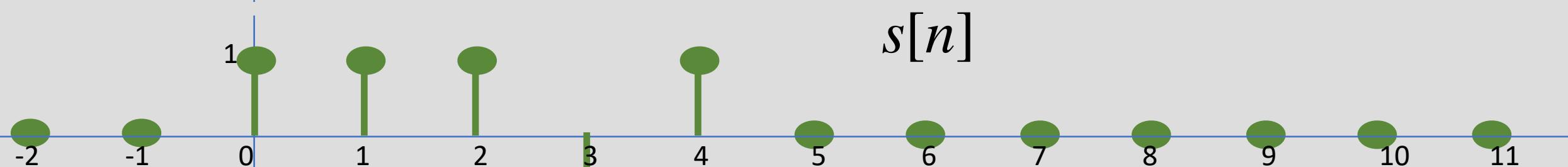
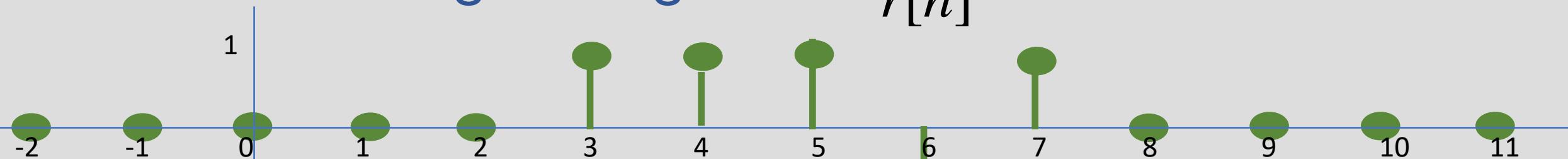
“Pattern Matching” of Signals



$$\langle r[n], s[n] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n] = \sum_{n=0}^{7} r[n]s[n] =$$



“Pattern Matching” of Signals

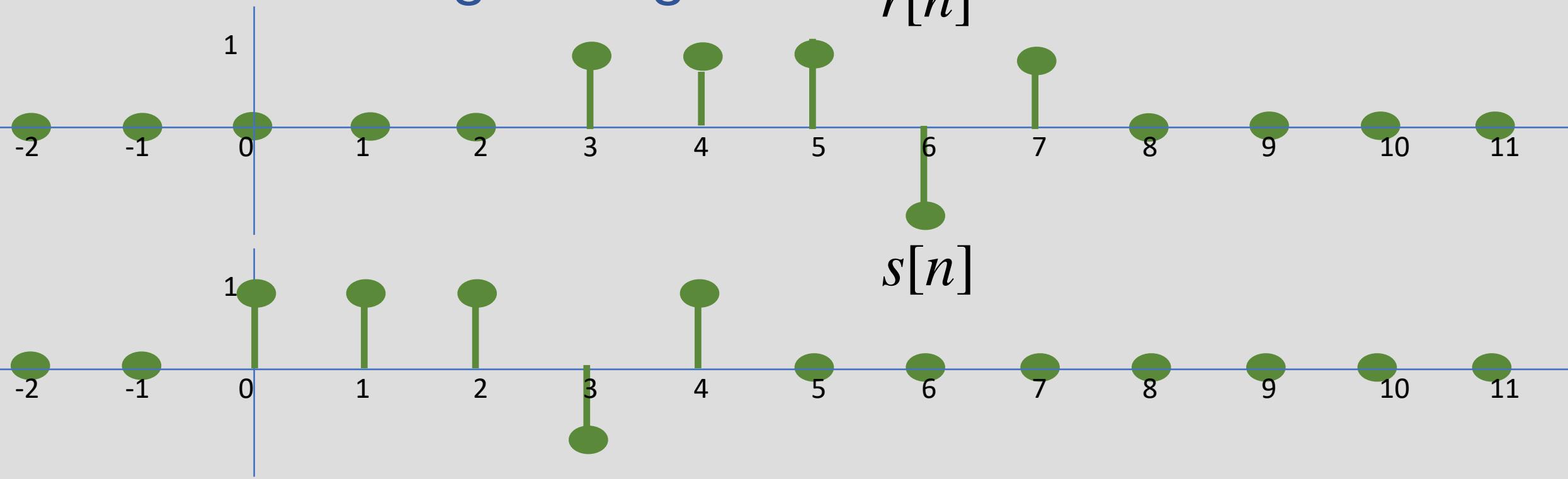


$$\langle r[n], s[n] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n] = \sum_{n=0}^{7} r[n]s[n] = \boxed{0,0,0,1,1,1,-1,1}$$

Q: How to match with shifted version?

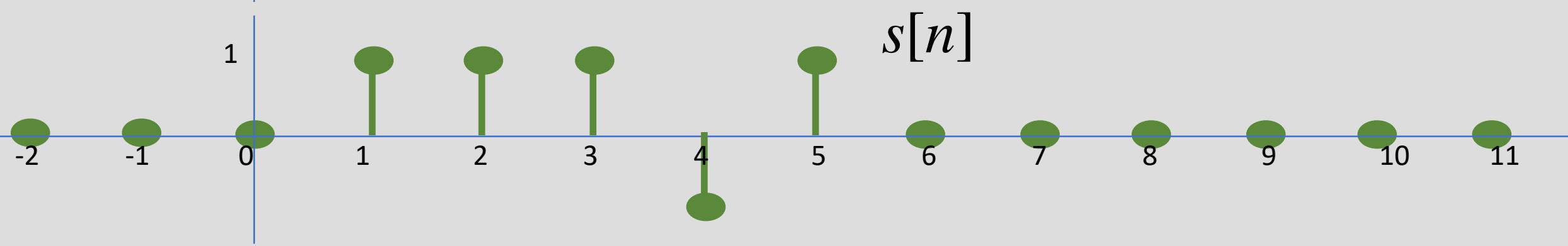
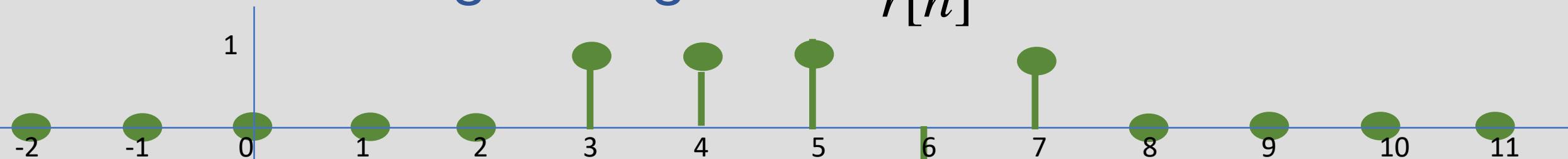
A: compute: $\langle r[n], s[n - 1] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n - 1]$

“Pattern Matching” of Signals



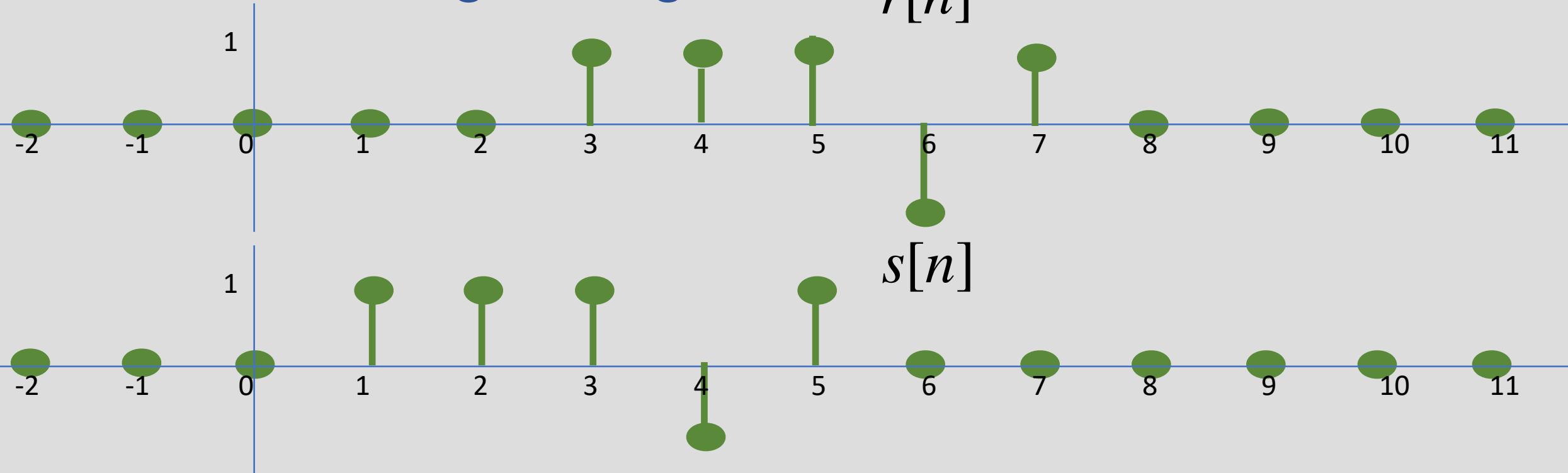
$$\langle r[n], s[n-1] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-1]$$

“Pattern Matching” of Signals

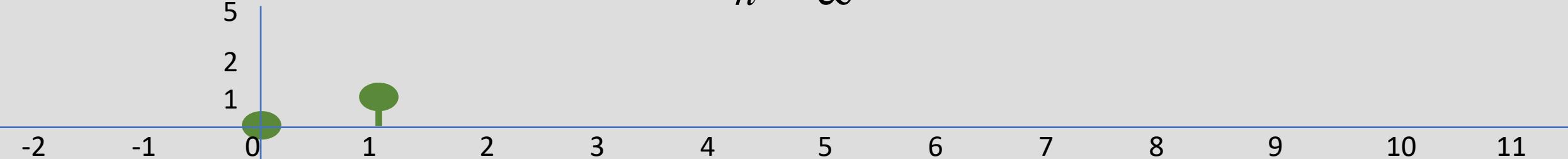


$$\text{corr}_{\vec{r}}(\vec{s})[1] = \langle r[n], s[n-1] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-1] = 1$$

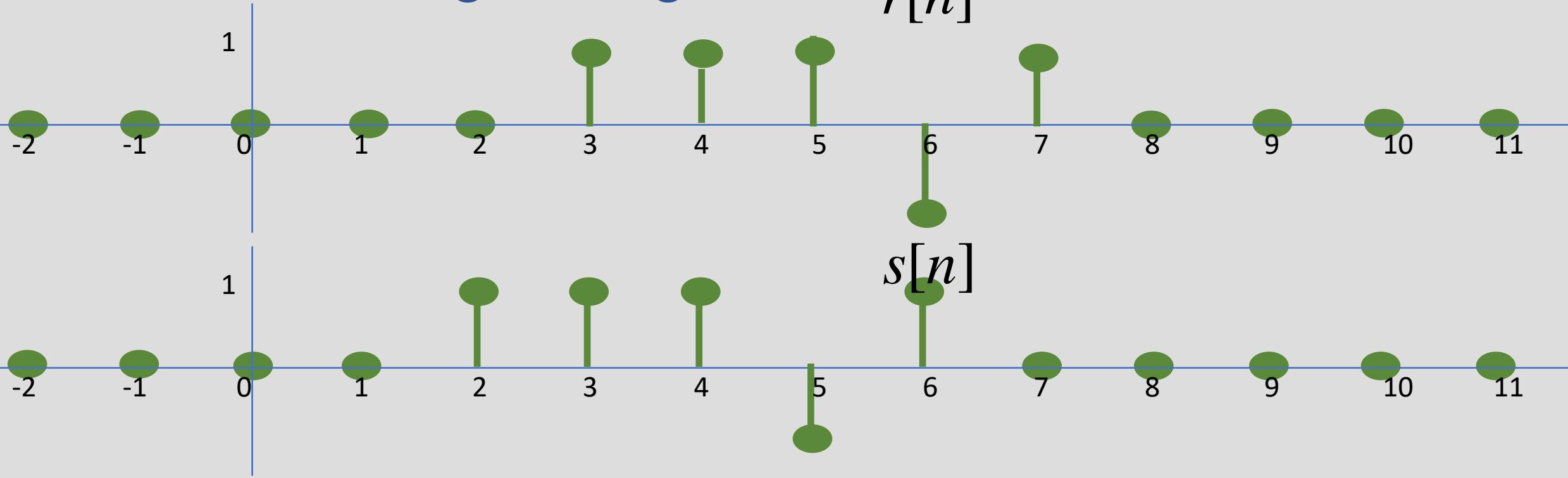
“Pattern Matching” of Signals



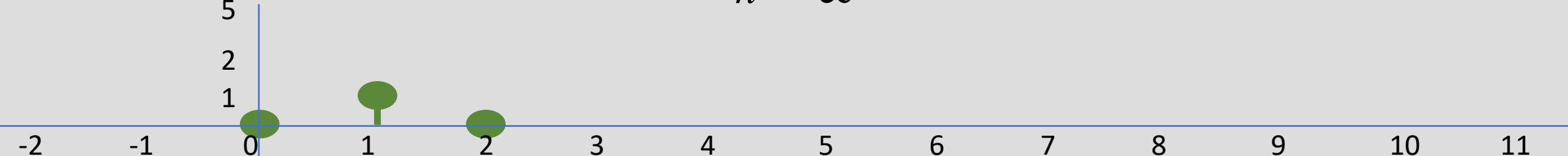
$$\text{corr}_{\vec{r}}(\vec{s})[1] = \langle r[n], s[n-1] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-1] = 1$$



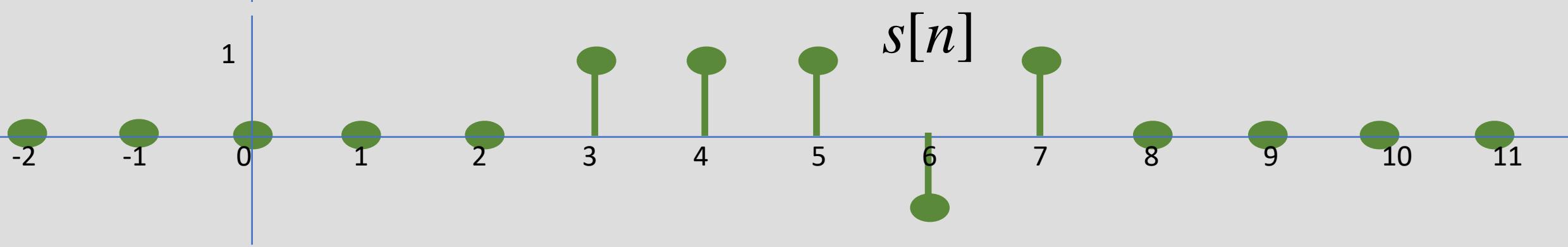
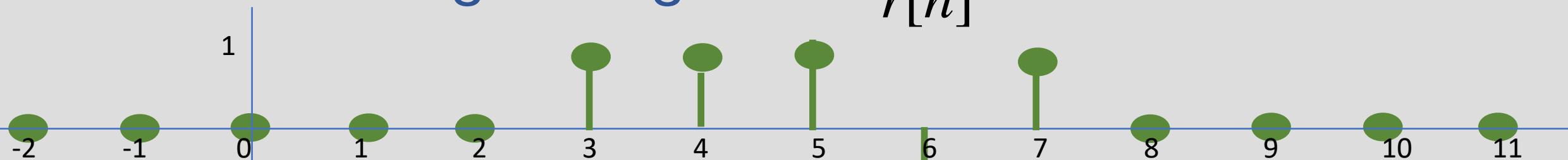
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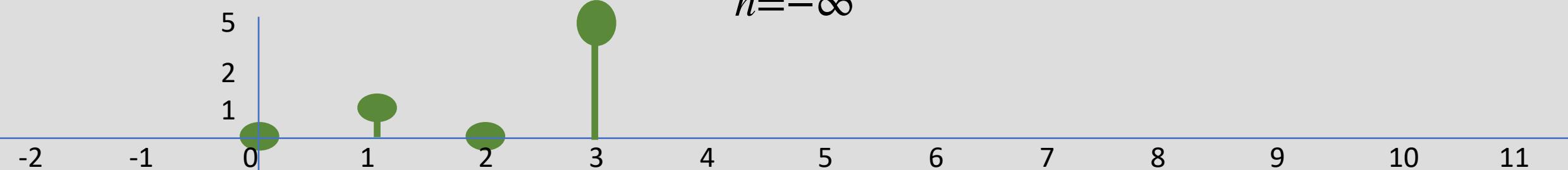
$$\text{corr}_{\vec{r}}(\vec{s})[2] = \langle r[n], s[n-2] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-2] = 0$$



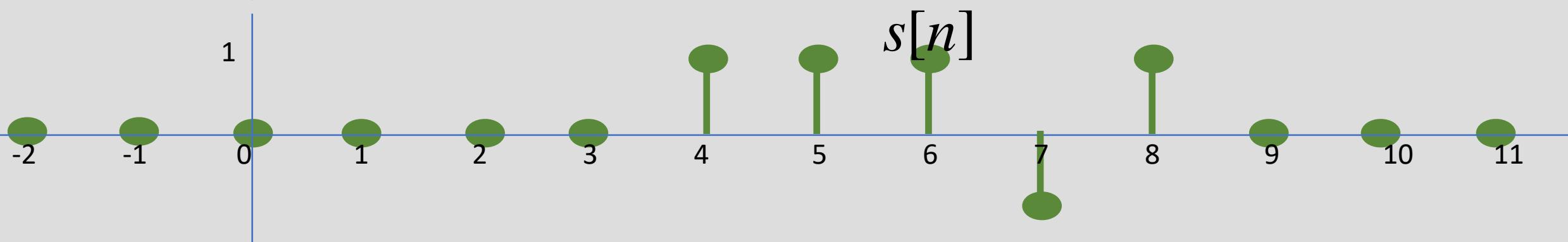
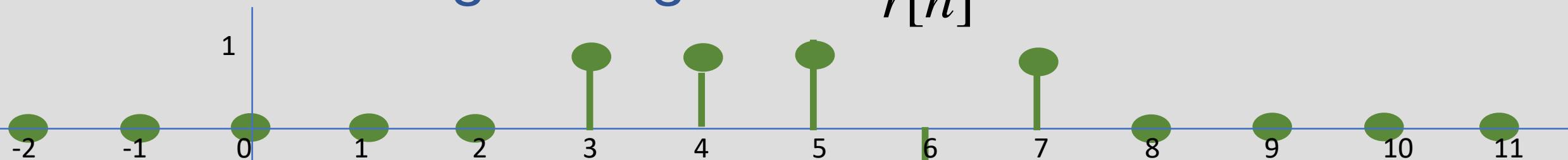
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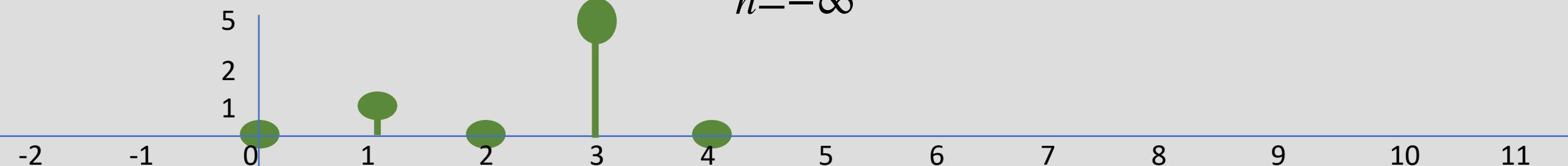
$$\text{corr}_{\vec{r}}(\vec{s})[3] = \langle r[n], s[n-3] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-3] = 5$$



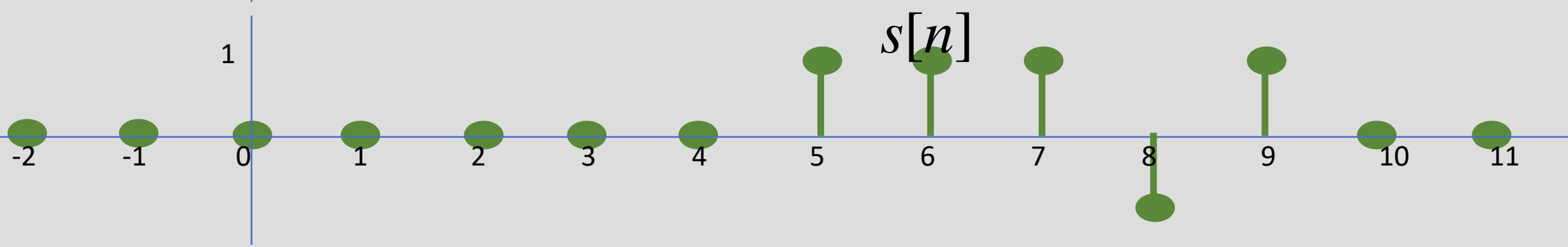
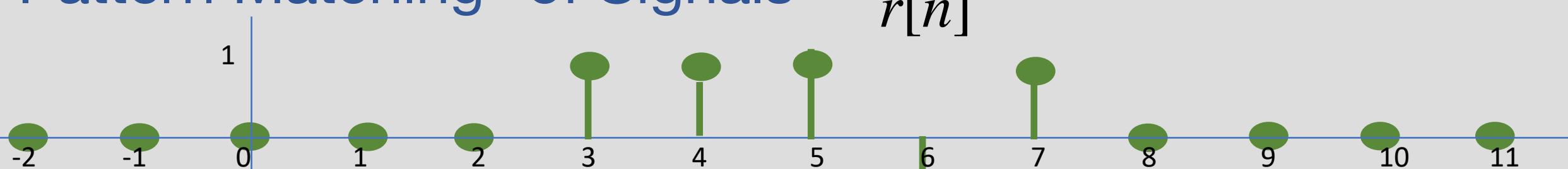
“Pattern Matching” of Signals



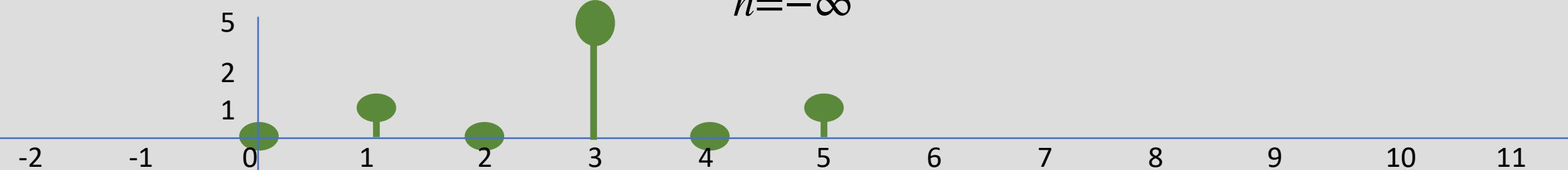
$$\text{corr}_{\vec{r}}(\vec{s})[4] = \langle r[n], s[n-4] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-4] = 0$$



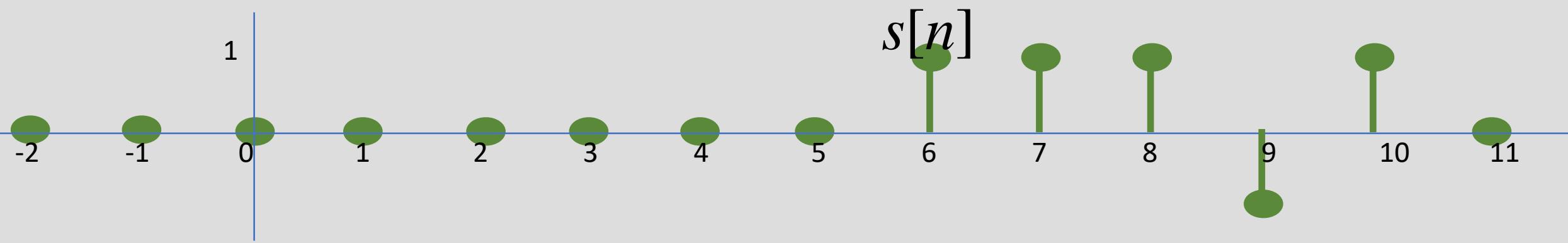
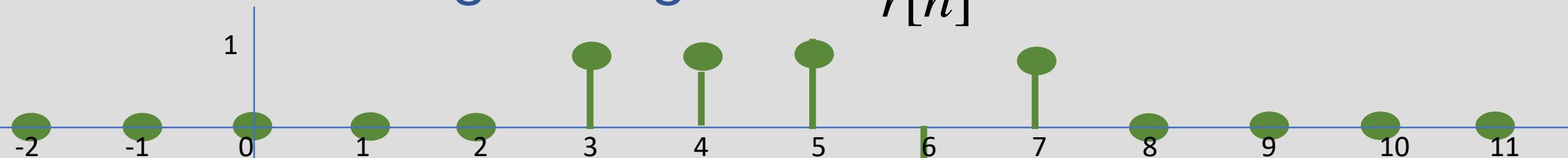
“Pattern Matching” of Signals



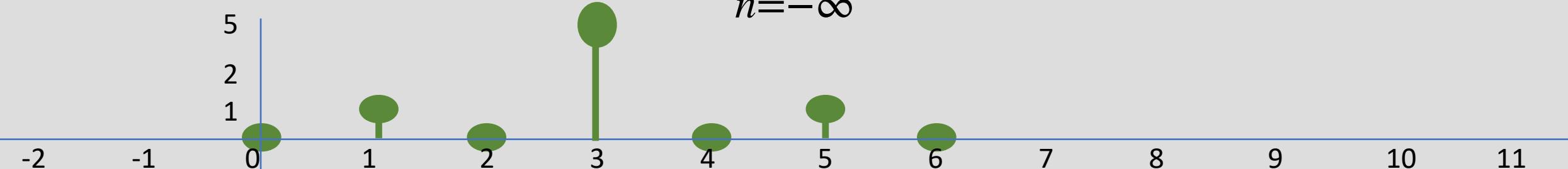
$$\text{corr}_{\vec{r}}(\vec{s})[5] = \langle r[n], s[n-5] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-5] = 1$$



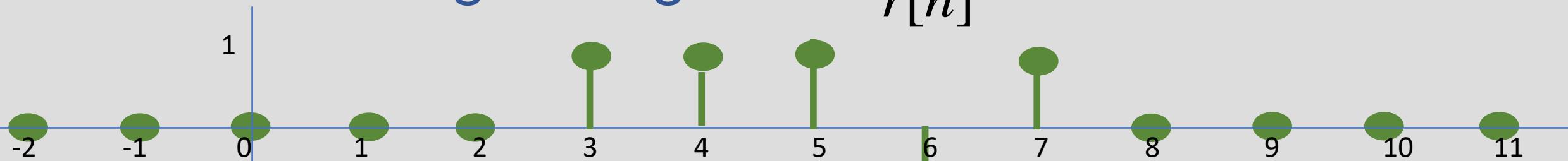
“Pattern Matching” of Signals



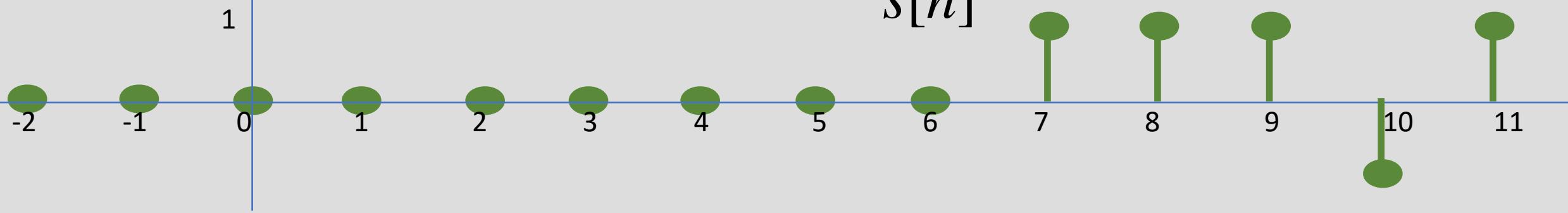
$$\text{corr}_{\vec{r}}(\vec{s})[6] = \langle r[n], s[n-6] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-6] = 0$$



“Pattern Matching” of Signals

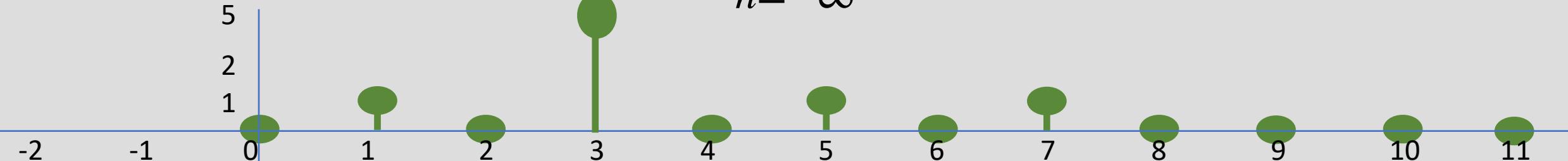


$r[n]$



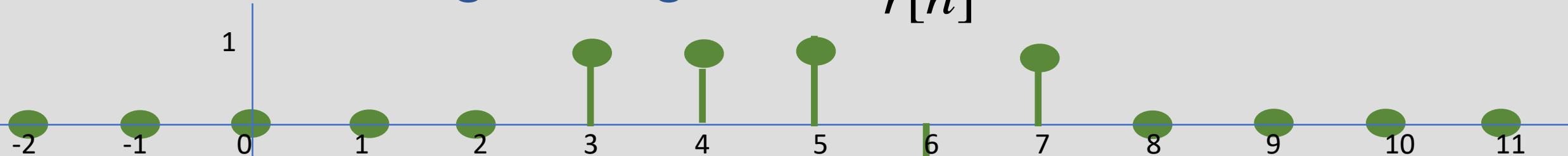
$s[n]$

$$\text{corr}_{\vec{r}}(\vec{s})[7] = \langle r[n], s[n-7] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-7] = 1$$



$n=-\infty$

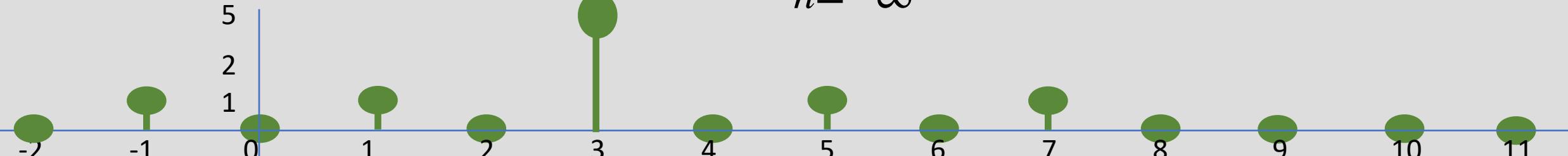
“Pattern Matching” of Signals



$r[n]$

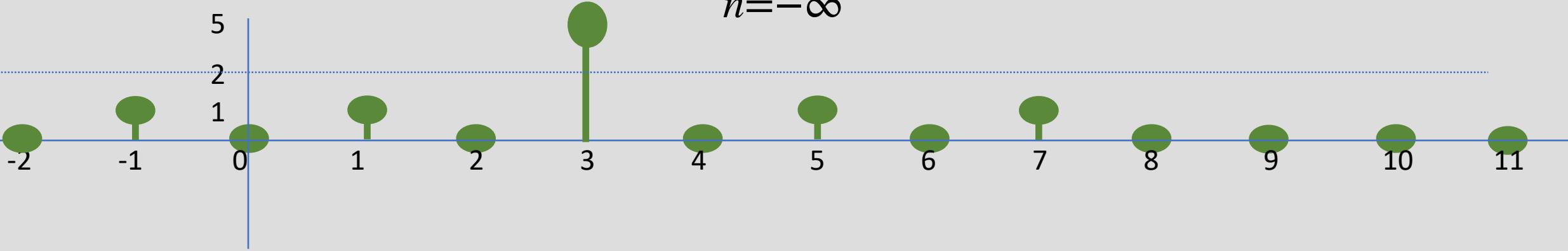
$s[n]$

$$\text{corr}_{\vec{r}}(\vec{s})[-1] = \langle r[n], s[n+1] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n+1] = 1$$



Cross Correlation

$$\text{corr}_{\vec{r}}(\vec{s})[k] = \langle r[n], s[n-k] \rangle = \sum_{n=-\infty}^{\infty} r[n]s[n-k]$$



$$k^* = \operatorname{argmax}_k \text{corr}_{\vec{r}}(\vec{s})[k]$$

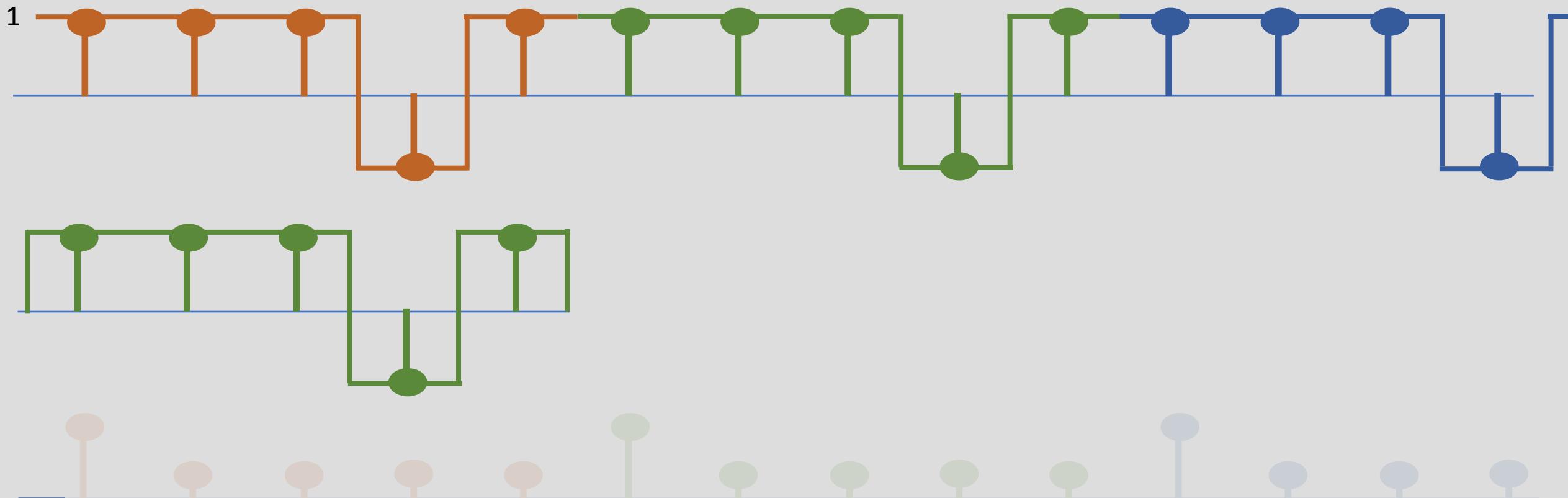
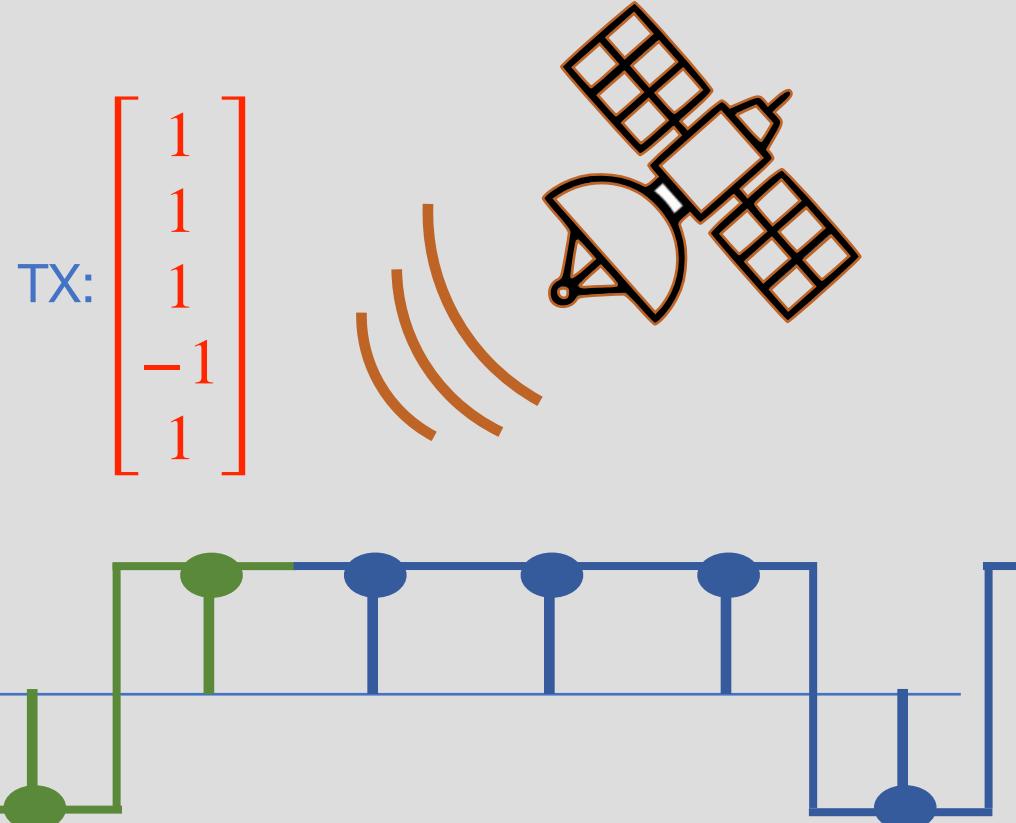
$$k^* = 3$$

Cross Correlation Properties

- If $\vec{x} \in \mathbb{R}^N$, and $\vec{y} \in \mathbb{R}^M$, then the length of $\text{corr}_{\vec{x}}(\vec{y})$ is $N + M - 1$
- $\text{corr}_{\vec{x}}(\vec{y}) \neq \text{corr}_{\vec{y}}(\vec{x})$
- $\text{corr}_{\vec{x}}(\vec{x})$ is called auto-correlation

Periodic Signals

- Satellites repeat the codes over and over
 - Cross correlation is “periodically expanded” instead of zero-padded
 - Result is periodic



Localization

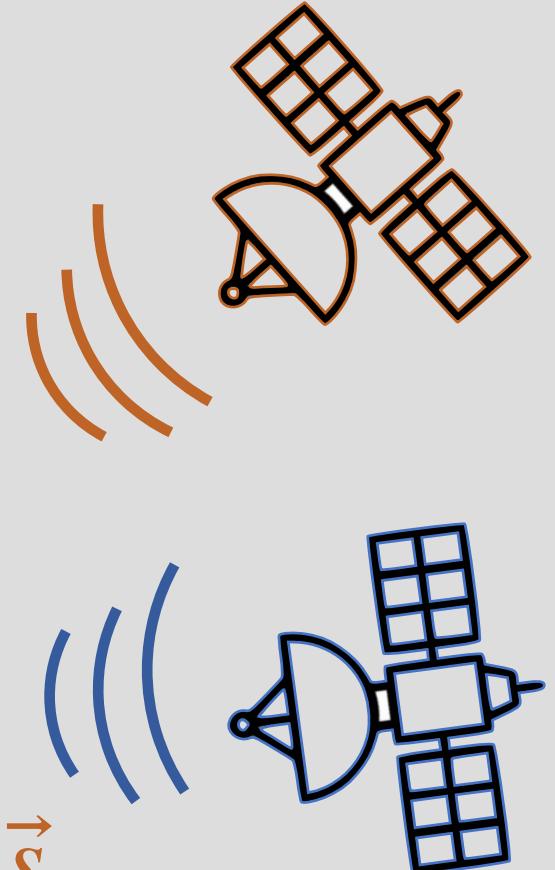
- Satellites transmit a unique code
 - Radio signal
- Signal is received and digitized by a receiver



Two problems:

1. Interference
2. Timing

What are good properties for the codes \vec{s}_i



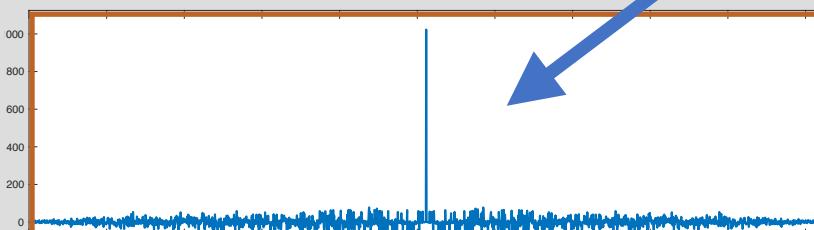
Received Signal

$$r[n] = s_1[n - \tau_1] + s_2[n - \tau_2] + w[n]$$

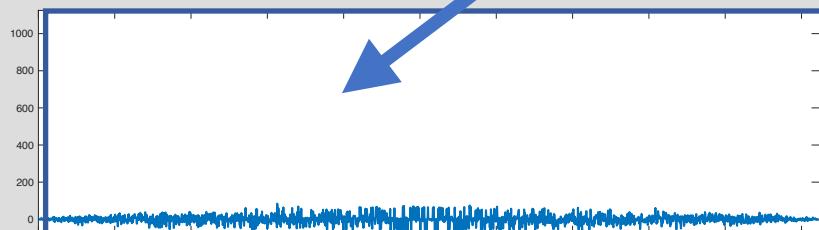
Correlate with $s_1[n]$:

$$\text{corr}_{\vec{r}}(\vec{s}_1)[k] = \langle r[n], s_1[n - k] \rangle$$

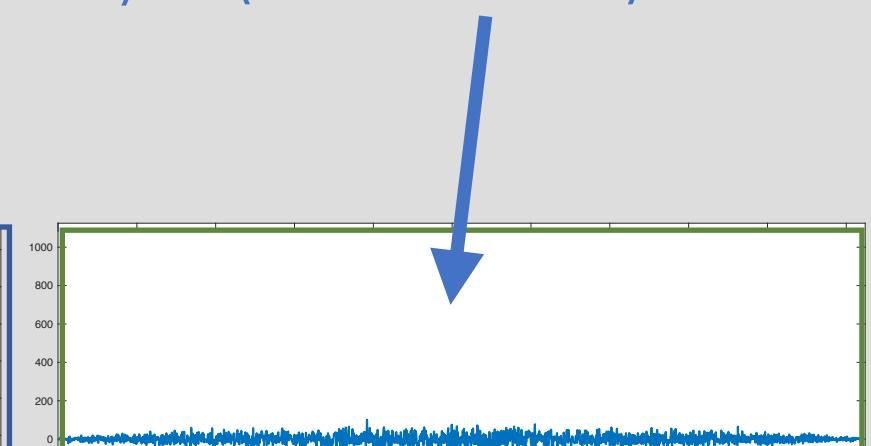
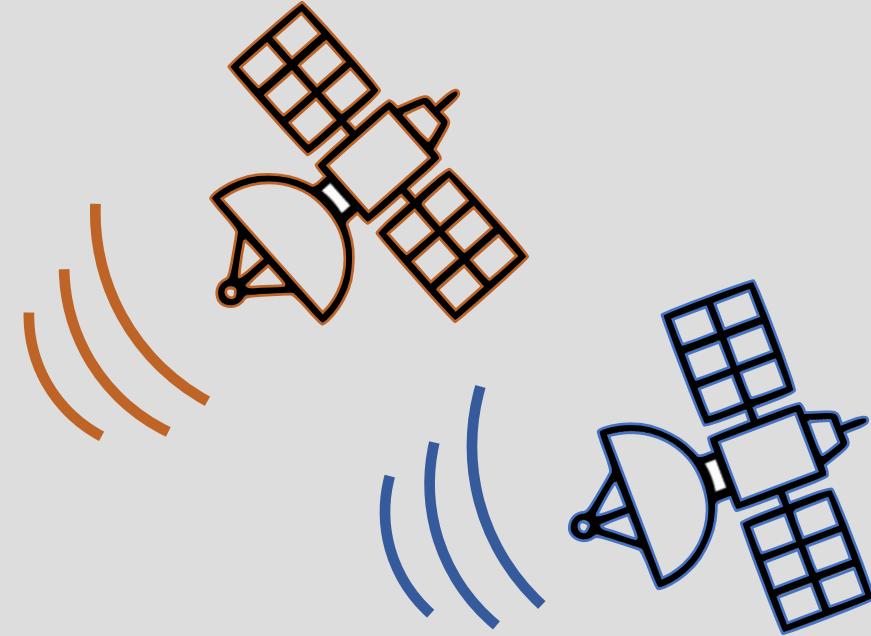
$$= \langle s_1[n - \tau_1], s_1[n - k] \rangle + \langle s_2[n - \tau_2], s_1[n - k] \rangle + \langle w[n], s_1[n - k] \rangle$$



Auto-correlation looks like an impulse



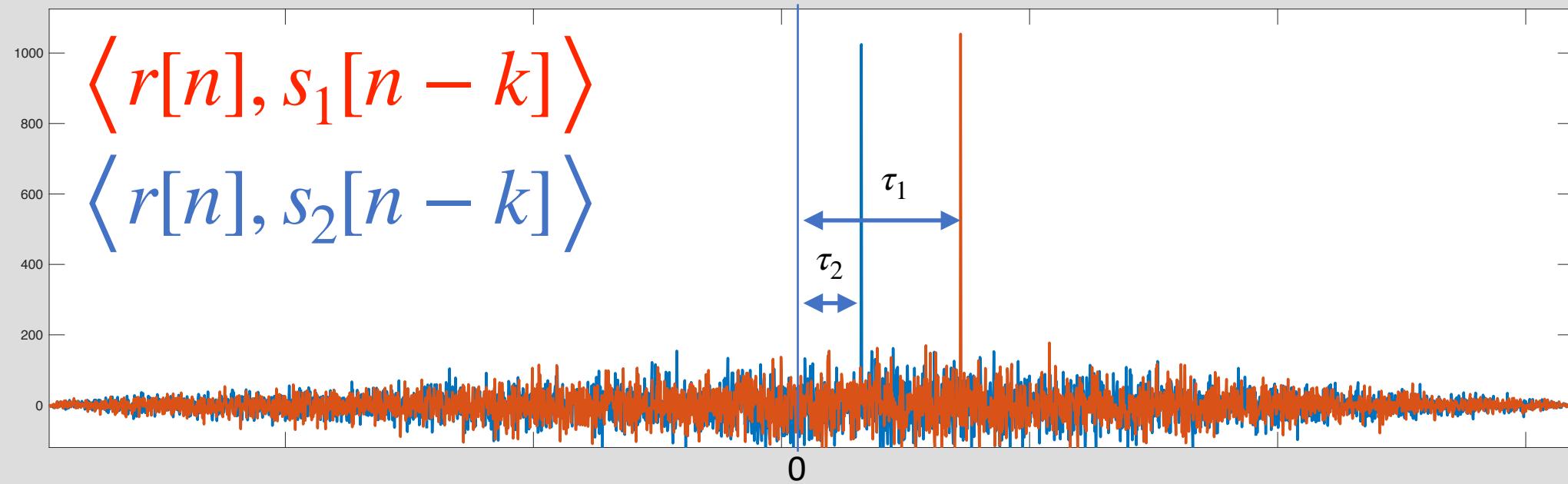
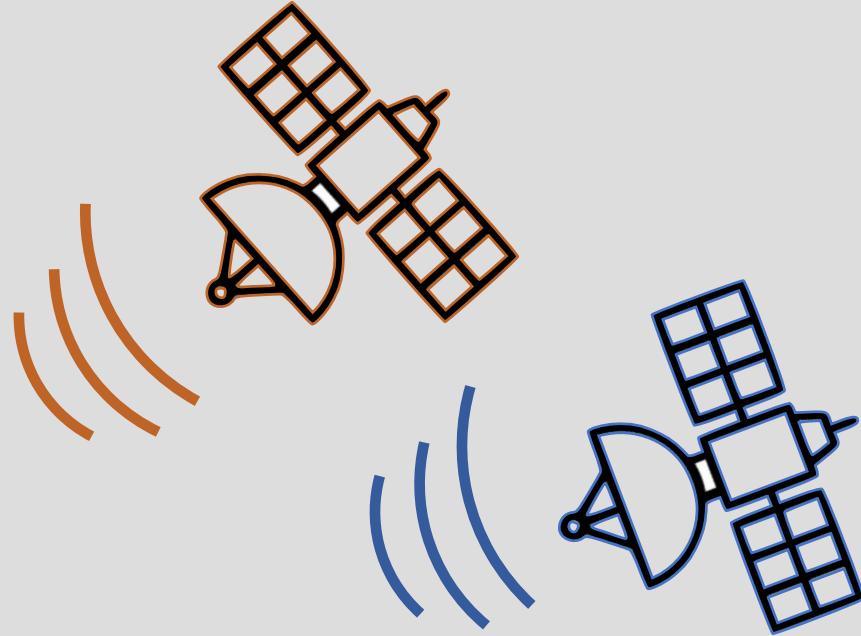
cross-correlation is small



cross-correlation with noise is small
(always true)

Received Signal

$$r[n] = s_1[n - \tau_1] + s_2[n - \tau_2] + w[n]$$

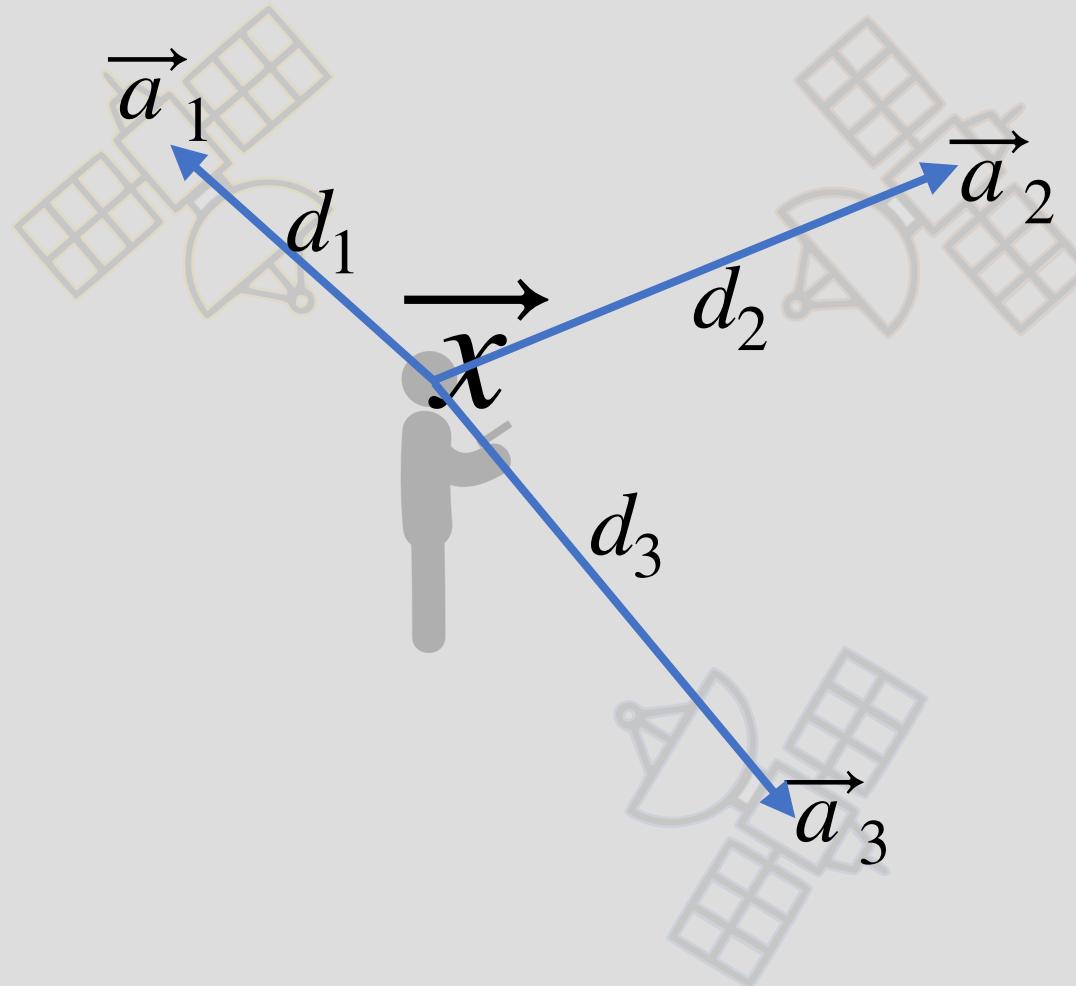


Trilateration

$$(1) \quad \|\vec{x} - \vec{a}_1\|^2 = d_1^2$$

$$(2) \quad \|\vec{x} - \vec{a}_2\|^2 = d_2^2$$

$$(3) \quad \|\vec{x} - \vec{a}_3\|^2 = d_3^2$$



$$d_1 = \tau_1 C$$

$$d_2 = \tau_2 C$$

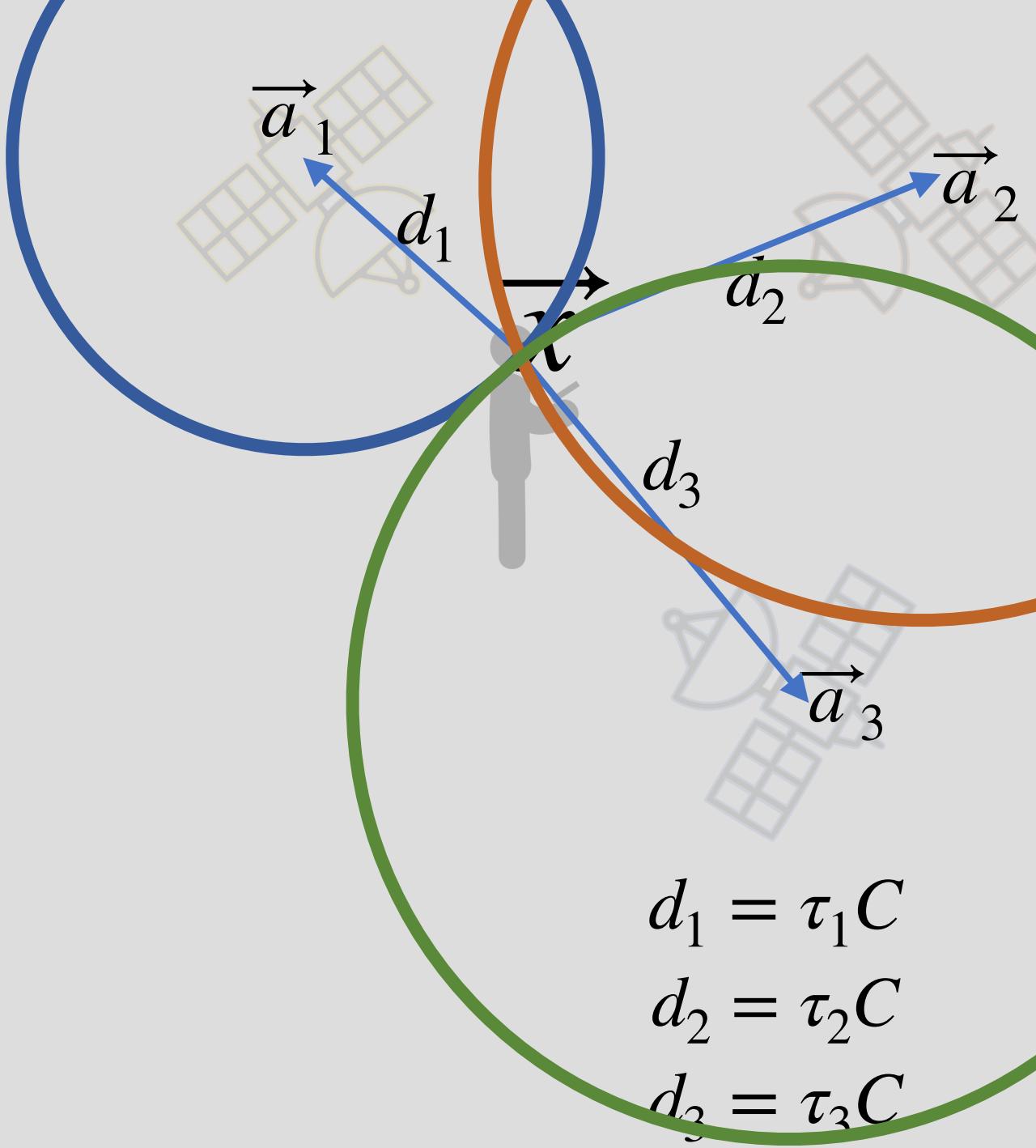
$$d_3 = \tau_3 C$$

Trilateration

$$(1) \quad \|\vec{x} - \vec{a}_1\|^2 = d_1^2$$

$$(2) \quad \|\vec{x} - \vec{a}_2\|^2 = d_2^2$$

$$(3) \quad \|\vec{x} - \vec{a}_3\|^2 = d_3^2$$



Trilateration

$$(1) \quad \|\vec{x} - \vec{a}_1\|^2 = d_1^2$$

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$$(3) \quad \|\vec{x} - \vec{a}_3\|^2 = d_3^2$$

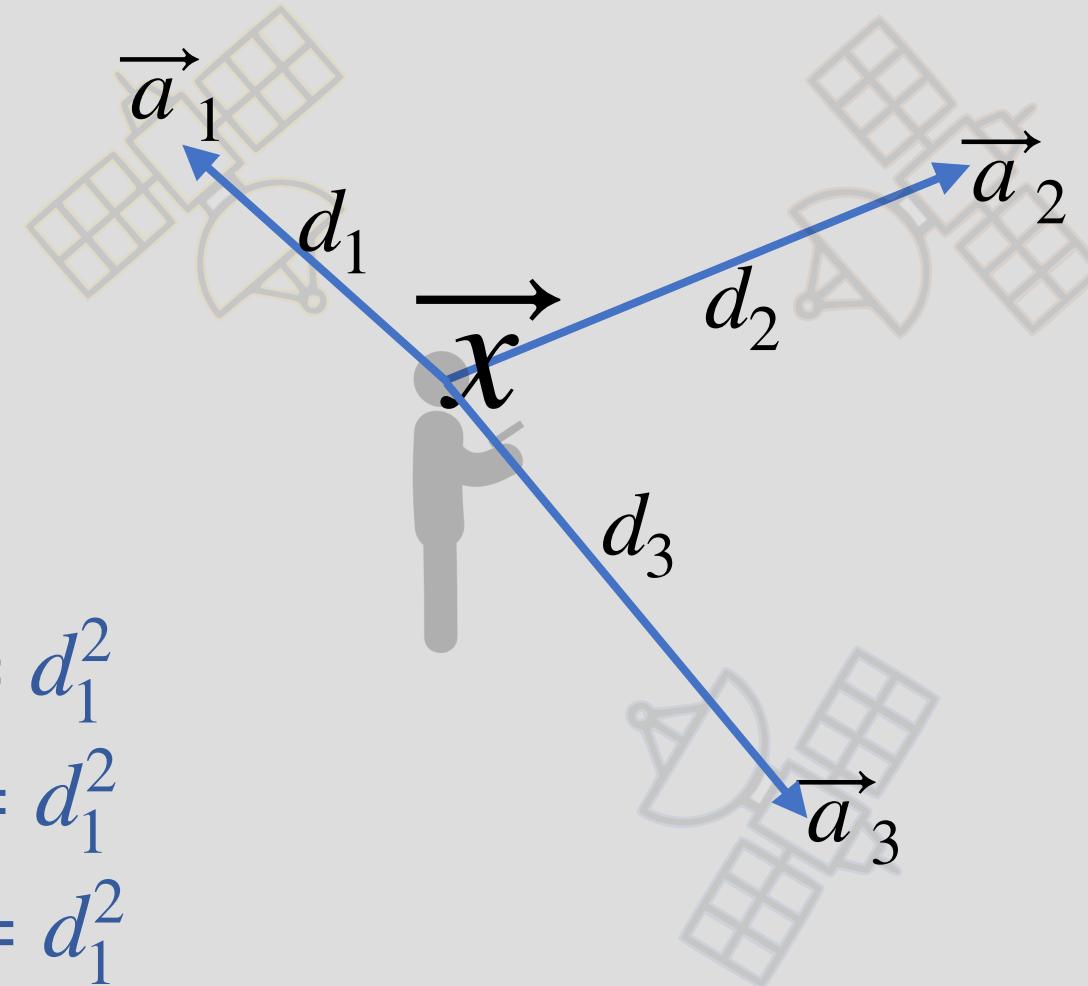
$$\|\vec{x} - \vec{a}_1\|^2 = d_1^2$$

$$(\vec{x} - \vec{a}_1)^T (\vec{x} - \vec{a}_1) = d_1^2$$

$$\vec{x}^T \vec{x} - \vec{a}_1^T \vec{x} - \vec{x}^T \vec{a}_1 + \vec{a}_1^T \vec{a}_1 = d_1^2$$

$$\|\vec{x}\|^2 - 2\vec{a}_1^T \vec{x} + \|\vec{a}_1\|^2 = d_1^2$$

$$\|\vec{x}\|^2 - 2\vec{a}_1^T \vec{x} + \|\vec{a}_1\|^2 = C^2 \tau_1^2$$



$$d_1 = \tau_1 C$$

$$d_2 = \tau_2 C$$

$$d_3 = \tau_3 C$$

Trilateration

$$(1) \quad \|\vec{x} - \vec{a}_1\|^2 = d_1^2$$

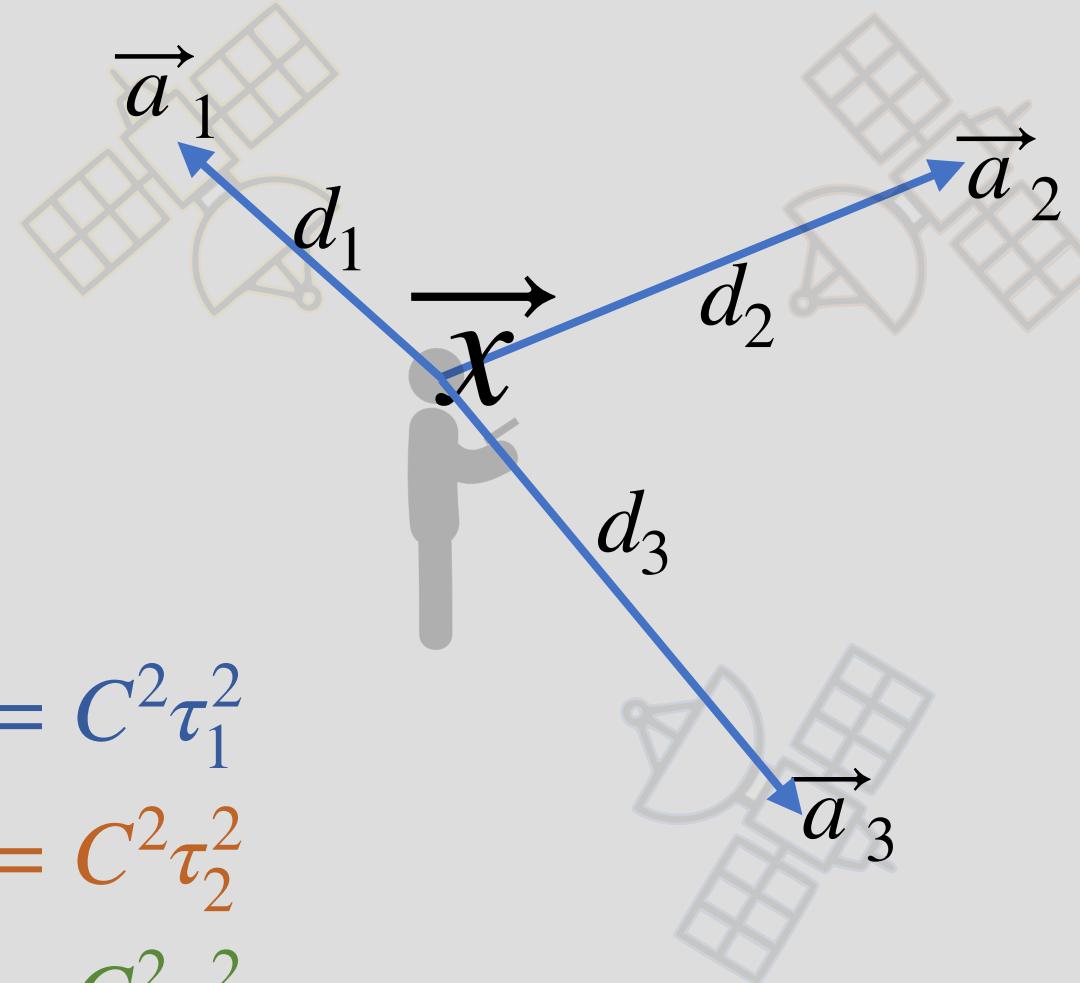
$$(2) \quad \|\vec{x} - \vec{a}_2\|^2 = d_2^2$$

$$(3) \quad \|\vec{x} - \vec{a}_3\|^2 = d_3^2$$

$$(1) \quad \|\vec{x}\|^2 - 2\vec{a}_1^T \vec{x} + \|\vec{a}_1\|^2 = C^2 \tau_1^2$$

$$(2) \quad \|\vec{x}\|^2 - 2\vec{a}_2^T \vec{x} + \|\vec{a}_2\|^2 = C^2 \tau_2^2$$

$$(3) \quad \|\vec{x}\|^2 - 2\vec{a}_3^T \vec{x} + \|\vec{a}_3\|^2 = C^2 \tau_3^2$$



$$d_1 = \tau_1 C$$

$$d_2 = \tau_2 C$$

$$d_3 = \tau_3 C$$

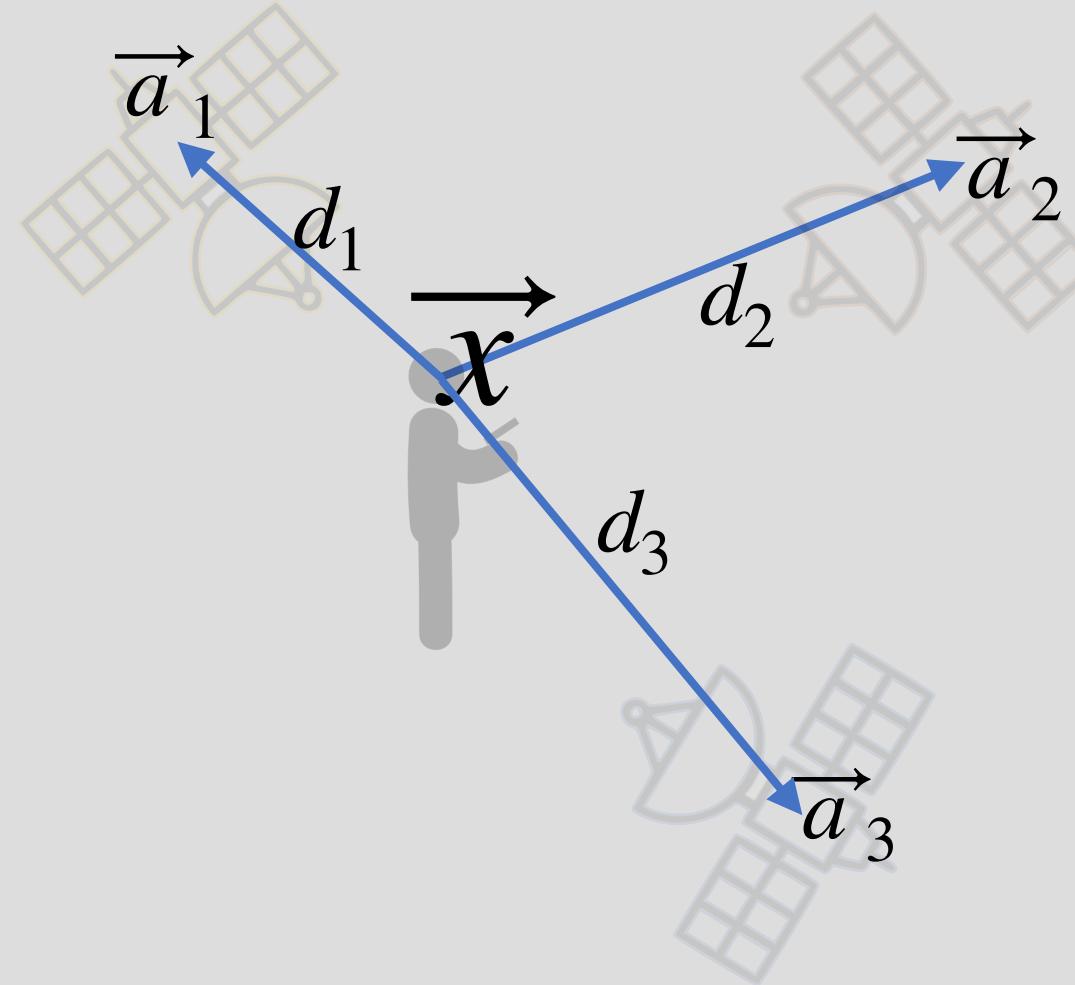
Trilateration

$$(1) \quad \|\vec{x}\|^2 - 2\vec{a}_1^T \vec{x} + \|\vec{a}_1\|^2 = C^2 \tau_1^2$$

$$(2) \quad \|\vec{x}\|^2 - 2\vec{a}_2^T \vec{x} + \|\vec{a}_2\|^2 = C^2 \tau_2^2$$

$$(3) \quad \|\vec{x}\|^2 - 2\vec{a}_3^T \vec{x} + \|\vec{a}_3\|^2 = C^2 \tau_3^2$$

$$(2) - (1)$$



Trilateration

$$(1) \|\vec{x}\|^2 - 2\vec{a}_1^T \vec{x} + \|\vec{a}_1\|^2 = C^2 \tau_1^2$$

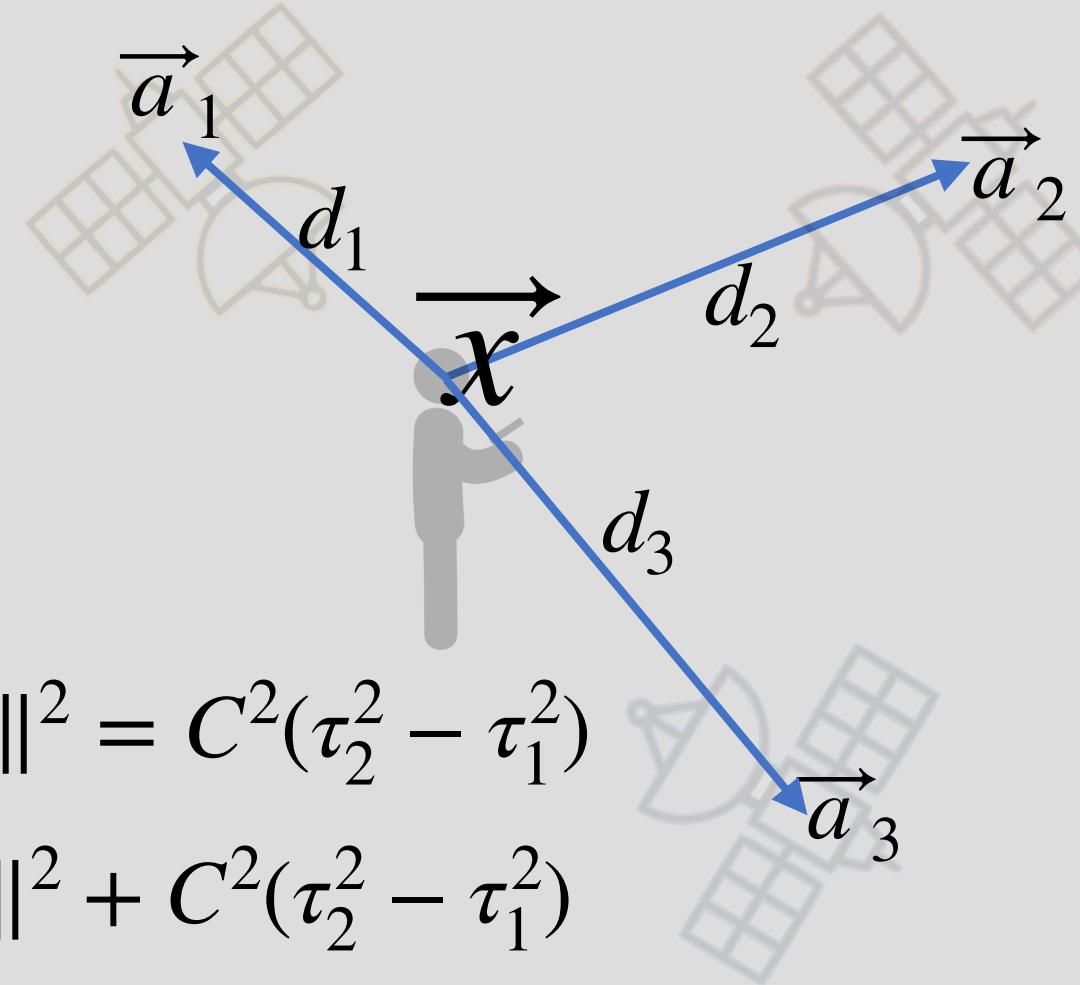
$$(2) \|\vec{x}\|^2 - 2\vec{a}_2^T \vec{x} + \|\vec{a}_2\|^2 = C^2 \tau_2^2$$

$$(3) \|\vec{x}\|^2 - 2\vec{a}_3^T \vec{x} + \|\vec{a}_3\|^2 = C^2 \tau_3^2$$

$$(2) - (1) -2\vec{a}_2^T \vec{x} + 2\vec{a}_1^T \vec{x} + \|\vec{a}_2\|^2 - \|\vec{a}_1\|^2 = C^2(\tau_2^2 - \tau_1^2)$$

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2^2 - \tau_1^2)$$

$$(3) - (1) 2(\vec{a}_1 - \vec{a}_3)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2(\tau_3^2 - \tau_1^2)$$

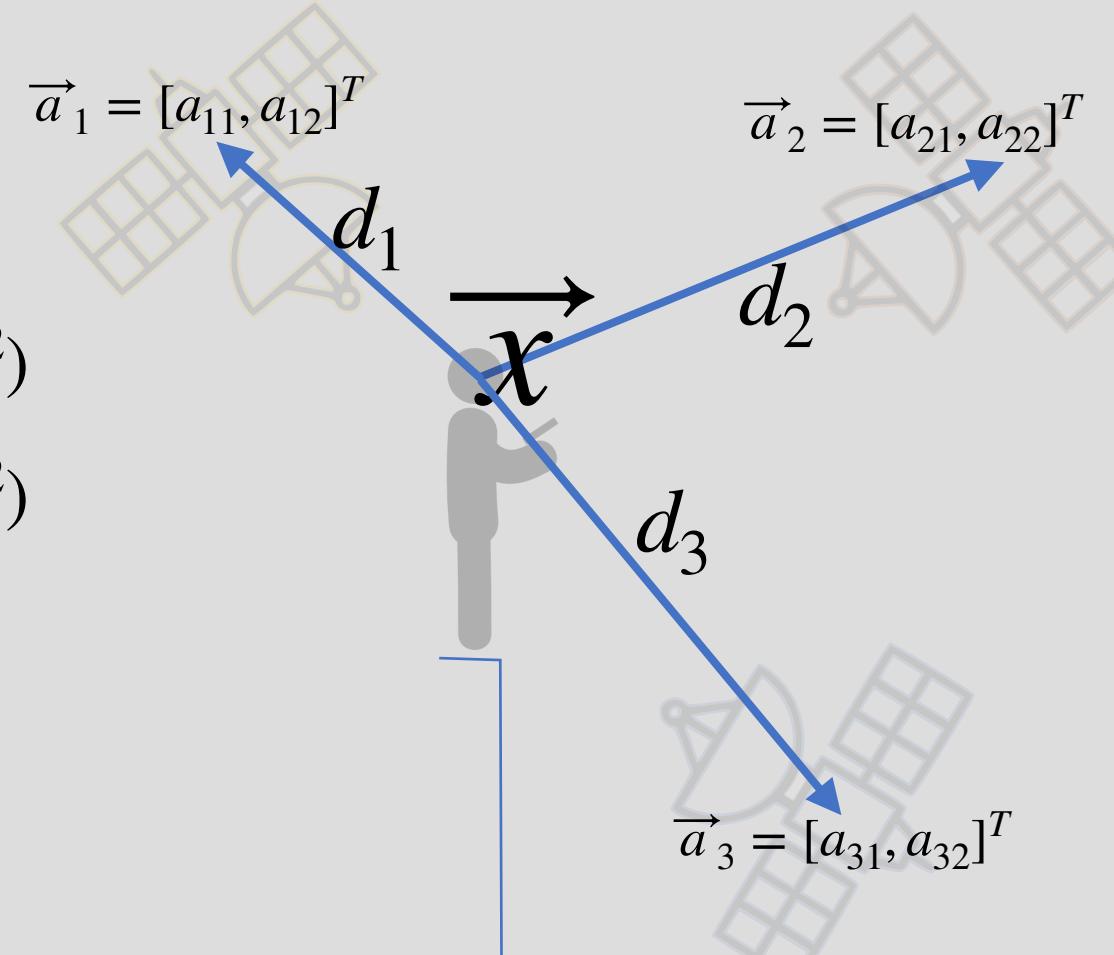


Trilateration

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2^2 - \tau_1^2)$$

$$2(\vec{a}_1 - \vec{a}_3)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2(\tau_3^2 - \tau_1^2)$$

$$\left[\begin{array}{c|c|c} & & \\ & & \\ & & \\ \hline & & \\ & & \\ & & \\ \hline & & \\ & & \\ & & \end{array} \right] = \left[\begin{array}{c} \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \end{array} \right]$$

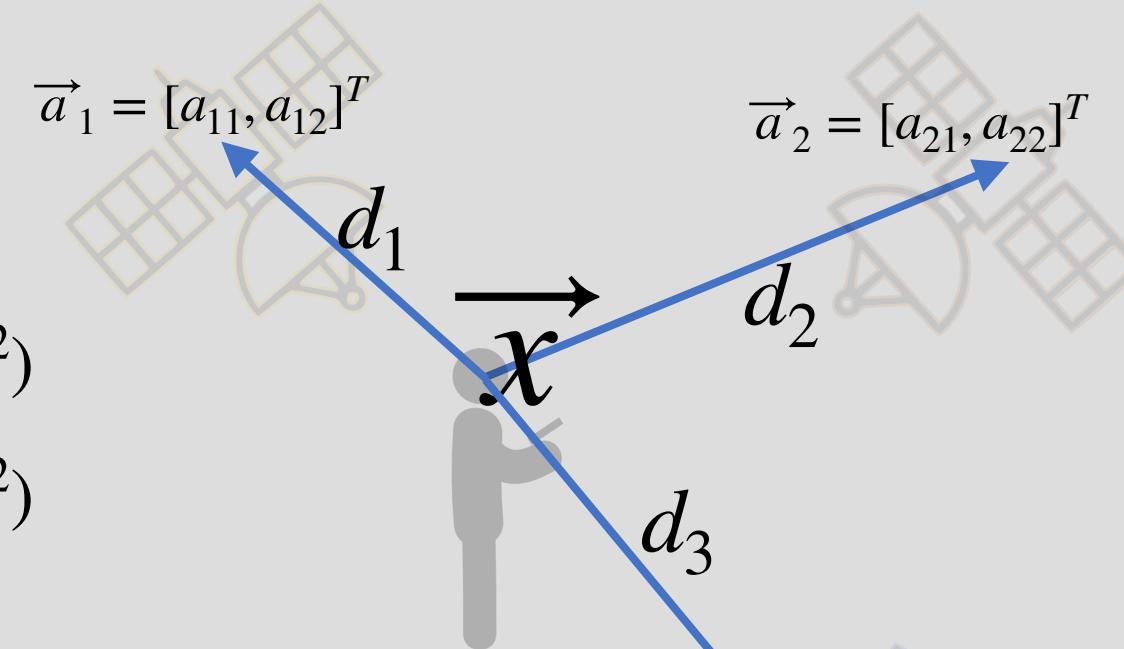


Trilateration

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2^2 - \tau_1^2)$$

$$2(\vec{a}_1 - \vec{a}_3)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2(\tau_3^2 - \tau_1^2)$$

$$2 \begin{bmatrix} a_{11} - a_{21} & a_{12} - a_{22} \\ a_{11} - a_{31} & a_{12} - a_{32} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2^2 - \tau_1^2) \\ \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2(\tau_3^2 - \tau_1^2) \end{bmatrix}$$

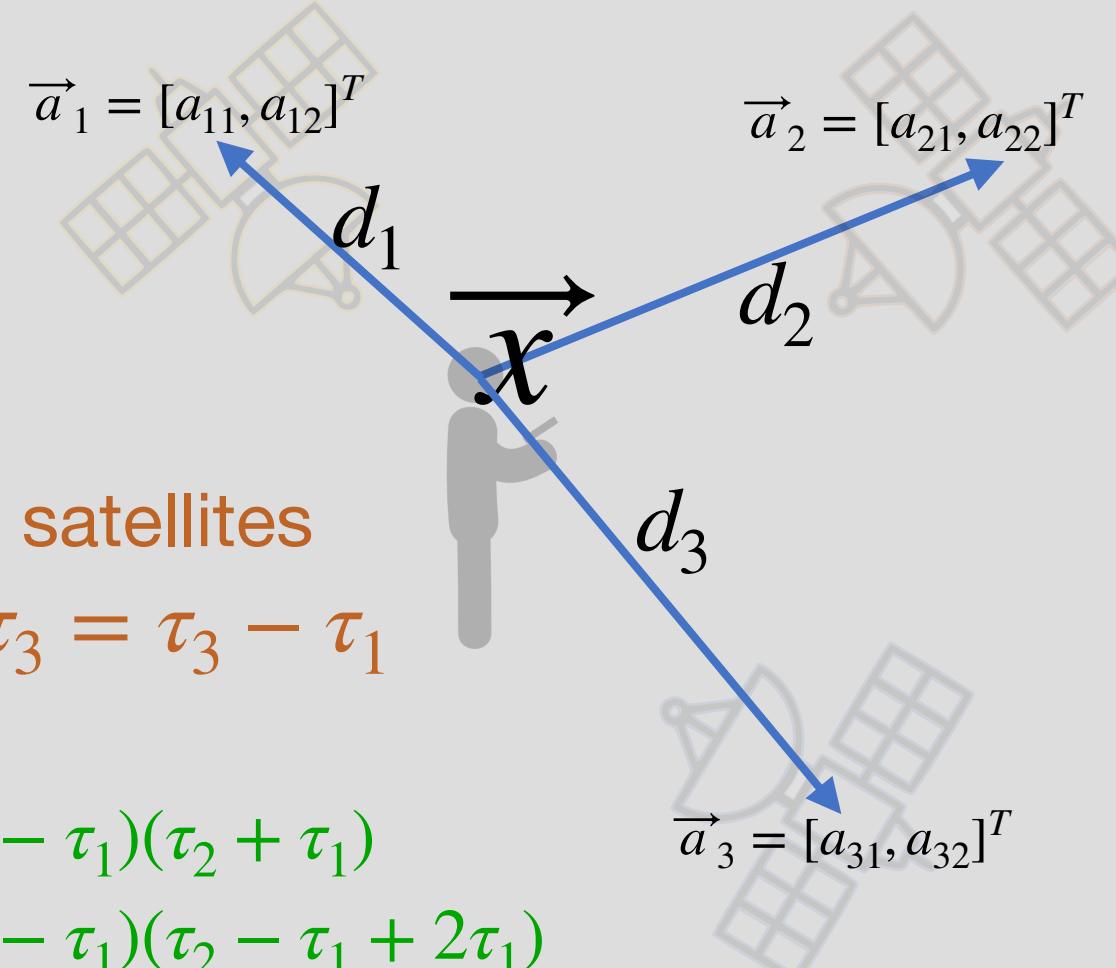


Solve via gaussian elimination!

Trilateration

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2^2 - \tau_1^2)$$

$$2(\vec{a}_1 - \vec{a}_3)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2(\tau_3^2 - \tau_1^2)$$



Problem – receiver clock is not synced to satellites

τ_1 is unknown, but $\Delta\tau_2 = \tau_2 - \tau_1$, and $\Delta\tau_3 = \tau_3 - \tau_1$ are known

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2 - \tau_1)(\tau_2 + \tau_1)$$

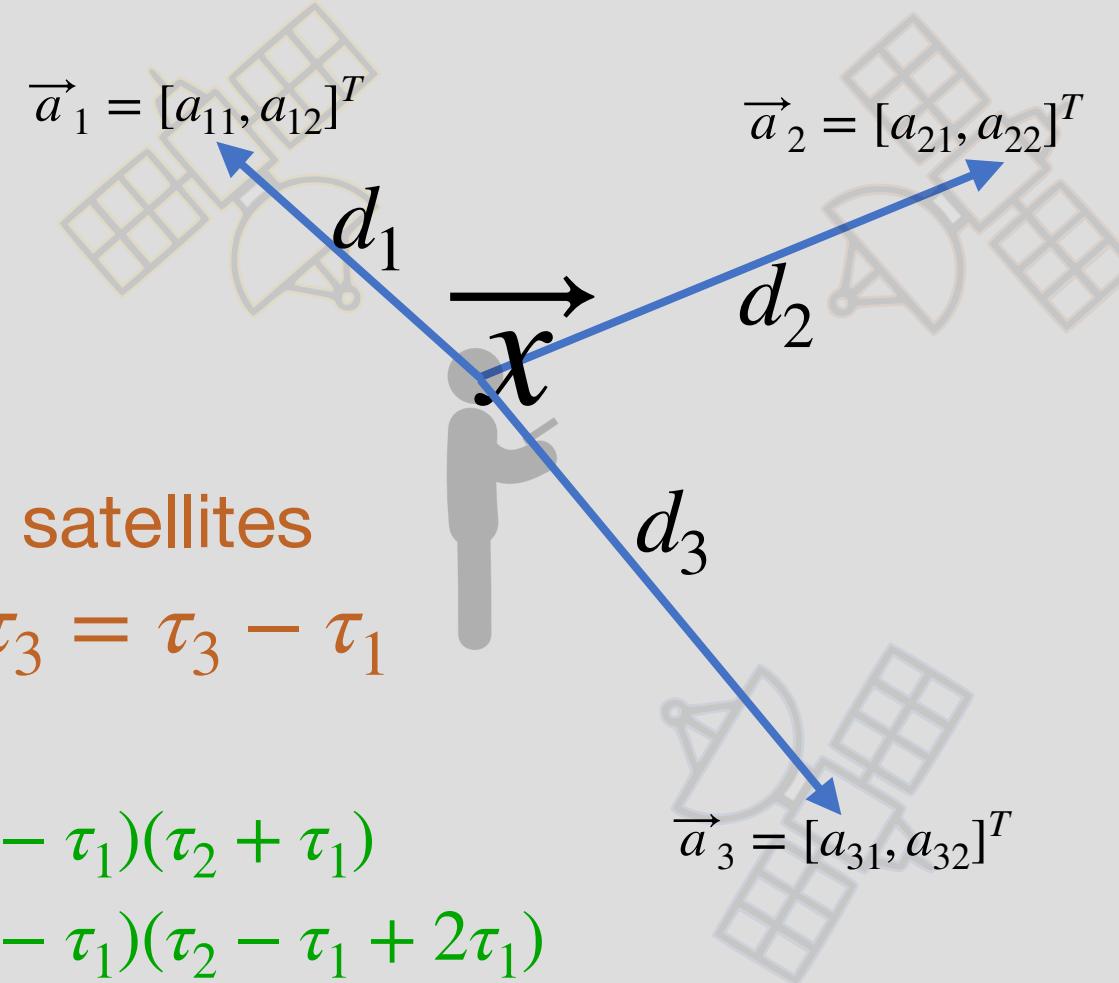
$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2 - \tau_1)(\tau_2 - \tau_1 + 2\tau_1)$$

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\Delta\tau_2)(\Delta\tau_2 + 2\tau_1)$$

Trilateration

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2^2 - \tau_1^2)$$

$$2(\vec{a}_1 - \vec{a}_3)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2(\tau_3^2 - \tau_1^2)$$



Problem – receiver clock is not synced to satellites

τ_1 is unknown, but $\Delta\tau_2 = \tau_2 - \tau_1$, and $\Delta\tau_3 = \tau_3 - \tau_1$ are known

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2 - \tau_1)(\tau_2 + \tau_1)$$

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\tau_2 - \tau_1)(\tau_2 - \tau_1 + 2\tau_1)$$

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\Delta\tau_2)(\Delta\tau_2 + 2\tau_1)$$

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} - 2C^2\Delta\tau_2 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\Delta\tau_2)^2$$

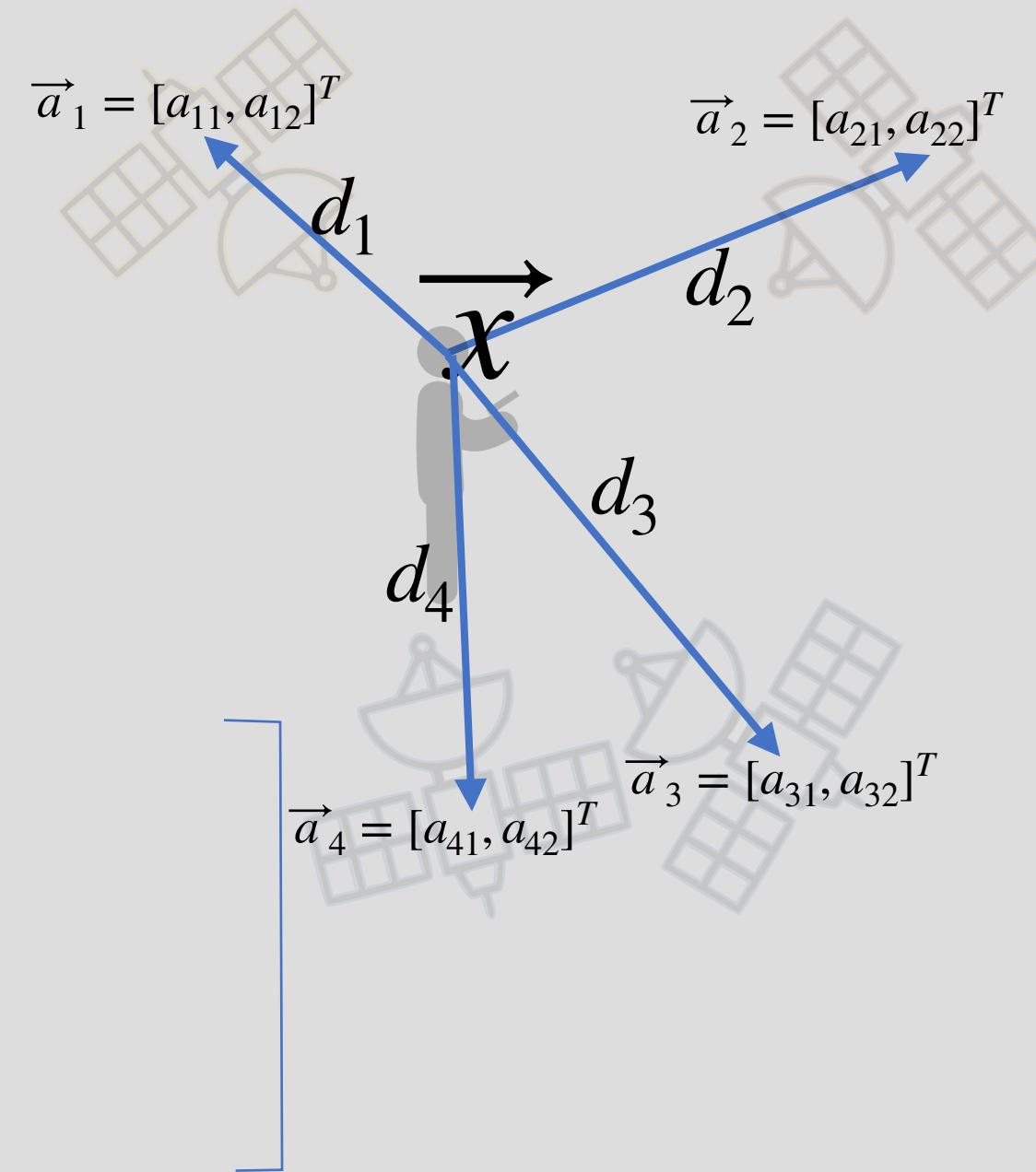
Another variable! Need 1 more equation (satellite)

Trilateration

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} - 2C^2 \Delta\tau_2 \textcolor{red}{\tau_1} = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2(\Delta\tau_2)^2$$

$$2(\vec{a}_1 - \vec{a}_3)^T \vec{x} - 2C^2 \Delta\tau_3 \textcolor{red}{\tau_1} = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2(\Delta\tau_3)^2$$

$$2(\vec{a}_1 - \vec{a}_4)^T \vec{x} - 2C^2 \Delta\tau_4 \textcolor{red}{\tau_1} = \|\vec{a}_1\|^2 - \|\vec{a}_4\|^2 + C^2(\Delta\tau_4)^2$$



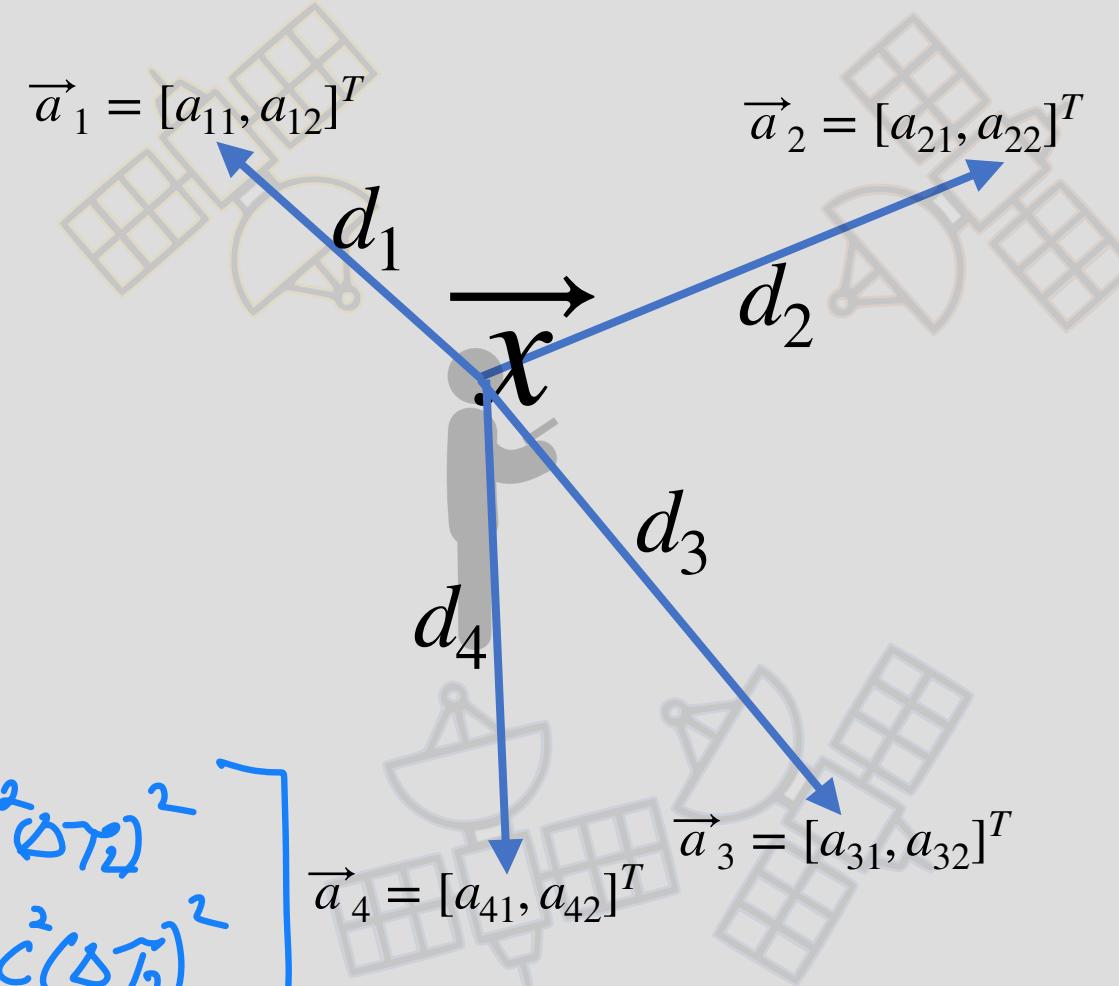
Trilateration

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} - 2C^2 \Delta \tau_2 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2 (\Delta \tau_2)^2$$

$$2(\vec{a}_1 - \vec{a}_3)^T \vec{x} - 2C^2 \Delta \tau_3 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2 (\Delta \tau_3)^2$$

$$2(\vec{a}_1 - \vec{a}_4)^T \vec{x} - 2C^2 \Delta \tau_4 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_4\|^2 + C^2 (\Delta \tau_4)^2$$

$$2 \begin{bmatrix} a_{11} - a_{21} & a_{12} - a_{22} & -C^2 \Delta \tau_2 \\ a_{11} - a_{31} & a_{12} - a_{32} & -C^2 \Delta \tau_3 \\ a_{11} - a_{41} & a_{12} - a_{42} & -C^2 \Delta \tau_4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \tau_1 \end{bmatrix} = \begin{bmatrix} \|a_1\|^2 - \|a_2\|^2 + C^2 (\Delta \tau_2)^2 \\ \|a_1\|^2 - \|a_3\|^2 + C^2 (\Delta \tau_3)^2 \\ \|a_1\|^2 - \|a_4\|^2 + C^2 (\Delta \tau_4)^2 \end{bmatrix}$$



Multi-Lateration

$$2(\vec{a}_1 - \vec{a}_2)^T \vec{x} - 2C^2 \Delta \tau_2 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_2\|^2 + C^2 (\Delta \tau_2)^2$$

$$2(\vec{a}_1 - \vec{a}_3)^T \vec{x} - 2C^2 \Delta \tau_3 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_3\|^2 + C^2 (\Delta \tau_3)^2$$

$$2(\vec{a}_1 - \vec{a}_4)^T \vec{x} - 2C^2 \Delta \tau_4 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_4\|^2 + C^2 (\Delta \tau_4)^2$$

$$2(\vec{a}_1 - \vec{a}_5)^T \vec{x} - 2C^2 \Delta \tau_5 \tau_1 = \|\vec{a}_1\|^2 - \|\vec{a}_5\|^2 + C^2 (\Delta \tau_5)^2$$

More equations than unknowns

Q: With noise, equations will be inconsistent!

A: Find closest solution with Least-Squares!

