

Utility Maximization in Cognitive Radio Networks using POMDP Approximate Value Iteration methods

Bharath Keshavamurthy, Nicolò Michelusi

Abstract—Cognitive radio technologies will be critical to the wireless communication infrastructure in the near future due to the increasingly incredible number of applications being added to the computer networking ecosystem, both in the commercial and the military spheres, resulting in increased pressure on the available spectrum which is a limited physical resource. In this paper, we propose a novel channel access strategy in networks with multiple licensed users wherein a cognitive radio node should learn the correlation model defining the occupancy behavior of the incumbents and devise an optimal strategy based on this correlation model by solving a utility maximization problem in a partially observable radio environment setting. Since the computational complexity associated with solving for the optimal channel access strategy scales exponentially with the number of spectrum bands under consideration, we propose a system employing approximate POMDP value iteration methods, namely, the PERSEUS algorithm. Furthermore, through system simulations, we compare the performance of standard MAP-based state estimators and correlation-coefficient based clustering algorithms in the state-of-the-art against our proposed system employing a customized PERSEUS algorithm, with respect to the secondary network throughput and the number of collisions with the incumbent transmissions.

Index Terms—Hidden Markov Model, POMDP, and the PERSEUS Algorithm

I. INTRODUCTION

With the advent of fifth-generation wireless communication networks, the problem of spectrum scarcity has been exacerbated [1]. For some time now, cognitive radio technologies have been in the spotlight as a potential solution to this problem in commercial and military applications [2]. Cognitive radio networks facilitate efficient spectrum utilization by intelligently accessing "white spaces" left unused by the sparse and infrequent transmissions of the licensed users. By intelligently accessing these spatial and temporal spectrum holes, the radio nodes in a cognitive radio network complete their allotted network flows subject to QoS requirements while ensuring that their transmissions do not interfere with the incumbents in the network [3]. This poses an interesting optimization problem of maximizing secondary network throughput while complying with strict non-interference with the transmissions of the primary, licensed users. A crucial aspect underlying the design of cognitive radio networks is the channel access protocol in the MAC layer of the stack. In this regard, the current state-of-the-art involves channel access strategies dictated by multi-armed bandits, reinforcement learning agents, and other custom heuristics. However, almost all these works assume

independence among channels in the discretized spectrum which is imprudent because licensed users exhibit both spatial and temporal correlation in their channel occupancy behavior: the primary users frequently occupy a set of adjacent channels implying that these channels would then be correlated spatially with respect to the occupancy behavior of a particular incumbent and furthermore, the primary users frequently use the same set of channels across time which further implies that these channels are correlated temporally. This pattern in occupancy behavior of the incumbents imputes very high levels of correlation among channels which need to be leveraged for more accurate predictions of spectrum holes in order to satisfy the QoS guarantees of flows in the secondary network while limiting collisions with incumbent transmissions. In this paper, we propose techniques to exploit the correlation model underlying the occupancy behavior of incumbents in the network—a parameter estimation algorithm to learn the aforementioned correlation model, a state estimation algorithm to infer channel occupancy from noisy and incomplete information, and an approach to solve for the optimal channel sensing and access policy to be followed by the cognitive radio node. We define the signal model in Section II of this document followed by the formulations, approaches, and algorithms for each of the three interconnected sub-problems detailed earlier, in Section III, and finally, in Section IV of this document, we present an evaluation of the system along with comparisons with other approaches in the state-of-the-art.

The existing state-of-the-art in this domain primarily involve data-driven approaches wherein the cognitive radio nodes collect occupancy behavior information of the incumbents in the radio environment and use this gathered data to determine the correlation factor between two given channels. One such approach is detailed in [4] in which the authors propose employing a Minimum Entropy Merging (MEM) technique in order to choose the best occupancy estimation outcome between the one provided by a Channel Correlation based Estimation technique (CCE) which incorporates heuristic channel clustering algorithms based on the pre-determined channel correlation factors and the one provided by a Markov Process based Estimation technique (MPE) which leverages the temporal correlation in incumbent occupancy behavior to estimate the occupancy of un-sensed channels. This work assumes that the SU nodes have prior knowledge about the correlation model underlying the occupancy of the channels in the radio environment and furthermore, this work does not provide a consolidated solution which considers the correlation across channels and correlation over time simultaneously. Moreover, the system model laid down in [4] assumes a perfect, noise-

This research has been funded by —.

The authors are with the School of Electrical and Computer Engineering, Purdue University. email: {bkeshava,michelusi}@purdue.edu.

free observation model which is impractical. The work detailed in [4] comes the closest in solving the problem tackled by us in the paper: An optimal spectrum sensing and access strategy for cognitive radio nodes driven by the need to reduce the channel sensing consumption of these nodes. However, in this paper, we make no assumptions about the prior knowledge capabilities of the SUs, i.e., we build a system that learns the correlation model of the channels in the radio environment both across the channel indices and across the time indices online and uses this learned model to arrive at an optimal sensing and access strategy that leverages both spatial and temporal correlation.

The work detailed in [5] involves a Channel Set Management system that employs pre-loaded data from a database as the "Channel History" and uses the observations from this training set to estimate the parameters of the HMM using the Baum-Welch Algorithm. Consequently, the next state of the channel is predicted using the Forward Algorithm. However, this work does not take into consideration the correlation among channels in the radio environment and not unlike [4], this work assumes prior knowledge in terms of an already available dataset. Similarly, the work described in [6] involves a data-driven approach incorporating previously learnt channel correlation coefficients in order to solve the joint spectral-temporal spectrum prediction problem from incomplete observations by framing it as a matrix completion problem.

In [7], the authors solve for a cooperative spectrum sensing policy employed by multiple spatially diverse cognitive radio nodes in order to efficiently utilize the available spectrum. However, the authors in this work assume independence among the channels in the radio environment which is seldom the case because incumbents tend to occupy adjacent channels for this transmissions thereby inducing correlation across channels and moreover, the incumbents tend to occupy the same set of channels over time thereby inducing correlation across time indices. Other works such as [8] also solve for the optimal sensing and access strategy by assuming independence among channels. Reference [9] formulates the spectrum sensing and access problem in a distributed multiple SU radio environment as a partially observable stochastic game and solves it using SARSA along with a gradient descent based linear function approximation technique. Although the formulation of the problem in this work is similar to ours, the model detailed in [9] does not consider correlation across channel indices and furthermore, the authors present a belief update heuristic based on the observations without estimating the state transition probabilities. However, in this paper, we outline a rigorous framework for estimating the transition model of the underlying MDP online and a fragmented PERSEUS algorithm with simplified beliefs to account for the computational complexity associated with the large state and action space of the problem at hand.

II. SYSTEM MODEL

A. Signal Model

We consider a network consisting of P licensed users termed the Primary Users (PUs) and one cognitive radio node

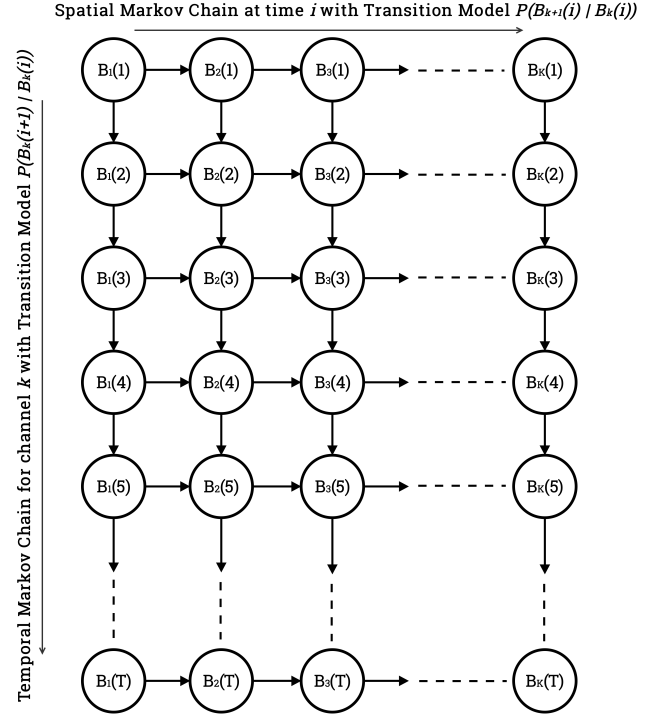


Fig. 1. The correlation model across channel indices and across time indices underlying the occupancy behavior of incumbents in the network

termed the Secondary User (SU) equipped with a spectrum sensor. The objective of the SU is to opportunistically access portions of the spectrum left unused by the PUs in order to maximize its own throughput. To this end, the SU should learn how to intelligently access spectrum holes (white-spaces) intending to maximize its throughput while maintaining strict non-interference compliance with incumbent transmissions. The wideband signal received at the SU receiver at time n is denoted as $y(n)$ and is given by

$$y(n) = \sum_{p=1}^P \sum_{l=0}^{L_p-1} h_p(l)x_p(n-l) + v(n), \quad (1)$$

where $y(n)$ is expressed as a convolution of the signal $x_p(n)$ of the p th PU with the channel impulse response $h_p(n)$, and $v(n)$ denotes additive white Gaussian noise (AWGN) with variances σ_v^2 . Eq. (1) can be written in the frequency domain by taking a K -point DFT which decomposes the observed wideband signal into K discrete narrow-band components as

$$Y_k(i) = \sum_{p=1}^P H_{p,k}(i)X_{p,k}(i) + V_k(i), \quad (2)$$

where $i \in \{1, 2, 3, \dots, T\}$ represents the time index; $k \in \{1, 2, 3, \dots, K\}$ represents the index of the components in the frequency domain; $V_k(i) \sim \mathcal{CN}(0, \sigma_v^2)$ represents a circularly symmetric additive complex Gaussian noise sample, i.i.d across frequency and across time; $X_{p,k}(i)$ is the signal of the p th PU in the frequency domain, and $H_{p,k}(i)$ is its frequency domain channel. The noise samples are assumed to be independent of the occupancy state of the channels. We further assume that the P PUs employ an orthogonal access to the

spectrum (e.g., OFDMA) so that $X_{p,k}(i)X_{q,k}(i) = 0, \forall p \neq q$. Thus, letting p_k be the index of the PU that contributes to the signal in the k th spectrum band (possibly, $p_k = 0$ if no PU is transmitting in the k th spectrum band), and letting $H_k(i) = H_{p_k,k}(i)$ and $X_k(i) = X_{p_k,k}(i)$, we can rewrite (2) as

$$Y_k(i) = H_k(i)X_k(i) + V_k(i). \quad (3)$$

Thus, $H_k(i)$ represents the k th DFT coefficient of the impulse response $h_{p_k,k}(n)$ of the channel in between the PU operating on the k th spectrum band and the SU, at time i ; we model it as a zero-mean circularly symmetric complex Gaussian random variable with variance σ_H^2 , $H_k \sim \mathcal{CN}(0, \sigma_H^2)$, i.i.d. across frequency bands, over time, and independent of the occupancy state of the channels.

B. PU Spectrum Occupancy Model

We now introduce the model of PU occupancy over time and across the frequency domain. We model each $X_k(i)$ as

$$X_k(i) = \sqrt{P_{tx}}B_k(i)S_k(i), \quad (4)$$

where P_{tx} is the transmission power of the PUs, $S_k(i)$ is the transmitted symbol modelled as a constant amplitude signal, $|S_k(i)| = 1$, i.i.d. over time and across frequency bands;¹ $B_k(i) \in \{0, 1\}$ is the binary spectrum occupancy variable, with $B_k(i) = 1$ if the k th spectrum band is occupied by a PU at time i , and $B_k(i) = 0$ otherwise. Therefore, the PU occupancy behavior in the entire wideband spectrum of interest at time i , discretized into narrow-band frequency components can be modeled as the vector

$$\vec{B}(i) = [B_1(i), B_2(i), B_3(i), \dots, B_K(i)]^T \in \{0, 1\}^K. \quad (5)$$

PUs join and leave the spectrum at random times. To capture this temporal correlation in the spectrum occupancy dynamics of PUs, we model the spectrum occupancy dynamics as a Markov process: given $\vec{B}(i)$, the spectrum occupancy state at time index i , $\vec{B}(i+1)$ is independent of the past, $\vec{B}(j)$, $j < i$; $j, i \in \{1, 2, 3, \dots, T\}$, i.e.

$$\mathbb{P}(\vec{B}(i+1)|\vec{B}(j), \forall j \leq i) = \mathbb{P}(\vec{B}(i+1)|\vec{B}(i)). \quad (6)$$

Additionally, when joining the spectrum pool, PUs occupy a number of adjacent spectrum bands, and may vary their spectrum needs depending on traffic demands, channel conditions, etc. To capture this behavior, we model $\vec{B}(i)$ as having Markovian correlation across the bands as,

$$\begin{aligned} & \mathbb{P}(\vec{B}(i+1)|\vec{B}(i)) \\ &= \mathbb{P}(B_1(i+1)|B_1(i)) \prod_{k=2}^K \mathbb{P}(B_k(i+1)|B_k(i), B_{k-1}(i+1)). \end{aligned} \quad (7)$$

That is, the spectrum occupancy at time $i+1$ in frequency band k , $B_k(i+1)$, depends on the occupancy state of the adjacent spectrum band at the same time, $B_{k-1}(i+1)$, and that of the same spectrum band k in the previous time index i , $B_k(i)$ as shown in Fig. 1.

¹In the case where $S_k(i)$ does not have constant amplitude, we may approximate $H_k(i)S_k(i)$ as complex Gaussian with zero mean and variance $\sigma_H^2 \mathbb{E}[|S_k(i)|^2]$, without any modification to the subsequent analysis.

C. Spectrum Sensing Model

In order to detect the available spectrum holes, the SU performs spectrum sensing. However, owing to physical design limitations at the SU's spectrum sensor [10], not all channels in the discretized spectrum can be sensed at once. Therefore, due to limited sensing capabilities, the SU can sense only κ out of K spectrum bands at any given time, with $1 \leq \kappa \leq K$. Let $\mathcal{K}_i \subseteq \{1, 2, \dots, K\}$ with $|\mathcal{K}_i| \leq \kappa$ be the set of indices corresponding to the spectrum bands sensed by the SU at time i , which is part of our design. Then, we define the observation vector

$$\vec{Y}(i) = [Y_k(i)]_{k \in \mathcal{K}_i}, \quad (8)$$

where $Y_k(i)$ is given by (3). The true states $\vec{B}(i)$ encapsulate the actual occupancy behavior of the PU and the measurements at the SU are noisy observations of these true states which are modeled to be the observed states of a Hidden Markov Model (HMM). Given the spectrum occupancy vector $\vec{B}(i)$ and the set of sensed spectrum bands \mathcal{K}_i , the probability density function of $\vec{Y}(i)$ is expressed as

$$f(\vec{Y}(i)|\vec{B}(i), \mathcal{K}_i) = \prod_{k \in \mathcal{K}_i} f(Y_k(i)|B_k(i)), \quad (9)$$

owing to the independence of channels, noise, and transmitted symbols across frequency bands. Moreover, from (3) we find that

$$Y_k(i)|B_k(i) \sim \mathcal{CN}(0, \sigma_H^2 P_{tx} B_k(i) + \sigma_V^2). \quad (10)$$

D. POMDP Agent Model

In this section, we model the spectrum access scheme of the SU as a Partially Observable Markov Decision Process (POMDP) wherein the goal of the POMDP agent is to devise an optimal sensing and access policy in order to maximize its throughput while maintaining strict non-interference compliance with incumbent transmissions. In fact, the agent's limited sensing capabilities coupled with its noisy observations result in an increased level of uncertainty at the agent's end about the occupancy state of the spectrum under consideration and the exact effect of executing an action on the radio environment. The transition model of the underlying MDP as described by (7), is denoted by \mathbf{A} and is learnt by the agent by interacting with the radio environment. The emission model is denoted by \mathbf{M} and is given by (9), with $f(Y_k(i)|B_k(i))$ given by (10). We model the POMDP as a tuple $(\mathcal{B}, \mathcal{A}, \mathcal{Y}, \mathbf{P}, \mathbf{M})$ where $\mathcal{B} \equiv \{0, 1\}^K$ represents the state space of the underlying MDP with states \vec{B} given by all possible realizations of the spectrum occupancy vector as described by (3), \mathcal{A} represents the action space of the agent, given by all $\binom{K}{\kappa}$ possible combinations in which the κ spectrum bands are chosen to be sensed out of K at any given time; and \mathcal{Y} represents the observation space of the agent based on the signal model outlined in the previous subsection. The state of the POMDP at time i is given by the *prior belief* β_i , which represents the probability distribution of the underlying MDP state $\vec{B}(i)$, given the information collected by the agent up to time i , but before collecting the new information in slot i . At the beginning of each time index i , given β_i , the agent selects κ

spectrum bands out of K according to a policy $\pi(\beta_i)$, thus defining the sensing set \mathcal{K}_i , performs spectrum sensing on these spectrum bands, observes $\vec{Y}(i) \in \mathcal{Y}$, and updates its *posterior belief* $\hat{\beta}_i$ of the current spectrum occupancy $\vec{B}(i)$ as

$$\begin{aligned}\hat{\beta}_i &= \mathbb{P}(\vec{B}(i) = \vec{B}' | \vec{Y}(i), \mathcal{K}_i, \beta_i) \\ &= \frac{\mathbb{P}(\vec{Y}(i) | \vec{B}', \mathcal{K}_i) \beta_i(\vec{B}')}{\sum_{\vec{B}'' \in \{0,1\}^K} \mathbb{P}(\vec{Y}(i) | \vec{B}'', \mathcal{K}_i) \beta_i(\vec{B}'')},\end{aligned}\quad (11)$$

where $\mathbb{P}(\vec{Y}(i) | \mathcal{K}_i, b_{i-1})$ is the normalization constant and b_{i-1} represents the belief of the agent prior to the observation $\vec{Y}(i)$, defined as a probability distribution over all possible states. Given the posterior belief $\hat{\beta}_i$, we employ a threshold based detection mechanism to determine whether a given channel is occupied ($\phi_k(\hat{\beta}_i) = 1$) or idle ($\phi_k(\hat{\beta}_i) = 0$). If a channel k is deemed to be idle by this threshold based detection mechanism, the SU accesses it for the delivery of its network flows. On the other hand, if a channel k is deemed to be occupied, the SU leaves it untouched. Furthermore, for utility evaluation, since relying on feedback from the radio environment will introduce unforeseen variables and additional dynamics into the problem, we use the state estimator detailed in Section III to determine the reference occupancy metrics of the incumbents (the reference estimated state vector) $\hat{\vec{B}}$ based on the observations $\vec{Y}(i)$ obtained from the POMDP agent's sensing action, i.e. κ_i . Based on the number of *truly idle* bands detected by the SU accounting for the throughput maximization aspect of the agent's end-goal and a penalty for *missed detections* accounting for the incumbent non-interference constraint, the reward to the agent is modeled as

$$R(\hat{\vec{B}}(i), \hat{\beta}_i) = \sum_{k=1}^K (1 - \hat{B}_k(i))(1 - \phi_k(\hat{\beta}_i)) - \lambda \hat{B}_k(i)(1 - \phi_k(\hat{\beta}_i)), \quad (12)$$

where $\lambda > 0$ represents the cost term penalizing the agent for missed detections, i.e. interference with the incumbent. After performing data transmission, the SU computes the prior belief for the next slot as **[NM: write $\beta(i+1)$ as a function of $\hat{\beta}(i)$.]** The action policy π of the agent maps the beliefs β_i to actions \mathcal{K}_i at time i and is characterized by a Value Function **[NM: Before defining the value function, you need to define your goal, i.e., what is the optimization that you are trying to solve? What is the cost function that you are trying to maximize?]**

$$V^\pi(\beta) = \mathbb{E}_\pi \left[\sum_{i=0}^{\infty} \gamma^i R(\beta_i, \pi(\beta_i) | \beta_0 = \beta) \right], \quad (13)$$

where $0 < \gamma < 1$ is the discount factor, $\pi(b_i)$ is the action taken by the agent at time i under policy π , and b_0 is the initial belief. The optimal policy π^* specifies the optimal action to take at the current time index assuming that the agent behaves optimally at future time indices as well. It is evident from equation (13) that we have an infinite-horizon discounted reward problem formulation and in order to solve

for the optimal policy we need to solve the Bellman equation given by

$$V^*(\beta) = \max_{\mathcal{K} \in \mathcal{A}} \left[\sum_{\vec{B} \in \mathcal{B}} R(\vec{B}, \mathcal{K}) \beta(\vec{B}) + \gamma \sum_{\vec{Y} \in \mathcal{Y}} \mathbb{P}(\vec{Y} | \mathcal{K}, \beta) V^*(\beta_{\mathcal{K}}^{\vec{Y}}) \right]. \quad (14)$$

Given the high dimensionality of the spectrum sensing and access problem, i.e. the number of states of the underlying MDP scales exponentially with the number of bands in the spectrum, solving equation (14) using Exact Value Iteration and Policy Iteration algorithms is computationally infeasible. Additionally, solving for the optimal policy from equation (14) requires prior knowledge about the underlying MDP's transition model. Therefore, in this paper we present a framework to estimate the transition model of the underlying MDP and then utilize this learned model to solve for the optimal policy by employing Randomized Point-Based Value Iteration techniques, namely, the PERSEUS algorithm.

III. APPROACHES AND ALGORITHMS

A. Occupancy Behavior Transition Model Estimation

In real-world implementations of cognitive radio systems, the transition model of the occupancy behavior of the PUs is not known to the SUs in the network and needs to be learnt over time. The learnt model then needs to be fed back to the POMDP agent in order to solve for the optimal policy. The system can learn the model either before triggering or during the operation of the POMDP agent. Inherently, the approach constitutes solving a parameter estimation problem formulated as

$$A^* = \arg \max_A \mathbb{P}([\vec{Y}(i)]_{i=1}^\tau | A), \quad (15)$$

which is a Maximum Likelihood Estimation (MLE) problem, where A is defined as $\mathbb{P}(\vec{B}(i+1) | \vec{B}(i))$ and τ refers to the learning period of the parameter estimator: this, as mentioned earlier, can be equal to the entire duration of the POMDP agent's interaction with the radio environment or can be a predefined parameter learning period before triggering the POMDP agent. In order to facilitate better readability, for the description of this parameter estimator, we denote $[\vec{Y}(i)]_{i=1}^\tau$ as \mathbf{Y} and $[\vec{B}(i)]_{i=0}^\tau$ as \mathbf{B} . Re-framing (15) as an optimization of the log-likelihood, using the definition of marginal probability, and focusing on the joint instead of the conditional, we get,

$$A = \arg \max_A \log \left(\sum_{\mathbf{B}} \mathbb{P}(\mathbf{B}, \mathbf{Y}, A) \right) \quad (16)$$

In order to exploit the characteristics of the stated Markov model, we multiply and divide the operand of the logarithm by β which from the equality constraint of Jensen's Inequality turns out to be $\mathbb{P}(\mathbf{B} | \mathbf{Y}, \hat{A})$. The optimization problem in (16) is then restated as,

$$A = \arg \max_A \sum_{\mathbf{B}} \mathbb{P}(\mathbf{B} | \mathbf{Y}, \hat{A}) \log(\mathbb{P}(\mathbf{B}, \mathbf{Y}, A)) \quad (17)$$

Applying the characteristics of the Markov model discussed in Section II, we write (17) as

$$A = \arg \max_A \sum_{\mathbf{B}} \mathbb{P}(\mathbf{B}|\mathbf{Y}, \hat{A}) \sum_L \sum_R \sum_{i=1}^{\tau} \sum_{j=1}^{\tau} \mathcal{I} \log(M_R(\vec{Y}(i))) + \mathcal{J} \log(a_{LR}), \quad (18)$$

where $L, R \in \{0, 1\}^K$ represent iterables for the occupancy state vectors,

$$M_R(\vec{Y}(i)) = \mathbb{P}(\vec{Y}(i) | \vec{B}(i) = R), \quad (19)$$

represents the emission model outlined in (9),

$$a_{LR} = \mathbb{P}(\vec{B}(i) = R | \vec{B}(i-1) = L, \hat{A}), \quad (20)$$

represents the unknown transition model which is the subject of this estimation, and \mathcal{I} and \mathcal{J} detailed below are indicator random variables introduced to bring in specificity into the estimation procedure.

$$\mathcal{I} = \begin{cases} 1, & \text{if } \vec{Y}(i) \text{ and } \vec{B}(i) = R \\ 0, & \text{otherwise} \end{cases} \quad (21)$$

$$\mathcal{J} = \begin{cases} 1, & \text{if } \vec{B}(i-1) = L \text{ and } \vec{B}(i) = R \\ 0, & \text{otherwise} \end{cases} \quad (22)$$

We impose a constraint on the transition probability in (18) as

$$\sum_R a_{LR} = 1, \quad (23)$$

and formulate the Lagrangian as

$$\mathcal{L} = \left\{ \sum_{\mathbf{B}} \mathbb{P}(\mathbf{B}|\mathbf{Y}, \hat{A}) \sum_L \sum_R \sum_{i=1}^{\tau} \sum_{j=1}^{\tau} \mathcal{I} \log(M_R(\vec{Y}(i))) + \mathcal{J} \log(a_{LR}) \right\} + \sum_L \lambda_L (1 - \sum_R a_{LR}). \quad (24)$$

Solving for a_{LR} , we get,

$$a_{LR} = \frac{\sum_{i=1}^{\tau} \mathbb{P}(\mathbf{Y}, A, \vec{B}(i) = R, \vec{B}(i-1) = L)}{\sum_{i=1}^{\tau} \mathbb{P}(\mathbf{Y}, A, \vec{B}(i-1) = L)}. \quad (25)$$

In order to further simplify (25) and bring it into an iterative algorithmic form, we introduce the forward and backward probabilities. We define the forward probability as

$$F(i, R) = \mathbb{P}([\vec{Y}(t)]_{t=1}^i, \vec{B}(i) = R) = \sum_L \mathbb{P}(\vec{B}(i) = R, \vec{Y}(i) | \vec{B}(i-1) = L) F(i-1, L), \quad (26)$$

and the backward probability as

$$D(i, L) = \mathbb{P}([\vec{Y}(t)]_{t=i}^{\tau} | \vec{B}(i-1) = L) = \sum_R \mathbb{P}(\vec{B}(i) = R, \vec{Y}(i) | \vec{B}(i-1) = L) D(i+1, R). \quad (27)$$

Using these definitions, (25) can be rewritten as,

$$a_{LR} = \frac{\sum_{i=1}^{\tau} F(i-1, L) M_R(\vec{Y}(i)) a_{LR} D(i+1, R)}{\sum_R \sum_{i=1}^{\tau} F(i-1, L) M_R(\vec{Y}(i)) a_{LR} D(i+1, R)}. \quad (28)$$

B. Occupancy Behavior State Estimation

During the reward evaluation phase of the POMDP agent at time i , the observations made based on the sensing action \mathcal{K}_i are employed at a state estimator to determine the occupancy state of the spectrum bands. Based on this estimated state vector, we formulate the false alarm and missed detection metrics which allow us to capture the throughput maximization and PU non-interference requirements essential for the operation of our POMDP agent. We formulate the state estimation problem as

$$\vec{B}(i)^* = \arg \max_{\vec{B}} \mathbb{P}(\vec{B} | \vec{Y}(i)), \quad (29)$$

which is a Maximum-A-Posteriori (MAP) estimation problem. This optimization problem can be restated in terms of the value functions as

$$V_{i,k}^r = \max_{\tilde{\mathbf{B}}_{[t-1, k-1]}} \mathbb{P}(\tilde{\mathbf{Y}}_{[t-1, k-1]}, \tilde{\mathbf{B}}_{[t-1, k-1]}, Y_k(i), B_k(i)), \quad (30)$$

where for the estimation of the occupancy in spectrum band k at time i ,

$$\tilde{\mathbf{Y}}_{[t-1, k-1]} \equiv \{ [\vec{Y}_v(i)]_{v=1}^{k-1}, [\vec{Y}_k(t)]_{t=1}^{i-1} \} \quad (31)$$

denotes the set of all essential past observations which for readability purposes is denoted simply as $\tilde{\mathbf{Y}}$ and

$$\tilde{\mathbf{B}}_{[t-1, k-1]} \equiv \{ [\vec{B}_v(i)]_{v=1}^{k-1}, [\vec{B}_k(t)]_{t=1}^{i-1} \} \quad (32)$$

denotes set of all essential past states which is henceforth simply referred to as $\tilde{\mathbf{X}}$ for readability. Applying the characteristics of the Markov model detailed in (7) to (30), we get

$$V_{i,k}^r = M_r(Y_k(i)) \max_{\mathbf{B}} \mathbb{P}(B_k(i) = r | B_k(i-1) = m, B_{k-1}(i) = n) \mathbb{P}(\tilde{\mathbf{Y}}, \tilde{\mathbf{B}}), \quad (33)$$

which can be simplified further to show that,

$$V_{i,k}^r = M_r(Y_k(i)) \max_{m,n} a_{mnr} V_{i-1,k}^m V_{i,k-1}^n, \quad (34)$$

where

$$a_{mnr} = \mathbb{P}(B_k(i) = r | B_k(i-1) = m, B_{k-1}(i) = n), \quad (35)$$

which can be evaluated from the estimated transition model. Equation (34) corresponds to the forward recursion aspect of the double Markov chain Viterbi algorithm. Next, similar to the backtracking procedure in the one dimensional (single Markov chain) Viterbi algorithm, the Trellis diagram is traversed backwards to recover the most probable previous neighbours of $B_k(i)$. This is done recursively until the entire Trellis diagram has been traversed to yield the most probable state sequence, i.e. the Viterbi path. Mathematically, the backtracking step with respect to the neighbours of $B_k(i)$ is represented as

$$m^*, n^* = \arg \max_{m,n} a_{mnr} V_{i-1,k}^m V_{i,k-1}^n, \quad (36)$$

where m^* is the most probable state of channel k in time index $i-1$ and n^* is the most probable state of channel $k-1$ in time index i , given that channel k in time index i is in state r ; $m, n, r \in \{0, 1\}$.

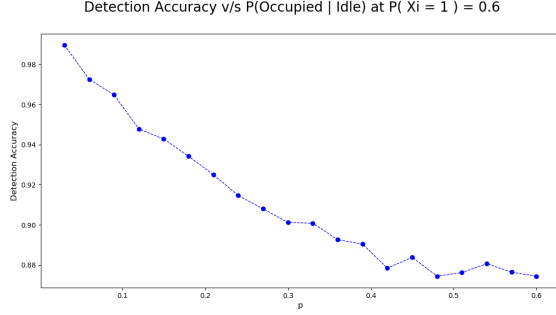


Fig. 2. The detection accuracy of the unconstrained Viterbi algorithm over varying values of $\mathbb{P}(\text{Occupied}|\text{Idle})$

Detection Accuracy v/s $\mathbb{P}(\text{Occupied} | \text{Idle})$ for 18 channels at $\mathbb{P}(X_i = 1) = 0.6$ with varying uniform channel sensing strategies

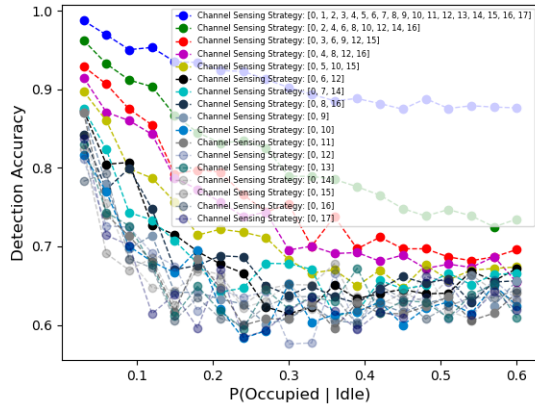


Fig. 3. The detection accuracies of the constrained Viterbi algorithm for different sensing strategies over varying values of $\mathbb{P}(\text{Occupied}|\text{Idle})$

Detection Accuracy v/s $\mathbb{P}(\text{Occupied} | \text{Idle})$ for 18 channels at $\mathbb{P}(X_i = 1) = 0.6$ with uniform channel sensing strategy [0, 2, 4, 6, 8, 10, 12, 14, 16]

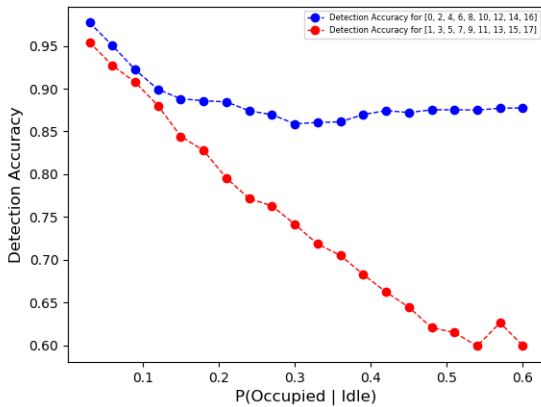


Fig. 4. The detection accuracies of the constrained Viterbi algorithm for sensed and un-sensed channels under a given channel sensing strategy

C. The PERSEUS Algorithm

As discussed in Section II of this article, solving the Bellman equation (14) for POMDPs with large state and action space using exact value iteration and policy iteration techniques [11] is computationally infeasible [12]. Hence, we resort to approximate value iteration techniques [11] to ensure that the system scales well to a large number of bands in the spectrum of interest. For infinite-horizon POMDPs, V^* in (14) can be approximated by a Piece-Wise Linear and Convex function (PWLC) [11]. The core idea behind the PERSEUS algorithm is that the value function in time index i can be parameterized by a set of hyperplanes $\{\vec{\alpha}_i^u\}$, $u = \{0, 1, 2, \dots, |V_i|\}$, each of which represents a region of the belief space for which it is the maximizing element. The backup step is defined as the process of determining the optimal hyperplane out of the set of available hyperplanes in time index i as

$$\vec{\alpha}_i(b) = \arg \max_{\vec{\alpha}_i^u} b \cdot \vec{\alpha}_i^u, \quad (37)$$

which implies that,

$$\begin{aligned} V_i(b) &= \max_{\vec{\alpha}_i^u} b \cdot \vec{\alpha}_i^u, \\ \pi_i(b) &= a(\vec{\alpha}_i^{u*}), \end{aligned} \quad (38)$$

where $a(\vec{\alpha}_i^{u*})$ refers to the action corresponding to the optimal hyperplane. The PERSEUS algorithm constitutes an Exploration phase wherein the POMDP agent randomly interacts with the radio environment to collect a set of so-called reachable beliefs which are to be improved and numerous iterative Backup stages. In each backup stage, the agent samples an unimproved belief point b uniformly at random from the set of unimproved points and performs a backup on this sampled belief point according to (37) to determine the optimal hyperplane $\vec{\alpha}$. Considering an arbitrary time index i , if $V_{i+1}(b) = b \cdot \vec{\alpha} \geq V_i(b)$, then the belief point b is said to be improved along with any other belief points b' in the unimproved set for which $V_{i+1}(b') = b' \cdot \vec{\alpha} \geq V_i(b')$. If $V_{i+1}(b) = b \cdot \vec{\alpha} < V_i(b)$, then a copy of the maximizing hyperplane for $V_i(b)$ is used for $V_{i+1}(b)$ and the belief point b is then removed from the set of unimproved points. The backup stage continues until the set of unimproved points is empty and the agent performs a series of backup stages until there the number of policy changes between iterations is below a specified threshold η .

The belief update procedure outlined in (11) is an essential aspect of the PERSEUS algorithm which can turn into a performance bottleneck for large state spaces due to the inherent iteration over all possible states. In order to circumvent this problem, we fragment the spectrum into much smaller, independent sets of correlated channels and then run the PERSEUS algorithm on these fragments by leveraging multi-processing and multi-threading tools available at our disposal in software frameworks. Furthermore, we avoid iterating over all possible states and allow only those state transitions we deem to be the most probable - for example, we allow only those state transitions that involve a Hamming distance of up to 3 between the previous state vector and the current state vector in an 18 channel radio environment.

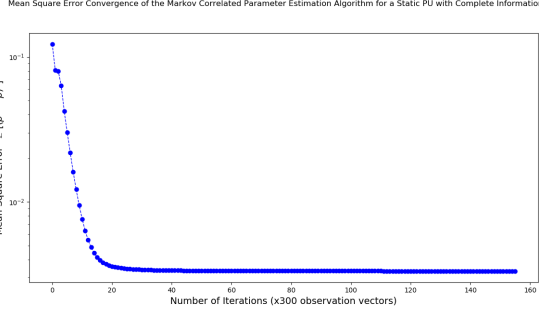


Fig. 5. The mean square error convergence plot of the parameter estimation algorithm

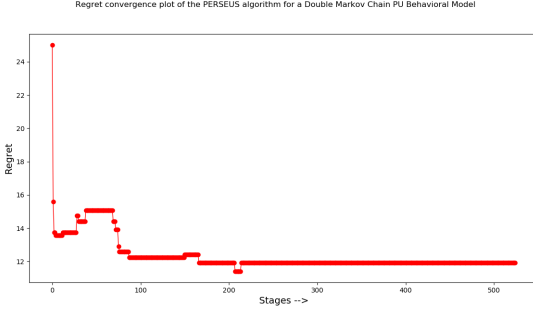


Fig. 6. The regret convergence plot of the PERSEUS algorithm over several backup and wrapper stages

IV. NUMERICAL EVALUATION

The given framework is simulated in Python for a system with 18 channels and a channel model comprising an SNR of 19dB when an incumbent occupies a specific channel. The same Markovian transition model, emission model, and steady-state model is employed across both the channel indices and the time indices. The plot depicted in Fig. 2 consists of the detection accuracy of the Viterbi algorithm wherein the SU makes observations of all the channels in the radio environment and estimates the occupancy states of these channels over varying values of $\mathbb{P}(\text{Occupied}|\text{Idle})$, i.e., as the channels transition toward independence. Fig. 3 illustrates the detection accuracies of the Viterbi algorithm wherein the SU makes observations of only the channels in the given channel sensing strategy and estimates the occupancy states of both the sensed and the un-sensed channels over varying values of $\mathbb{P}(\text{Occupied}|\text{Idle})$. In Fig. 4, the plot depicted shows the detection accuracies of the estimation of sensed and un-sensed channels for the constrained Viterbi algorithm in which the SU only senses channels whose indices correspond to the multiples of 2 and uses these channels to estimate the occupancy behavior of the incumbents across all 18 channels over varying values of $p = \mathbb{P}(\text{Occupied}|\text{Idle})$.

The plot depicted in Fig. 5 shows the Mean Square Error convergence of the Parameter Estimation algorithm while determining the transition model of the MDP underlying the problem under consideration. The EM algorithm detailed in Section III converges to the actual transition model of $\{0 : \{0 : 0.7, 1 : 0.3\}, 1 : \{0 : 0.2, 1 : 0.8\}\}$ over numerous iterations,

#policy_changes during the course of the PERSEUS algorithm for a Double Markov Chain PU Behavioral Model

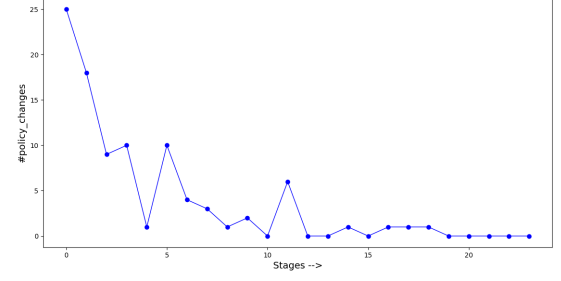


Fig. 7. The number of policy changes involved in the PERSEUS algorithm as it transitions toward convergence over numerous backup and wrapper stages

each iteration corresponding to an averaging operation of 300 observation vectors. Fig. 6 illustrates the Regret Convergence plot of the PERSEUS algorithm over several backup and wrapper stages wherein the regret metric corresponds to the difference in utility between the PERSEUS algorithm in a certain stage and an Oracle which has complete information with respect to the occupancy behavior of the incumbents in the radio environment. Furthermore, the algorithm involves an online estimation of the transition model of the underlying MDP and a random exploration strategy to gather the initial set of reachable beliefs. Fig. 7 depicts the number of policy changes involved in each individual stage of the PERSEUS algorithm as it moves toward convergence. The termination condition for the PERSEUS algorithm is that the number of policy changes over several consecutive backup stages should be zero.

REFERENCES

- [1] C. Pradhan, K. Sankhe, S. Kumar, and G. R. Murthy, "Revamp of enodeb for 5g networks: Detracting spectrum scarcity," in *2015 12th Annual IEEE Consumer Communications and Networking Conference (CCNC)*, Jan 2015, pp. 862–868.
- [2] M. Danneberg, R. Datta, A. Festag, and G. Fettweis, "Experimental testbed for 5g cognitive radio access in 4g lte cellular systems," in *2014 IEEE 8th Sensor Array and Multichannel Signal Processing Workshop (SAM)*, June 2014, pp. 321–324.
- [3] F. Xu, L. Zhang, Z. Zhou, and Y. Ye, "Architecture for next-generation reconfigurable wireless networks using cognitive radio," in *2008 3rd International Conference on Cognitive Radio Oriented Wireless Networks and Communications (CrownCom 2008)*, May 2008, pp. 1–5.
- [4] M. Gao, X. Yan, Y. Zhang, C. Liu, Y. Zhang, and Z. Feng, "Fast spectrum sensing: A combination of channel correlation and markov model," in *2014 IEEE Military Communications Conference*, Oct 2014, pp. 405–410.
- [5] C. Park, S. Kim, S. Lim, and M. Song, "Hmm based channel status predictor for cognitive radio," in *2007 Asia-Pacific Microwave Conference*, Dec 2007, pp. 1–4.
- [6] G. Ding, J. Wang, Q. Wu, L. Yu, Y. Jiao, and X. Gao, "Joint spectral-temporal spectrum prediction from incomplete historical observations," in *2014 IEEE Global Conference on Signal and Information Processing (GlobalSIP)*, Dec 2014, pp. 1325–1329.
- [7] J. Oksanen, V. Koivunen, J. LundÄä'n, and A. Huttunen, "Diversity-based spectrum sensing policy for detecting primary signals over multiple frequency bands," in *2010 IEEE International Conference on Acoustics, Speech and Signal Processing*, March 2010, pp. 3130–3133.
- [8] S. Chaudhari, V. Koivunen, and H. V. Poor, "Autocorrelation-based decentralized sequential detection of ofdm signals in cognitive radios," *IEEE Transactions on Signal Processing*, vol. 57, no. 7, pp. 2690–2700, July 2009.

- [9] J. LundÄn, S. R. Kulkarni, V. Koivunen, and H. V. Poor, "Multiagent reinforcement learning based spectrum sensing policies for cognitive radio networks," *IEEE Journal of Selected Topics in Signal Processing*, vol. 7, no. 5, pp. 858–868, Oct 2013.
- [10] S. Maleki, S. P. Chepuri, and G. Leus, "Energy and throughput efficient strategies for cooperative spectrum sensing in cognitive radios," in *2011 IEEE 12th International Workshop on Signal Processing Advances in Wireless Communications*, June 2011, pp. 71–75.
- [11] L. P. Kaelbling, M. L. Littman, and A. R. Cassandra, "Planning and acting in partially observable stochastic domains," *Artif. Intell.*, vol. 101, no. 1-2, pp. 99–134, May 1998. [Online]. Available: [http://dx.doi.org/10.1016/S0004-3702\(98\)00023-X](http://dx.doi.org/10.1016/S0004-3702(98)00023-X)
- [12] J. Pineau, G. Gordon, and S. Thrun, "Point-based value iteration: An anytime algorithm for pomdps," in *Proceedings of the 18th International Joint Conference on Artificial Intelligence*, ser. IJCAI'03. San Francisco, CA, USA: Morgan Kaufmann Publishers Inc., 2003, pp. 1025–1030. [Online]. Available: <http://dl.acm.org/citation.cfm?id=1630659.1630806>