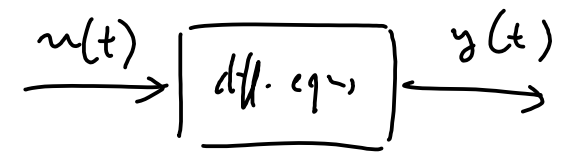
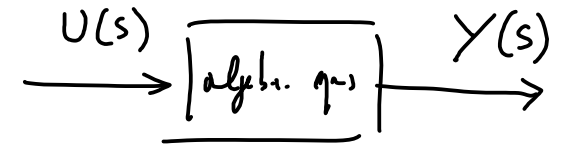


① Obtain linear diff. eqns



② LAPLACE TRANSFORM diff. eqns



③ To find the response of the system, y(t), to an input u(t)

③a LT u(t) into U(s)


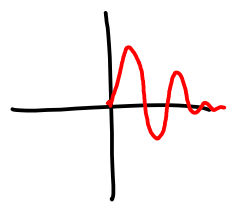

③b Compute Y(s) algebraically

③c Inverse LT Y(s) into y(t)

Recall

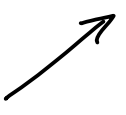
LT of $f(t)$

$$\mathcal{L}\{f(t)\} = F(s) = \int_{0^-}^{\infty} f(t) e^{-st} dt$$

$f(t)$	$F(s)$	$f(t)$	$F(s)$
IMPULSE 	1	 $e^{-at} \sin \omega t$	$\frac{\omega}{(s+a)^2 + \omega^2}$
STEP 	$\frac{1}{s}$	$e^{-at} \cos \omega t$	$\frac{s+a}{(s+a)^2 + \omega^2}$
e^{-at}	$\frac{1}{s+a}$	$\frac{\omega_n}{\sqrt{1-\xi^2}} e^{-\xi \omega_n t} \sin(\sqrt{1-\xi^2} \omega_n t)$	$\frac{\omega_n^2}{s^2 + 2\xi \omega_n s + \omega_n^2}$
$\sin \omega t$	$\frac{\omega}{s^2 + \omega^2}$	$\xi < 1$	
$\cos \omega t$	$\frac{s}{s^2 + \omega^2}$		

PROPERTIES OF LT

- ① LINEARITY $\mathcal{L}\{\alpha f(t) + \beta g(t)\} = \alpha F(s) + \beta G(s)$
- ② TRANSLATION in t $\mathcal{L}\{f(t-\tau)\} = e^{-\tau s} F(s)$
in s $\mathcal{L}\{e^{\alpha t} f(t)\} = F(s-\alpha)$
- ③ DIFFERENTIATION in t $\mathcal{L}\left\{\frac{d}{dt} f(t)\right\} = s F(s) - f(0^-)$
in s $\mathcal{L}\{t f(t)\} = -\frac{dF(s)}{ds}$
- ④ INTEGRATION $\mathcal{L}\left\{\int_{0^-}^t f(\tau) d\tau\right\} = \frac{1}{s} F(s)$
- ⑤ CONVOLUTION $\mathcal{L}\{f_1(t) * f_2(t)\} = F_1(s) F_2(s)$

$$\int_{-\infty}^{\infty} f_1(\tau) f_2(t-\tau) d\tau$$


6

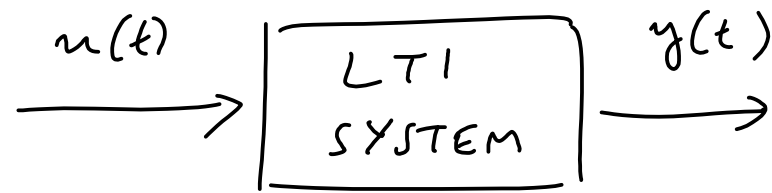
INITIAL VALUE THEOREM

$$f(0^+) = \lim_{s \rightarrow \infty} s F(s)$$

FINAL VALUE THEOREM

$$f(\infty) = \lim_{s \rightarrow 0} s F(s)$$

LONG-TERM BEHAVIOR
WITHOUT SOLVING
THE DIFF. EQN.
DESCRIBING THE
SYSTEM



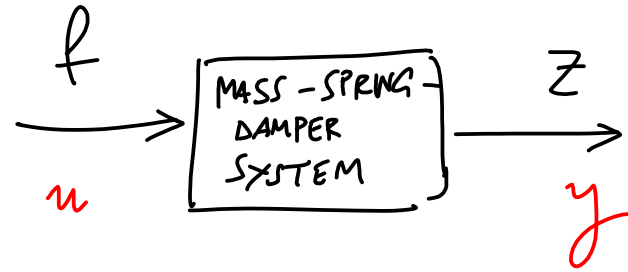
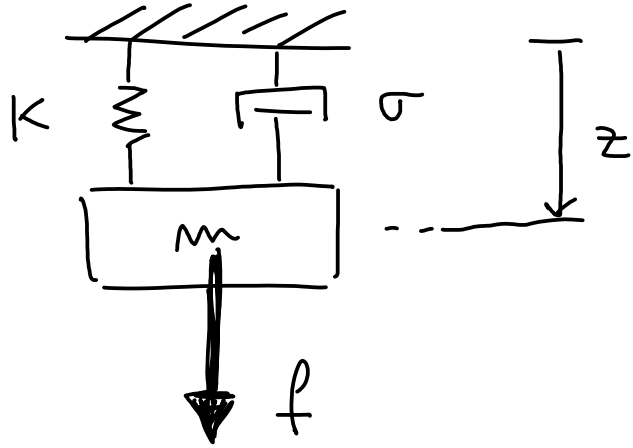
The transfer function (TF) of an LTI system is the ratio $\boxed{\frac{Y(s)}{U(s)}}$, where the LT is taken assuming zero initial conditions.

NO STATE

I/O RELATION

$$x(0) = 0$$

EXAMPLE



$$m \ddot{z} + \sigma \dot{z} + K z = f$$

$$m \ddot{y} + \sigma \dot{y} + K y = u \xrightarrow{\text{LT}} \underbrace{m s^2 Y(s)}_{\substack{\text{Zero} \\ \text{I.C.}}} + \underbrace{\sigma s Y(s)}_{=} + K Y(s) = U(s)$$

$$\frac{Y(s)}{U(s)} = \frac{1}{m s^2 + \sigma s + K}$$

T.F. $G(s)$ is (real) rational if

$$G(s) = \frac{b_m s^{\textcircled{m}} + \dots + b_1 s + b_0}{s^{\textcircled{n}} + a_{n-1} s^{n-1} + \dots + a_1 s + a_0}, \quad a_i, b_i \in \mathbb{R}$$

$G(s)$ is proper $\lim_{s \rightarrow \infty} G(s)$ exists in \mathbb{C}

(if $G(s)$ is rational : $n \geq m$)

$G(s)$ is strictly proper $\lim_{s \rightarrow \infty} G(s) = 0$

(if $G(s)$ is rational : $n > m$)

$p \in \mathbb{C}$ is a pole of $G(s)$ if $\lim_{s \rightarrow p} |G(s)| = \infty$

$z \in \mathbb{C}$ is a zero of $G(s)$ if $\lim_{s \rightarrow z} G(s) = 0$

$G(s)$ rational & numerator and denominator are coprime

poles of $G(s)$ are roots of denominator

zeros of $G(s)$ ——— numerator

Factorization in "zero-pole-gain" form:

$$G(s) = \underset{\substack{\uparrow \\ \text{GAIN}}}{K} \frac{(s - \overset{\downarrow}{z_1}) \dots (s - \overset{\downarrow}{z_m})}{(s - \underset{\uparrow}{p_1}) \dots (s - \underset{\uparrow}{p_n})}$$

EXAMPLE

ASSUME

$$m = 1$$

$$\sigma = 3$$

$$k = 2$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{1}{ms^2 + \sigma s + k}$$

$$= \frac{1}{s^2 + 3s + 2}$$

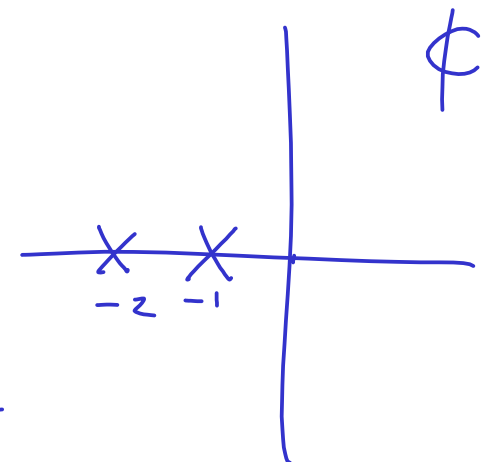
$G(s)$ is rational

proper
strictly proper $\begin{pmatrix} m = 2 \\ m = 0 \end{pmatrix}$

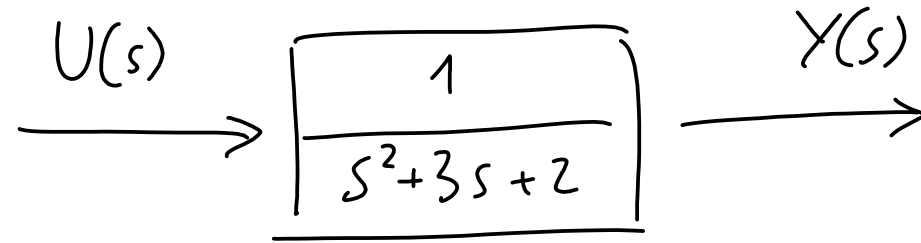
$$G(s) = \frac{(1)}{(s+1)(s+2)} \rightarrow$$

No zeros

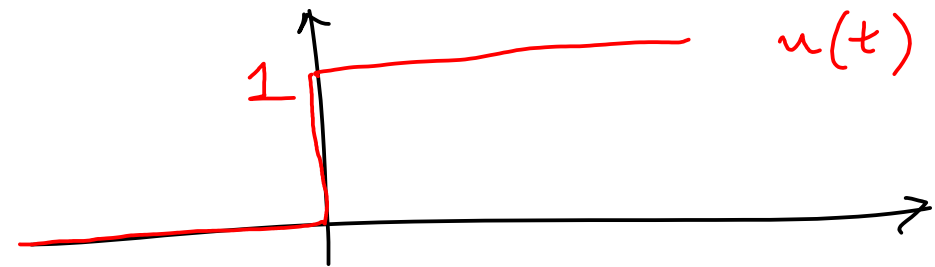
$$p_1 = -1, p_2 = -2$$



EXERCISE



Compute the response, $y(t)$, of the mass-spring-damper system to a step force input



- (a) LT $u(t) \rightarrow U(s)$
- (b) $Y(s) = G(s) \cdot U(s) \left(= \frac{Y(s)}{U(s)} U(s) \right)$ ALGEBRAIC!
- (c) Inverse LT $Y(s) \rightarrow y(t)$