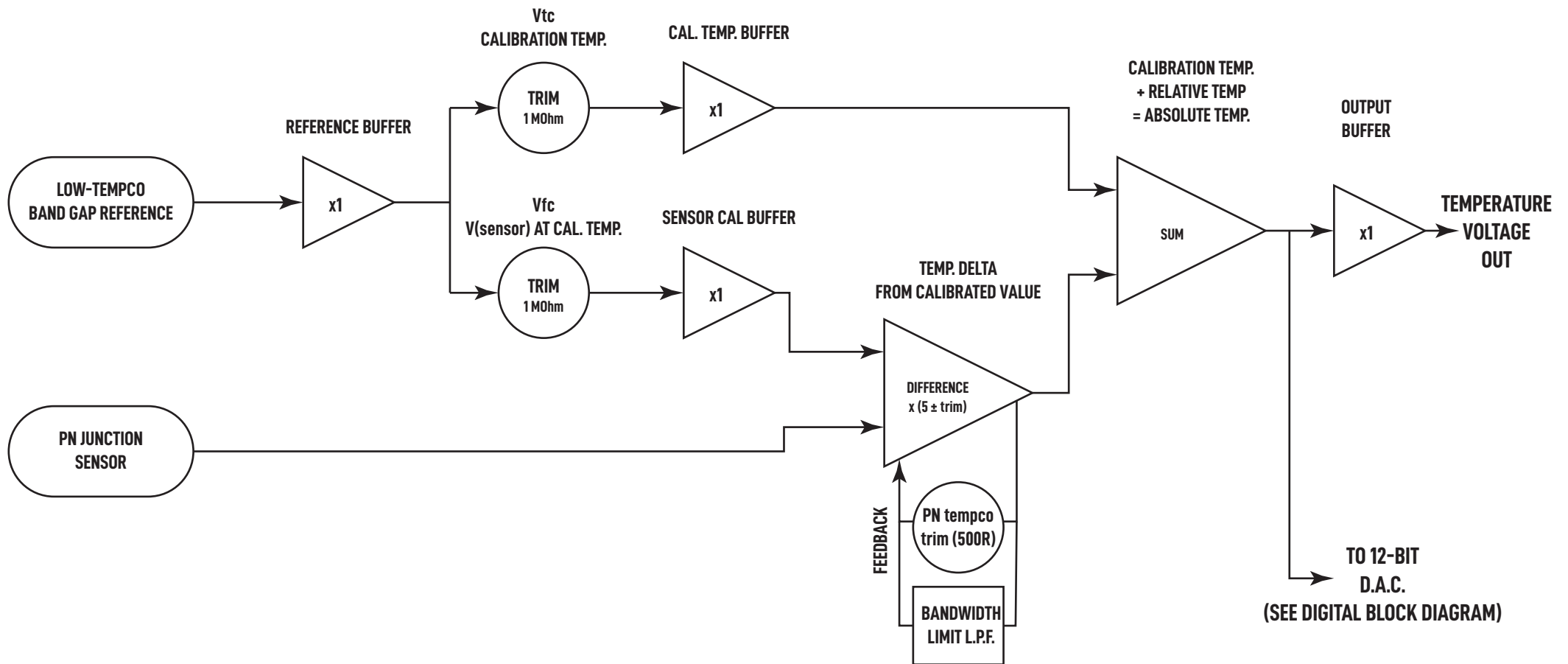


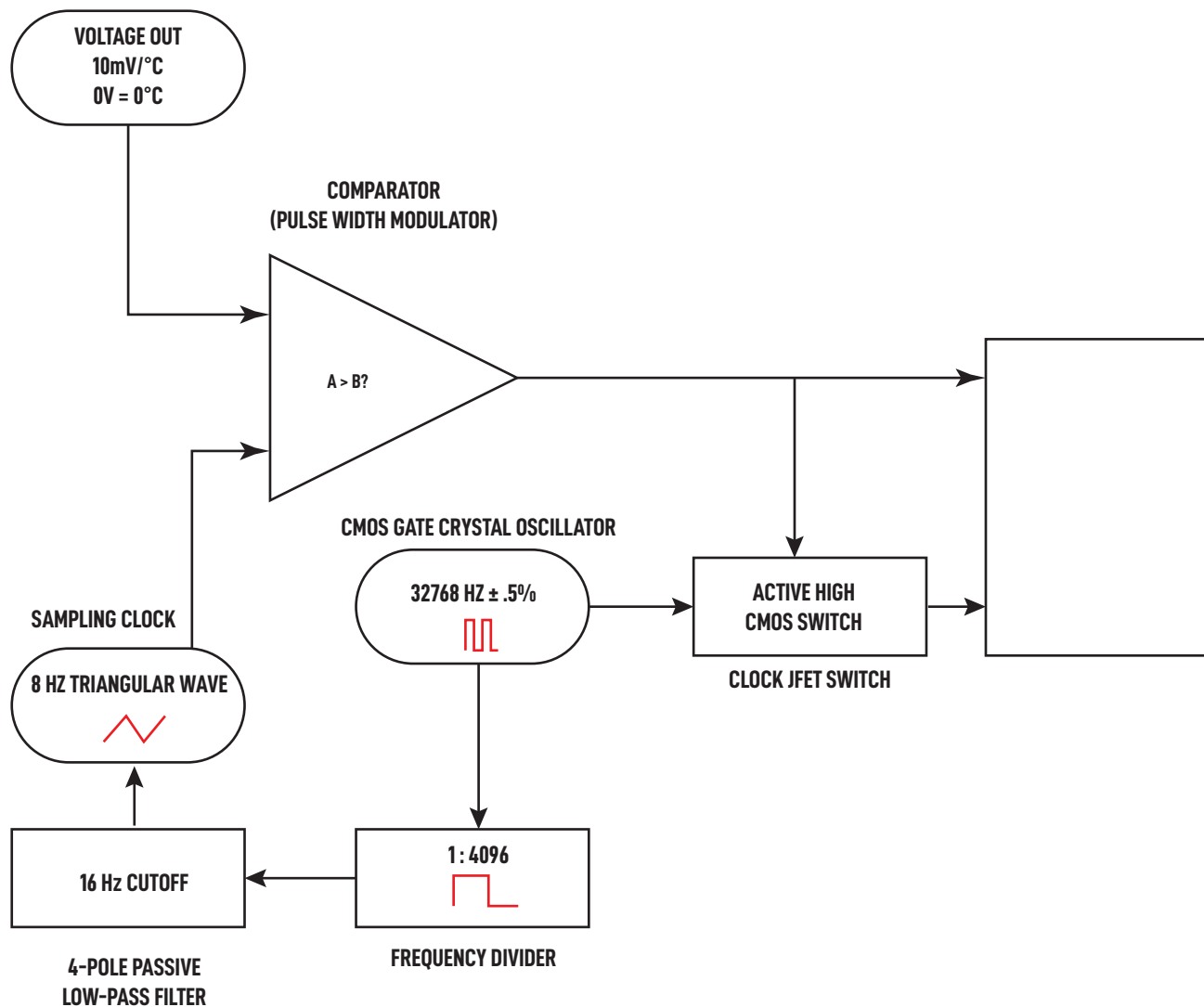
SYSTEM BLOCK DIAGRAM

ANALOG SIGNAL PATH



SYSTEM BLOCK DIAGRAM

13-BIT ANALOG-DIGITAL CONVERTER



The diagram shows a 100 kHz oscillator circuit. A +3V supply is connected to the VDD pin (pin 8) of the LMC7660 (IC1). The output of the oscillator (pin 7, OUT) is connected to a -3V supply through a 47μF capacitor (C2). The LMC7660 is also connected to a 470pF capacitor (C2) between pins 3 (GND) and 7 (OSC). The LMC7660 is connected to a 15μF Tantalum capacitor (C1) between pins 2 (CAP+) and 4 (CAP-).

+3

R2
820R
Carbon film (not metal film)

J1
P1086

IBIAS
1mA

CONN1
Sensor+

SENS1
2N3904
b/e junction diode,
TO-18 preferred

CONN2
Sensor-

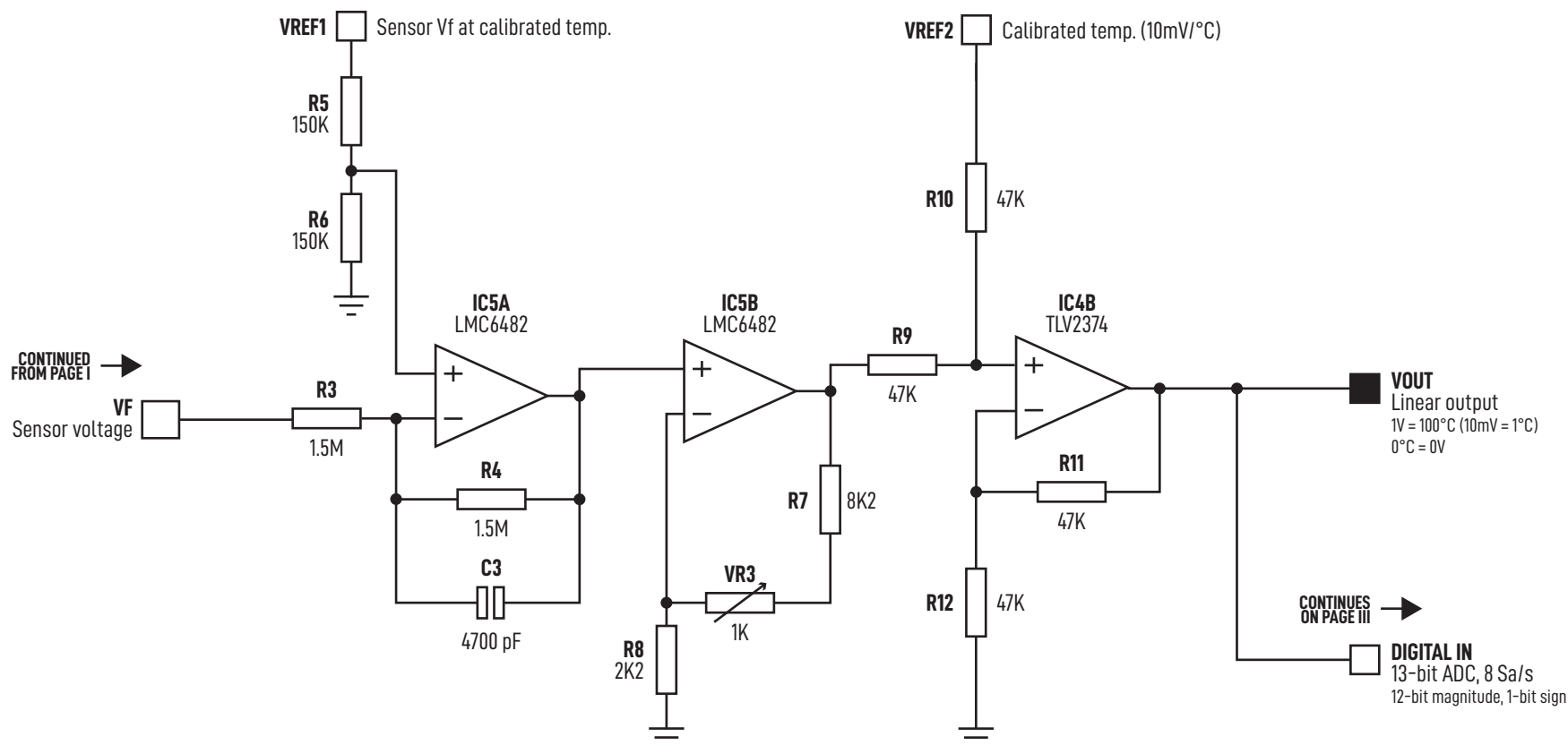
VF
Sensor voltage

CONTINUED ON PAGE II

JFETs have a negative temperature coefficient; their transconductance decreases with increased temperature. Carbon film resistors, too, have a negative temperature coefficient, causing a decrease in resistance – or an increase in conductance. Therefore, the two temperature coefficients balance out making the sensor drive constant current source stable across use ambient temperatures.

TO-18 devices are preferred for the sensor transistor diode (over TO-92s) since their metal case permit superior thermal contact to the sensing junction.

The calibration is two-part; first, the ambient temperature is measured, and **VR2** adjusted so **VREF2** (measured at the op amp output) is equal to $T_{AMB} \cdot 10mV/^{\circ}C$. Then, **VR1** is adjusted so that **VREF1** = **VF**. Thus, the sensor value at a given temperature and the temperature are the calibration values. Do not touch **SEN1** at any point during the calibration.



BANDWIDTH CONTROL

Including **C3** in the feedback loop creates a single-pole (~6dB/oct.) low pass filter with a cutoff frequency of approximately 20Hz, which reduces high frequency noise in the measurement.

SENSOR TEMPCO TRIM

Sensor temperature coefficient trimmable from -2.12 mV/°C to -1.93 mV/°C using **VR3**

AMPLIFIER OFFSET VOLTAGES

No actual op amp is ideal; therefore, each device has an input offset voltage that is added (or subtracted) to the input. This offset is amplified equally with the input. Therefore, the lowest-offset device (**IC5**, with a typical offset of 100µV) is used for subtracting the sensor voltage from **VREF1** and multiplying this by ~5 (trimmable), giving a maximal offset of ~500µV in the gain stage, corresponding to ±0.05°C, without specific trimming of **VREF1** to adjust for the offset voltage of **IC5A/B**.

