

# 電子類元件 零件承認書文件 CHECK LIST

零件廠商：VISHAY(Weikeng)

品名規格：Power IC

技嘉料號：10IFD-400788-01R

項次	文件項目
<b>Data Sheet 檢核項目</b>	
1	DATASHEET (含機構尺寸、 <b>端子腳鍍層材質、MSL Report</b> )
2	零件 Making 文字面說明
3	零件 Part Number 說明
4	零件 Qualification Test Report
5	料件包裝方式及包裝 Label 之零件 Part number 說明
6	UL Safety Report (If Request )
7	零件耐溫焊接 Profile(包含最高耐焊溫度,時間, 焊接/過爐次數與曲線圖)。 <b>註 2</b>
8	零件樣品 20PCS(Chipset 等高單價, 至少 1PCS)
9	<b>電子零件承認基本調查表。註 3</b>
10	以上資料電子檔為 PDF 檔, 且是同 1 個 File
<b>GSCM 綠色產品管理系統-物料管制文件檢核清單</b>	
物料管制文件 1	GSCM 綠色產品管理系統：零件照片
物料管制文件 2	GSCM 綠色產品管理系統：不使用禁用物質證明書 (保證書)。 <b>註 4</b>
物料管制文件 3	GSCM 綠色產品管理系統：Data Sheet
<b>GSCM 綠色產品管理系統-MCD 表格</b>	
MCD 表格	物質內容宣告表格 (Material Content Declaration, MCD)
<b>其他文件 (僅適用電阻、電容類之系列元件)</b>	
附件 1	危害物質測試報告 Test Report of Hazardous Substances。 <b>註 5</b>
附件 2	元件調查表 Component Composition Table

- ※ 1. 各項說明文件內容應明訂零件交貨時之規格、方式；如捲盤、文字印刷為雷射或油墨等
- ※ 2. 零件耐溫焊接 Profile 需附相關測試報告 (國際認證之實驗室資格單位所出具之測試報告)
- 2.1. 基本需符合 JEDEC 規範
- 2.2. Ambient Temp. (Reflow Temp endure): >225°C, 70 sec. 零件塑膠材質需 PA9T(含)等級以上
- 2.3. PASTE IN HOLE 零件塑膠材質需 PA9T(含)等級以上
- ※ 3. 電子零件適用(技嘉)料號：**積體電路(IC) 10H\*,10T\*,10I\*,10D\*,10G\*,11T\***  
**非 IC 類：10C\*,11C\*,10L\*,11L\*,10X\*,11X\*,10R\*,11B\***
- ※ 4. 物料管制文件 2：網通事業群之所屬料件須一併提交 “不使用禁用物質證明書(保證書)+ REACH 調查表”
- ※ 5. 危害物質測試報告 Test Report of Hazardous Substances：泛指為具有 ISO/IEC 17025 國際認證之實驗室資格單位所出具之測試報告

## 電子零件承認基本調查表

一、原物料規格/來源			
項次	部位名稱/規格	材質	原物料來源產地
1	Bonding wire	Metal	SINGAPORE
2	Clip	Metal	Taiwan
3	Die Attach Adhesive	Adhesives	USA
4	Lead Frame	Metal	Taiwan
5	Molding Compound	Epoxy Resin	Japan
6	Silicon Wafer	Wafer	Germany
7	Solder Plating	Metal materials plating layer	China
8	Wafer back side coating	Epoxy	USA

二、晶圓廠(非 IC 類免填)					
項次	工廠名稱	生產產地	Wafer (吋)	投產率(%)	自有/外包
1	TOWER 以色列	以色列	8"	>50%	Outsource
2	Vanguard 台灣	台灣	8"	>50%	Outsource

三、封裝廠( IC 類)；成品之生產製造工廠(非 IC 類)				
項次	工廠名稱	生產產地	投產比率(%)	自有/外包
1	Vishay Siliconix TW	TW	>80%	Own
3				

四、產能	
總產能(月/PCS)	可供技嘉產能(月/PCS)
3M/M	400K/M

- ※ 1. IC 類之晶圓廠、封裝廠之所有 AVL 請均表列，並提供相關資訊與文件。當異動 (包含 AVL 或相關資訊文件之異動)時，請主動通知技嘉 Sourcer 與 RD 承認單位，並更新文件
- ※ 2. 非 IC 類零件之成品生產製造工廠之所有 AVL 請均表列，並提供相關資訊與文件。當異動 (包含 AVL 或相關資訊文件之異動)時，請主動通知技嘉 Sourcer 與 RD 承認單位，並更新文件
- ※ 3. 以上資訊欄位若有不足，可自行增加行數

## 50 A VRPower® Integrated Power Stage

### DESCRIPTION

The SiC788 and SiC788A are integrated power stage solutions optimized for synchronous buck applications to offer high current, high efficiency, and high power density performance. Packaged in Vishay's proprietary 6 mm x 6 mm MLP package, SiC788 and SiC788A enable voltage regulator designs to deliver up to 50 A continuous current per phase.

The internal power MOSFETs utilize Vishay's state-of-the-art Gen IV TrenchFET technology that delivers industry benchmark performance to significantly reduce switching and conduction losses.

The SiC788 and SiC788A incorporate an advanced MOSFET gate driver IC that features high current driving capability, adaptive dead-time control, an integrated bootstrap Schottky diode, a thermal warning (THWn) that alerts the system of excessive junction temperature, and skip mode (SMOD#) to improve light load efficiency. The drivers are also compatible with a wide range of PWM controllers and supports tri-state PWM, 3.3 V (SiC788A) / 5 V (SiC788) PWM logic.

### FEATURES

- Thermally enhanced PowerPAK® MLP66-40L package
- Vishay's Gen IV MOSFET technology and a low-side MOSFET with integrated Schottky diode
- Delivers up to 50 A continuous current
- 95 % peak efficiency
- High frequency operation up to 1.5 MHz
- Power MOSFETs optimized for 12 V input stage
- 3.3 V (SiC788A) / 5 V (SiC788) PWM logic with tri-state and hold-off
- SMOD# logic for light load efficiency improvement
- Low PWM propagation delay (< 20 ns)
- Thermal monitor flag
- Faster enable / disable
- Under voltage lockout for  $V_{CIN}$
- Material categorization: for definitions of compliance please see [www.vishay.com/doc?99912](http://www.vishay.com/doc?99912)



**RoHS**  
COMPLIANT  
HALOGEN  
**FREE**

### APPLICATIONS

- Multi-phase VRDs for CPU, GPU, and memory

### TYPICAL APPLICATION DIAGRAM

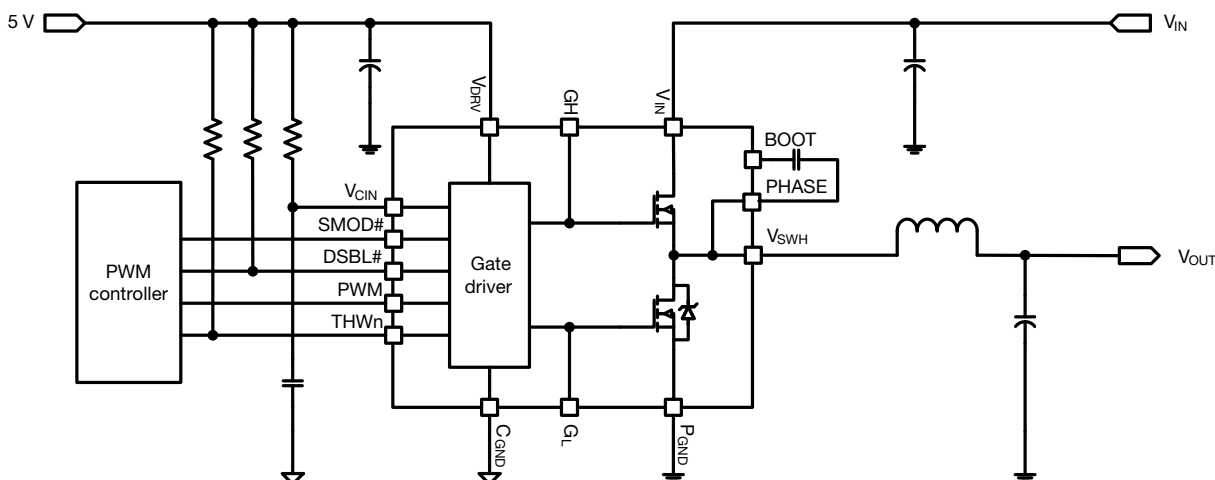
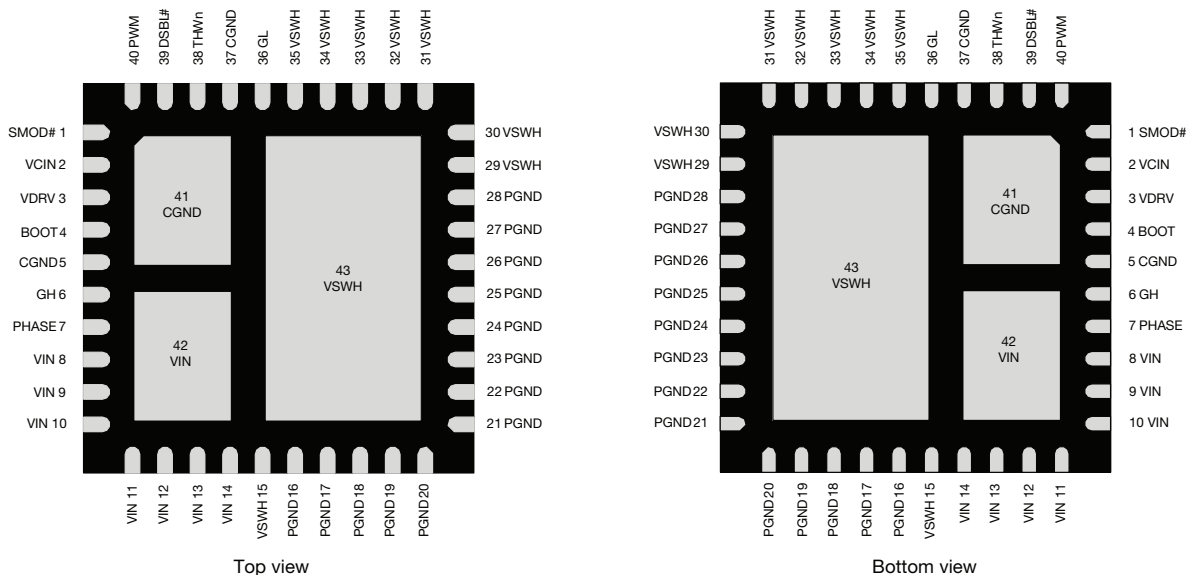


Fig. 1 - SiC788 and SiC788A Typical Application Diagram

## PINOUT CONFIGURATION



**Fig. 2 - SiC788 and SiC788A Pin Configuration**

PIN DESCRIPTION		
PIN NUMBER	NAME	FUNCTION
1	SMOD#	Low-side gate turn-off logic. Active low
2	V <sub>CIN</sub>	Supply voltage for internal logic circuitry
3	V <sub>DRV</sub>	Supply voltage for internal gate driver
4	BOOT	High-side driver bootstrap voltage
5, 37, 41	C <sub>GND</sub>	Analog ground for the driver IC
6	GH	High-side gate signal
7	PHASE	Return path of high-side gate driver
8 to 14, 42	V <sub>IN</sub>	Power stage input voltage. Drain of high-side MOSFET
15, 29 to 35, 43	V <sub>SWH</sub>	Switch node of the power stage
16 to 28	P <sub>GND</sub>	Power ground
36	GL	Low-side gate signal
38	THW <sub>n</sub>	Thermal warning open drain output
39	DSBL#	Disable pin. Active low
40	PWM	PWM control input

ORDERING INFORMATION			
PART NUMBER	PACKAGE	MARKING CODE	OPTION
SiC788ACD-T1-GE3	PowerPAK® MLP66-40L	SiC788A	3.3 V PWM optimized
SiC788CD-T1-GE3		SiC788	5 V PWM optimized
SiC788ADB and SiC788DB	Reference board		

**ABSOLUTE MAXIMUM RATINGS**

ELECTRICAL PARAMETER	CONDITIONS	LIMIT	UNIT
Input Voltage	$V_{IN}$	-0.3 to +25	V
Control Logic Supply Voltage	$V_{CIN}$	-0.3 to +7	
Drive Supply Voltage	$V_{DRV}$	-0.3 to +7	
Switch Node (DC voltage)	$V_{SWH}$	-0.3 to +25	
Switch Node (AC voltage) <sup>(1)</sup>		-8 to +30	
BOOT Voltage (DC voltage)	$V_{BOOT}$	32	
BOOT Voltage (AC voltage) <sup>(2)</sup>		38	
BOOT to PHASE (DC voltage)	$V_{BOOT- PHASE}$	-0.3 to +7	
BOOT to PHASE (AC voltage) <sup>(3)</sup>		-0.3 to +8	
All Logic Inputs and Outputs (PWM, DSBL#, and THWn)		-0.3 to $V_{CIN} + 0.3$	
Output Current, $I_{OUT(AV)}$ <sup>(4)</sup>	$f_S = 300 \text{ kHz}$ , $V_{IN} = 12 \text{ V}$ , $V_{OUT} = 1.8 \text{ V}$	50	A
	$f_S = 1 \text{ MHz}$ , $V_{IN} = 12 \text{ V}$ , $V_{OUT} = 1.8 \text{ V}$	40	
Max. Operating Junction Temperature	$T_J$	150	°C
Ambient Temperature	$T_A$	-40 to +125	
Storage Temperature	$T_{stg}$	-65 to +150	
Electrostatic Discharge Protection	Human body model, JESD22-A114	5000	V
	Charged device model, JESD22-C101	1000	

**Notes**

- (1) The specification values indicated “AC” is  $V_{SWH}$  to  $P_{GND}$ , -8 V (< 20 ns, 10  $\mu$ J), min. and 30 V (< 50 ns), max.
- (2) The specification value indicates “AC voltage” is  $V_{BOOT}$  to  $P_{GND}$ , 36 V (< 50 ns) max.
- (3) The specification value indicates “AC voltage” is  $V_{BOOT}$  to  $V_{PHASE}$ , 8 V (< 20 ns) max.
- (4) Output current rated with testing evaluation board at  $T_A = 25^\circ\text{C}$  with natural convection cooling. The rating is limited by the peak evaluation board temperature,  $T_J = 150^\circ\text{C}$ , and varies depending on the operating conditions and PCB layout. This rating may be changed with different application settings.

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**RECOMMENDED OPERATING RANGE**

ELECTRICAL PARAMETER	MINIMUM	TYPICAL	MAXIMUM	UNIT
Input Voltage ( $V_{IN}$ )	4.5	-	18	V
Drive Supply Voltage ( $V_{DRV}$ )	4.5	5	5.5	
Control Logic Supply Voltage ( $V_{CIN}$ )	4.5	5	5.5	
BOOT to PHASE ( $V_{BOOT-PHASE}$ , DC voltage)	4	4.5	5.5	
Thermal Resistance from Junction to PAD	-	1	-	°C/W
Thermal Resistance from Junction to Case	-	2.5	-	

**ELECTRICAL SPECIFICATIONS**(DSBL# = SMOD# = 5 V,  $V_{IN}$  = 12 V,  $V_{DRV}$  and  $V_{CIN}$  = 5 V,  $T_A$  = 25 °C)

PARAMETER	SYMBOL	TEST CONDITION	LIMITS			UNIT
			MIN.	TYP.	MAX.	
POWER SUPPLY						
Control Logic Supply Current	I <sub>VCIN</sub>	V <sub>DSBL#</sub> = 0 V, no switching, V <sub>PWM</sub> = FLOAT	-	85	-	μA
		V <sub>DSBL#</sub> = 5 V, no switching, V <sub>PWM</sub> = FLOAT	-	290	-	
		V <sub>DSBL#</sub> = 5 V, f <sub>S</sub> = 300 kHz, D = 0.1	-	295	-	
Drive Supply Current	I <sub>VDRV</sub>	f <sub>S</sub> = 300 kHz, D = 0.1	-	9	15	mA
		f <sub>S</sub> = 1 MHz, D = 0.1	-	30	-	
		V <sub>DSBL#</sub> = 0 V, no switching	-	30	-	μA
		V <sub>DSBL#</sub> = 5 V, no switching	-	55	-	
BOOTSTRAP SUPPLY						
Bootstrap Diode Forward Voltage	V <sub>F</sub>	I <sub>F</sub> = 2 mA			0.4	V
PWM CONTROL INPUT (SiC788)						
Rising Threshold	V <sub>TH_PWM_R</sub>		3.4	3.7	4.0	V
Falling Threshold	V <sub>TH_PWM_F</sub>		0.72	0.9	1.1	
Tri-state Voltage	V <sub>TRI</sub>	V <sub>PWM</sub> = FLOAT	-	2.3	-	
Tri-state Rising Threshold	V <sub>TRI_TH_R</sub>		0.9	1.15	1.38	
Tri-state Falling Threshold	V <sub>TRI_TH_F</sub>		3.1	3.35	3.6	
Tri-state Rising Threshold Hysteresis	V <sub>HYS_TRI_R</sub>		-	225	-	mV
Tri-state Falling Threshold Hysteresis	V <sub>HYS_TRI_F</sub>		-	325	-	
PWM Input Current	I <sub>PWM</sub>	V <sub>PWM</sub> = 5 V	-	-	350	μA
		V <sub>PWM</sub> = 0 V	-	-	-350	
PWM CONTROL INPUT (SiC788A)						
Rising Threshold	V <sub>TH_PWM_R</sub>		2.2	2.45	2.7	V
Falling Threshold	V <sub>TH_PWM_F</sub>		0.72	0.9	1.1	
Tri-state Voltage	V <sub>TRI</sub>	V <sub>PWM</sub> = FLOAT	-	1.8	-	
Tri-state Rising Threshold	V <sub>TRI_TH_R</sub>		0.9	1.15	1.38	
Tri-state Falling Threshold	V <sub>TRI_TH_F</sub>		1.95	2.2	2.45	
Tri-state Rising Threshold Hysteresis	V <sub>HYS_TRI_R</sub>		-	225	-	mV
Tri-state Falling Threshold Hysteresis	V <sub>HYS_TRI_F</sub>		-	275	-	
PWM Input Current	I <sub>PWM</sub>	V <sub>PWM</sub> = 3.3 V	-	-	225	μA
		V <sub>PWM</sub> = 0 V	-	-	-225	
TIMING SPECIFICATIONS						
Tri-State to GH/GL Rising Propagation Delay	t <sub>PD_TRI_R</sub>	No load, see fig. 4	-	30	-	ns
Tri-state Hold-off Time	t <sub>TSHO</sub>		-	130	-	
GH - Turn Off Propagation Delay	t <sub>PD_OFF_GH</sub>		-	18	-	
GH - Turn On Propagation Delay (Dead time rising)	t <sub>PD_ON_GH</sub>		-	15	-	
GL - Turn Off Propagation Delay	t <sub>PD_OFF_GL</sub>		-	12	-	
GL - Turn On Propagation Delay (Dead time falling)	t <sub>PD_ON_GL</sub>		-	8	-	
DSBL# Low to GH/GL Falling Propagation Delay	t <sub>PD_DSBL#_F</sub>	Fig. 5	-	15	-	
DSBL# High to GH/GL Rising Propagation Delay	t <sub>PD_DSBL#_R</sub>	Fig. 5	-	20	-	
PWM Minimum On-time	t <sub>PWM_ON_MIN</sub>		30	-	-	
DSBL# SMOD# INPUT						
DSBL# Logic Input Voltage	V <sub>IH_DSBL#</sub>	Input logic high	2	-	-	V
	V <sub>IL_DSBL#</sub>	Input logic low	-	-	0.8	
SMOD# Logic Input Voltage	V <sub>IH_SMOD#</sub>	Input logic high	2	-	-	
	V <sub>IL_SMOD#</sub>	Input logic low	-	-	0.8	

**ELECTRICAL SPECIFICATIONS**(DSBL# = SMOD# = 5 V,  $V_{IN} = 12$  V,  $V_{DRV}$  and  $V_{CIN} = 5$  V,  $T_A = 25$  °C)

PARAMETER	SYMBOL	TEST CONDITION	LIMITS			UNIT
			MIN.	TYP.	MAX.	
PROTECTION						
Under Voltage Lockout	V <sub>UVLO</sub>	V <sub>CIN</sub> rising, on threshold	-	3.7	4.1	V
		V <sub>CIN</sub> falling, off threshold	2.7	3.1	-	
Under Voltage Lockout Hysteresis	V <sub>UVLO_HYST</sub>		-	575	-	mV
THWn Flag Set <sup>(2)</sup>	T <sub>THWn_SET</sub>		-	160	-	°C
THWn Flag Clear <sup>(2)</sup>	T <sub>THWn_CLEAR</sub>		-	135	-	
THWn Flag Hysteresis <sup>(2)</sup>	T <sub>THWn_HYST</sub>		-	25	-	
THWn Output Low	V <sub>OL THWn</sub>	I <sub>THWn</sub> = 2 mA	-	0.02	-	V

**Notes**

(1) Typical limits are established by characterization and are not production tested.

(2) Guaranteed by design.

**DETAILED OPERATIONAL DESCRIPTION****PWM Input with Tri-state Function**

The PWM input receives the PWM control signal from the VR controller IC. The PWM input is designed to be compatible with standard controllers using two state logic (H and L) and advanced controllers that incorporate tri-state logic (H, L and tri-state) on the PWM output. For two state logic, the PWM input operates as follows. When PWM is driven above  $V_{PWM\_TH\_R}$  the low-side is turned OFF and the high-side is turned ON. When PWM input is driven below  $V_{PWM\_TH\_F}$  the high-side is turned OFF and the low-side is turned ON. For tri-state logic, the PWM input operates as previously stated for driving the MOSFETs when PWM is logic high and logic low. However, there is a third state that is entered as the PWM output of tri-state compatible controller enters its high impedance state during shut-down. The high impedance state of the controller's PWM output allows the SiC788 and SiC788A to pull the PWM input into the tri-state region (see definition of PWM logic and tri-state, fig. 4). If the PWM input stays in this region for the tri-state hold-off period,  $t_{TSHO}$ , both high-side and low-side MOSFETs are turned OFF. The function allows the VR phase to be disabled without negative output voltage swing caused by inductor ringing and saves a Schottky diode clamp. The PWM and tri-state regions are separated by hysteresis to prevent false triggering. The SiC788A incorporates PWM voltage thresholds that are compatible with 3.3 V logic and the SiC788 thresholds are compatible with 5 V logic.

**Disable (DSBL#)**

In the low state, the DSBL# pin shuts down the driver IC and disables both high-side and low-side MOSFETs. In this state, standby current is minimized. If DSBL# is left unconnected, an internal pull-down resistor will pull the pin to  $C_{GND}$  and shut down the IC.

**Pre-Charger Function**

When DSBL# is driven from below  $V_{IL\_DSBL\#}$  to above  $V_{IH\_DSBL\#}$  the low-side is turned ON for a short duration (60 ns typical) to refresh the BOOT capacitor in case it has been discharged due to the driver being in standby for a long period of time.

**Diode Emulation Mode (SMOD#)**

When SMOD# is logic low diode emulation mode is enabled and the low-side is turned OFF. This is a non-synchronous conversion mode that improves light load efficiency by reducing switching losses. Conducted losses that occur in synchronous buck regulators when inductor current is negative can also be reduced. Circuitry in the external controller IC detects when inductor current crosses zero and drives SMOD# below  $V_{IL\_SMOD\#}$  turning the low-side MOSFET OFF. The function can be also be used for a pre-biased output voltage. If SMOD# is left unconnected, an internal pull up resistor will pull the pin to  $V_{CIN}$  (logic high) to disable the SMOD# function.

**Thermal Shutdown Warning (THWn)**

The THWn pin is an open drain signal that flags the presence of excessive junction temperature. Connect, with a maximum of 20 k $\Omega$ , to  $V_{CIN}$ . An internal temperature sensor detects the junction temperature. The temperature threshold is 160 °C. When this junction temperature is exceeded the THWn flag is set. When the junction temperature drops below 135 °C the device will clear the THWn signal. The SiC788 and SiC788A do not stop operation when the flag is set. The decision to shutdown must be made by an external thermal control function.

**Voltage Input ( $V_{IN}$ )**

This is the power input to the drain of the high-side power MOSFET. This pin is connected to the high power intermediate BUS rail.



### Switch Node ( $V_{SWH}$ and PHASE)

The switch node,  $V_{SWH}$ , is the circuit power stage output. This is the output applied to the power inductor and output filter to deliver the output for the buck converter. The PHASE pin is internally connected to the switch node,  $V_{SWH}$ . This pin is to be used exclusively as the return pin for the BOOT capacitor. A 20 k $\Omega$  resistor is connected between GH and PHASE to provide a discharge path for the HS MOSFET in the event that  $V_{CIN}$  goes to zero while  $V_{IN}$  is still applied.

### Ground Connections ( $C_{GND}$ and $P_{GND}$ )

$P_{GND}$  (power ground) should be externally connected to  $C_{GND}$  (control signal ground). The layout of the printed circuit board should be such that the inductance separating  $C_{GND}$  and  $P_{GND}$  is minimized. Transient differences due to inductance effects between these two pins should not exceed 0.5 V.

### Control and Drive Supply Voltage Input ( $V_{DRV}$ , $V_{CIN}$ )

$V_{CIN}$  is the bias supply for the gate drive control IC.  $V_{DRV}$  is the bias supply for the gate drivers. It is recommended to separate these pins through a resistor. This creates a low pass filtering effect to avoid coupling of high frequency gate drive noise into the IC.

### Bootstrap Circuit (BOOT)

The internal bootstrap diode and an external bootstrap capacitor form a charge pump that supplies voltage to the BOOT pin. An integrated bootstrap diode is incorporated so that only an external capacitor is necessary to complete the bootstrap circuit. Connect a bootstrap capacitor with one leg tied to BOOT pin and the other tied to PHASE pin.

### Shoot-Through Protection and Adaptive Dead Time

The SiC788 and SiC788A have an internal adaptive logic to avoid shoot through and optimize dead time. The shoot through protection ensures that both high-side and low-side MOSFETs are not turned ON at the same time. The adaptive dead time control operates as follows. The high-side and low-side gate voltages are monitored to prevent the MOSFET turning ON from turning ON until the other MOSFET's gate voltage is sufficiently low (< 1 V). Built-in delays also ensure that one power MOSFET is completely OFF, before the other can be turned ON. This feature helps to adjust dead time as gate transitions change with respect to output current and temperature.

### Under Voltage Lockout (UVLO)

During the start up cycle, the UVLO disables the gate drive, holding high-side and low-side MOSFET gates low, until the supply voltage rail has reached a point at which the logic circuitry can be safely activated. The SiC788 and SiC788A also incorporate logic to clamp the gate drive signals to zero when the UVLO falling edge triggers the shutdown of the device. As an added precaution, a 20 k $\Omega$  resistor is connected between GH and PHASE to provide a discharge path for the HS MOSFET.

## FUNCTIONAL BLOCK DIAGRAM

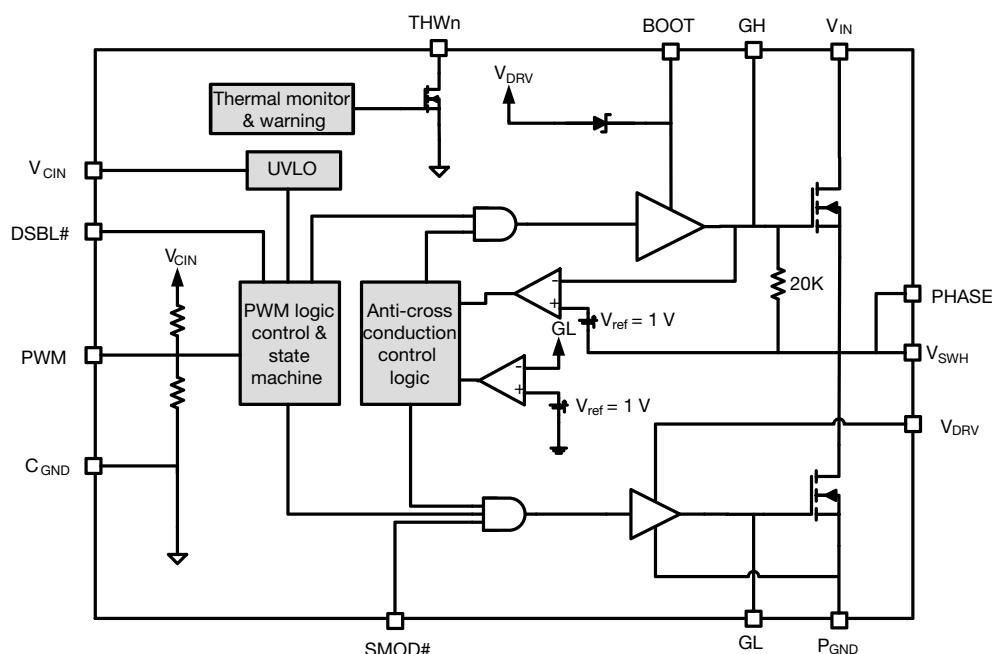


Fig. 3 - SiC788 and SiC788A Functional Block Diagram



DEVICE TRUTH TABLE				
DSBL#	SMOD#	PWM	GH	GL
Open	X	X	L	L
L	X	X	L	L
H	L	L	L	L
H	L	H	H	L
H	L	Tri-state	L	L
H	H	L	L	H
H	H	H	H	L
H	H	Tri-state	L	L

## PWM TIMING DIAGRAM

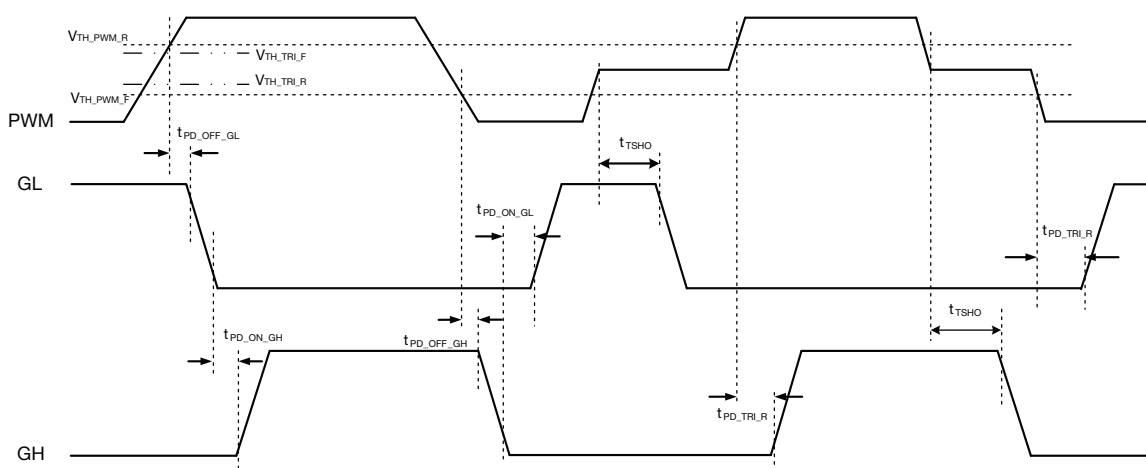
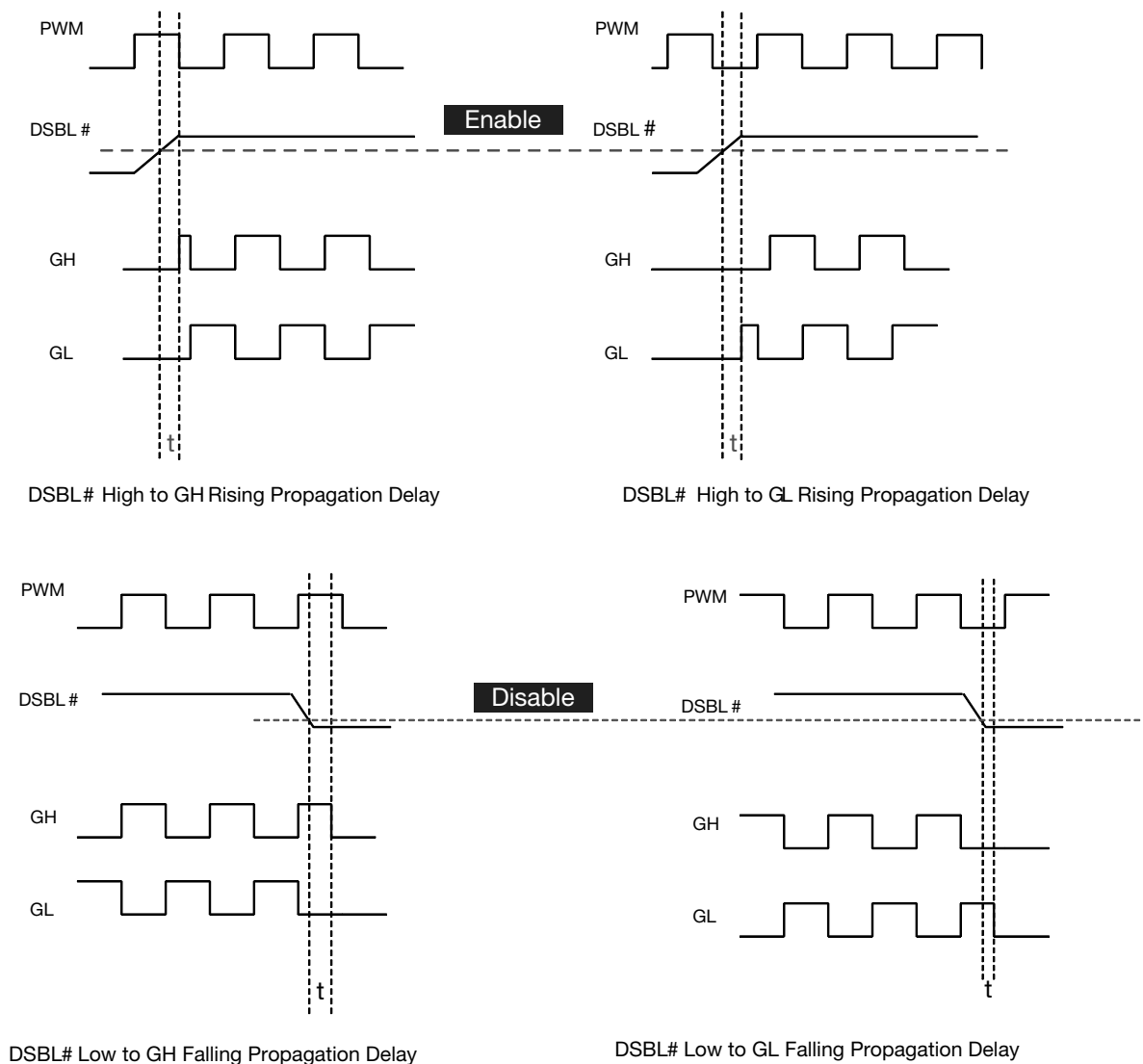
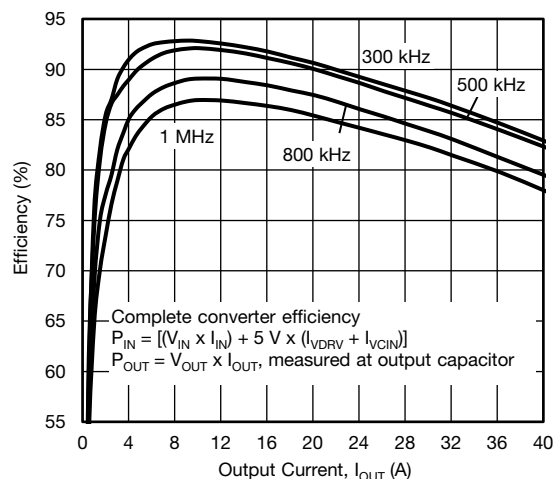


Fig. 4 - Definition of PWM Logic and Tri-State

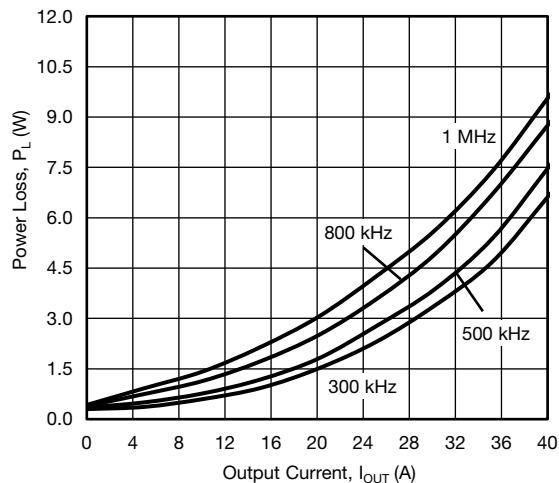
**OPERATION TIMING DIAGRAM: DSBL#**

**Fig. 5 - DSBL# Propagation Delay**

## ELECTRICAL CHARACTERISTICS

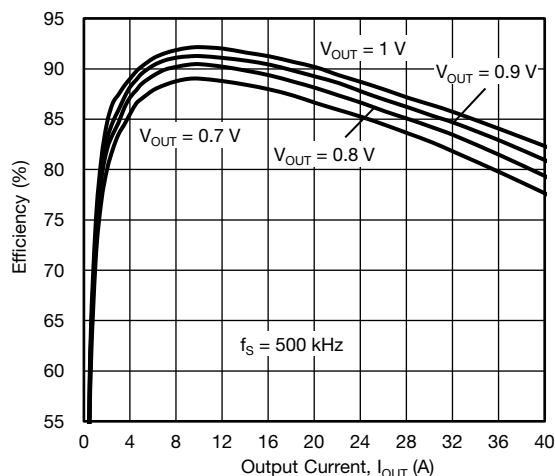
Test condition:  $V_{IN} = 12\text{ V}$ ,  $V_{DRV} = V_{CIN} = 5\text{ V}$ ,  $DSBL\# = SMOD\# = 5\text{ V}$ ,  $V_{OUT} = 1.8\text{ V}$ ,  $L_{OUT} = 250\text{ nH}$  (DCR = 0.32 mΩ),  $T_A = 25\text{ °C}$ , natural convection cooling (All power loss and normalized power loss curves show SiC788 and SiC788A losses only unless otherwise stated)



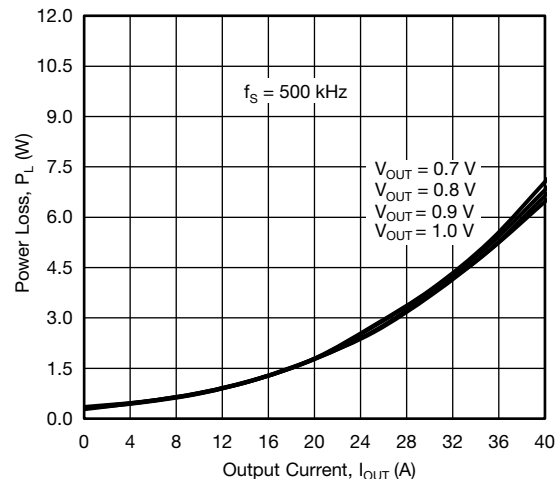
**Fig. 6 - Efficiency vs. Output Current**



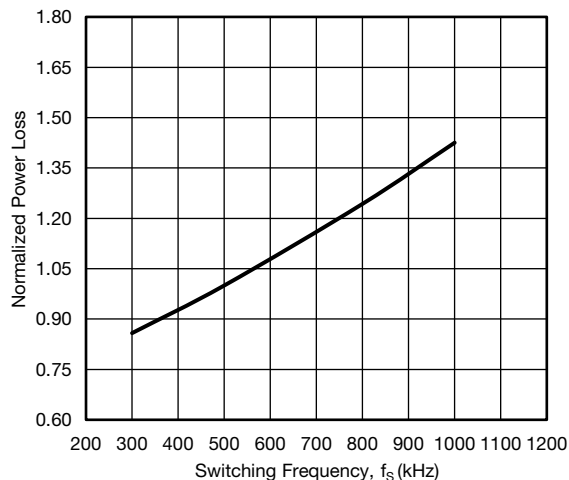
**Fig. 9 - Power Loss vs. Output Current**



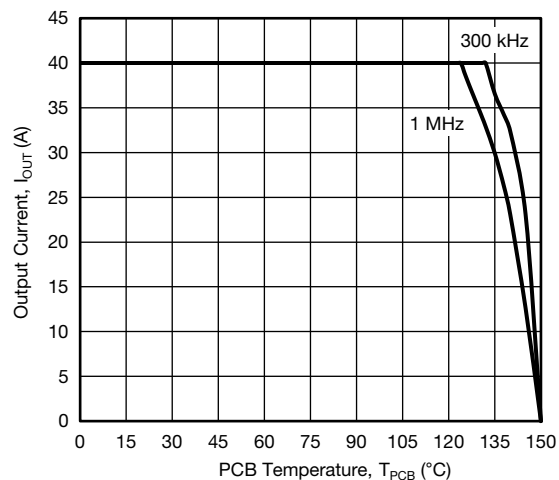
**Fig. 7 - Efficiency vs. Output Current**



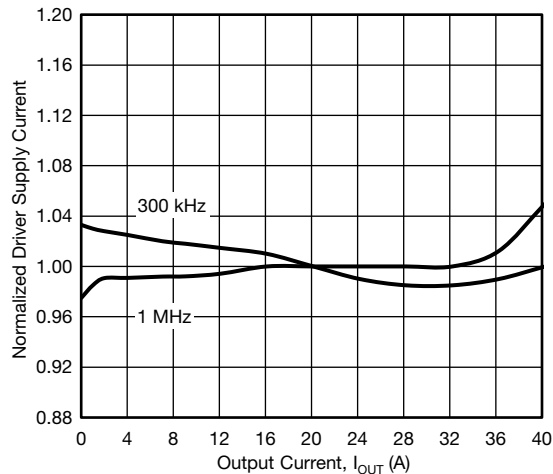
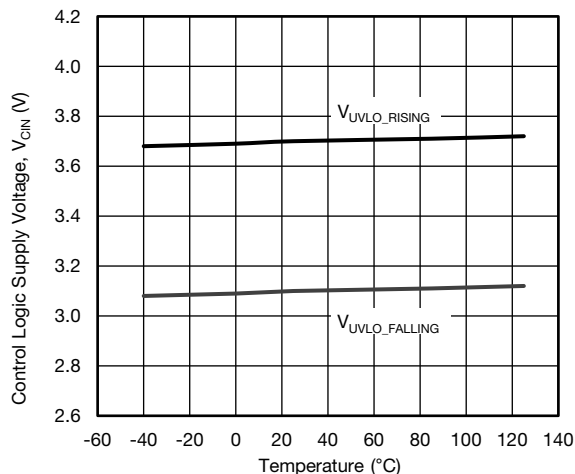
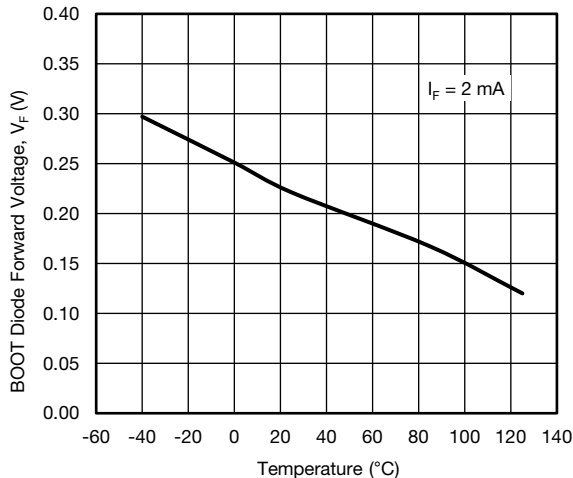
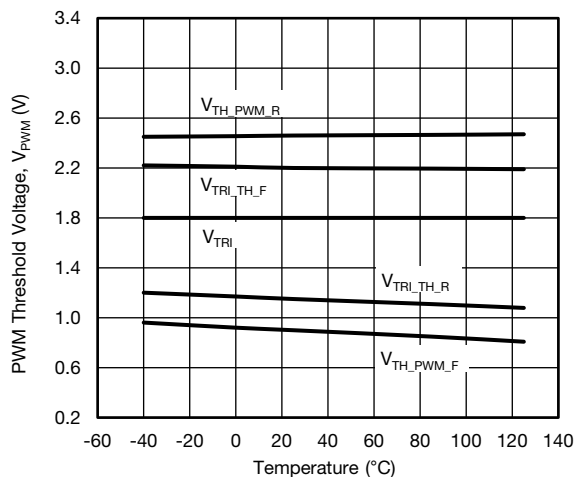
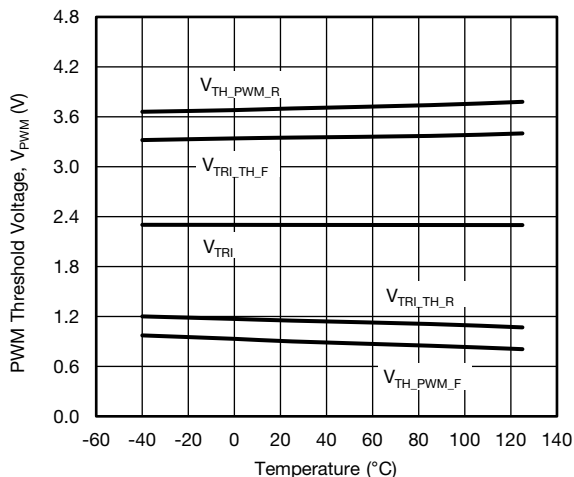
**Fig. 10 - Power Loss vs. Output Current**

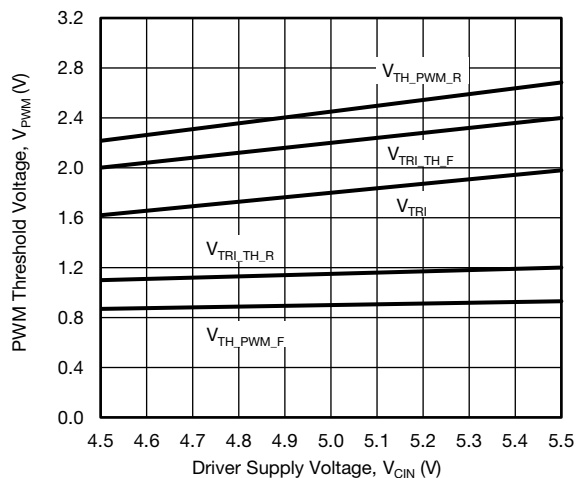
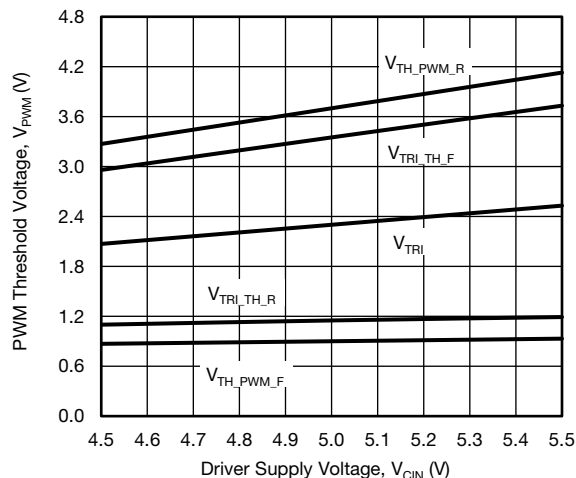
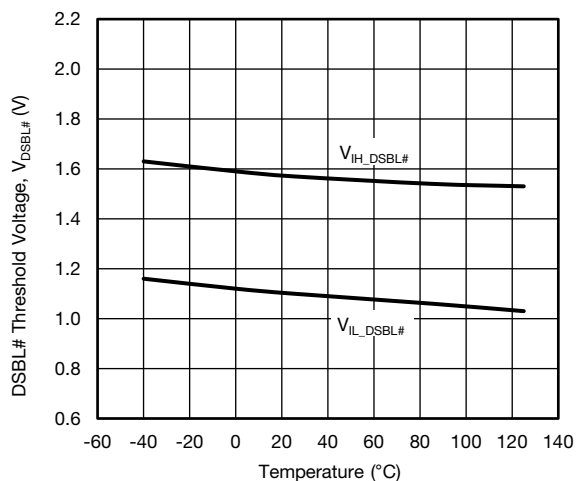
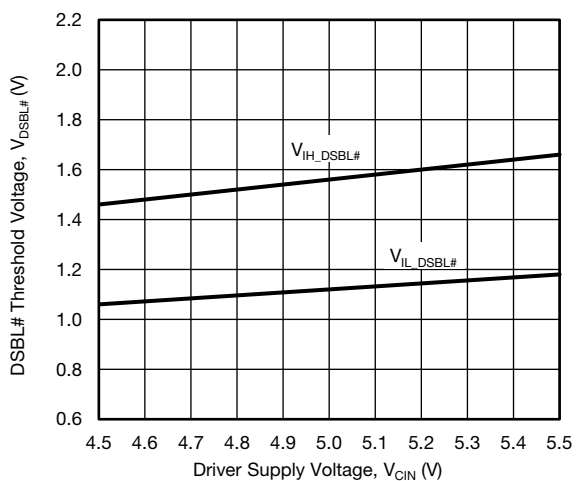
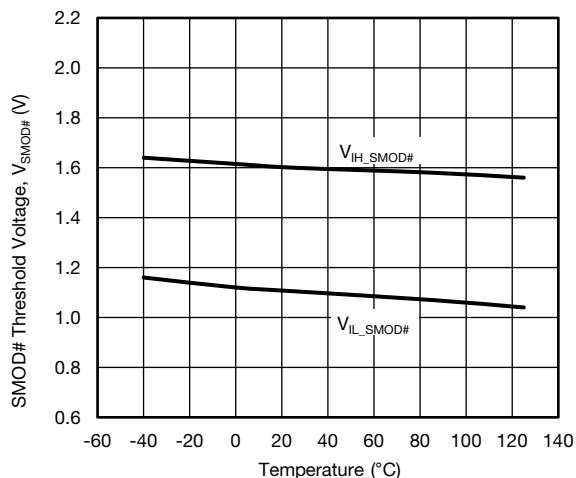
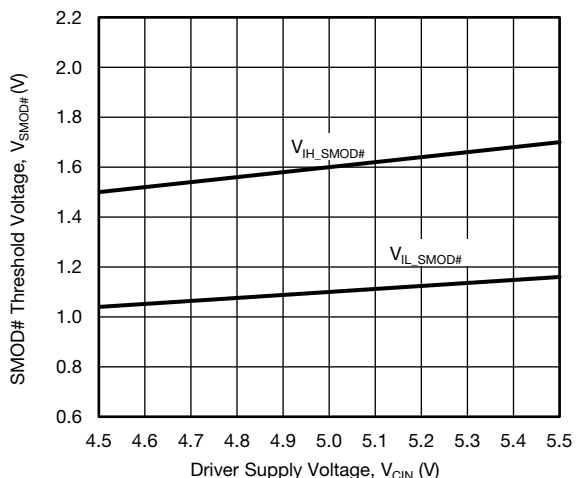


**Fig. 8 - Power Loss vs. Switching Frequency**



**Fig. 11 - Safe Operating Area**


**Fig. 12 - Driver Supply Current vs. Output Current**

**Fig. 13 - UVLO Threshold vs. Temperature**

**Fig. 15 - BOOT Diode Forward Voltage vs. Temperature**

**Fig. 14 - PWM Threshold vs. Temperature (SiC788A)**

**Fig. 16 - PWM Threshold vs. Temperature (SiC788)**


**Fig. 17 - PWM Threshold vs. Driver Supply Voltage (SiC788A)**

**Fig. 20 - PWM Threshold vs. Driver Supply Voltage (SiC788)**

**Fig. 18 - DSBL# Threshold vs. Temperature**

**Fig. 21 - DSBL# Threshold vs. Driver Supply Voltage**

**Fig. 19 - SMOD# Threshold vs. Temperature**

**Fig. 22 - SMOD# Threshold vs. Driver Supply Voltage**

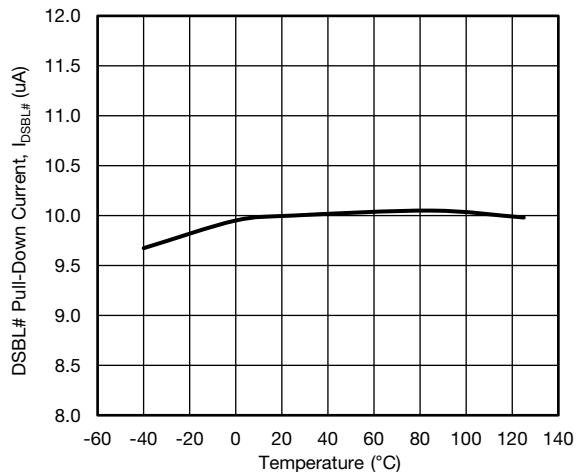


Fig. 23 - DSBL# Pull-down Current vs. Temperature

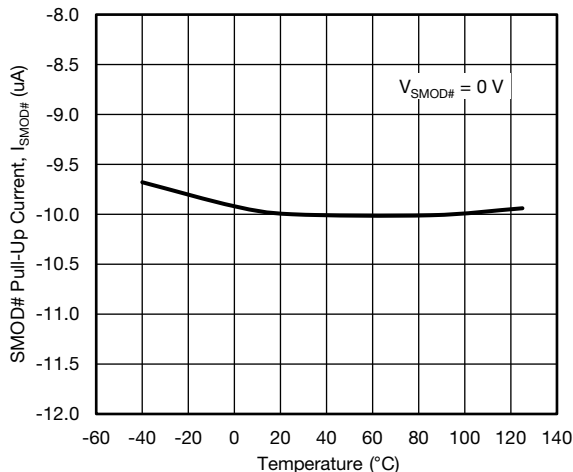


Fig. 25 - SMOD# Pull-up Current vs. Temperature

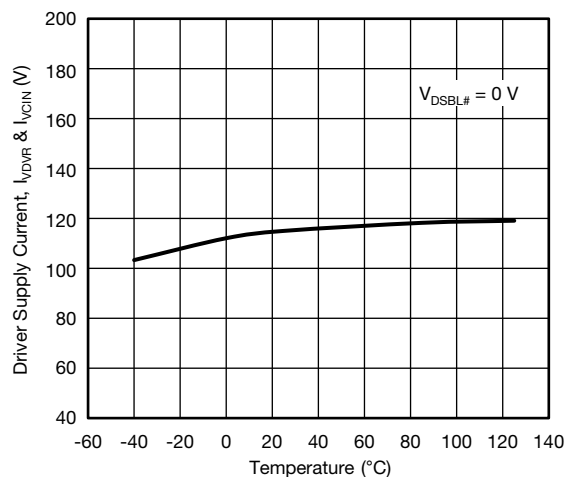


Fig. 24 - Driver Quiescent Current vs. Temperature

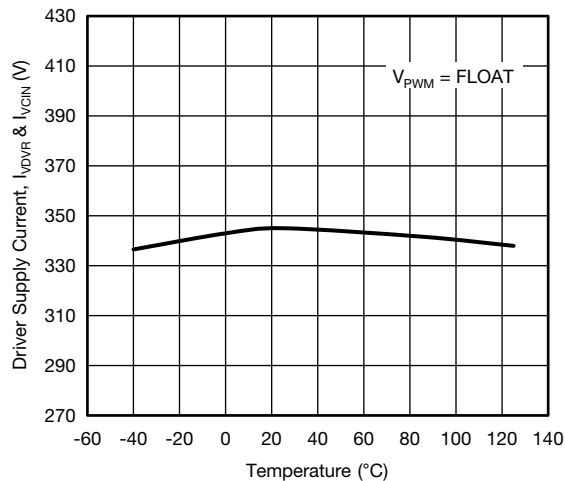
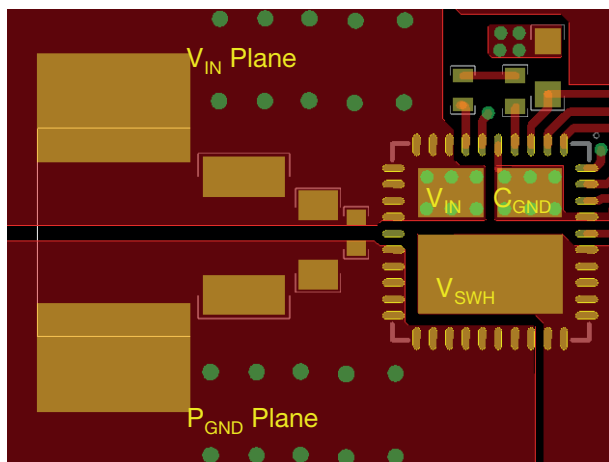


Fig. 26 - Driver Quiescent Current vs. Temperature

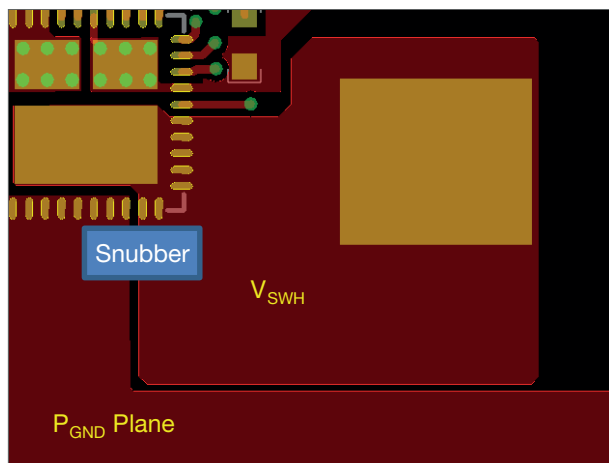
## PCB LAYOUT RECOMMENDATIONS

### Step 1: $V_{IN}$ / $P_{GND}$ Planes and Decoupling



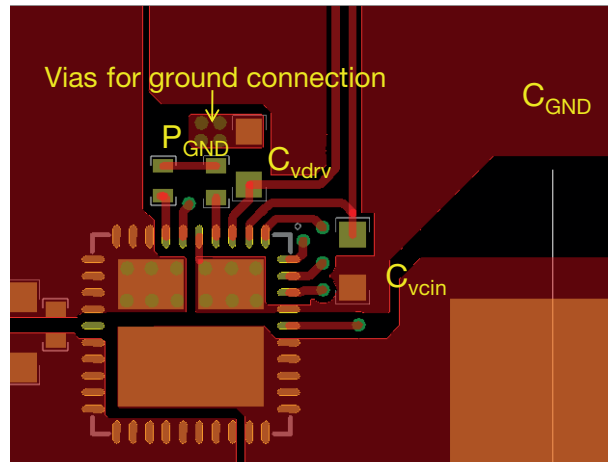
1. Layout  $V_{IN}$  and  $P_{GND}$  planes as shown above
2. Ceramic capacitors should be placed directly between  $V_{IN}$  and  $P_{GND}$ , and as close as possible to IC for best decoupling effect
3. Different ceramic capacitor values and packages should be used to cover entire decoupling spectrum, e.g. 1210, 0805, 0603, and 0402
4. Smaller capacitance values, placed closer to the IC's  $V_{IN}$  pin(s), result in better high frequency noise absorbing

### Step 2: $V_{SWH}$ Plane



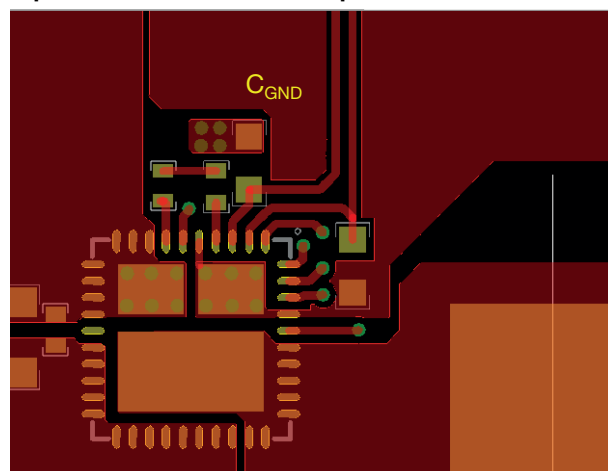
1. Connect output inductor to IC with large plane to lower resistance
2.  $V_{SWH}$  plane also serves as a heat-sink for low-side MOSFET. Please make the plane wide and short to achieve best thermal path
3. If a snubber network is required, place components as shown above

### Step 3: $V_{CIN}$ / $V_{DRV}$ Input Filter



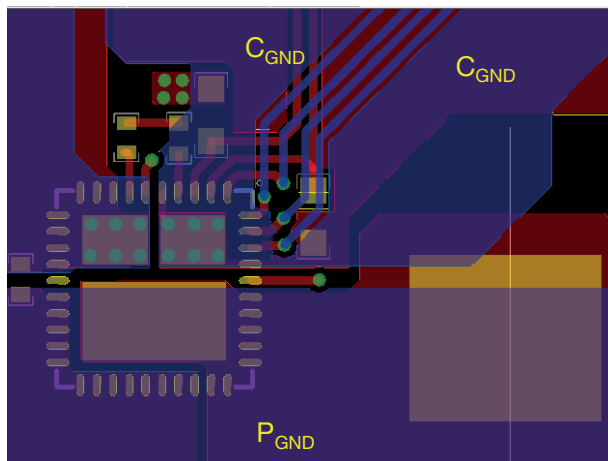
1.  $V_{CIN}$  /  $V_{DRV}$  input filter ceramic capacitors should be placed as close as possible to IC. It is recommended to connect two capacitors separately
2.  $V_{CIN}$  capacitor should be placed between pin 2 and pin 37 ( $C_{GND}$  of driver IC) to achieve best noise filtering
3.  $V_{DRV}$  capacitor should be placed between pin 3 and  $P_{GND}$  to provide maximum instantaneous driver current for low-side MOSFET during switching cycle.  $P_{GND}$  can be connected to inner ground plane through vias, as shown above
4. Pin 5 and pin 37 should be connected with  $C_{GND}$  pad, as shown above
5. For connecting  $V_{CIN}$  to  $C_{GND}$ , it is recommended to use a large plane to reduce parasitic inductance

### Step 4: BOOT Resistor and Capacitor Placement

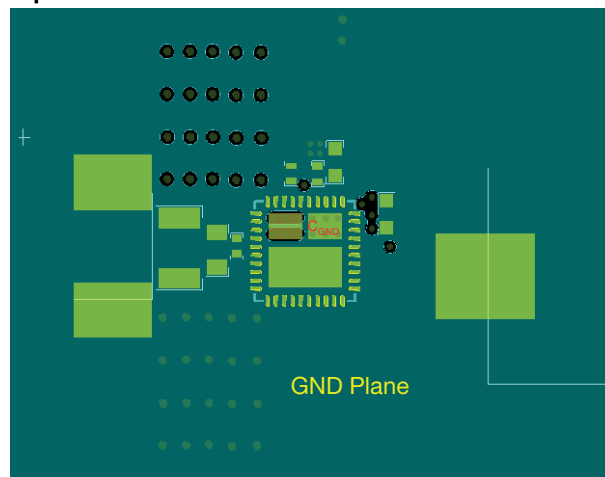


1. The components need to be placed as close as possible to IC, directly between PHASE (pin 7) and BOOT (pin 4)
2. To reduce parasitic inductance, 0402 package size can be used

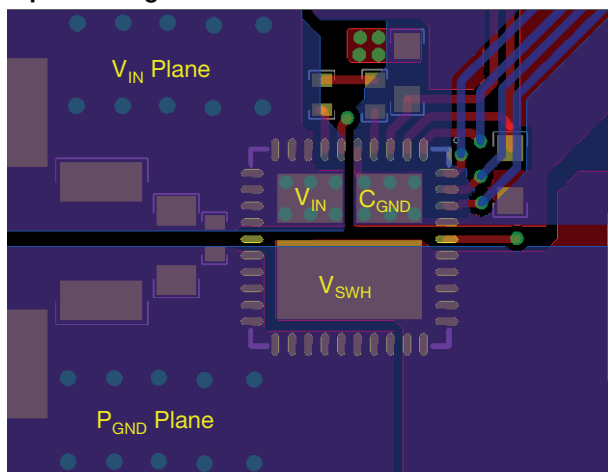


**Step 5: Signal Routing**


1. Route the PWM, SMOD#, DSBL#, and THWN signal traces out of the top right corner, next to pin 1
2. The PWM signal is a very important signal, both signal and return traces should not cross any power nodes on any layer
3. It is best to “shield” these traces from power switching nodes, e.g.  $V_{SWH}$ , with a GND island to improve signal integrity

**Step 7: Ground Connection**


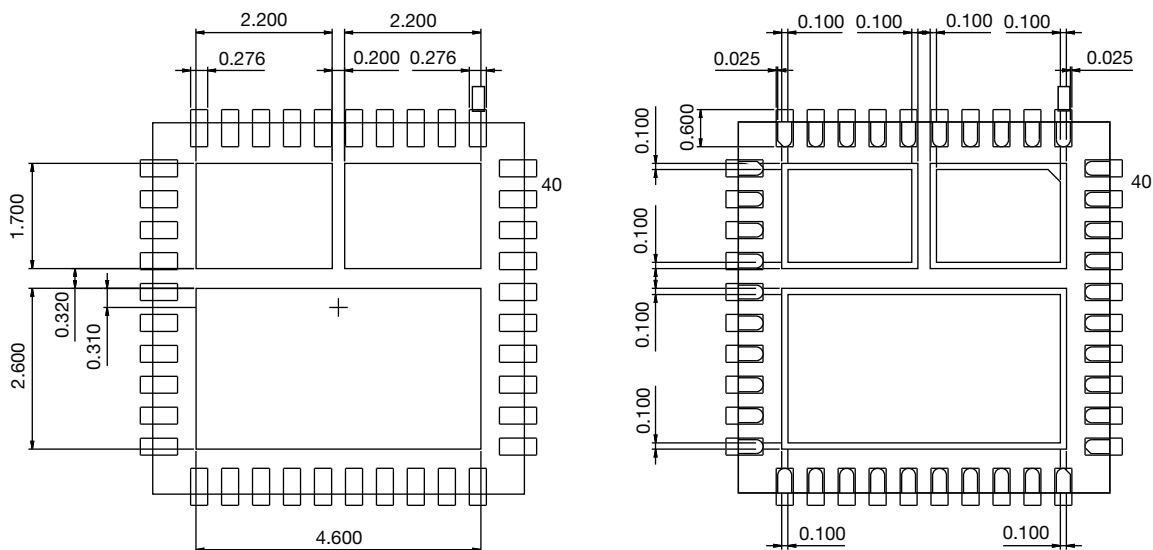
1. It is recommended to make the entire first inner layer (below top layer) the ground plane
2. The ground plane provides analog ground and power ground connections
3. The ground plane provides shielding between noise source on top layer and signal traces on bottom layer

**Step 6: Adding Thermal Relief Vias**


1. Thermal relief vias can be added to the  $V_{IN}$  and  $C_{GND}$  pads to utilize inner layers for high-current and thermal dissipation
2. To achieve better thermal performance, additional vias can be added to  $V_{IN}$  and  $P_{GND}$  planes
3. The  $V_{SWH}$  pad is a noise source and it is not recommended to place vias on this pad
4. 8 mil vias for pads and 10 mils vias for planes are the optimal sizes. Vias on pad may drain solder during assembly and cause assembly issues. Please consult with the assembly house for guidelines

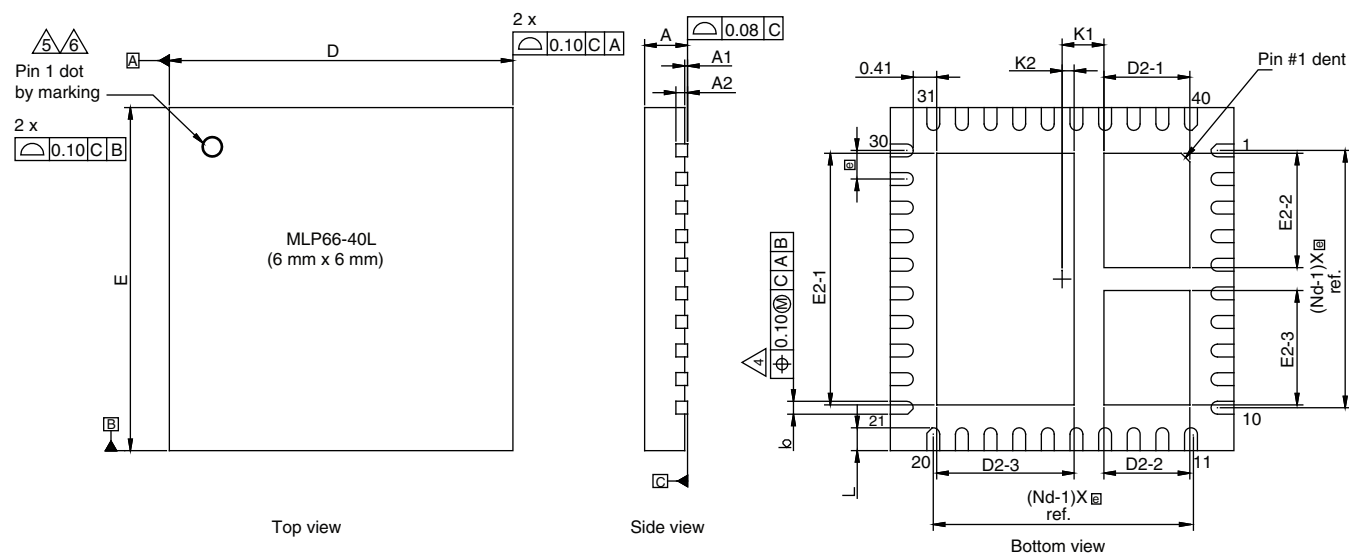


**RECOMMENDED LAND PATTERN PowerPAK® MLP66-40L** in millimeters





## PACKAGE OUTLINE DRAWING MLP66-40L

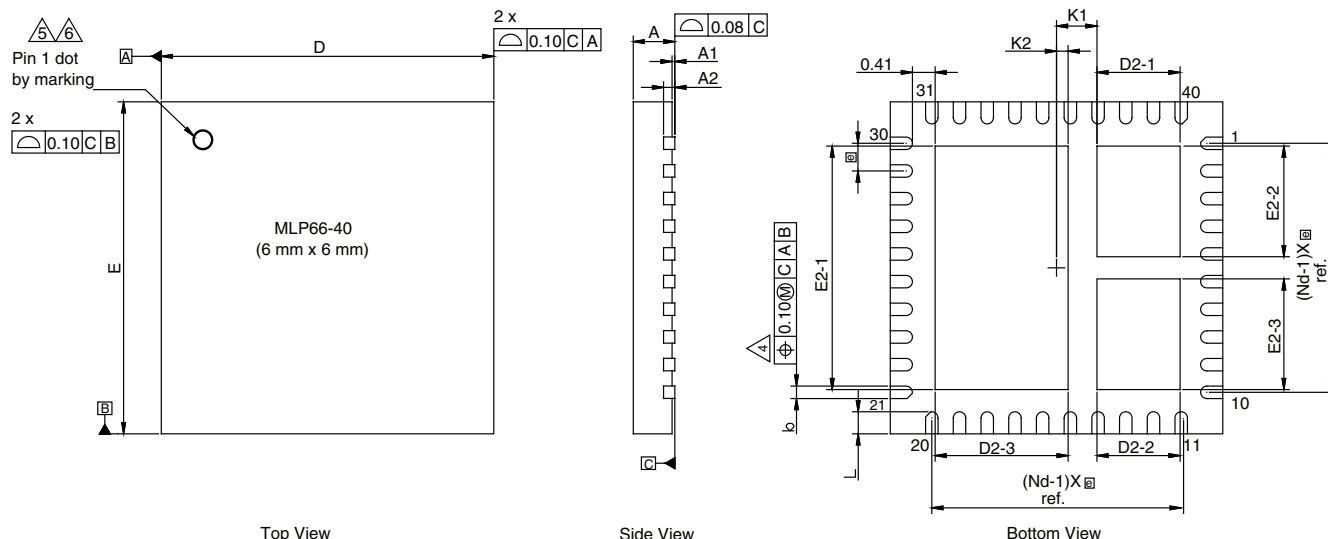


DIM.	MILLIMETERS			INCHES		
	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.
A	0.70	0.75	0.80	0.027	0.029	0.031
A1	0.00	-	0.05	0.000	-	0.002
A2	0.20 ref.			0.008 ref.		
b	0.20	0.25	0.30	0.078	0.098	0.111
D	6.00 BSC			0.236 BSC		
e	0.50 BSC			0.019 BSC		
E	6.00 BSC			0.236 BSC		
L	0.35	0.40	0.45	0.013	0.015	0.017
N	40			40		
Nd	10			10		
Ne	10			10		
D2-1	1.45	1.50	1.55	0.057	0.059	0.061
D2-2	1.45	1.50	1.55	0.057	0.059	0.061
D2-3	2.35	2.40	2.45	0.095	0.094	0.096
E2-1	4.35	4.40	4.45	0.171	0.173	0.175
E2-2	1.95	2.00	2.05	0.076	0.078	0.080
E2-3	1.95	2.00	2.05	0.076	0.078	0.080
K1	0.73 BSC			0.028 BSC		
K2	0.21 BSC			0.008 BSC		

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## PowerPAK® MLP66-40 Case Outline



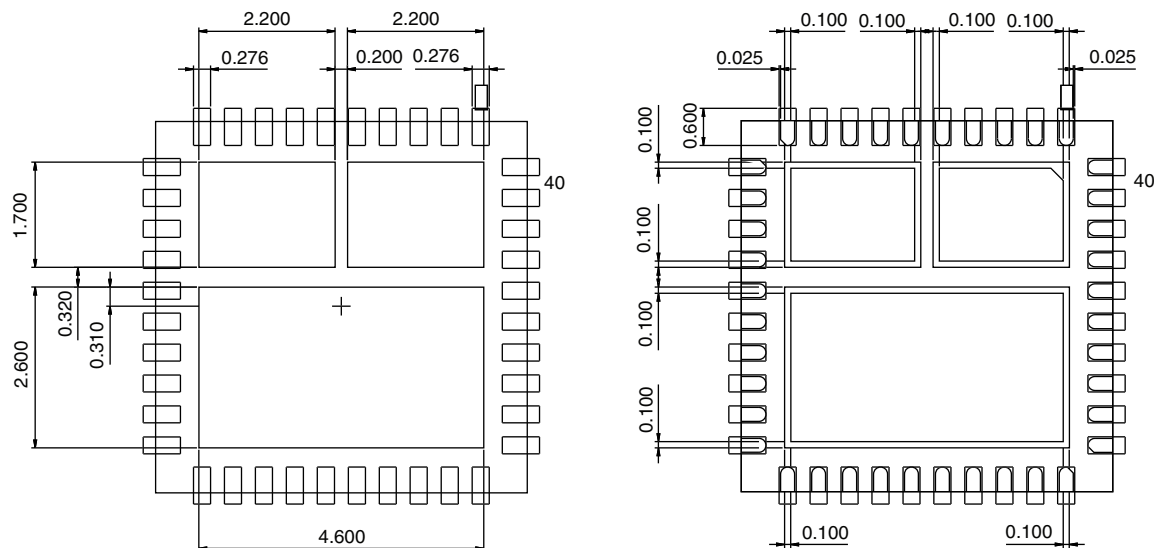
DIM.	MILLIMETERS			INCHES		
	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.
A <sup>(8)</sup>	0.70	0.75	0.80	0.027	0.029	0.031
A1	0.00	-	0.05	0.000	-	0.002
A2	0.20 ref.			0.008 ref.		
b <sup>(4)</sup>	0.20	0.25	0.30	0.078	0.098	0.011
D	6.00 BSC			0.236 BSC		
e	0.50 BSC			0.019 BSC		
E	6.00 BSC			0.236 BSC		
L	0.35	0.40	0.45	0.013	0.015	0.017
N <sup>(3)</sup>	40			40		
Nd <sup>(3)</sup>	10			10		
Ne <sup>(3)</sup>	10			10		
D2-1	1.45	1.50	1.55	0.057	0.059	0.061
D2-2	1.45	1.50	1.55	0.057	0.059	0.061
D2-3	2.35	2.40	2.45	0.095	0.094	0.096
E2-1	4.35	4.40	4.45	0.171	0.173	0.175
E2-2	1.95	2.00	2.05	0.076	0.078	0.080
E2-3	1.95	2.00	2.05	0.076	0.078	0.080
K1	0.73 BSC			0.028 BSC		
K2	0.21 BSC			0.008 BSC		
ECN: T14-0826-Rev. B, 12-Jan-15						
DWG: 5986						

## Notes

1. Use millimeters as the primary measurement
2. Dimensioning and tolerances conform to ASME Y14.5M. - 1994
3. N is the number of terminals. Nd is the number of terminals in X-direction and Ne is the number of terminals in Y-direction
4. Dimension b applies to plated terminal and is measured between 0.20 mm and 0.25 mm from terminal tip
5. The pin #1 identifier must be existed on the top surface of the package by using indentation mark or other feature of package body
6. Exact shape and size of this feature is optional
7. Package warpage max. 0.08 mm
8. Applied only for terminals



## Recommended Land Pattern PowerPAK® MLP66-40L



All Dimensions are in millimeters



## Disclaimer

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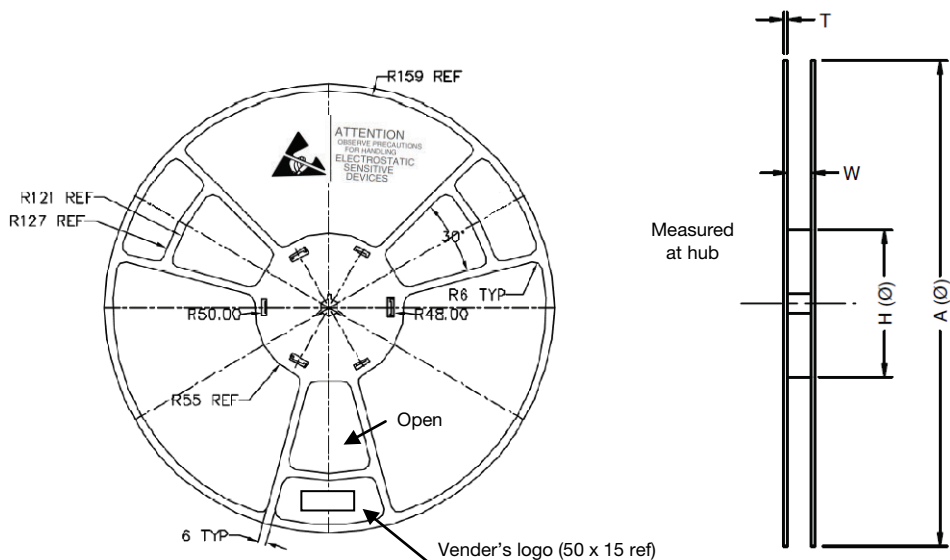
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## Reel

### 330 mm Reel (Lock Reel)

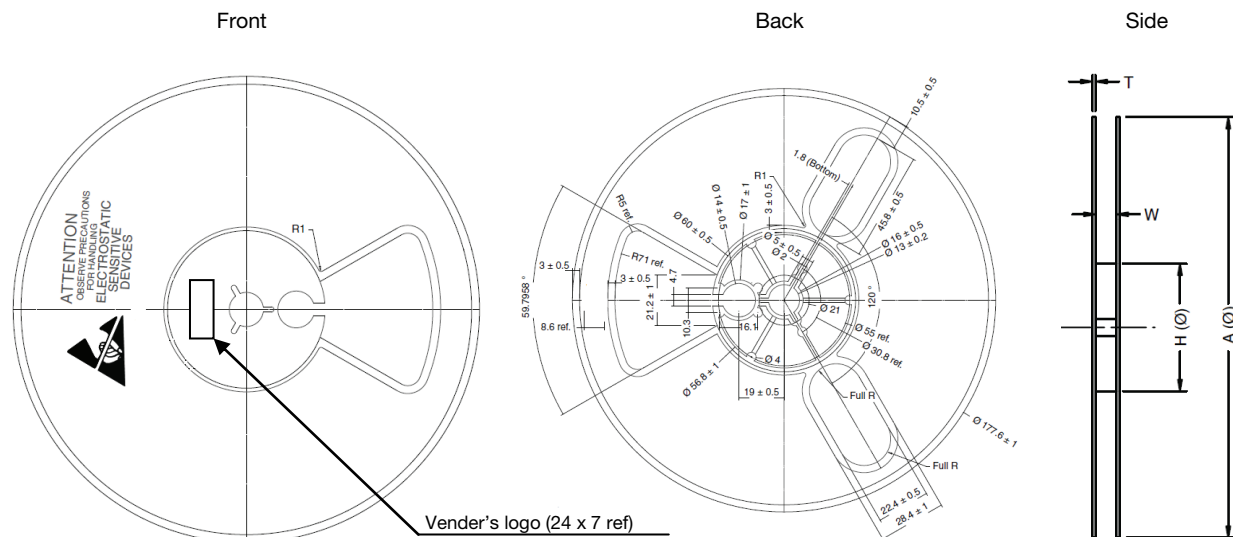


#### Notes

1. Material: antistatic or conductor plastic
2. All dimensions in mm
3. ESD-surface resistivity  $-10^4 \Omega$  to  $10^{11} \Omega$
4. Color: black

VER	APPLICATION		A	W	TAPE WIDTH	H	T
- 1	SOIC-14/16 TO-251 (Short Lead) TO-252/TO-252 (Reverse Lead) PLCC-20 TSSOP-8/14/16/20/28 SSOP-24 SOIC-16 (W)	PowerPAK MLF 9 x 9 PowerPAK MLP 6 x 6 MLF 8 x 8 PowerPAK 8 x 8L PowerPAK 8 x 8 MLP77	$330 \pm 2$	16.4 $\begin{smallmatrix} +2.0 \\ -0 \end{smallmatrix}$	16	$100 \pm 1$	$2.5 \pm 0.5$
- 2	SOIC-8 (N), SOIC-8 (N) epad MSOP-8/10 PowerPAK® SO-8 PowerPAK 1212 PowerPAK 1212-8W MICRO FOOT® MLP33-5, MLP33-8, MLP33-10 QFN (4 x 4)/(3 x 3)/DFN-10 (3 x 3)/ MLP44-16L MLP65-18/20L	PolarPAK® MLP55 PowerPAIR® 6 x 5 PowerPAIR 6 x 3 J PowerPAIR SO-8L PolarPAK1215 PowerPAIR 6 x 3.7 PowerPAK SO-8L MLP4.5 x 3.5-22L PKSO8DCWL	$330 \pm 2$	12.4 $\begin{smallmatrix} +2.0 \\ -0 \end{smallmatrix}$	12	$100 \pm 1$	$2.5 \pm 0.5$
- 4	SOT-23/143 SC70 MICRO FOOT	TSOP-6, 1206-8 ChipFET PowerPAK SC70, PowerPAK SC75	$330 \pm 2$	8.4 $\begin{smallmatrix} +1.5 \\ -0 \end{smallmatrix}$	8.4	$100 \pm 1$	$2.5 \pm 0.5$
- 5	SOIC-20W/24W D²PAK SSOP-28 QSOP-36	PowerPAK MLF 10 x 10	$330 \pm 2$	24.4 $\begin{smallmatrix} +2.0 \\ -0 \end{smallmatrix}$	24	$100 \pm 1$	$2.5 \pm 0.5$
- 8	KGD		$330 \pm 2$	16.4 $\begin{smallmatrix} +2.0 \\ -0 \end{smallmatrix}$	16	$130 \pm 1$	$2.5 \pm 0.5$

### 178 mm Reel (Complete Reel)



## Notes

1. Material: antistatic or conductor plastic
2. All dimensions in mm
3. ESD-surface resistivity  $-10^4 \Omega$  to  $10^{11} \Omega$
4. Color: black

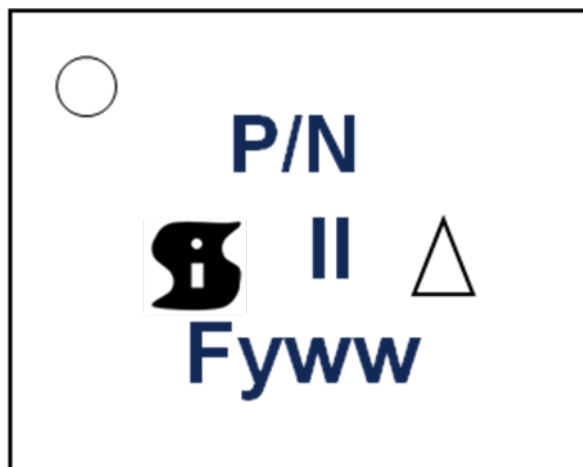
VER	APPLICATION		A	W	TAPE WIDTH	H	T
- 3	SOT-23/143 TSOP-5/6/SC70JW-8L 1206-8 ChipFET® SC70/SC75A/SC89 MICRO FOOT SC-89 (SOT-666) SOT23-5, 6 KGD WCSP PowerPAK 0806 PowerPAK SC70	PowerPAK SC75 MiniQFN PowerPAK MLP22-5 PowerPAK ChipFET PowerPAK SC75-6L (PIC) PowerPAK TSC75-6L (PIC) TDFN4 1.2 x 1.6, TDFN8 2 x 2 Thin PowerPAK SC-70 Thin PowerPAK SC-75 µDFN-6L 1 x 1	178 ± 2	8.4 +1.5 -0	8.4	62 ± 2	1.5 ± 0.5
- 7	MICRO FOOT PowerPAK 2 x 5	KGD	178 ± 2	12.4 +2.0 -0	12	55 ± 2	1.6 ± 0.25

ECN: T17-0308-Rev. BT, 03-Jul-17  
DWG: 93-5211-X



## 1

- **MK - PPAK MLP 7x7, MLF 6x6, MLP 5x5 and MLP45x35**



Format:

LINE 1: P/N

LINE 2: SILICONIX LOGO + LOT CODE + ESD SYMBOL

LINE 3: FACTORY CODE + YEAR CODE + WORK WEEK CODE

## For YEAR CODE:

YEAR	CODE	YEAR	CODE
2010	0	2020	0
2011	1	2021	1
2012	2	2022	2
2013	3	2023	3
2014	4	2024	4
2015	5	2025	5
2016	6	2026	6
2017	7	2027	7
2018	8	2028	8
2019	9	2029	9

## For WORK WEEK CODE:

WEEK	CODE
1 to 6	06
7 to 12	12
13 to 18	18
19 to 24	24
25 to 30	30
31 to 36	36
37 to 42	42
43 to 48	48
49 to 52	52