

Fiber Optics Communication

Page No. 27
 Ch. 22

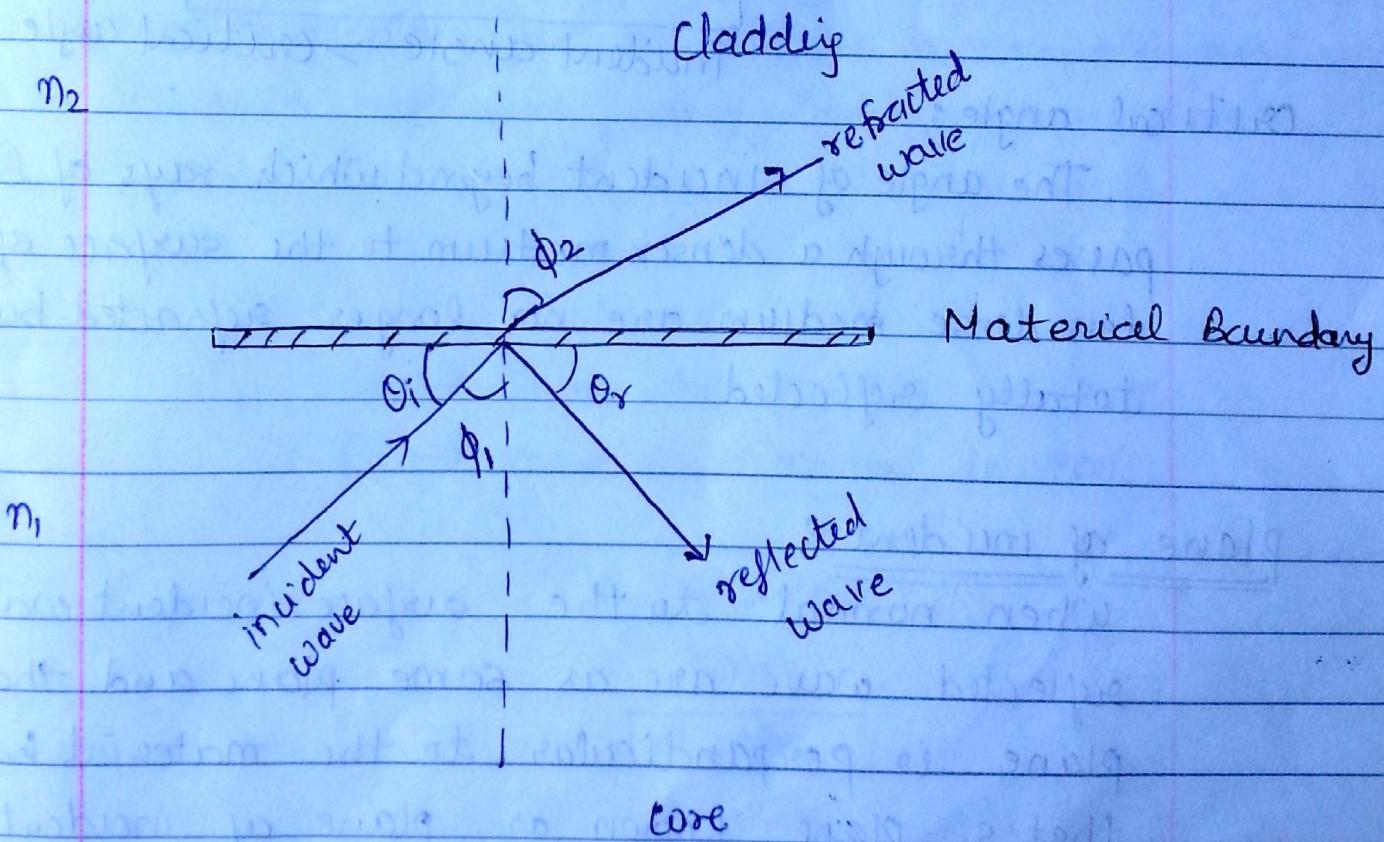
Optics theory:

- Reflection / Refraction
- Refraction Index

$$n = \frac{c}{v} \rightarrow \text{free space velocity of light}$$

$v \rightarrow \text{velocity of light in medium}$

Ray model:



- $n_2 < n_1$
- $q_2 > q_1$

ϕ_1 → angle of incident
 ϕ_2 → angle of refracted

Snell's law:

$$n_1 \sin \phi_1 = n_2 \sin \phi_2$$

- When light passes through denser to lighter medium velocity will increases ($n_2 < n_1$)

law of reflection:

$$\angle \phi_i = \angle \phi_r$$

when $\angle \phi_i >$ critical angle

incident angle → critical angle

critical angle:

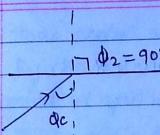
The angle of incident beyond which rays of light passes through a denser medium to the surface of a less dense medium are no longer refracted but totally reflected.

plane of incident:

When normal to the surface incident wave and reflected wave are in same plane and that plane is perpendicular to the material boundary that's plane known as plane of incident.

critical angle of incident:

When $\phi_2 = 90^\circ$, then incident angle is ϕ_c .



- When $\phi_1 > \phi_c \rightarrow$ then reflection occurs (TIR)
in eqn (A) $n_1 \sin \phi_c = n_2 \sin(90^\circ)$
 $\phi_c = \sin^{-1} \left(\frac{n_2}{n_1} \right)$

only reflection occurs when $n_2 < n_1$

- Internal reflection ($n_2 < n_1$ denser → less denser)
- External reflection ($n_2 > n_1$ less denser → denser)
 - Due to reflection energy and phase loops occur.

phase shift:

$s_p \rightarrow$ phase shift parallel to POI

$s_n \rightarrow$ phase shift normal to POI

$$\tan \frac{s_p}{2} = \frac{n \sqrt{n^2 \cos^2 \phi_i - 1}}{\sin \phi_i}$$

$$\phi_i = \phi_2 - \phi_c \quad \tan \frac{s_n}{2} = \frac{\sqrt{n^2 \cos^2 \phi_i - 1}}{n \sin \phi_i}$$

- phase shift is totally dependent on incident angle.

Characterization of light source

- wavelength of light (transmitting the information perpendicularly)
- intensity of light
- spectral width of source (optical source will be able to transmit the range of wavelength $\frac{1}{\text{wavelength}}$)

$$\text{wavelength} = 660 \text{ nm}$$

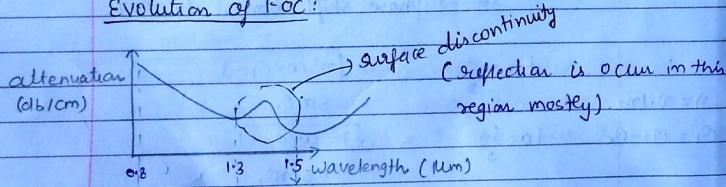
$$400 \text{ nm} / 700 \text{ nm} \rightarrow \text{spectral width}$$

$\frac{1}{300 \text{ nm}}$

Understanding the Behaviour of light

- Ray Model (Reflection, Refraction, TIR, Snell, PCT, refractive index)
 - Wave Model
 - Quantum Model
- 400 to 700 nm \rightarrow Visible spectrum

Evolution of FOc:



Second genr (no. of repeaters are less, repeater generates the information again after certain interval, bcz of reflection, scattered information may loss so we use seq. Repeaters)

third generation (1.5 μm)

- By changing the signal wavelength or dimension and object dimensions are fixed, then reflection may differ or changeable. A large amount of reflection occurs when wavelength of signal is less as compare to object dimension.

elements of an optical fiber transmission link

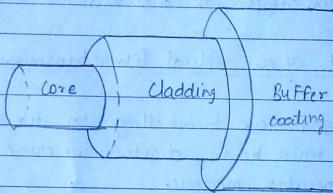
(From slide)

- all signals are transmitting in fiber transmission are must be in digital. we done this by using regenerator module.

Structure of optical fiber:

$$n_2 \text{ (cladding)} < n_1 \text{ (core)}$$

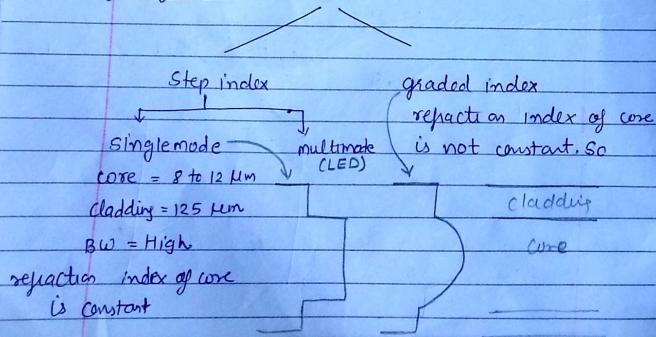
- Working principle : TIR



Why cladding is necessary :

- To minimize the scattering loss occurred due to unequal surface of core and gives the mechanical strength also cladding is used.

Types of Fiber



- multimode fiber requires more size of cladding

Ex8- Design consideration for step index fiber

$$n_2 = n_1(1-\Delta)$$

$$\Delta = \text{core cladding index diff} = 0.01$$

multimode fibers : $\Delta = 1\% \text{ to } 3\%$,

singlemode fibers : $\Delta = 0.2\% \text{ to } 1\%$.

what happens when Δ kept large ?

- for multimode we required core size is more to cladding bcoz in multimode more no. of waves will propagate.

$$\Delta = \frac{\text{change in refractive index}}{\text{refractive index of core}}$$

$$= \frac{n_1 - n_2}{n_1}$$

(n ₁) Core Glass(1.5)	(n ₂) Cladding Air (1)	$\Delta (\text{in } \%)$ 33.34%	Phase shift (N)	Parallel	Single mode	Multimode
2.69	1.049	44.60%	1.23	4.46		
1.60	1.49	8.835%			3.894	
1.55	1.48					
1.4776	1.461					

$$\Delta = \frac{n_1 - n_2}{n_1}$$

$$\theta_c = \sin^{-1}(n_2/n_1)$$

$$\tan \frac{\theta_c}{2} = \sqrt{n_2^2 \cos^2 \alpha - 1}$$

$$n_2 \sin \theta_c = n_1 \cos \alpha$$

$$m = n_1/n_2$$

$$\theta_i = \theta_c - \theta_c$$

- When ϕ_c is higher than ϕ_i (Phase shift is less)
- when core-cladding difference is large reflection is less
- core-cladding difference is more, reflection is ~~more~~^{less}
- critical angle is less \rightarrow angle of incident is more
- $\phi_c = \sin^{-1}(n_2/n_1) \quad n_1 > n_2$

when Δ is large,

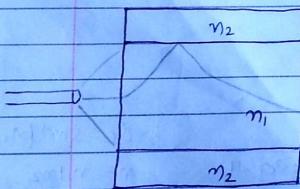
Critical angle is less and reflection is more.

from slide:

Intrinsic absorption: absorption of photons inside fiber material

Extrinsic absorption: caused by dissolved water in the glass or hydroxyl (OH^-) ions

Launching light into fibers



Ray A $\rightarrow \phi$

Ray B $\rightarrow \alpha > \phi$

Ray C $\rightarrow \alpha_i > \alpha$

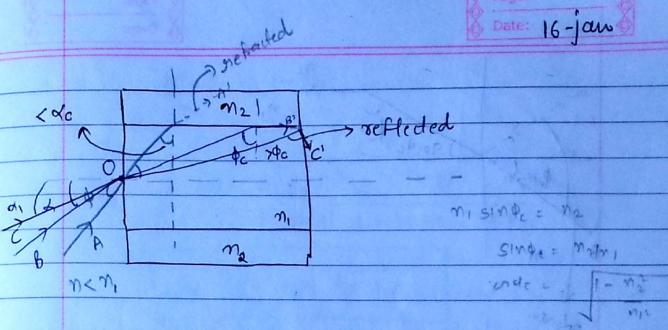
$$\text{By applying Snell's law for ray B: } n \sin(\alpha_{\max}) = n_1 \sin(90 - \phi_c) = n_1 \cos \phi_c$$

$$n \sin(\alpha_{\max}) = \frac{n_1 \sqrt{n_1^2 - n_2^2}}{n_1}$$

$$n \sin(\alpha_{\max}) = \sqrt{n_1^2 - n_2^2}$$

$$NA = \sqrt{n_1^2 - n_2^2}$$

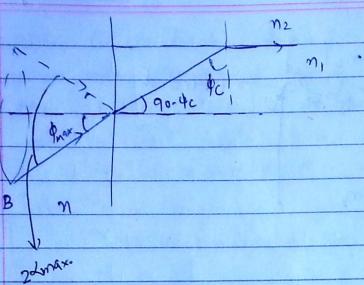
If light is ^{incident} propagated in the range of $\alpha < \alpha_{\max}$, then light can propagate in fiber.



$$\sin(\alpha_{\max}) = f(n_1, n_2, n)$$

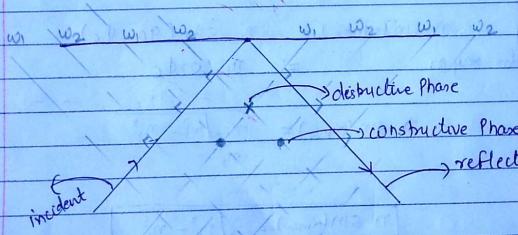
max. acceptance angle

$$\text{and } \alpha_{\max} = \sqrt{1 - \frac{n_2^2}{n_1^2}}$$



- If wave is not propageted in fiber while ^{incident} angle is 52° it is due to only phase shift.

*



\Rightarrow

$$NA = f(\Delta) = \sqrt{n_1^2 - n_2^2}$$

$$\therefore n_2 = n_1(1-\Delta)$$

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$$\frac{n_2}{n_1} = (1-\Delta)$$

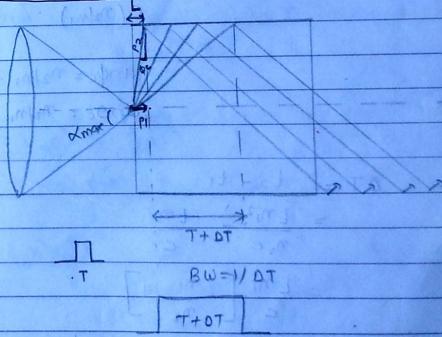
$$\therefore NA = \sqrt{n_1^2 - n_2^2(1-\Delta)^2}$$

$$= n_1 \sqrt{1 - (1+\Delta^2 - 2\Delta)}$$

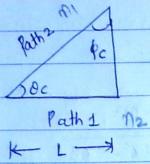
$$NA \approx n_1 \sqrt{2\Delta}$$

Significance :- We can launched the light, at max. angle of acceptance.

Numerical Aperture



If multiple rays is transmitted at θ_{max} then multiple reflections occur. pulse is spread due to reflection.



$$\cos \theta_c = \frac{L}{x}$$

path 1: Time taken to travel Path 1

$$t_1 = \frac{L}{c/n_1} = \frac{L n_1}{c}$$

Path 2: Time taken to travel Path 2

$$t_2 = \frac{L/\cos \theta_c}{c/n_2} = \frac{L n_2}{c \cos \theta_c}$$

$$= \frac{L n_2}{c(n_2/n_1)} = \frac{L n_2^2}{n_1 c}$$

$$\because \sin \theta_c = n_2/n_1 \Rightarrow \sin(90 - \alpha) = ?$$

$$\cos \theta_c = n_1/n_2$$

$$\Delta T = t_2 - t_1$$

$$= \frac{L n_2^2}{n_1 c} - \frac{L n_1}{c}$$

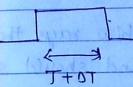
$$= \frac{L n_1}{c} \left[\frac{n_1}{n_2} - 1 \right]$$

$$\boxed{\Delta T = \frac{L n_1}{c} \left(\frac{n_1 - n_2}{n_2} \right)}$$

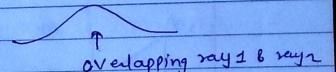
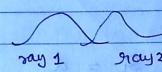
$$BW = \frac{1}{\Delta T}$$

- Single mode having the higher BW, as compare to

multimode & Single mode \rightarrow NA is High
 $\Delta T \rightarrow$ max. allowable spreading

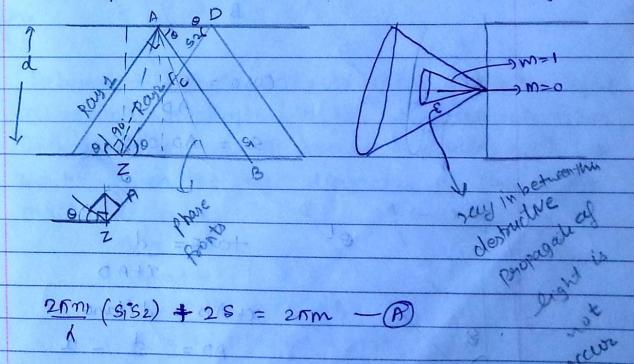


\rightarrow if the spreading is more than this then overlapping b/w rays would occur.



- for max. spreading $\rightarrow (n_1 - n_2)$ should be small.

\Rightarrow Wave representation in a Dielectric slab waveguide



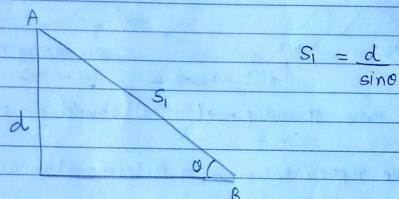
$$2\pi n_1 (S_1 S_2) + 2S = 2\pi m \quad \text{--- (A)}$$

- as we know that if angle of incidence between 2 modes then ray must be propagated, but it's not

always occur due to destructive phase.

- The rays can collide constructive or destructive when sent at an angle $\leq \text{NA}$.

To get constructive collision b/w rays the angle should be multiple of π and should satisfy the condition given (Eqn A)



$$\cos \theta = \frac{s_2}{AD}$$

$$s_2 = AD \cos \theta$$

$$\tan \theta = \frac{d}{x + AD}$$

$$x + AD = d / \tan \theta$$

$$AD = \frac{d}{\tan \theta} - x$$

$$AD = \frac{d}{\tan \theta} - d \tan \theta'$$

$$\frac{2\pi n_1}{\lambda} (s_1 s_2) + \frac{2s}{\lambda} = 2\pi m$$

$$\frac{2\pi n_1}{\lambda} \left[\frac{d - AD \cos \theta}{\sin \theta} \right] + \frac{2s}{\lambda} = 2\pi m$$

$$\frac{2\pi n_1}{\lambda} \left[\frac{d}{\sin \theta} - \underbrace{\left[\frac{d - AD \cos \theta}{\sin \theta} \cos \theta \right]}_{2d \sin \theta} \right] + 2s = 2\pi m$$

$$\frac{2\pi n_1}{\lambda} d \sin \theta + s = \pi m \quad m = 0, 1, 2, \dots$$

It is only valid for electric field, SN

This equation would satisfy when constructive collision is occurring and wave is propagated in the fiber optic cable.

Q10 - fan

Ex: Determine the set of angle (θ) for which the light at 1550nm wavelength will be propagated into fiber with refractive index of core and cladding with 1.5 and 1.4. Given the core diameter of 100 μm for multimode fiber?

d10.8

$$n_1 = 1.5$$

$$n_2 = 1.4$$

$$m = n_1/n_2 = 1.5/1.4 = 1.07$$

$$\theta_c = \sin^{-1}(n_1/n_2) = 68.96^\circ$$

$$\theta_i = \pi/2 - \theta_c = 21.03^\circ$$

Normal to the plane of incident

$$\tan \delta_{\text{av}} = \frac{\sqrt{n^2 \cos^2 \theta_i - 1}}{n \sin \theta_i}$$

rays is reflected back, so we have taken the same phase shift at SN, at θ angle.

$$SN = \frac{2\pi n_1}{\lambda} d \sin\theta + \delta = \frac{2\pi m}{\lambda}$$

$$\Rightarrow \frac{2\pi n_1}{\lambda} d \sin\theta + \delta = \frac{2\pi m}{\lambda}$$

$$m=0$$

$$\frac{2\pi n_1}{\lambda} d \sin\theta + \frac{2\pi \tan^{-1}(\sqrt{n^2 \cos^2\theta - 1})}{n \sin\theta} = 0$$

$$\frac{2\pi(1.5) \times 100 \times 10^{-9}}{1550 \times 10^{-9}} d \sin\theta + \frac{2\pi \tan^{-1}(\sqrt{n^2 \cos^2\theta - 1})}{n \sin\theta} = 0$$

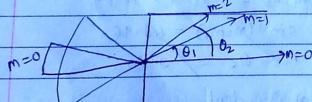
$$\frac{6.08 \times 10^5}{2} d \sin\theta = -\tan^{-1}\left(\frac{\sqrt{n^2 \cos^2\theta - 1}}{n \sin\theta}\right)$$

$$\tan^{-1}(304.02 \sin\theta) = -\frac{\sqrt{n^2 \cos^2\theta - 1}}{n \sin\theta}$$

$$n \sin\theta \tan^{-1}(304.02 \sin\theta) = -\sqrt{n^2 \cos^2\theta - 1}$$

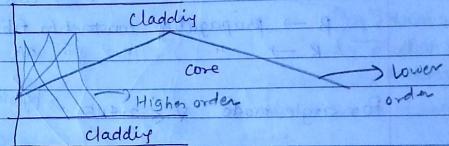
$$n^2 \sin^2\theta \tan^{-2}(304.02 \sin\theta) = n^2 \cos^2\theta - 1$$

$$\theta \approx 15^\circ$$

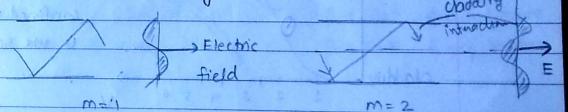


- $m=0 \rightarrow$ single mode

- Higher order & lower order

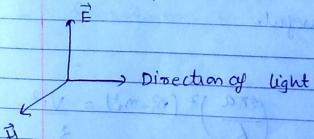


$M \rightarrow$ indicates no. of zeros in field pattern



Wave Model:

$$\psi(x, t) = A e^{j(wt - \beta x)}$$



V number (normalised frequency)

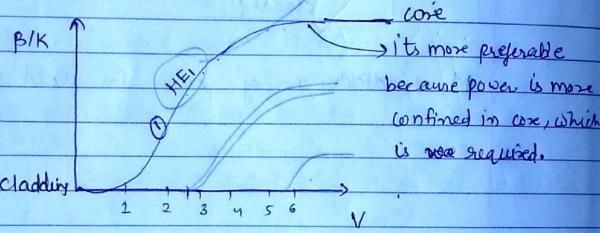
- How many modes a fiber can support.

$$V = \frac{2\pi a}{\lambda} (\eta_1^2 - \eta_2^2)^{1/2}$$

$a \rightarrow$ core diameter radius

How much power $\beta \rightarrow$ propagation constant in fiber
link with $K \rightarrow$ " " " at
the cladding

For single mode $V < 2.405 \rightarrow$ hybrid transition mode



- less amount of electric field we require to link with cladding.
- Higher amount of electric field link with core \rightarrow its requirement is useful.

No. of modes (m):

$$M \approx \frac{1}{2} \left(\frac{2\pi a}{\lambda} \right)^2 (\eta_1^2 - \eta_2^2) = \frac{V_1^2}{2}$$

Power coupled in cladding (P_{clad}):

$$P_{clad} = \frac{4}{\pi M} P \quad P \rightarrow \text{total optical power}$$

$$\boxed{\text{for a particular mode } \frac{P_{clad}}{V} \propto \frac{1}{V} \quad (\text{from } \textcircled{1})}$$

Higher mode power confined in cladding is more which is normally not preferable.

Radiation modes

- modes which are not trapped in core
- radiation trapped in cladding
- modes in cladding
- power launched outside λ_{max} refracted out from core

leaky modes

- power leaks into cladding due to tunneling effect.
- condition for leaky/guided mode
- $\beta > \eta_2 k$ guided mode
- $\beta < \eta_2 k$ leaky mode

cut off wavelength

- minimum wavelength for single mode fiber
- Rearranging the eqn of V no. we can get

$$\boxed{\lambda_c = \frac{2\pi a \eta_1}{V_c} (2.405)^{1/2}}$$

Example

$$\begin{aligned} \eta_1 &= 1.51 \text{ nm} = 50 \text{ nm} \\ \eta_2 &= 1.36 \text{ nm} \\ \lambda &= 1300 \text{ nm} \\ V_c &= 26.6 \\ \lambda_c &= ? \end{aligned}$$

$$NA = \frac{V\lambda}{2\pi a}$$

$$= \frac{26.6 \times 1300 \times 10^{-9}}{2 \times 3.14 \times 25 \times 10^{-6} \times 2}$$

$$= \frac{0.220}{2} \quad \text{Ans.}$$

$$= 0.110 \quad \text{Ans}$$

$$NA = \frac{V\lambda}{2\pi a}$$

$$= \frac{26.6 \times 1300 \times 10^9}{2 \times 3.14 \times 25 \times 10^{-6} \times L}$$

$$= 0.11$$

Example: find cut off wavelength, $n_1 = 1.46$

Step Index & graded index fiber structure

- Step index

- refractive index of core is constant.

$$\downarrow_{2a} - n_2 = n_1(1-\Delta)$$

- Graded index

- refractive index of core is not constant.

$$\begin{cases} n_1 [1 - 2\Delta (\frac{r}{a})^\alpha]^{1/2} & ; 0 \leq r \leq a \\ n_1 (1 - 2\Delta)^{1/2} & ; r \geq a \end{cases}$$

$$\text{where } \Delta = \frac{n_1 - n_2}{n_1}$$

- multimode fibers are designed with the graded index fiber structure because in graded index dispersion is less compared to single mode.

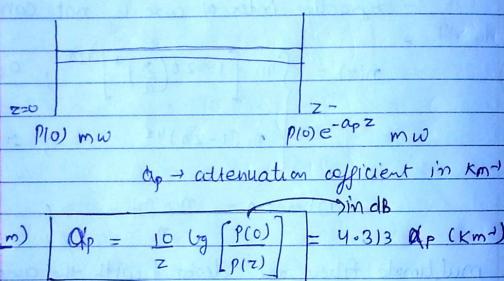
Manufacturing of fiber optic cable

(Self study)

Signal degradation in optical fiber

Attenuation & degradation:

power loss across a fiber

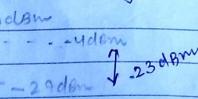


$$* \quad \text{dBm} = 10 \log \left(\frac{P}{1 \text{ mW}} \right)$$

if $P = 10 \text{ mW}$

$$\begin{aligned} \text{in dBm} &= 10 \\ \text{in dB} &= -20 \quad (-20+20) \end{aligned}$$

for ex: emitter power = -4 dBm
Rx sensitivity = -27 dBm



Case I	Tx	7.3	4.8	6.9	Rx = <u>✓</u>
				-4	$-19 \text{ dBm} = -23 \text{ dBm}$
Case II	Tx	8.5	+ 7.2	+ 8.6	Rx = <u>X</u>
				24.3 dBm	$-4 - 24.3 = -28.3 \text{ dBm}$

in Case I, at Rx side power is detected.

in Case II, at Rx side signal is not detected because power $> -27 \text{ dB}$

$$\bullet \quad P(z) [\text{dBm}] = P(0) [\text{dBm}] - \alpha (\text{dB/km}) z (\text{km})$$

\downarrow

power at a distance

Ex 8- Consider a 30 km long optical fiber that has an attenuation of 0.8 dB/km . find the optical O/P Power Out (in absolute and dBm) if 200Hw of optical power launched into the fiber.

$$P_0 = 200 \text{ mW}$$

$$\text{dBm} \rightarrow -6.98 \text{ dBm}$$

$$\text{dB} \rightarrow -3.698 \text{ dB}$$

$$\alpha = 0.8 \text{ dB/km}$$

$$z = 30 \text{ km}$$

$$P(z) = -6.98 - 0.8 \times 30 = -30.98 \text{ dBm} \quad ①$$

$$\alpha = \frac{10}{z} \log \left(\frac{P(0)}{P(z)} \right) \Rightarrow P(z) = \frac{P(0)}{e^{\alpha z}}$$

$$= \frac{200}{e^{0.8 \times 30}} \mu\text{W}$$

$$= 30.99 \text{ dBm} \quad ②$$

Ex8 It is observed that for establishing a particular fiber link of 2 Km, the optical signal power would decrease by 75% of launched optical fiber. Determine the attenuation coefficient for given data. Comment how can we establish the fiber link of 20 Km by maintaining the same IIP & OIP power relationship as mentioned.

Sol8

$$z = 2 \text{ Km}$$

$$\alpha = \frac{10}{z} \log \left(\frac{P(0)}{P(z)} \right) = \frac{10}{2} \log \left(\frac{100}{25} \right) = \frac{6}{2} = 3.0 \text{ dB/Km}$$

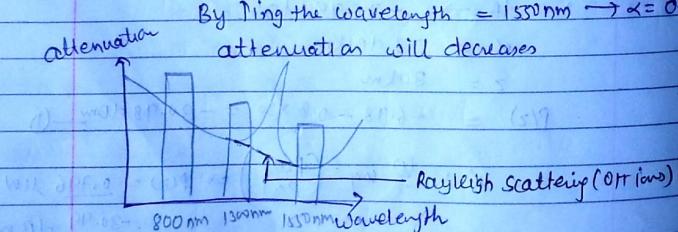
for, when $z = 20 \text{ Km}$, then $\alpha = 30 \text{ dB/Km}$

from the graph of wavelength v/s attenuation

at wavelength $= 800 \text{ nm} \rightarrow \alpha = 3.0 \text{ dB/Km}$



By Tting the wavelength $= 1550 \text{ nm} \rightarrow \alpha = 0.3 \text{ dB/Km}$
attenuation will decreases



* 1st window : 850 nm, attenuation 2 dB/km
2nd window ; 1300 nm, attenuation 0.5 dB/km
3rd window ; 1550 nm, attenuation 0.3 dB/km

Example :

When the optical power launched into an 8 Km length of fiber is 120 mW. the optical power at the OIP is 3 mW. Determine:

i) overall signal attenuation in decibels through the po fiber assuming there are no connectors.

$$(i) \alpha = \frac{10}{z} \log \left[\frac{P(0)}{P(z)} \right]$$

$$= \frac{10}{8} \log \left[\frac{120}{3} \right]$$

$$= 2 \text{ dB/km}$$

Overall attenuation per km.

(ii)

$$\text{attenuation} = 20.68 \text{ dB}$$

$$(10 \log 120) \text{ in dB} - (10 \log 3) \text{ in dB}$$

$$= 16.02 \text{ dB}$$

(iii)

the overall signal attenuation for a 10km optical link using the same fiber with splices at 1 Km intervals, each giving an attenuation of 1 dB.

2 dB/km means - for 1 km \rightarrow 2 dB attenuation
 10 km \rightarrow 20 dB attenuation
 bcoz of splicers added attenuation is there which we have to add.

$$\text{overall attenuation} = 20 + 9 = 29 \text{ dB}$$

for link $\square \square \square \square \rightarrow$

\uparrow with the total 9 splicers (fiber conn) along the splicers link each with an attenuation of 1 dB.

(v) the OIP & IIP power ratio in (c).

$$\alpha_{dB} = 10 \log \left[\frac{P(0)}{P(2)} \right]$$

$$10^{\alpha/10} = P_{in}/P_{out}$$

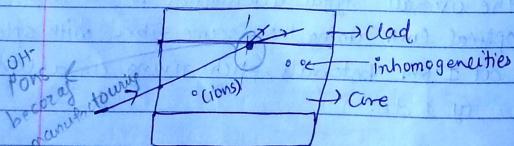
$$P_{out}/P_{in} = 10^{2.5} \times 10^{-3}$$

attenuation depends on: λ wavelength, surface discontinuity, absorption

(2) Scattering loss :

$$\text{rayleigh scattering } R_s S_0 \propto \frac{1}{\lambda^4}$$

radiation in cladding bcoz of scattering is known as $R_s S_0$

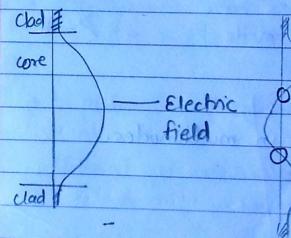


Bending :

Macro

radius of core $<$ radius of bend curvature (bending radius)

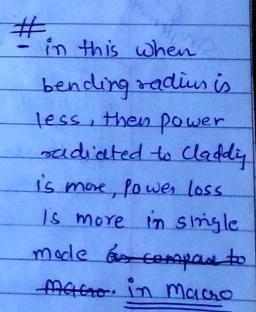
power coupling to higher order modes that are more lossy.



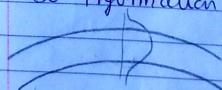
Micro

radius of core $>$ bending radius

power coupling to lower order modes that are more lossy.

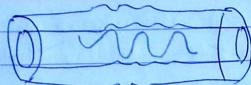


- In macro bending by reducing the bending radius, power associated with clad is more so information loss is more.

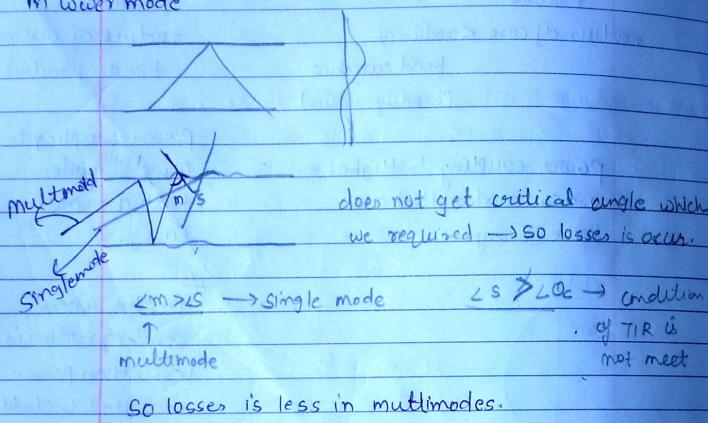


Scanned by CamScanner

micro bend loss



in lower mode



\Rightarrow

Attenuation
scattering (80% loss)

Bend loss (20% loss)

$\lambda \uparrow$

\downarrow

\times

$\lambda \downarrow$

\times

\checkmark

we are taking the higher wavelength to reduce the losses in fiber.

$$P_{\text{clad}} \propto L \quad \frac{P}{R} \rightarrow \text{bendy loss}$$

and is P_{clad} , Power associated with clad is more. So power loss is more.

Dispersion

- intramodal dispersion (chromatic Dispersion)

- Single mode fiber

- intermodal dispersion (Modal Dispersion)

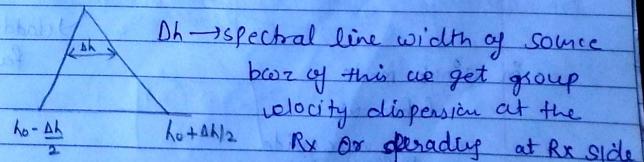
- multimode fiber

$$\Delta T = \frac{L}{C} \frac{n_1 (n_1 - n_2)}{n_2}$$

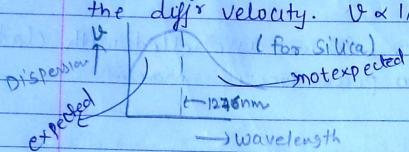
- pulse is spreaded in dispersion

$$V \propto \frac{1}{\lambda} \quad (\text{Wavelength & velocity are inversely proportional})$$

chromatic dispersion in single mode: as the source like LED
it will not only transmit the single wavelength but it will transmit multiple wavelengths of signals so spreading will occur.



- group delay is occur due to all wavelengths having the diff. velocity. $v \propto 1/\lambda$

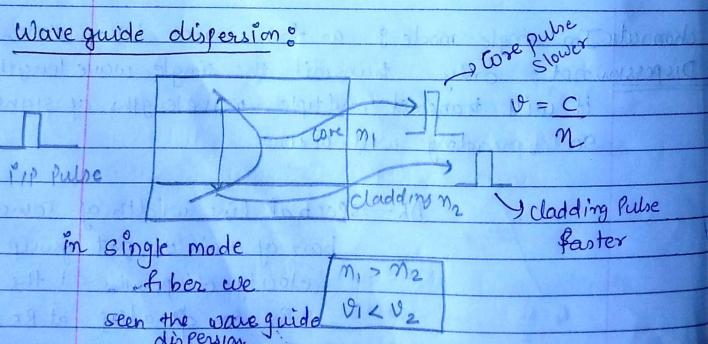


In a particular material as silica, the wavelength 1276nm, as λ is less than λ_0 then v is less as $v \propto \lambda$. Frequency is less * lower freq. travel faster through core-cladding loss: dispersive medium.

$$\text{attenuation coefficient} = \frac{\alpha_{\text{clad}}}{P_{\text{clad}}} + \frac{\alpha_{\text{core}}}{P_{\text{core}}} \rightarrow 20\% + 80\%$$

α_{av} → total attenuation coeff. in fiber.

Wave guide dispersion:



in single mode

$$\begin{cases} n_1 > n_2 \\ v_1 < v_2 \end{cases}$$

Due to pulse travelling in core & cladding core diff so pulse spreading will occur → Due to that dispersion waveguide Dispersion is there.

wavelength dispersion depends on the mode field distribution in the core & Cladding wavelength

pulse spreading limitation towards bit rate

In multimode, Pintersymbol of information signal will occur. Due to that information may loss. In this now we can see that how much wavelength spreading will allowed into fiber so that loss not occurs or signal detect easily.

$$\Delta T = \frac{L}{c} \frac{n_1(n_1 - n_2)}{n_2}$$

$$\text{10 bits} \rightarrow 10 \times 100ns = 1/100ns$$

$$\Delta T \rightarrow$$

$$\Delta T < T$$

$$\Delta T < 1$$

(bit rate) B

$B \cdot \Delta T < 1$ → for not getting the intersymbol interference

$$\text{Bit rate (B)} = \frac{BW \log_2(1+SNR)}{\approx BW}$$

$$B \cdot \Delta T = \frac{BW \cdot L}{c} \frac{n_1(n_1 - n_2)}{n_2} < 1$$

Unit of channel cap is bits/sec & Bit rate is also bits/s

$$BW \cdot L < \frac{C n_2}{n_1(n_1 - n_2)}$$

Bandwidth ^{distance} product is defined in eqn ①

- 100 MHz for 100m
- 10 MHz · km
- 10 Mbps & length is 1 km
- 1 KHz for 10m
- If BW is increased by 10 then Length will be decreased by 10.
- In single mode fiber, BW is high because $n_1 - n_2$ change in refractive index is very less. bcoz NA is the range of angle, in which we can launched the light in fiber.
- for single mode, we required less incident angle bcoz By which light can propagate in the axis of core (10°)

Example L = 1 Km

$$NA = 0.295$$

$$n_{core} = 1.407$$

Calculate how much will a light pulse spread after travelling along 1 km.

$\approx 1\%$

$$(NA)^2 = n_1^2 - n_2^2$$

$$n_2^2 = (1.407)^2 - (0.295)^2$$

$$n_2 = 1.46$$

Page: 31 Date: 31-Jan

Important Keywords 8

① plane wave

$$\psi(x, t) = A e^{j(kx - \omega t)}$$

$k \rightarrow$ propagation constant = $n(\lambda) \cdot 2\pi / \lambda$

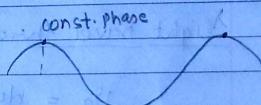
$\omega = 2\pi c / \lambda$

② phase velocity

$$Kx - \omega t = \varphi$$

$$v_p = \frac{dx}{dt} = \frac{\omega}{k}$$

phase velocity



phase wave & phase velocity are associated

③ Wave Packets

$$\psi(x, t) = \sum_{n=1}^N A_n e^{j(k_n x - \omega_n t)}$$

④ Group velocity

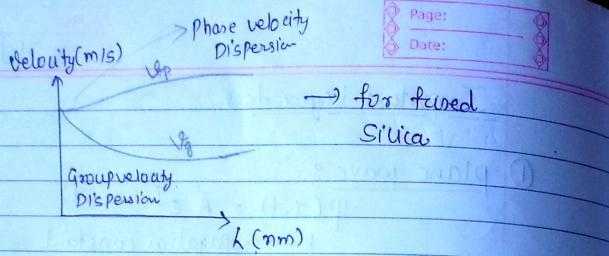
$$v_g = \frac{d\omega}{dx}$$

$$v_p \& v_g$$

$$v_p = c/n \rightarrow \text{single } \lambda$$

$$v_g = c/n \rightarrow \text{due to thin dispersion}$$

dispersion effect will occur at Rx



Any information signal is a wave packet, then it travels in group velocity, not in phase velocity.

Light travel through dispersive medium

$$V_g = \frac{dw}{dk}$$

$$k = n(\lambda) \cdot \frac{2\pi}{\lambda} \quad \text{and} \quad \omega = \frac{2\pi c}{\lambda}$$

$$\frac{dk}{dh} = -\frac{2\pi}{\lambda^2} n(\lambda) + \frac{2\pi}{\lambda} \frac{dn}{dh} \quad \text{and} \quad \frac{dw}{dh} = -\frac{2\pi c}{\lambda^2} \frac{dn}{dh}$$

$$\frac{dw}{dk} = \frac{c}{\frac{dn(\lambda)}{dh} - \frac{n(\lambda)}{\lambda}}$$

$$V_g = \frac{c}{n(\lambda) - \lambda \frac{dn(\lambda)}{dh}}$$

$$n_g = n(\lambda) - \lambda \frac{dn(\lambda)}{dh}$$

$$\text{so; } V_g = c/n_g$$

$$V_p = c/n(\lambda)$$

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for wave packet, refractive index is uses due to group velocity.

GVD (Group velocity dispersion) (for single mode)

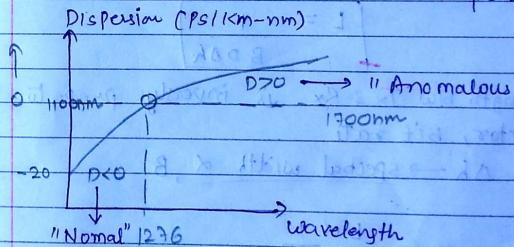
$$T = L/V_g$$

if $\Delta h \rightarrow$ spectral width of an optical fiber

pulse broadening due to a differential time delay

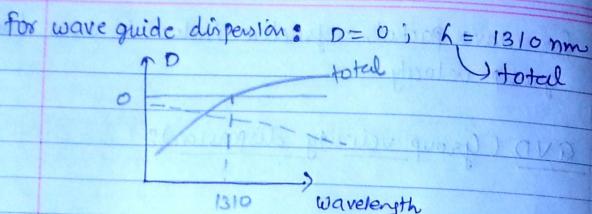
$$\Delta T = L D \Delta h$$

$$D = \frac{d(1/V_g)}{dh} \rightarrow \text{dispersion parameter}$$



$\left\{ \begin{array}{l} \text{for silica} \rightarrow \text{at wavelength} = 1276 \rightarrow D=0 \\ D<0 \rightarrow \text{pulse compressed} \\ D>0 \rightarrow \text{pulse broadening} \end{array} \right.$

- Upper dispersion is material dispersion bcoz dispersion is varies as wavelength is varies



$$\text{total chromatic Dispersion} = D_{\text{mod}} + D_{\text{wg}}$$

Material Wave

$$\Delta T = L D \Delta \lambda$$

$$\frac{1}{(\text{bit rate})B} = L D \Delta \lambda$$

$$L = \frac{1}{B D \Delta \lambda}$$

length bw Tx & Rx is inversely prop. to Disp factor, bit rate

L, $\Delta \lambda \rightarrow$ spectral width $\propto B$

$$\Delta \lambda \propto \frac{1}{B^2}$$

We can say that By $\frac{\text{Disp}}{\text{bit rate}}$ the effect of chromatic dispersion is decreases by 16.

so if we multiply insertion loss with respect to dispersion in dB then it will be

$$\Delta T = \frac{L}{c} \frac{n_1(n_1 - n_2)}{n_2} = \frac{9.16 \times 10^{-3} \times 10^3}{10^8} = 91.6 \text{ ns}$$

for a link: $10 \text{ Mbps} \Rightarrow T = 100 \text{ ns}$

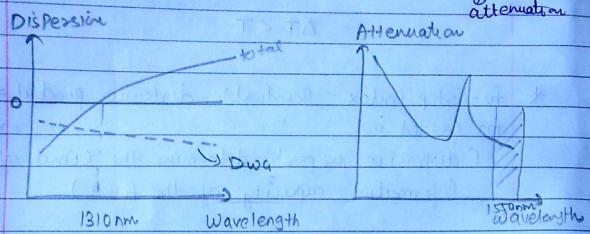
$$\Delta T < T$$

* for step index Bandwidth-distance product = $20 \text{ MHz} \cdot \text{km}$
 for graded " " " " " = 2.5 GHz km
 (dispersive properties determine the limit of the information capacity of the fibres.)

- measure of information capacity of an optical waveguide is usually specified by the BW-Distance product in $\text{MHz} \cdot \text{km}$.

What is our expectation?

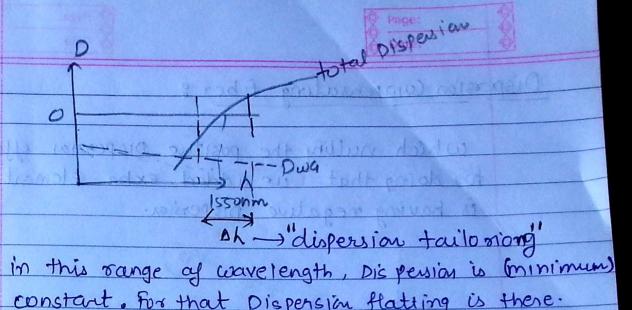
As we know that $\lambda = 1310 \text{ nm} \rightarrow D = 0$
 $\lambda = 1550 \text{ nm} \rightarrow \alpha = 0.3 \text{ dB/km}$



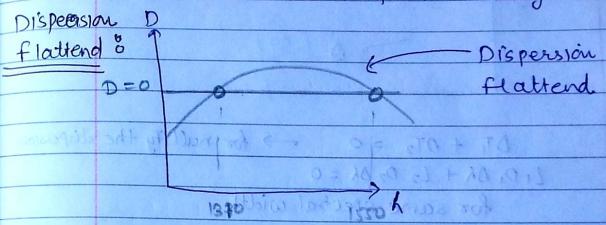
So we have to minimize the Dispersion & attenuation @ 1550 nm, so we have to shift the graph by dispersion at 1550 nm wavelength.

- ↳ material dispersion (refractive index & wavelength)
- ↳ waveguide dispersion (change in refractive index and radius of core)

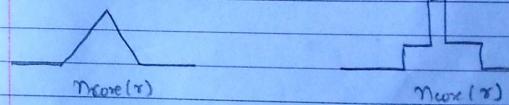
By changing the waveguide dispersion, we can shift the $D = 0$, @ $\lambda = 1550 \text{ nm}$. And this will done by the change the pattern of core refractive index.



in this range of wavelength, Dispersion is minimum constant. For that Dispersion flattening is there.



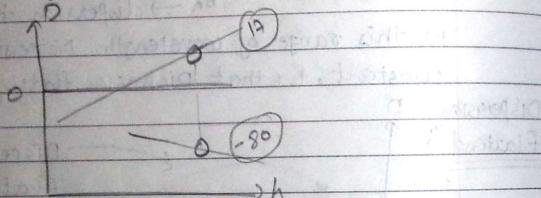
for dispersion flattening, we changing the pattern of core refractive index.



- Dispersion flattening in which n_{core} is changing this process is known as Dispersion Tailoring.

Dispersion compensating fiber:

which nullify the positive dispersion effect.
for doing that we added extra element, which
is having negative dispersion.



$$DT_1 + DT_2 = 0 \rightarrow \text{for nullify the dispersion}$$

$$L_1 D_1 \Delta\lambda + L_2 D_2 \Delta\lambda = 0$$

for same spectral width

$$L_1 D_1 = -L_2 D_2$$

↑ ↓ more

Distance more Distance less

Polarization:

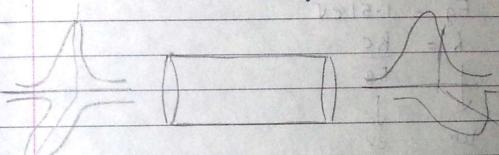
- Polarization varies along the fiber length.
- pulse broadening due to the orthogonal polarization modes

$$n = \frac{c}{v}$$

$$B = |n_x - n_y|$$

refractive index
in x-dir

B → birefringence birefringence index



With respect to central axis of fiber, varil dev in refractive index → causes velocity diff → causes polarization dispersion.

n in x & y dir should small value as $\approx 10^{-6}$ to 10^{-5}

LED:

- photons \rightarrow recombination of e⁻ & holes

$$E_g = \frac{hc}{\lambda}$$

L) $E_g = 1.421 + 1.266x + 0.266x^2$
(Ga_{1-x}Al_xAs)

When $x = 0.07$

$$E_g = 1.51 \text{ eV}$$

$$\lambda = \frac{hc}{E_g}$$

↓
λ (μm)
↓
λ (nm)

$$\lambda (\mu\text{m}) = \frac{1.240}{E_g (\text{eV})} = 0.82 \text{ μm}$$

↳ transmitting wave length

- if we need low loss window then $\lambda = 1550 \text{ nm}$

$$\begin{array}{c} \downarrow \\ E_g (\text{eV}) \\ \downarrow \\ \lambda \rightarrow (\text{GaAlAs}) \end{array}$$

Desirable char of optical sources:

- response time min (when we apply electric field, then OIP will comes as soon as possible)
- quantum efficiency
- minimum spectral width

↳ min. time for convert the electric field to optical

Radiative & non-radiative:

PN Diode \rightarrow 1 junction

LED \rightarrow multiple junctions (Heterojunction)

Active region \rightarrow recombination of e⁻ & holes
↓
Photons

Recombination:

Radiative:

e⁻ & holes are recombine and photons will emitted.

(Light)

Quantum efficiency = $\frac{\text{radiation recom.}}{\text{total recombination}}$

Non radiative:

e⁻ & holes are recombine but light don't emit but in the form of heat.

(Heat)

internal \rightarrow light is within the material / not yet emitted.

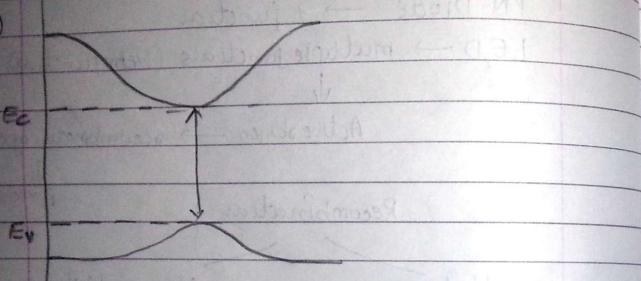
external \rightarrow $= \frac{\text{light emitted out}}{\text{total light generated}}$

- Diff' b/w LED & PN junction

- Based on material
- junctions

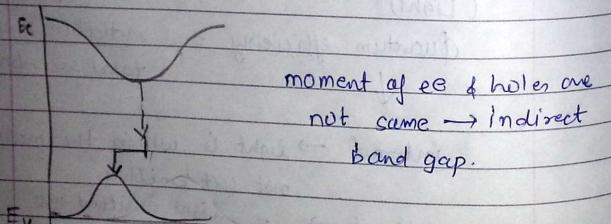
Direct Band gap & Indirect Band gaps

①



moment of e & holes are same, and lowest of conduction band, max. of valence
→ Direct Band gap
which is used in light source.

②



Charge confinement:

is dealing with the energy gap b/w charges.
we arrange the energy gap b/w junctions in
a such a way that the max. recombination occurs
in Active regions

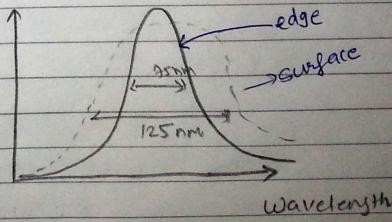
optical confinements:

in the form of refractive index.
we kept the center at the highest refractive index
due to that in that region max. recombination
of e & holes will occur → due to that
light (photons) emitted in that region is
high.

- We required that because we only realized the minimum spectral width, and concentrating light in a particular dis^c → so that we can use multiple junctions → and by which providing diff. band gap & diff. n, we get the max. light in Active Region.

LED spectral patterns:

- Surface emitting LED (195 nm)
- Edge-emitting LED (75 nm)



- spectral width is less in edge-emitting LED
so we are using this.

Internal quantum efficiency:
fraction of e^- & holes
radiatively:

$$\eta_{int} = \frac{R_r}{R_r + R_{nr}}$$

$$R_r \rightarrow \text{radiative recomb.} = \frac{n}{\tau_r} \rightarrow e^- \text{ density } (\text{cm}^{-3})$$

$$R_{nr} \rightarrow \frac{n}{\tau_{nr}}$$

$$\eta_{int.} = \frac{1}{1 + \tau_r / \tau_{nr}} \rightarrow \text{non radiative recomb.}$$

\downarrow
radiative
recomb. life time

- In order to generate recomb., we have supplied some reversed bias current.

R → rate at which total recomb. take place

$$R_r + R_{nr} = \frac{I}{q}$$

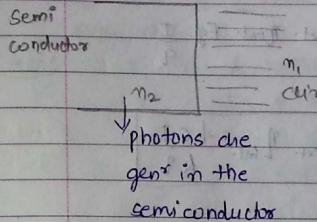
$$\begin{cases} I = q/t \\ I/q = 1/t = \text{rate} \end{cases}$$

$$\eta_{int.} = \frac{R_r}{R_r + R_{nr}} = \frac{R_r \cdot \alpha}{I}$$

$$R_r = \frac{I}{q} \eta_{int.}$$

If $\eta_{int.} \rightarrow 0.6$ means 60% recombination is done which are radiative and 40% \rightarrow non radiative

External quantum efficiency:



- all photons which are gen' in the semiconductor material, are not completely emitted outside (or in the air). its based on the refractive index, which amount of photons are emitted.

$$\eta_{ext} = \frac{\text{no. of photons emitted}}{\text{no. of photons generated}}$$

let RI of semiconductor material = n

$$\eta_{ext.} = \frac{1}{n(n+1)^2}$$

then and then only when emitted in the air $n=1$.

$$\text{power emitted by the LED} \quad P_{optical} = \eta_{ext.} P_{int.}$$

$$P_{int} = \frac{E_g}{time}$$

Power confined wth
the semiconductor
= $E_g \cdot R_s$
= $E_g \cdot \frac{I}{q} \cdot \eta_{int}$

$$P_{optical} = \eta_{ext} \cdot \eta_{int} \cdot E_g \cdot \frac{I}{q}$$

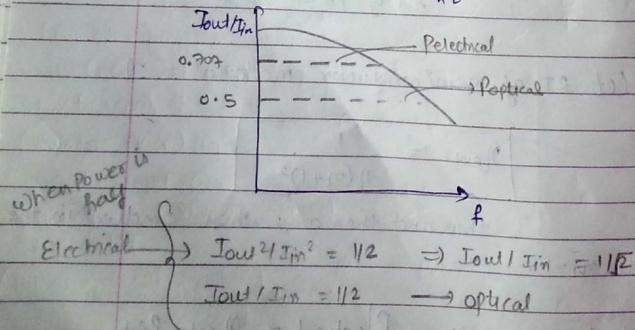
$$P_{optical} = \eta_{ext} \cdot \eta_{int} \cdot \frac{hcI}{\lambda^2}$$

Power origin
the circuit

Optical BW / Electrical BW

$$\text{Electrical Power: } P_{elec} = I^2 / R \rightarrow P \propto I^2$$

$$\text{Optical power: } P_{optical} = \eta_{ext} \cdot \eta_{int} \cdot \frac{hcI}{\lambda^2} \rightarrow P \propto I$$



Ex8- A Double heterojunction InGaAsP LED emits at peak wavelength 1310nm has radiance and non-radiance time of 30ns & 100ns respectively. The drive current is 40mA & R_E is 3.5. find int quantum eff., total optical Power gen & power emitted?

Sols

$$\lambda = 1310\text{nm}$$

$$t_{R} = 30\text{ns}$$

$$t_{NR} = 100\text{ns}$$

$$I = 2V \cdot 40\text{mA}$$

$$\eta_{int} = \frac{1}{1 + t_R/t_{NR}} = \frac{1}{1 + 0.3/1.0} = \frac{1}{1.3} = 0.76$$

$$\eta_{ext} = \frac{1}{n(1+M)^2} = \frac{1}{3.5(1+3.5)^2} = 0.014$$

$$P_{int} = \eta_{int} \cdot \frac{hcI}{\lambda^2}$$

$$= 0.76 \times \frac{1.84 \times 40 \times 10^{-23}}{1.6 \times 10^{-19} \times 1310 \times 10^{-6}}$$

$$= 0.76 \times 6.67 \times 10^{-34} \times 3 \times 10^8 \times 40 \times 10^{-9}$$

$$= 9.29 \times 10^{-1}$$

$$= 9.29 \text{ mw}$$

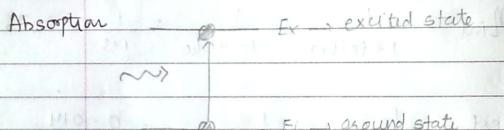
$$P_{ext} = P_{int} \cdot \eta_{ext}$$

$$= 4.088 \text{ mw}$$

LASER: Light amplification through stimulated emission of radiation

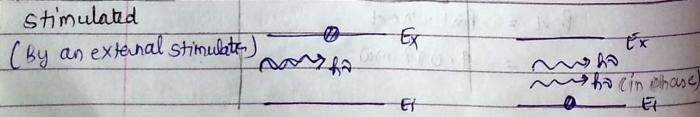
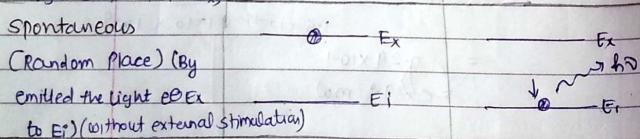
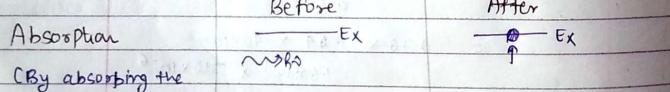
- spectral linewidth
 - BW
- response time < 1ms
- operating principle - Stimulated emission
 - spontaneous emission
 - absorption emission

Plank's law



By absorption of energy ϵ_0 travel from E_i to $E_x \rightarrow$

Absorption

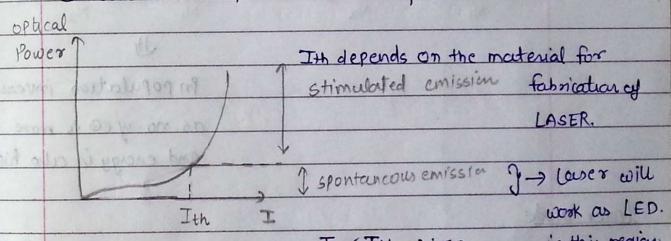


Condition for stimulated emission

Population of excited e^- > population of ground state e^-

To achieve this, population inversion is done where e^- at ground state are provided energy so that they move to $E_2 \Rightarrow$ stimulated emission takes place. (By injecting the e^- in mat. at the device contact to fill the lower energy states of the conduction band)

- necessary amount of e^- present at E_2 with sufficient amount of energy so that they are stable at E_2 . - more amount of e^- in excited state \Rightarrow more no. of photons emitted. Population $\rightarrow e^-$ present at E_2



Types of laser diode:

(1) Fabry Perot cavity

$$\Delta h = \Delta \lambda = \frac{c}{\Delta f}$$

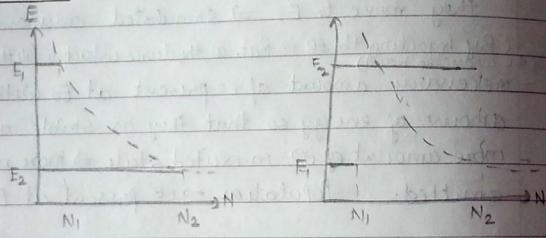
$\Delta \lambda$ = spectral width

In stimulated emission:

- identical energy \rightarrow identical wavelength \rightarrow narrow linewidth

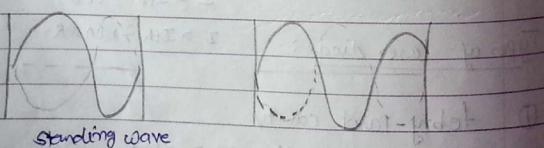
- identical direction \rightarrow narrow beam width
- identical phase \rightarrow coherence & identical polarization

population inversions:



In population inversion
as no. of e⁺ is more
and energy is also high

Laser Diode



Resonator
wave length

non-resonator
wave length

Fabry-Pérot cavity resonator only passes the standing wave.

1) Distributed feedback laser

Gratings \rightarrow present so that we can have
Bragg reflector
Very less linewidth \Rightarrow only required wavelength is transmitted.

power launching and coupling:

→ C depends on NA, size of core, n_1 , change in n of core-clad, Radiance)
 Launching of power → source to fiber
 Coupling of power → (source to fiber)
 (fiber to fiber)

* Coupling efficiency = $\frac{\text{Power coupled into fiber}}{\text{Power emitted from fiber}}$

- ↳ int quantum efficiency → radiative recombination int.
(source)
- ↳ ext. quantum efficiency → radiative emission
(source to medium)
- ↳ coupling efficiency → source → med → fiber

* Radiance → brightness

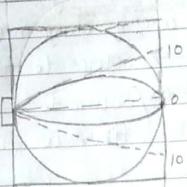
$$\text{Radiance} = \frac{\text{Optical Power}}{\text{optical power radiated} \times \text{solid angle} \times \text{surface area emitted}}$$

(W/sr cm²)

for unit area emitting surface

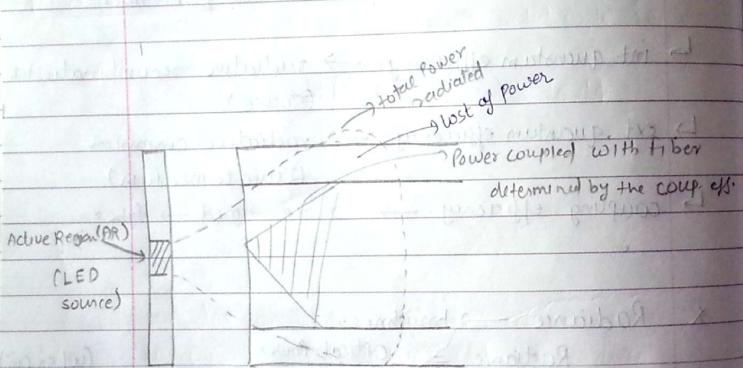
Surface emitting LED radiation pattern
(lambertian)

surface emitting LED are characterized by their lambertian op pattern, which means the source is equally bright when viewed from any direction.



$$B(\theta, \phi) = B_0 \cos \theta$$

(LED)



$$\text{radiance} = \frac{\text{Optical Power}}{\text{solid angle} \times \text{area}}$$

$$\begin{aligned} O_o P_o &= \text{Rad.} \times \text{solid angle} \times \text{area} \\ &= \text{Rad.} \times \pi F \times A_l \end{aligned}$$

\downarrow
core diameter of
numerical fiber
(antireflection)
apperture

$$P_f = \int_{AF} \left[\int_{SF} B(\theta_s, \phi_s) d\Omega_s \right] dA_s$$

↓
Power
coupled
with fiber

$$= \int_0^{\pi} \int_0^{2\pi} \left[\int_0^{2\pi} \int_{\theta_s, \max} B(\theta, \phi) \sin \theta d\phi d\theta \right] d\Omega_s \cdot r dr$$

Power coupled from point
source

$$\begin{aligned} g_m &= a \quad \left\{ \begin{array}{l} g_s > a \\ g_s < a \end{array} \right. \\ &= g_s \quad \uparrow \end{aligned}$$

source emitted
surface radius

$$= \int_0^{\pi} \int_0^{2\pi} \left[\int_0^{2\pi} \int_{\theta_s, \max} 2B_0 \cos \theta \sin \theta d\phi d\theta \right] d\Omega_s \cdot r dr$$

$$= \frac{B_0}{2} \int_0^{\pi} \int_0^{2\pi} \int_{\theta_s, \max} \left[\frac{\cos \theta}{2} \right] d\phi d\theta \cdot r dr$$

$$= \frac{B_0 \cdot 2\pi}{2} \int_0^{\pi} \int_0^{2\pi} r dr d\theta (1 - \cos \theta_{\max})$$

$$= \frac{B_0 \cdot 2\pi \cdot \pi s^2 \cdot 2\pi \cdot (1 - \cos \theta_{\max})}{2}$$

$$= \frac{B_0 \pi^2 \pi s^2 (1 - \cos \theta_{\max})}{2} = \frac{B_0 \pi^2 \pi s^2 \cdot 2 \sin^2 \theta_{\max}}{2}$$

$$P_F = \pi^2 B_0 g_s^2 (NA)^2$$

$g_s < a$

$$P_F = \pi^2 a^2 (NA)^2 B_0 ; g_s > a$$

power emitted from source (P_s) :

$$P_s = \text{radiant solid angle} \times A_{\text{area}}$$

$$= \left[\int_0^{2\pi} \int_0^{\pi/2} B_0 \cos \theta \sin \phi d\theta d\phi \right] A_s$$

$$P_s = \pi^2 g_s^2 B_0$$

Ex: Consider an LED that has circular emitting area with radius 35 μm and Lambertian pattern 150 W/cm² sr at given driver. Calculate current. Compare OP coupled into a step index fiber. One has co-radius of 25 μm & other has co-radius of 50 μm. NA = 0.2

Sol:

$$i) g_s = 35 \Rightarrow a = 35 \mu\text{m}$$

$$P_F = \pi^2 B_0 g_s^2 (NA)^2$$

$$= \pi^2 \frac{(150)}{10^{-4}} \times (35)^2 \times (0.2)^2 \times 10^{-12}$$

$$= 36.84 \times 10^{-9}$$

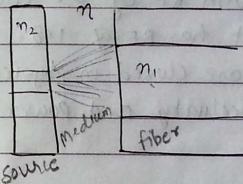
$$ii) g_s = 35 < a = 50$$

$$P_F = \pi^2 a^2 (NA)^2 B_0$$

$$= \pi^2 (150) (50)^2 (0.2)^2$$

$$= 74.02 \times 10^{-5}$$

→ Reflectivity %



$$R = \left(\frac{n_1 - n}{n_1 + n} \right)^2 = \text{fresnel reflection (reflectivity)}$$

$$g_1 = \left(\frac{n_1 - n}{n_1 + n} \right) = \text{reflection coefficient}$$

rate at which power coupled into fiber reduces w.r.t. emitted power from source

$$R = f (P_s, P_t)$$

$$R = \frac{P_s - P_F}{P_s}$$

— When there is no space b/w fiber & source then

$$R = \left(\frac{n_2 - n_1}{n_2 + n_1} \right)^2$$

Coupling effi → how much power is transmit
in fiber

Reflectivity → its provide the cause of
power loss in fiber

Ex8- GaAs Optical source with RF of 3.6 coupled
with silica fiber that has RF of 1.48. If the
fiber and the source are closed in physical
contact calculate Reflectivity and Power loss ratio

Sol:

$$n_2 = 3.6$$

$$n_1 = 1.48$$

$$R = \left(\frac{n_2 - n_1}{n_2 + n_1} \right)^2 = \text{reflectivity}$$

$$= 0.194$$

$$\text{NA} = 3.28$$

$$P_F = \pi^2 B_0 g_{se}^2 (\text{NA})^2$$

$$= P_s (\text{NA})^2$$

$$\frac{P_F}{P_s} = \frac{(3.28)^2}{1.48^2} = 10.28 \quad R = \frac{P_s - P_F}{P_s}$$

$$10 \log \left(\frac{P_F}{P_s} \right) = 10.28 \text{ dB}$$

NA is not given; for fiber

$$P_F = (1-R) P_s$$

$$10 \log \frac{P_F}{P_s} = -0.80 \text{ dB}$$

* Power loss L in decibels (from book)

$$L = -10 \log \left(\frac{P_F}{P_s} \right)$$

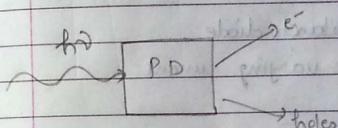
Chp-6

Photodetector

To receive the optical signal at Rx side
principal:

- Optical to electric signal converted by the photo-detector.

- Optical \longrightarrow Electrical
(power) \longrightarrow (time varying current)



- photons striking on the photo-detector and produce the e⁻ & holes pairs.

- quantum efficiency of $= \frac{\text{No. of } e^- \& \text{holes pairs}}{\text{No. of incident photons}}$

in Pin diode \rightarrow 30 - 45% quantum efficiency
which means that 100 photons
strikes on the PD and effectively
photons 30 to 45 are makes the
hole-e- pair.

- We required Quantum efficiency as high as possible for that some factors are there:

- E_g (energy band gap)

$\hbar\omega > E_g$ (for changing the E)

- Dimension of material (thickness of region)

- λ (wavelength)

By changing the E_g & k , Current eff. may be varied as per the requirement.

Types of Photodiodes

(1) Semiconductor

Pin photodetector Avalanche diode

- they gives the time varying current.

(2) Photoelectric

photons \rightarrow heat \rightarrow Dielectric length \rightarrow capacitor

- they gives the time varying capacitors.

(3) Photo multipliers

Their large size and High Volt. reqd; make them unsuitable for optical fiber sls.

PIN Diode

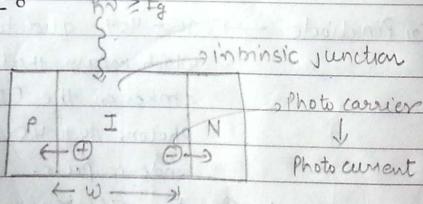
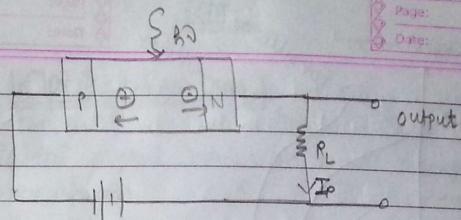
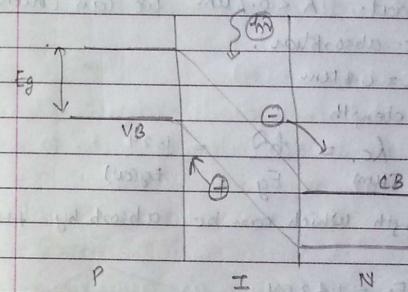


Photo carrier \rightarrow free charges (as electrons) which are due to photons.



- it gives the time varying current



Power in intrinsic mat.

$$P(x) = P_0 (1 - e^{-\alpha_s(x)x})$$

$x \rightarrow$ depth of intrinsic material

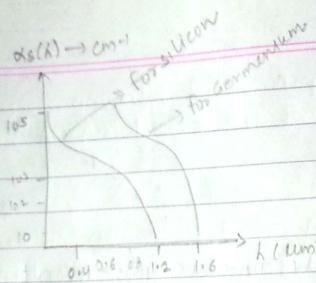
depth is as high as possible so that our intrinsic material can absorb the max. optical power

if $x \rightarrow \infty$; $P(x) = P_0$ which is not practically feasible.

$P_0 \rightarrow$ fin power, which we strike on the mat.

$\alpha_s(x) \rightarrow$ attenuation coefficient absorption

- we required the largest value of $\alpha_s(x)$, so for that



- for silicon mat. $\lambda < 0.6 \mu\text{m}$, we can choose for the max. absorbtion.
- for Ge $\rightarrow \lambda = 0.8 \mu\text{m}$
- cut off wavelength

$$\lambda_c = \frac{hc}{E_g} = \frac{1.24}{E_g(\text{eV})}$$

Max wavelength which can be absorb by the mat.

For ex: if $E_g = 1.12 \text{ eV}$

then $\lambda_c \approx 800 \text{ nm}$

chp-7

Page: _____ Date: 10-March

Optical Receivers

- Digital signal emission and reception
- Error sources
- prob. of error
- Quantum limit

① PD (Photo detectors)

Amp

② Error

int. Ext

Short thermal noise

noise (random motion of e^- in conductor)

(random arrival rate of photons) \rightarrow thermal noise

bias resistance

noise

PD \rightarrow Amps \rightarrow equalizer

shot noise (Quantum)

PD gives the current variation

through the resistance

current variation converted into voltage

- noise is generated by theamps
- thermnoise associated with the biased resistance

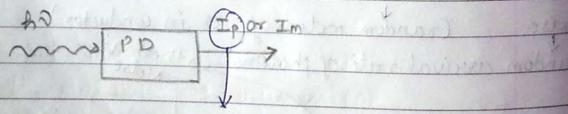
Shot noise:

- random arrival rate of photons
- random arrival rate of photons at the PD side, due to that random variation of photocurrent will occurs
- Carrier Multiplication (M)

In an avalanche photo diode, carrier multiplication factor is M, is not const. Variation of factor M will then due to that noise will occurs.

- discontinuity in the variation in current is shot noise.
- Shot noise is also known as quantum noise.

Mathematical representation of shot noise:



Variation of I_p depends on the shot noise

$I_m \rightarrow$ multiplied current

$I_p \rightarrow$ generated current

avg. $e^- - h$ pairs (\bar{N}) generated in given time T for optical power $p(t)$

$$\bar{N} = \frac{n}{h\nu} \int_0^T p(t) dt$$

prob. of occurring K event

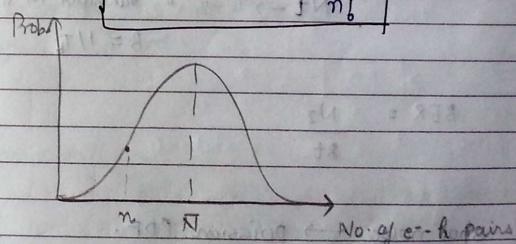
$$P(K) = \frac{\bar{N}^K e^{-\bar{N}}}{K!}$$

↓
mean

actual n $e^- - h$ pairs gen

prob. of that is

$$p(n) = \frac{\bar{N}^n e^{-\bar{N}}}{n!}$$



for RAPP (reach-through Avalanche photo diode)

$$\text{noise factor } F(M) \approx M \quad x \rightarrow 0.5 \leq 1$$

depends on material

$\leftarrow T_b \rightarrow$



Receiver performance :

- prob of error (BER) $= 10^{-9} = \frac{1}{10^9}$

for optical fiber for transmitted the 10^9 bits
only 1 bit error will there.

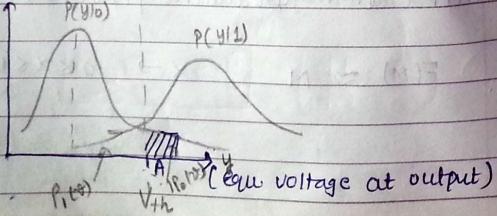
- $BER = \frac{N_e}{Nt} \rightarrow$ no of bits having error in time t
 $Nt \rightarrow$ " " " transmitted in time t
 $B = 1/T_b$

$$BER = \frac{N_e}{Bt}$$

- shot noise \rightarrow poission PDF.

thermal noise \rightarrow gaussian PDF

Prob:



(equ voltage at output)

A: $P_0(V_{th}) = \int_{V_{th}}^{\infty} P(y|0) dy \rightarrow$ error, when we transmit 0, Rx $\rightarrow y$

$P_1(V_{th}) = \int_{-\infty}^{V_{th}} P(y|1) dy \rightarrow$ error, when we transmit 1 & Rx $\rightarrow y$

\downarrow error prob. when $V_{th} > V$, while sending logic 1

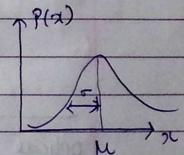
$$P_e = a P_0(V_{th}) + b P_1(V_{th})$$

$$a = b = 0.5$$

Unbiased

14-March

* $f(x) = \frac{1}{\sqrt{2\pi\sigma^2}} e^{-(x-\mu)^2/2\sigma^2}$



\hookrightarrow For eqn A

$$\begin{aligned} \bar{x}_{OFF} &= \text{Var} \\ b_{OFF} &= \text{mean} \end{aligned} \quad \left. \right\} - \text{transmit logic} = 0$$

for eqn B

$$\sigma^2_{ON} = \text{Varianne}$$

$$b_{ON} = \text{mean}$$

so; $P_0(V_{th}) = \int_{V_{th}}^{\infty} \frac{1}{\sqrt{2\pi\sigma_{OFF}^2}} e^{-(x-\bar{x}_{OFF})^2/2\sigma_{OFF}^2} dx$

$$P_e(V_{th}) = \int_{-\infty}^{V_{th}} \frac{1}{\sqrt{2\pi}\sigma_{on}} e^{-(x-b_{on})^2/2\sigma_{on}^2} dx$$

$$\text{BER} = P_e(\phi) = \frac{1}{\sqrt{2\pi}} \frac{e^{-\phi^2/2}}{\alpha}$$

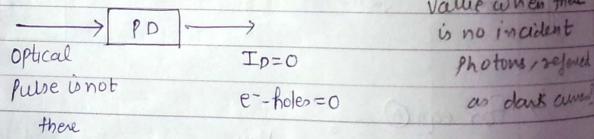
Where $\phi = \frac{V_{th} - b_{off}}{\sigma_{off}} = \frac{b_{on} - V_{th}}{\sigma_{on}}$

Quantum limit

- in Ideal condition

$$\eta = 1$$

that means, no dark current (means current



- (avg. \bar{N})

$$P_s(n) = \bar{N}^n e^{-\bar{N}}$$

$$P_s(0) = e^{-\bar{N}} \quad \left(\frac{n_e E}{h\nu} = \bar{N} \right)$$

$$\bar{N} = \frac{n_e P_s \tau}{h\nu} = \frac{P_s \tau}{h\nu}$$

$\tau \rightarrow$ time when photons strikes on Photo detector

Ex: if BER = 10^{-9} & Oppulse is not present

(a) then find $\bar{N} = ?$

Sol:

$$P_e(0) = 10^{-9} = e^{-\bar{N}}$$

$$\bar{N} = 20.72 \approx 21$$

(b) min. incident optical power, that must fall on Photo diode to achieve 10^{-9} BER. The given data rate is 10 Mbps, for simple binary level signalling

Sol:

$$\tau = \frac{2}{B} \quad B \rightarrow \text{data rate}$$

$$B = 2 \text{ Mbps}$$

$$\tau = \frac{2}{10M} = 0.2 \mu s$$

$$\text{as we know that } \bar{N} = \frac{P_s \tau}{h c / \lambda}$$

$$P_s = \frac{\bar{N} h c}{\lambda \times 2}$$

$$= \frac{21 \times 6.62 \times 10^{-34} \times 3 \times 10^8}{850 \times 10^{-9} \times 0.2 \times 10^{-6}}$$

$$= 2.4 \times 10^{-11} \text{ Watt}$$

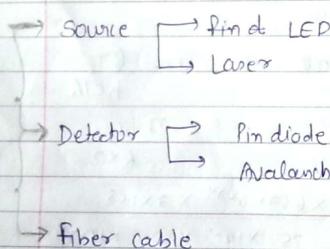
$$= 24.2 \text{ PW}$$

Digital links

- Splicer \rightarrow mountable joint (internally) (permanent)
- Connector \rightarrow dismountable joint (permanent)
- $P_s = -20 \text{ dBm}$
 $P_f = -10 \text{ dBm}$
 $P_R = -30 \text{ dBm}$

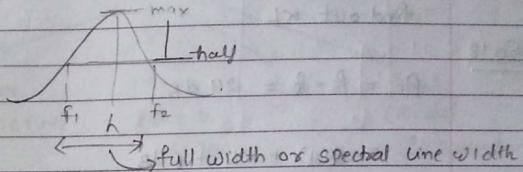
$$\boxed{P_T = P_s - P_R}$$

If amount of loss, we can accept in fiber link and that losses due to various splicers, connector and due to fiber cable. And this known as "link budget".

Criteria for Selecting the fiber

- BW
- Cost
- Single mode / multimode
- Step / graded

- Selecting the optical source
 $(FWHM) \rightarrow$ full width half max



- Selecting the detector
 Avalanche photodetector we may choose bcz in this small amount of optical power can be detected and it converts can gen' the photo carriers

Optical power loss model:

$\Delta LSP \rightarrow$ power loss occurs bcz of splicer

$\alpha \rightarrow$ attenuation coeff

$L \rightarrow$ loss occurred by fiber

$\Delta Lc \rightarrow$ loss occurred by connector

SIS margin \rightarrow margin required for future expansion of fiber cable

$$\boxed{P_T = P_s - P_R = m_{lc} + n_{lsp} + \alpha L + \text{SIS margin}}$$

if splicer loss is included in cable loss, and no connector in blw

$$\boxed{P_T = \Delta Lc + \alpha L + \text{system margin}}$$

Ex: if $P_R = -42 \text{ dBm}$; $P_s = -13 \text{ dBm}$

connector loss : 1 dB at both Rx & Tx

SIS margin: 6dB

find out αL

So 18

$$P_T = P_S - R = 29 \text{ dB}$$

$$= Q(1) + \alpha L + 6 \text{ dB}$$

$$\text{so, } \alpha L = 21 \text{ dB}$$

↳ if $\alpha = 3.5 \text{ dB/Km}$, then 6 Km transmission path is possible.

↳ if $\alpha = 2 \text{ dB/Km}$, then $L = \frac{21}{2} = 10.5 \text{ Km}$

(By changing the wavelength, we can change the attenuation factor)

17-march

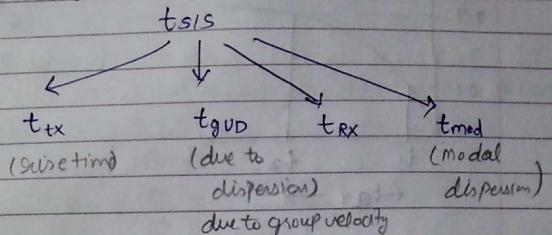
Rise time Budget

- Determine Dispersion (limitation of optical link)
- t_{sys}

$$\hookrightarrow < 0.7 t_b \quad (\text{NRZ})$$

$$< 0.35 t_b \quad (\text{RZ})$$

- rise time occurs at the Tx side.



$$t_{sys} = [t_{tx}^2 + t_{gud}^2 + t_{mod}^2 + t_{rx}^2]^{1/2}$$

$$\hookrightarrow t_{gud} \approx \frac{L |D| \Delta \lambda}{c} \quad \begin{matrix} \text{spectral line width (nm)} \\ \text{km} \end{matrix}$$

$$\hookrightarrow t_{mod} = \frac{44 D}{B_m} \quad \begin{matrix} \text{(ms)} \\ B_m \rightarrow \text{BW for M, mode} \text{ (MHz)} \end{matrix}$$

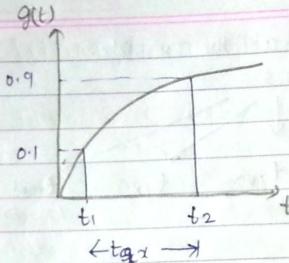
$$\boxed{t_{mod} \approx \frac{0.44}{B_m}}$$

$$B_m = \frac{B_0}{L} \quad \begin{matrix} \text{BW for 1 km} \\ L \rightarrow 0.5 \text{ to } 1 \end{matrix}$$

↳ length of fiber

↳ empirical parameter (practical value 201)

$$\hookrightarrow t_{rx}$$



$$g(t) = (1 - e^{-2\pi B_{rx}xt}) \quad U(t)$$

↓ 3dB t ≥ 0

$$\begin{aligned} g(t_1) &= 1 - e^{-2\pi B_{rx}xt_1} & g(t_2) &= 0.9 \\ 0.1 &= 1 - e^{-2\pi B_{rx}xt_1} & e^{-2\pi B_{rx}xt_2} &= 0.1 \\ e^{-2\pi B_{rx}xt_1} &\approx 0.9 \rightarrow \textcircled{1} & e^{-2\pi B_{rx}xt_2} &\approx 0.1 \rightarrow \textcircled{2} \end{aligned}$$

$$3 \text{ dB} \rightarrow t_{rx} = t_2 - t_1 = f(B_{rx})$$

electrical BW

$t_{rx}(\text{ns}) =$	350
$B_{rx} (\text{MHz})$	

Example 8 Risetime Budget for multimode 6 Km fiber link
with 20Mbps

$t_{rx} \Rightarrow T_{rx}$ Rise time 15ns

D - Dispersion of 0.08 ns/mm-km

Oh - Spectral line width of source 40nm

$B_{rx} - B_{rx} = 400 \text{ MHz/km}$

$B_{rx} - \text{Received } B_w = 400 - 25 \text{ MHz}$

$$\begin{aligned} a &= 0.7, t_{sys} = ? \\ \text{So } t_{rx} &= 15 \text{ ns} \\ t_{mod} &= L \cdot D \cdot \Delta \lambda \\ &= 6 \text{ km} \times 0.08 \times 40 \text{ nm} = 192 \text{ ns} \\ t_{mod} &= \frac{0.44}{B_m} = 3.9 \text{ ns} \\ B_m &= \frac{B_0}{L^2} = \frac{400 \times 10^6}{(6 \text{ km})^{0.9}} \end{aligned}$$

$$t_{rx} = \frac{350}{25} = 14 \text{ ns}$$

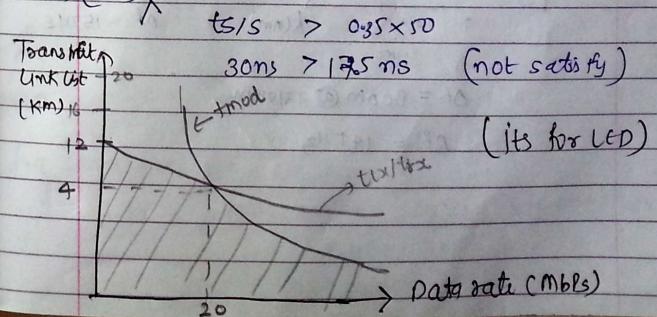
So, $t_{sys} = 30 \text{ ns}$

↳ 20Mbps (NRZ) ✓

$$t_b = \frac{1}{20 \text{ M}} = 0.05 \times 10^{-6} = 50 \text{ ns}$$

$$\begin{aligned} t_{sys} &< 0.7 \cdot t_b \\ &< 0.7 \times 50 \\ &< 35 \text{ ns} \end{aligned}$$

↳ (RZ) X

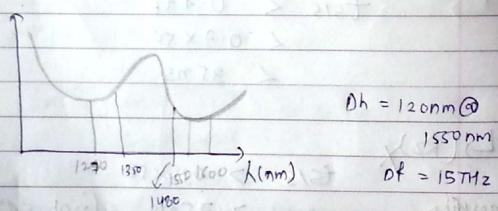
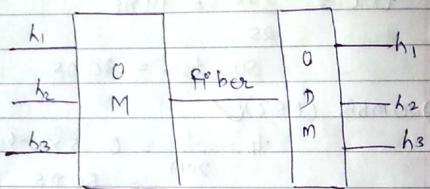


WDM: (wavelength division Multiplexing)

In TDM, it's difficult to handle in the fiber because freq. is in THz.

But in WDM, for respective wavelength at that freq. is very less compare to TDM or FDM.

$$\lambda_{1310} \quad \lambda_{1550}$$



$$\Delta\lambda = 80 \text{ nm} @ 1310 \text{ nm}$$

$$\Delta f = 14 \text{ THz}$$

Page: 7
Date: 7-10-2023

Page: _____
Date: _____

$$f = \frac{c}{\lambda}$$

$$\frac{df}{d\lambda} = -\frac{c}{\lambda^2}$$

$$\Delta f = -\frac{c}{\lambda^2} d\lambda$$

$$|\Delta f| = \frac{c}{\lambda^2} (\Delta\lambda)$$

available wavelength range

center wavelength

Ex: find out no. of channels possible on 1525 mm to 1565 nm range. if reg. Spacing is 100 GHz.

$$\text{Sol: } \Delta\lambda = \frac{1565 + 1525}{100 \text{ GHz}} = 4 \text{ nm}$$

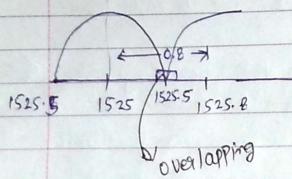
$$100 \text{ GHz} = \frac{3 \times 10^8}{\text{nm}^2} (4 \text{ nm})$$

$$1525 \quad 1565 \rightarrow \lambda (\text{nm})$$

freq. spacing is $\Delta f = 100 \text{ GHz}$

$$\Delta\lambda = \frac{3 \times 10^8}{100 \times 10} \frac{100 \times 10^9 \times (1545)^2 (10^{-18})}{3 \times 10^8}$$

$$= 0.8 \text{ nm}$$



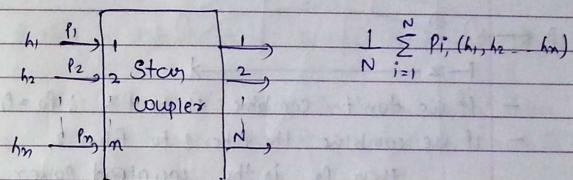
(Q) Find out the max spectral linewidth available in the source, such that it accommodates 50 channels.

Sol:

Passive Components :

- Working in optical domain
- Variable Star coupler

Star coupler :

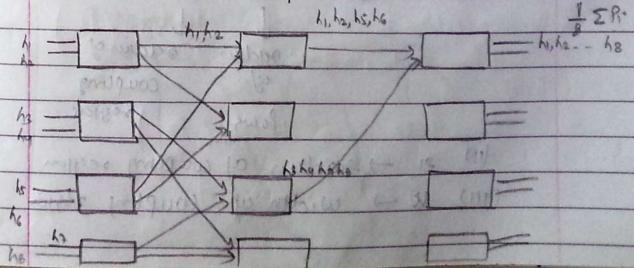


$$\sum_{i=1}^n p_i$$

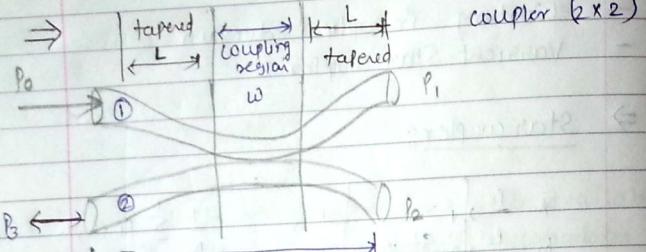
- Input power and output power is same.
- Output power contains at a particular channel is the avg. of the input power at every wavelength.
- Split and combine

8x8 Star coupler

2x2 star coupler



$\Rightarrow N \times N \rightarrow \frac{N \log_2 N}{2} \rightarrow$ needs the amount of coupler ($k \times 2$)



- if we don't combine f_1 to f_2 ; $P_0 = P_1$
- if we combine the fiber 1 to fiber 2

then P_2 is the coupling power

$P_3 \rightarrow$ cross talk, that means, some power is reflected from the P_2 . So at terminals we don't require any power is to given.

Coupling region

through that we can obtain the P_1 & P_2 that depends on

$$(I) \Delta r = r_1 - r_2$$

\downarrow \downarrow
 radius radius of
 of coupling
 power region

$$(II) \Delta r = r_1 - r_2$$

$$(III) W \rightarrow \text{width of coupling region}$$

Tapered region

$$V = \pi/4$$

a abrupt change in radius of fiber \rightarrow
 as radius is changes $\rightarrow V$ is changes
 due to that $P_{\text{taper}} \propto \frac{4}{3\sqrt{m}}$; $m \propto V$

P_{taper} is there and WSS is there at P_3

- so by smoothly change in radius of fiber we can change the power at P_3 which is loss.

Power coupled into node 2

$$P_2 = P_0 \sin^2(Kz)$$

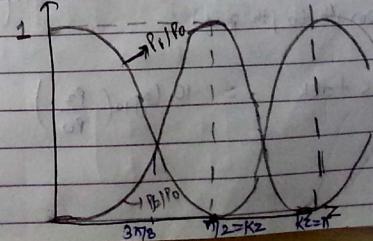
$K \rightarrow$ coupling coefficient

$z \rightarrow$ distance $K = \frac{2\pi n}{\lambda}$

$$\text{lossless, } P_1 = P_0 - P_2$$

$$\text{condition, } P_1 = P_0 \cos^2(Kz)$$

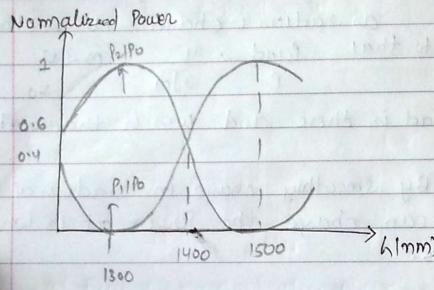
Normalised Power



$$\text{at } kZ = \pi/4 ; \quad P_1 = 1/2 P_0 = P_2$$

$$\text{at } kZ = \frac{3\pi}{2} ; \quad P_1 = P_2$$

$$\text{at } kZ = \pi/2 ; \quad P_1 = 0 \\ P_2 = \text{max.}$$



⇒

$$\text{- splitting ratio} = \frac{P_2}{P_1 + P_2} \times 100 \\ (\text{coupling ratio})$$

$$\text{- excess loss} = 10 \log_{10} \left(\frac{P_0}{P_1 + P_2} \right)$$

$$\text{- insertion loss} = 10 \log_{10} \left(\frac{P_i}{P_j} \right) \\ (\text{from } i^{\text{th}} \text{ to } j^{\text{th}} \text{ port})$$

$$\text{- cross talk} = 10 \log_{10} \left(\frac{P_3}{P_0} \right)$$

- 7 - April
- ① - for splitting the power $kZ = \pi/4 \rightarrow 3dB \text{ coupler}$
 - ② - cross bar (in that all power is $P_2 = P_0$)
 $P_1 = 0 \text{ at } k = \pi/2$

Scattering Matrix:

relation b/w $a_1 p$ and $01p$ power in terms of magnitude and phase.

$$1) \quad \begin{array}{c|cc|c} a_1 & S_{11} & & b_1 \\ \hline a_2 & & S_{22} & b_2 \end{array}$$

$$2) \quad \begin{array}{ccc|cc} a_1 & & & S_{11} & b_1 \\ a_2 & & & S_{21} & b_2 \\ \hline & & & S_{12} & \\ & & & S_{22} & \end{array} \quad (\text{in context of coupling})$$

$$\begin{bmatrix} b_1 \\ b_2 \end{bmatrix} = \begin{bmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \end{bmatrix}$$

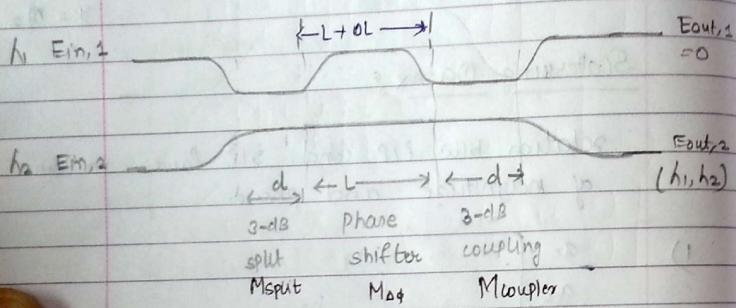
$$b_1 = a_1 S_{11} + a_2 S_{12}$$

$$b_2 = S_{21} a_1 + S_{22} a_2$$

$$\text{scattering matrix } S = \begin{bmatrix} \sqrt{1-\epsilon} & j\sqrt{\epsilon} \\ j\sqrt{\epsilon} & \sqrt{1-\epsilon} \end{bmatrix} \\ \epsilon \rightarrow \text{proportionate power}$$

MZI (Mach-Zehnder interferometer) Multiplexer:

- wavelength multiplexer



at $E_{out,1} \rightarrow$ destructive phase (h_1 and h_2 added in such a way that $E_{out,1} = 0$)

at $E_{out,2} \rightarrow$ constructive phase

- ΔL is added to provide the phase shift.

For 3 dB split (Msplit)

3 dB means half

$$M_{split} = M_{coupler} = S = \begin{bmatrix} \sqrt{1-\epsilon} & j\sqrt{\epsilon} \\ j\sqrt{\epsilon} & \sqrt{1-\epsilon} \end{bmatrix}$$

$\epsilon = 0.5$

$$= \frac{1}{\sqrt{2}} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix}$$

Phase shifter:

$$\Delta\phi = \frac{2\pi n_1}{\lambda} (L + \Delta L) - \frac{2\pi n_2}{\lambda} (L)$$

↓
Phase shift by the
shift 1

provided by the Path 1

assume: when both fiber are having the same

RF. $n_1 = n_2 = n_{eff}$.

$$\Delta\phi = \frac{2\pi n_{eff}}{\lambda} (\Delta L)$$

= $K \Delta L$

↳ Proportional constant

$$M_{\Delta\phi} = \begin{bmatrix} e^{jK\Delta L/2} & 0 \\ 0 & e^{-jK\Delta L/2} \end{bmatrix}$$

$$M = M_{split} M_{\Delta\phi} \cdot M_{split} \cdot M_{coupler}$$

$$= \frac{1}{2} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix} \begin{bmatrix} e^{jK\Delta L/2} & 0 \\ 0 & e^{-jK\Delta L/2} \end{bmatrix}$$

$$= \frac{1}{2} \begin{bmatrix} 0 & 2j \\ 2j & 0 \end{bmatrix} \begin{bmatrix} e^{jK\Delta L/2} & 0 \\ 0 & e^{-jK\Delta L/2} \end{bmatrix}$$

$$= \frac{1}{2} \begin{bmatrix} 0 & 2je^{jK\Delta L/2} \\ 2je^{-jK\Delta L/2} & 0 \end{bmatrix}$$

$$= j \begin{bmatrix} 0 & e^{j\frac{\alpha L}{2}} \\ e^{j\frac{\alpha L}{2}} & 0 \end{bmatrix}$$

$$= \frac{1}{2} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix} \begin{bmatrix} e^{j\alpha} & 0 \\ 0 & e^{-j\alpha} \end{bmatrix} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix}$$

$$= \frac{1}{2} \begin{bmatrix} 1 & j \\ j & 1 \end{bmatrix} \begin{bmatrix} e^{j\alpha} & j e^{j\alpha} \\ j e^{-j\alpha} & e^{-j\alpha} \end{bmatrix}$$

$$= \frac{1}{2} \begin{bmatrix} e^{j\alpha} - e^{-j\alpha} & j(e^{j\alpha} + e^{-j\alpha}) \\ j(e^{j\alpha} + e^{-j\alpha}) & (e^{-j\alpha} - e^{j\alpha}) \end{bmatrix}$$

$$= \frac{1}{2} \begin{bmatrix} 2j \sin \alpha & 2j \cos \alpha \\ 2j \cos \alpha & -2j \sin \alpha \end{bmatrix}$$

$$= j \begin{bmatrix} \sin \alpha & \cos \alpha \\ \cos \alpha & -\sin \alpha \end{bmatrix}$$

$$M = j \begin{bmatrix} \sin(\frac{\alpha L}{2}) & \cos(\frac{\alpha L}{2}) \\ \cos(\frac{\alpha L}{2}) & -\sin(\frac{\alpha L}{2}) \end{bmatrix}$$

$$\begin{bmatrix} E_{out,1} \\ E_{out,2} \end{bmatrix} = j \begin{bmatrix} \sin(\frac{\alpha L}{2}) & \cos(\frac{\alpha L}{2}) \\ \cos(\frac{\alpha L}{2}) & -\sin(\frac{\alpha L}{2}) \end{bmatrix} \begin{bmatrix} E_{in,1} \\ E_{in,2} \end{bmatrix}$$

$$k = \frac{2\pi n_{eff}}{\lambda h}$$

at λ_1 and λ_2 wavelength

$$\begin{bmatrix} E_{out,1} \\ E_{out,2} \end{bmatrix} = j \begin{bmatrix} \sin(\frac{k_1 \alpha L}{2}) & \cos(\frac{k_2 \alpha L}{2}) \\ \cos(\frac{k_1 \alpha L}{2}) & -\sin(\frac{k_2 \alpha L}{2}) \end{bmatrix} \begin{bmatrix} E_{in,1} \\ E_{in,2} \end{bmatrix}$$

$$E_{out,1} = j \left\{ \sin\left(\frac{k_1 \alpha L}{2}\right) E_{in,1}(\lambda_1) + \cos\left(\frac{k_2 \alpha L}{2}\right) E_{in,2}(\lambda_2) \right\}$$

$$E_{out,2} = j \left[\cos\left(\frac{k_1 \alpha L}{2}\right) E_{in,1}(\lambda_1) + \sin\left(\frac{k_2 \alpha L}{2}\right) E_{in,2}(\lambda_2) \right]$$

$E \rightarrow$ intensity

if we want to find the Power $P = E * E^*$

$$P_{out,1} = |E_{out,1}|^2$$

$$P_{out,2} = |E_{out,2}|^2$$

$$P_{out,1} = P_{in,1} \sin^2\left(\frac{k_1 \alpha L}{2}\right) + \cos^2\left(\frac{k_2 \alpha L}{2}\right) P_{in,2} \quad (1)$$

$$P_{out,2} = P_{in,1} \cos^2\left(\frac{k_1 \alpha L}{2}\right) + P_{in,2} \sin^2\left(\frac{k_2 \alpha L}{2}\right) \quad (2)$$

↳ for getting the $P_{out,1} = 0$
 $P_{out,2} = P_{in,1} + P_{in,2}$

$$\frac{P_{out,1}}{P_{out,2}} \text{ so, } \sin^2\left(\frac{k_1 \Delta L}{2}\right) = 0 \quad \& \quad \cos^2\left(\frac{k_1 \Delta L}{2}\right) = 1$$

$$\frac{k_1 \Delta L}{2} = \pi$$

$$\cos^2\left(\frac{k_2 \Delta L}{2}\right) = 0 \quad \& \quad \sin^2\left(\frac{k_2 \Delta L}{2}\right) = 1$$

$$\frac{k_2 \Delta L}{2} = \frac{\pi}{2}$$

$$\text{so, } k_1 \Delta L = 2\pi$$

$$k_2 \Delta L = \pi$$

$$(k_1 - k_2) \Delta L = +\pi$$

$$\left(\frac{2\pi n_{eff}}{\lambda_1} - \frac{2\pi n_{eff}}{\lambda_2} \right) \Delta L = \pi$$

$$\Delta L = \left\{ \frac{2\pi n_{eff}}{\lambda_1} \left(\frac{1}{\lambda_1} - \frac{1}{\lambda_2} \right) \right\}^{-1}$$

this is extra path which is provided
 for getting the $\text{① terminal 1 } \overset{\text{derby phase}}{P_{out,1}} = 0$ and
 $\text{② terminal 2 } P_{out,2} = P_{in,1} + P_{in,2}$
 const phase

$$\Delta f = \frac{C}{\lambda_1} - \frac{C}{\lambda_2}$$

$$\Delta L = \frac{C}{2n_{eff} \Delta f}$$

Example: Assume that input wavelengths for 2×2 silicon MZI are separated by 10 GHz with $n = 1.5$. determine length diff b/w interferometers arm for given wavelength.

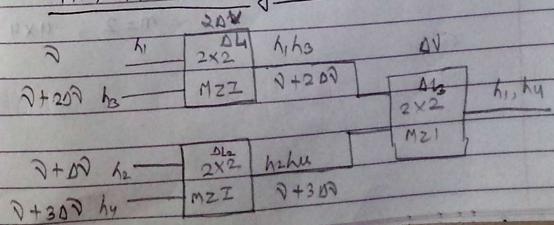
Sol:

$$\begin{aligned} \Delta L &= \frac{C}{2n_{eff} \Delta f} \\ &= \frac{3 \times 10^8}{2 \times 1.5 \times 10^9} \\ &= \frac{0.1}{10} \\ &= 10 \text{ mm} \end{aligned}$$

$$\begin{aligned} \Delta L &= 10 \text{ mm for } \Delta f = 10 \text{ GHz} \\ \Delta L &= 0.22 \text{ mm for } \Delta f = 130 \text{ GHz} \end{aligned}$$

By Tting the freq. or BW, path diff will \downarrow se.

$\Rightarrow 4 \times 4 \text{ MZI using } 2 \times 2 \text{ MZI }$



Stage : 0

Stage : 1

- each of mux having the same freq. separation.
- Stage: 0

$$\text{freq. separation} = \Delta f$$

$$\Delta f_1 = \Delta f_2 = \frac{C}{\text{Neff}(\Delta f)}$$

Stage: 1

$$\Delta f_3 = \frac{C}{\text{Neff}(\Delta f)}$$

NxN Mux

$$2^n, n \geq 1$$

$$\boxed{\Delta f_j = \frac{C}{2^{n-j} \text{Neff}(\Delta f)}}$$

Stage no.

$$0 \leq j \leq n,$$

$$N = 2^n$$

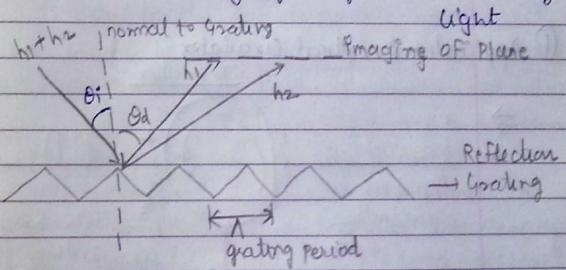
$$m = 2; 4 \times 4 \text{ Mux}$$

QUESTION

Page: _____
Date: 10 - April

Grating

- Grating is a periodic structure or variation in a material.
- depend on the wavelength \rightarrow operation
- for design the optical filter, we req. grating principle for suppressing and enhanced the wavelengths by filters.
- working principle of grating: Diffraction of light



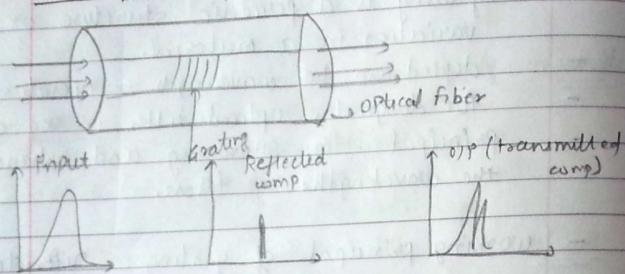
What is the slope of grating, which we required to split the wavelength as h_1 & h_2 into h_1 & h_2 .

Constructive

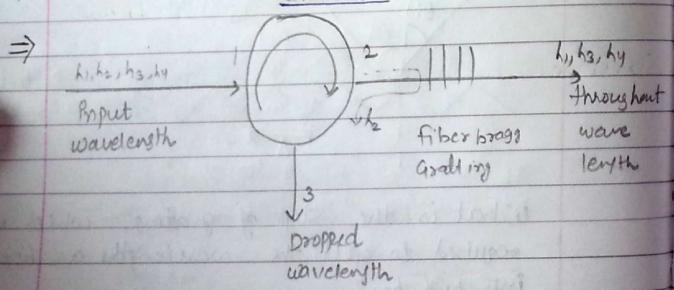
$$\text{interference: } \Lambda (\sin \theta_i - \sin \theta_d) = m\lambda$$

Grating incident diffraction angle
spare angle

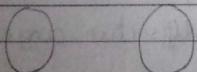
Development of fiber Bragg Gratings:



① Application: optical Circulator



② Applications: Demux & mux



Advantages:

- temp and stress sensitive
- can separate the wavelength

Tunable Laser sources:

Available choices for laser sources for WDM App:

- ① DFB laser (Distributed feedback)
- ② (very costly for tunable wavelengths)

② Wavelength tunable laser

(0.008 - 0.04 nm/m)

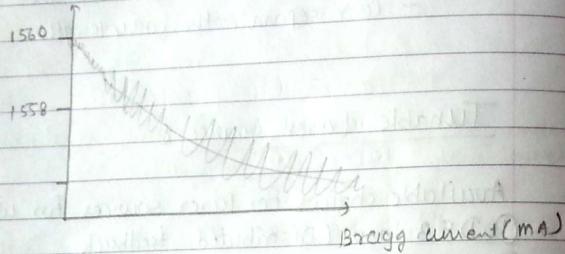
By changing the bias current, By changing the RI, we can change the optical wavelength "Thermo-optical" "Electric-optical"

Disad: time required for tune the wavelength is more.

③ Multi wavelength laser array

- we are making an array which one having the multiple wavelengths, By using the diff types of grading.

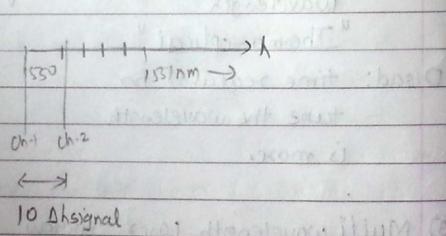
- Distributed bragg reflector (DBR) laser wavelength(nm)



By changing Bragg current, we can change the operating wavelength.

- channel spacing depends on the spectral line width

$$\Delta h = 0.01 \text{ nm}$$



Channel spacing = 10 times of the Δh_{signal}
If we don't take it, then interfacing b/w them is possible.

- How many channels can be possible in the system

$$1500 \quad 1600 \rightarrow h(\text{nm})$$

$$N = \frac{\Delta h_{\text{tunable}}}{\Delta h_{\text{channel}}} \quad \downarrow \text{no. of channels}$$

- ↳ tunability range is given by the (variation of RI) \Rightarrow

Tuning range

$$\frac{\Delta h_{\text{tune}}}{\lambda} = \frac{\Delta n_{\text{eff}}}{n_{\text{eff}}}$$

Operating wavelength

Example: max. index change of particular DFB laser operating at 1550 nm is 0.65 \times . Then find out Δh_{tune} , if source of 0.02 spectral line width is modulated with 2.59 bps signal then find out no of channels in tuning range.

Q10

$$\Delta h_{\text{tune}} = 0.65 \times \lambda$$

$$= 0.65 \times 1550 \text{ nm}$$

$$= 10.075 \text{ nm}$$

$$\Delta h_{\text{signal}} = 0.02 \text{ nm}$$

$$\begin{aligned}\Delta h_{\text{channel}} &= 10 \times 0.02 \\ &= 0.2 \text{ nm}\end{aligned}$$

$$\begin{aligned}N &= \frac{\Delta h_{\text{tune}}}{\Delta h_{\text{channel}}} \\ &= \frac{100.075 \times 10^{-9}}{0.2} \\ &= 500375 \approx 50 \text{ channels}\end{aligned}$$

Array of tunable lenses

$\Delta \lambda \rightarrow$ shifted

CH-1 [] | | | + $w \rightarrow$ width of grating

CH-2 [] | | $w_2 = w_1 + \Delta \lambda$

- Tuning range of such filters is around 60nm having channel BW 1nm (125 GHz @ 1550nm)

$$|\Delta \lambda| = \frac{c}{f^2} \Delta \lambda$$

$$\begin{aligned}&= \frac{3 \times 10^8 \times 1 \text{ nm}}{150 \times 150 \times (10^9)^2} \\ &= 1.24 \times 10^{-6} \times 10^8 \times 10^9 \\ &= 124 \text{ GHz}\end{aligned}$$