CDMA2000 Simulation Project

Nikita Teplitskiy

February 2020

1 Introduction

The submitted project is a simulation of the bit error rate resulting from passing binary phase shift keyed data over a time varying Rayleigh fading channel when using space-time block codes. The implementation was taken from techniques mentioned by Rappaport in his wireless communications textbook as well as from Alamouti's paper on space-time block codes. The main script to run the simulation is titled BER.Sim and executes in only a few seconds for $N=2^{20}$ samples. Space time encoding and decoding are brought out into their own functions and appropriately labeled files. An additional file called ST.Test was constructed to verify basic functionality of space-time codes and is of little interest.

2 Rayleigh Fading Channel Implementation

The Rayleigh channel was implemented using a technique described by Rappaport in his textbook "Wireless Communications Principles and Practice." This technique implements Clarke's model based on Smith's simple implementation of a Rayleigh fading channel. For a time domain Rayleigh signal of length N, this is accomplished by generating two sets of complex Gaussian random variable for N/2 positive bin frequencies ranging from some minimal frequency to the maximum Doppler frequency. The complex conjugate of these values are used to populate the corresponding negative frequency bins. A shaping spectrum specified by the formula

$$S_{E_z} = \frac{1.5}{\pi f_m \sqrt{1 - \frac{f - f_c}{f_m}^2}}$$

is applied to each of the two Gaussian noise spectra after which the IFFT is taken of the resulting shaped noise. It is important to note that S_{E_z} approaches values of positive infinity when approaching

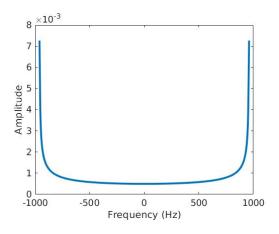


Figure 1: Example Shaping Spectrum S_{E_z}

 $\pm f_m$ which makes taking an IFFT impossible. This is dealt with by a approximating the extreme values using predicted values base on the difference between the n-1th and n-2th values. Since the probability of an incoming signal having a Doppler frequency offset of exactly $\pm f_m$ is zero, it is acceptable to approximate the extreme behavior instead. One of the resulting signals undergoes a phase shift of -90° which is done by multiplying it by -i. Both signals are then squared and summed together, after which the square root is taken. The resulting signal is a model of a time varying Rayleigh fading channel.

3 Space-Time Block Code Implementation

Space-time block codes using two antennae are generated from a time domain signal by splitting the signal into its odd and even elements. Let these elements be s_0 and s_1 representing even and odd respectively. Antenna 0 then alternates between transmitting a symbol from s_o and $-s_1^*$. Antenna 1 transmits symbols from s_1 then s_0^* .

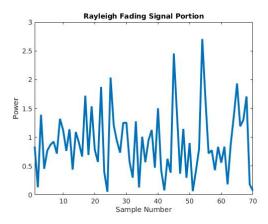


Figure 2: Example Rayleigh Fading Signal

The decoding scheme for the code depends on the number of receive antenna present. For a single antenna, the incoming data is again split into even and odd components. Let these be r_0 and r_1 . The received symbols are then determined by $\tilde{s_0} = h_0^* r_0 + h_1 r_1^*$ and $\tilde{s_1} = h_1^* r_0 - h_0 r_1^*$ where h represents the known Rayleigh fading channel. The Euclidean distance between each of these symbols and the known BPSK symbols is computed and the closest symbol is selected. After decision making the odd and even components of the signal are combined. Alternatively, the step of computing Euclidean distance can be eliminated by simply looking at the real component of the raw received symbol and checking if it is positive or negative.

To decode space-time codes when two receive antennae are used, the technique used for the simple single antenna case is extended to include data from both antennae. As previously, the received data is split into odd and even sets, however this now results in 4 sets as there are 2 data sources. Let these be r_0 , r_1 , r_2 , and r_3 . The received symbols are then determined by $\tilde{s_0} = h_0^* r_0 + h_1 r_1^* + h_2^* r_2 + h_3 r_3^*$ and $\tilde{s_0} = h_1^* r_0 + h_0 r_1^* + h_3^* r_2 + h_2 r_3^*$. The same Euclidean distance decoding scheme as well as data recombination method is used as in the previous case.

4 Sources

S. M. Alamouti, "A simple transmit diversity technique for wireless communications," in IEEE Journal on Selected Areas in Communications, vol. 16, no. 8, pp. 1451-1458, Oct. 1998.

T. S. Rappaport, Wireless Communications Princi-

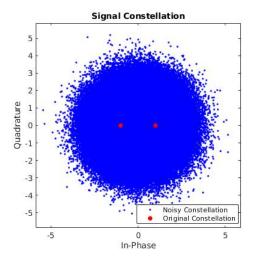


Figure 3: Very Noisy BPSK Signal

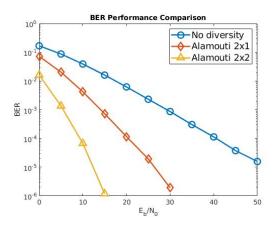


Figure 4: Simulated BER Results

ples and Practice, Prentice Hall, 2002.

J. I. Smith, "A Computer Generated Multipath Fading Simulation for Mobile Radio", IEEE Trans. Vehicular Technology, vol. 24, no. 3, pp. 39-40, Aug. 1975.