

# ENGR 065 Electric Circuits

## Lecture 18: The Steady-State Sinusoidal Responses

# Today's Topics

- ▶ Sinusoidal sources
- ▶ The total responses of RC and RL circuits
- ▶ The steady-state sinusoidal response of circuits
- ▶ Covered in Section 13.7

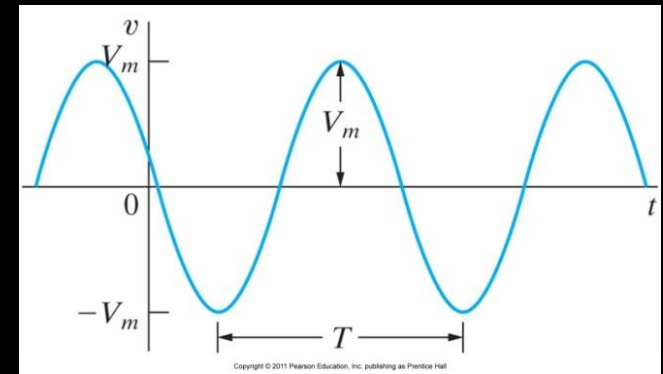
# Sinusoidal Sources

- ▶ A **sinusoidal voltage/current source** (independent or dependent) produces a voltage/current that varies sinusoidally with time.

**Period**,  $T$ , is the length of time required for the sinusoidal function to pass through all its possible values. Its unit is **seconds**.

The reciprocal of  $T$  is **frequency**,  $f$ . It gives the number of cycles per second. It has a unit of Herz, abbreviated **Hz**.

$$f = \frac{1}{T}.$$



$$v = V_m \cos(\omega t + \phi)$$

$\omega$  represents the **angular frequency** of the sinusoidal function, or

$$\omega = 2\pi f = \frac{2\pi}{T} \quad (\text{radians/second})$$

The coefficient  $V_m$  gives the **maximum amplitude** of the sinusoidal voltage.

The angle  $\phi$  is called the **phase angle** of the sinusoidal voltage. It determines the value of the sinusoidal function at  $t = 0$ ;

# The Total Response of RC Circuit

Suppose the voltage source is  $v_i = V_m \cos(\omega t + \phi)$  and the initial voltage  $v_o(t)$  in the circuit is zero. What is the response of  $v_o(t)$ ?

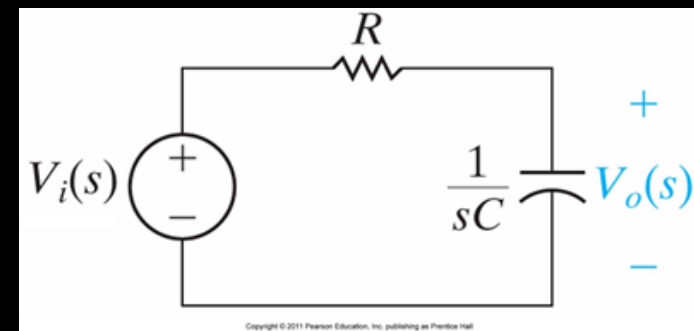
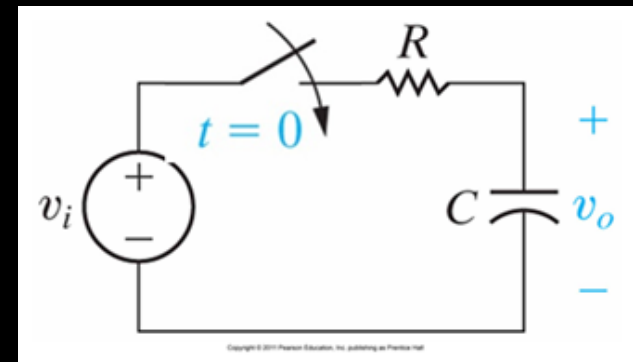
Applying KVL to the circuit on the right:

$$RC \frac{dv_o(t)}{dt} + v_o(t) = V_m \cos(\omega t + \phi)$$

Applying Laplace transform to the both sides of the above equation, we have:

$$RCsV_o(s) + V_o(s) = \frac{V_m(s \cos \phi - \omega \sin \phi)}{s^2 + \omega^2},$$

$$V_o(s) = \frac{\frac{V_m}{RC}(s \cos \phi - \omega \sin \phi)}{(s + \frac{1}{RC})(s^2 + \omega^2)}$$



# The Total Response of RC Circuit

$$V_o(s) = \frac{\frac{V_m}{RC} (s \cos \phi - \omega \sin \phi)}{\left(s + \frac{1}{RC}\right) (s^2 + \omega^2)} = \frac{k_1}{s + \frac{1}{RC}} + \frac{k_2}{s - j\omega} + \frac{k_2^*}{s + j\omega}$$

$$k_1 = V_o(s) \left(s + \frac{1}{RC}\right) \Big|_{s=-\frac{1}{RC}} = \frac{-V_m}{\sqrt{(\omega RC)^2 + 1}} \cos(\phi + \theta)$$

where  $\theta = -\arctan(\omega RC)$ .

$$k_2 = V_o(s)(s - j\omega) \Big|_{s=j\omega} = \frac{V_m}{2\sqrt{(\omega RC)^2 + 1}} e^{j(\phi + \theta)}$$

where  $\theta = -\arctan(\omega RC)$ .

# The Total Response of RC Circuit

The solution for  $v_o(t)$  is

$$v_o(t) = \underbrace{\frac{-V_m}{\sqrt{(\omega RC)^2 + 1}} \cos(\phi + \theta) e^{-\left(\frac{1}{RC}\right)t}}_{\text{Transient component}} + \underbrace{\frac{V_m}{\sqrt{(\omega RC)^2 + 1}} \cos(\omega t + \phi + \theta)}_{\text{steady-state component}}$$

Transient component

steady-state component

where  $\theta = -\arctan(\omega RC)$ .

**The total response = transient response + steady-state response.**

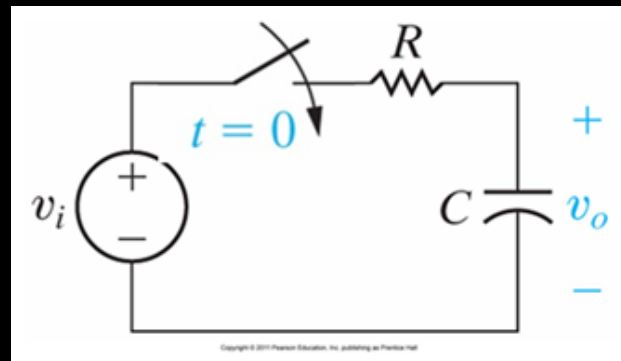
As  $t \rightarrow \infty$ ,  $\frac{-V_m}{\sqrt{(\omega RC)^2 + 1}} \cos(\phi + \theta) e^{-\left(\frac{1}{RC}\right)t} = 0$ , which means the transient response approaches to zero as  $t \rightarrow \infty$ .

# The Steady-State Sinusoidal Response

We denote the second part as  $v_{oss}$ . It is

$$v_{oss} = \frac{V_m}{\sqrt{(\omega RC)^2 + 1}} \cos(\omega t + \phi + \theta)$$

where  $\theta = -\arctan(\omega RC)$ . It is the voltage across the capacitor as  $t \rightarrow \infty$ .



## Four notes:

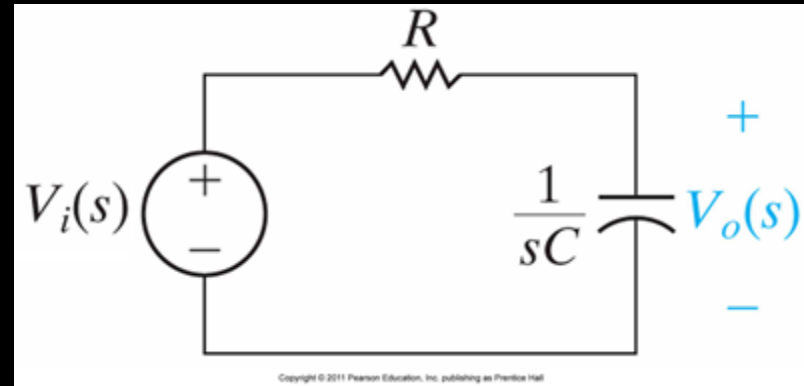
1. The steady-state response/output (the voltage across the capacitor) has the same frequency as the source/input.
2. The magnitude of the steady-state response is reduced by the factor of  $\sqrt{(\omega RC)^2 + 1}$ .
3. The phase angle of the steady-state response is lagged by a degree of  $\arctan(\omega RC)$ .
4. The magnitude and the phase angle of the response vary with frequency.

# The Steady-State Sinusoidal Response

Let's take a look at the transfer function of the circuit.

Here, the input is  $V_i(s)$ , and the output is  $V_o(s)$ , and the transfer function is:

$$H(s) = \frac{V_o(s)}{V_i(s)} = \frac{1/RC}{s + 1/RC}$$



Replacing  $s$  with  $j\omega$  ( $s = j\omega$ ), we have

$$H(j\omega) = \frac{1/RC}{j\omega + 1/RC} = \frac{1}{\sqrt{(\omega RC)^2 + 1}} e^{j[-\arctan(\omega RC)]} = |H(j\omega)| e^{j\angle H(j\omega)},$$

$$\text{where } |H(j\omega)| = \frac{1}{\sqrt{(\omega RC)^2 + 1}}, \text{ and } \angle H(j\omega) = -\arctan(\omega RC)$$



# The Steady-State Sinusoidal Response

Comparing the steady-state output  $v_{oss} = \frac{V_m}{\sqrt{(\omega RC)^2 + 1}} \cos(\omega t + \phi + \theta)$  to

$$|H(j\omega)| = \frac{1}{\sqrt{(\omega RC)^2 + 1}}, \text{ and } \angle H(j\omega) = -\arctan(\omega RC)$$

we know that the steady-state output is equal to

$$\begin{aligned} v_{oss} &= V_m \frac{1}{\sqrt{(\omega RC)^2 + 1}} \cos(\omega t + \phi + \theta) \\ &= V_m |H(j\omega)| \cos(\omega t + \phi + \angle H(j\omega)) \end{aligned}$$

# The Total Response of RL Circuit

Suppose the voltage source is  $v_s = V_m \cos(\omega t + \phi)$  and the initial current in the circuit is zero. What is the total response of  $i(t)$ ?

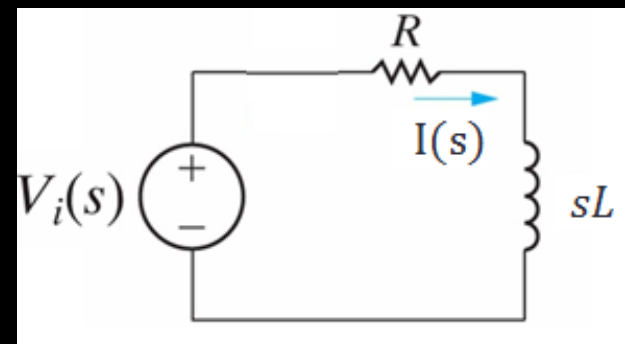
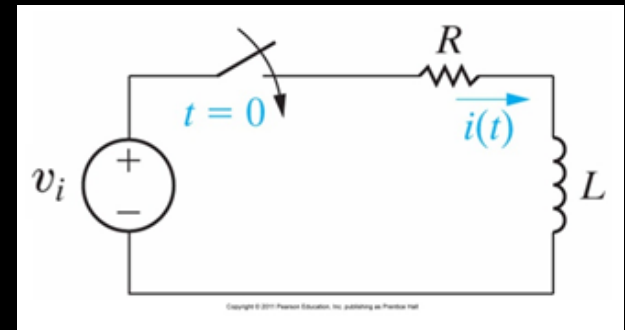
Applying KVL:

$$L \frac{di}{dt} + Ri = V_m \cos(\omega t + \phi)$$

Applying Laplace transform to the both sides of  
The above formula, we have:

$$LsI(s) + RI(s) = \frac{V_m(s \cos \phi - \omega \sin \phi)}{s^2 + \omega^2},$$

$$I(s) = \frac{\frac{V_m}{L}(s \cos \phi - \omega \sin \phi)}{(s + \frac{R}{L})(s^2 + \omega^2)}$$



# The Total Response of RL Circuit

By using the partial fraction expansion, we can find the solution for  $i(t)$  is

$$i(t) = \underbrace{\frac{-V_m}{\sqrt{R^2 + (\omega L)^2}} \cos(\phi + \theta) e^{-\left(\frac{R}{L}\right)t}}_{\text{Transient component}} + \underbrace{\frac{V_m}{\sqrt{R^2 + (\omega L)^2}} \cos(\omega t + \phi + \theta)}_{\text{steady-state component}}$$

Transient component

steady-state component

where  $\theta = -\tan^{-1}\left(\frac{\omega L}{R}\right)$

**The total response = transient response + steady-state response.**

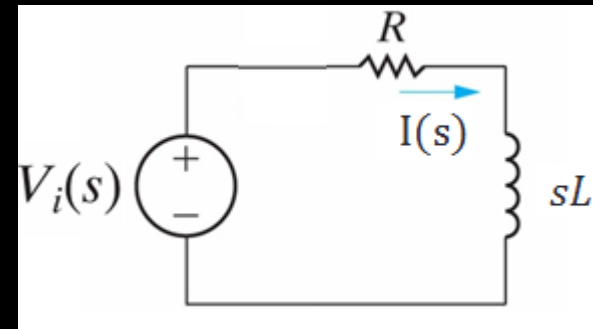
The transient response is approaching to zero as  $t \rightarrow \infty$ , which is

$$\lim_{t \rightarrow \infty} \frac{-V_m}{\sqrt{R^2 + (\omega L)^2}} \cos(\phi + \theta) e^{-\left(\frac{R}{L}\right)t} = 0.$$

# The Steady-State Sinusoidal Response

The transfer function of the circuit is

$$H(s) = \frac{1/L}{s + R/L},$$



$$H(j\omega) = \frac{1/L}{j\omega + R/L} = \frac{1}{j\omega L + R} = \frac{1}{\sqrt{R^2 + (\omega L)^2}} \angle -\tan^{-1}\left(\frac{\omega L}{R}\right),$$

$$|H(j\omega)| = \frac{1}{\sqrt{R^2 + (\omega L)^2}}, \quad \angle H(j\omega) = -\tan^{-1}\left(\frac{\omega L}{R}\right)$$

From the second term of  $i(t)$ , we know that

$$i_{ss}(t) = V_m |H(j\omega)| \cos(\omega t + \phi + \angle H(j\omega))$$

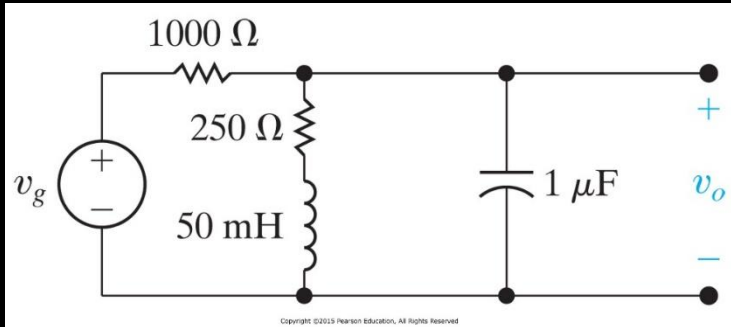
# The Steady-State Sinusoidal Response

We can extend the results from the RC and RL circuits to any circuits and conclude that, in general, if the input of a circuit is a sinusoidal source of  $x(t) = A \cos(\omega t + \phi)$ , the steady-state response (output) of the circuit is:

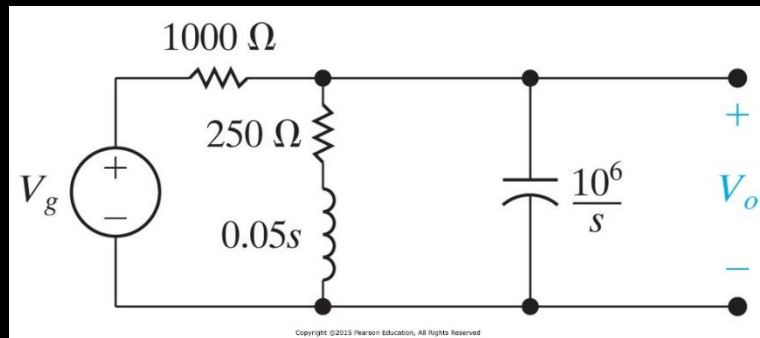
$$y_{ss}(t) = A|H(j\omega)|\cos [\omega t + \phi + \theta(\omega)]$$

where  $H(j\omega)$  is the transfer function,  $H(s) = \frac{Y(s)}{X(s)}$ , of the circuit when  $s = j\omega$ .  $H(j\omega) = |H(j\omega)|e^{j\theta(\omega)}$ , where  $|H(j\omega)|$  and  $\theta(\omega) = \angle H(j\omega)$  is the magnitude and phase angle of  $H(j\omega)$ , respectively.

# Example (p. 501)



No initial stored energy. The sinusoidal voltage source is  $120 \cos(5000t + 30^\circ) \text{ V}$ . Find the steady-state expression for  $v_o(t)$ .



Step 1: No initial stored energy

Step 2:  $v_g = 120 \cos(5000t + 30^\circ) \text{ V}$ .

Step 3. Draw the circuit model in the s domain

Step 4 Use the node-voltage method to find  $V_o(s)$

$$\frac{V_o - V_g}{1000} + \frac{V_o}{250 + 0.05s} + \frac{V_o s}{10^6} = 0$$

Solving for  $V_o$  yields

# Example (p. 501) – cont'd

$$V_o(s) = \frac{1000(s + 5000)V_g(s)}{s^2 + 6000s + 25 \times 10^6}$$

Step 5: The transfer function is

$$H(s) = \frac{V_o(s)}{V_g(s)} = \frac{1000(s + 5000)}{s^2 + 6000s + 25 \times 10^6}$$

Step 6: From  $v_g = 120 \cos(5000t + 30^\circ)$  V, we know

$$V_m = 120 \text{ V}, \omega = 5000 \text{ rad/s}, \varphi = 30^\circ$$

Step 7: Find  $H(j\omega)$

$$H(j5000) = \frac{1000(j5000 + 5000)}{(j5000)^2 + 6000(j5000) + 25 \times 10^6} = \frac{\sqrt{2}}{6} e^{-j45^\circ}$$

$$|H(j\omega)| = \frac{\sqrt{2}}{6}$$

$$\theta = \angle H(j\omega) = -45^\circ$$

# Example (p. 501) – cont'd

Step 8: Find the steady-state expression for  $v_o(t)$ .

$$v_{0ss}(t) = V_m |H(j\omega)| \cos(\omega t + \phi + \angle H(j\omega))$$

$$v_{0ss}(t) = 120 \times \frac{\sqrt{2}}{6} \cos(5000t + 30^\circ - 45^\circ)$$

$$= 20\sqrt{2} \cos(5000t - 15^\circ) \text{ V}$$



# Summary

The transfer function of a circuit can be used to find the circuit's steady-state response to a sinusoidal source.

If the source of the circuit is  $x(t) = A \cos(\omega t + \phi)$ , the steady-state response to the source is  $y_{ss}(t)$ , the transfer function is

$$H(s) = \frac{Y_{ss}(s)}{X(s)}, \text{ then}$$

$$y_{ss}(t) = A|H(j\omega)|\cos[\omega t + \phi + \theta(\omega)]$$