EN 3053 Digital Communication-I Semester 5 Assignment No. 1 7th November, 2021

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1. A certain signal s(t) assumes the following decomposition

$$s(t) = \sum_{k=0}^{N-1} x[k]\phi_k(t), \quad 0 \le t < NT,$$

where $x[k] \in \mathbb{C}$, $N \in \mathbb{Z}^+$, T > 0, and

$$\phi_k(t) = \frac{1}{\sqrt{NT}} e^{j\frac{2\pi}{NT}kt}$$

with $j = \sqrt{-1}$.

(a) Show that

$$\int_0^{NT} \phi_k(t) \phi_\ell^*(t) dt = \begin{cases} 1 & \text{if } k = \ell \\ 0 & \text{if } k \neq \ell \end{cases}$$

where $(\cdot)^*$ denotes the complex conjugate operator.

[2.5 marks]

(b) Hence show that the total energy content of s(t) is given by

[2.5 marks]

$$\mathcal{E} = \sum_{k=0}^{N-1} |x[k]|^2.$$

(c) Show that [10 marks]

$$\sum_{m=0}^{N-1} \phi_k(mT) \phi_\ell^*(mT) = \begin{cases} \frac{1}{T} & \text{if } k = \ell \\ 0 & \text{if } k \neq \ell. \end{cases}$$

[**Hint**: You may use the formula $\sum_{m=0}^{N-1} z^m = \frac{1-z^N}{1-z}$, where $z \in \mathbb{C}$ and $z \neq 1$.]

(d) Let us now obtain a discrete-time sequence of length N from the continuous-time signal s(t) through sampling. In particular, the discrete-time sequence s[m] can be written as

$$s[m] = s(mT), m = 0, 1, \dots, N - 1.$$

Show that [10 marks]

$$x[\ell] = T \sum_{m=0}^{N-1} s[m] \phi_{\ell}^*(mT), \quad \ell = 0, 1, \dots, N-1.$$

2. Assume that a set $\{\theta_i\}$ of M equally likely real vectors satisfies the equations

$$\boldsymbol{\theta}_i^T \boldsymbol{\theta}_\ell = \left\{ egin{array}{ll} 1 & i = \ell \\ \rho & i
eq \ell. \end{array} \right.$$

- (a) Prove that $-\frac{1}{M-1} \le \rho \le 1$. [Hint: Consider $\left| \sum_{k=1}^{M} \boldsymbol{\theta}_{k} \right|^{2}$.] [5 marks]
- (b) Show that the left-hand equality (i.e., $\rho = -\frac{1}{M-1}$) is achieved by the unit average energy simplex set.¹ [10 marks]
- (c) Prove for any allowable ρ that the signal set $\{\mathbf{s}_i\}$, with $\mathbf{s}_i = \sqrt{E_{\rho}}\boldsymbol{\theta}_i$ for all i, has the same error probability as the simplex signal set with energy [10 marks]

$$E_s = E_\rho \left(1 - \frac{1}{M} \right) (1 - \rho).$$

[Hint: First, determine the minimum energy set corresponding to the set $\{\theta_i\}$. Then show that up to a scaling factor the minimum energy set and the simplex set have the same geometry.]

3. One of the two equally likely signals $s_0 = -1$, $s_1 = 1$ is transmitted over a certain AWGN channel as shown in Fig. 1. The two noise random variables n_1 and n_2 are statistically independent of the transmitted signal and of each other. Consequently, we can write the received signals, for i = 0, 1, as

$$r_1 = h_1 s_i + n_1$$
$$r_2 = h_2 s_i + n_2.$$

Now it is convenient to rewrite the above equations in vector form as

$$\underbrace{\begin{pmatrix} r_1 \\ r_2 \end{pmatrix}}_{r} = \underbrace{\begin{pmatrix} h_1 \\ h_2 \end{pmatrix}}_{h} s_i + \underbrace{\begin{pmatrix} n_1 \\ n_2 \end{pmatrix}}_{n},$$

which in turn gives

¹Given a set of M equally likely orthogonal vectors $\{\boldsymbol{\beta}_i\}$, the set $\{\boldsymbol{\beta}_i - \mathbf{a}\}$ with $\mathbf{a} = \frac{1}{M} \sum_{k=1}^{M} \boldsymbol{\beta}_i$ is known as the simplex set.

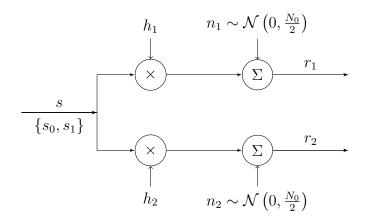


Figure 1: A diversity communication system.

$$r = hs_i + n, \quad i = 0, 1,$$

where $r, h \in \mathbb{R}^{2\times 1}$, and $n \sim \mathcal{N}_2\left(\mathbf{0}, \frac{N_0}{2}\mathbf{I}_2\right)$ with \mathbf{I}_2 denoting the 2×2 identity matrix.

(a) Show that the optimum MAP decision rule is given by

$$m{r}^T m{h} \overset{s_1}{\underset{s_0}{\gtrless}} 0$$

and draw the corresponding decision boundary on (r_1, r_2) plane. [5 marks]

(b) Show that

$$m{r}^Tm{h} \sim \mathcal{N}\left(\mu_{m{h}}, \sigma_{m{h}}^2\right)$$

where

$$\mu_{\boldsymbol{h}} = \mathbb{E}\left\{\boldsymbol{r}^T\boldsymbol{h}\right\} = ||\boldsymbol{h}||^2$$

$$\sigma_{\boldsymbol{h}}^2 = \operatorname{Var}\left(\boldsymbol{r}^T\boldsymbol{h}\right) = \mathbb{E}\left\{\left(\boldsymbol{r}^T\boldsymbol{h} - ||\boldsymbol{h}||^2\right)^2\right\} = \frac{N_0}{2}||\boldsymbol{h}||^2$$

with $||\cdot||$ denoting the Euclidean norm.

[10 marks]

(c) Hence show that the BER is given by

$$\Pr(\varepsilon) = \mathcal{Q}\left(\sqrt{2\frac{||\boldsymbol{h}||^2}{N_0}}\right)$$

where Q(z) denotes the Gaussian-Q function.²

[5 marks]

(d) Explain how the coherent BPSK BER formula is related to the above BER result. [5 marks]

 $^{{}^{2}\}mathcal{Q}(z) = \frac{1}{\sqrt{2\pi}} \int_{z}^{\infty} e^{-\frac{t^{2}}{2}} dt.$

4. (a) A transmitter uses the signals $\{s_i(t)\}$ to communicate one of 4 equally likely messages over an AWGN channel, where for i = 0, 1, 2, 3

$$s_i(t) = \begin{cases} \sqrt{\frac{2}{T}} \cos \left(2\pi \frac{t}{T} + \frac{\pi i}{2} \right), & 0 \le t < T \\ 0, & \text{elsewhere.} \end{cases}$$

Moreover, we adopt the following mapping to map bits to message waveforms:

$$(0,0) \mapsto s_0(t) \quad (0,1) \mapsto s_1(t) \quad (1,1) \mapsto s_2(t) \quad (1,0) \mapsto s_3(t)$$

where the bit interval T_b is related to the symbol interval via $T = 2T_b$. Now consider the binary sequence $S = \{1, 1, 0, 0, 1, 0, 0, 1, 1, 1, 1, 0, 0, 0, 1, 0\}$. Plot the sequence of signals (waveforms) corresponding to the binary sequence S for $0 \le t < 8T$.

(b) A certain quaternary minimum-shift keying (MSK) waveform can be written as

$$s(t) = \sqrt{\frac{2}{T}}\cos\left(2\pi\frac{t}{T} + \Phi_n(t)\right), \quad nT \le t \le (n+1)T$$

where

$$\Phi_n(t) = \underbrace{\frac{\pi}{2} \sum_{k=-\infty}^{n-1} X_k + \frac{\pi}{2} X_n \left(\frac{t}{T} - 1\right)}_{\theta_n}$$
$$= \theta_n + \frac{\pi}{2} X_n \left(\frac{t}{T} - 1\right)$$

and each $X_k \in \{-3, -1, +1, +3\}$ is chosen according to the mapping

$$(0,0) \mapsto -3 \quad (0,1) \mapsto -1 \quad (1,1) \mapsto +1 \quad (1,0) \mapsto +3.$$

Draw the quaternary MSK phase trajectory (i.e., $\Phi_n(t)$ vs. t) corresponding to the binary sequence S for $0 \le t \le 8T$. [10 marks]

(c) Plot the quaternary MSK waveform corresponding to the binary sequence S for $0 \le t \le 8T$. [10 marks]