A theoretical framework for CV-QKD

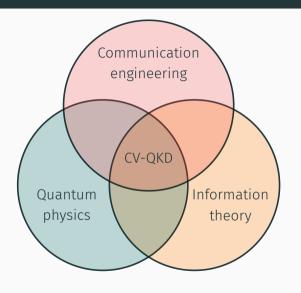
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You never know what you might discover by thinking outside the box that culture, conformity, and critics have tried to impose. — T.D. Jakes

Outline



- 1. Introduction
- 2. Coherent state transmitter
- 3. Coherent state receiver
- 4. Conclusion and outlook

Introduction

Secure transmission system

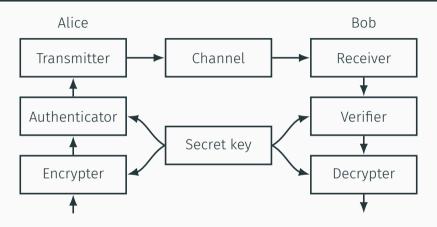


Figure 1: Block diagram of a secure transmission system.

Quantum key distribution system

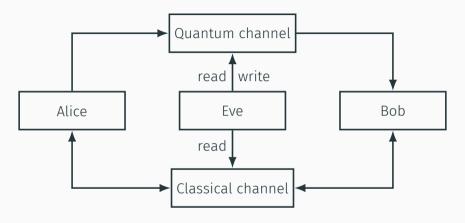


Figure 2: Block diagram of a quantum key distribution (QKD) system.

Discrete and continuous-variable QKD

Table 1: Comparison of discrete- and continuous-variable QKD.

	Discrete	Continuous
Visualization	Bloch sphere	Phase space
Hilbert space (dim)	Finite (two)	Countable (infinite)
Observable	$\hat{\mathbf{S}}(\mathbf{n}) = \hat{S}_i n^i$	$\hat{X}(\theta) = \frac{1}{\sqrt{2}} \left(\hat{a}e^{-i\theta} + \hat{a}^{\dagger}e^{+i\theta} \right)$
Standard basis	$\{ 0\rangle, 1\rangle\}$	$\{ x\rangle, p\rangle\}_{x,p\in\mathbb{R}}$

Continuous-variable QKD using coherent states

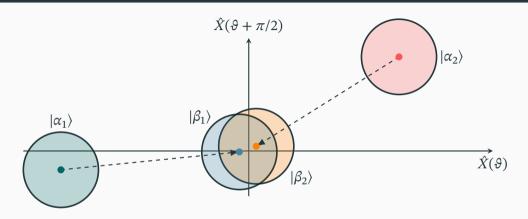


Figure 3: Phase space diagram of transmitted and received coherent states with mean (dots) and variances (opaque circles).

Continuous-variable QKD using continuous-time coherent states

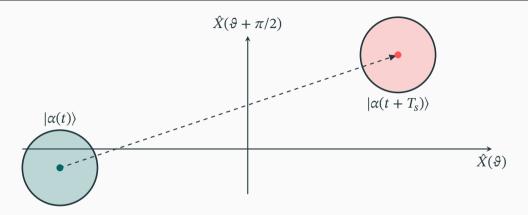


Figure 4: Phase space diagram of continuous-time coherent states with mean (dots) and variances (opaque circles) at two time instances.

Software-defined transmission system



Figure 5: Block diagram of the software-defined transmitter architecture.



Figure 6: Block diagram of software-defined receiver architecture.

Coherent state transmitter

Digital signal processing pipeline for symbol encoding

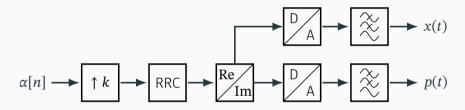


Figure 7: Block diagram of the digital signal processing for symbol encoding.

Symbol-encoding in the time domain

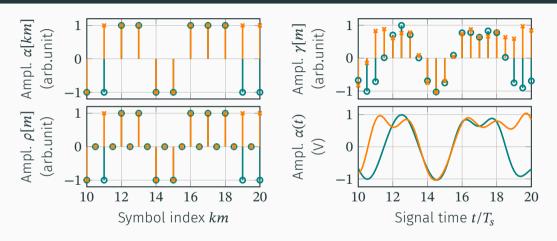


Figure 8: Symbol-encoding steps in the time domain for random QPSK sequence with real (orange) and imaginary (green) part.

Symbol-encoding steps in the frequency domain

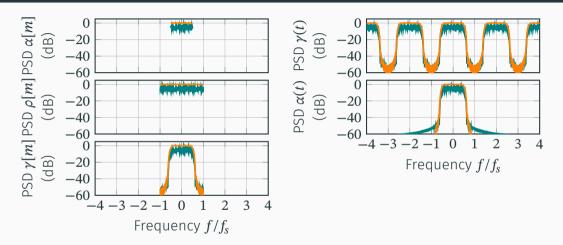


Figure 9: Symbol-encoding steps in the frequency domain for random symbols (green) and single symbol (orange).

Dual-quadrature upconversion in the time domain

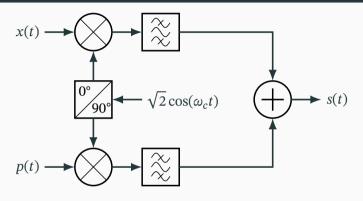


Figure 10: Power spectrum illustrating single-quadrature upconversion.

$$s(t) = x(t)\cos(\omega_c t) - p(t)\sin(\omega_c t) = \text{Re}\left[\alpha(t)e^{+i\omega_c t}\right] \quad \text{with} \quad \alpha(t) = x(t) + ip(t) \quad (1)$$

Dual-quadrature upconversion in the frequency domain

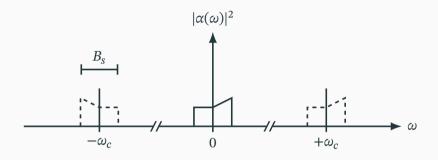


Figure 11: Power spectrum illustrating dual-quadrature upconversion.

$$s(t) = \operatorname{Re} \int_{\omega_0 - B_c/2}^{\omega_0 + B_c/2} \frac{\mathrm{d}\omega}{2\pi} \alpha(\omega - \omega_c) e^{+i\omega t}$$
 (2)

In-phase and quadrature modulator as dual-quadrature upconverter

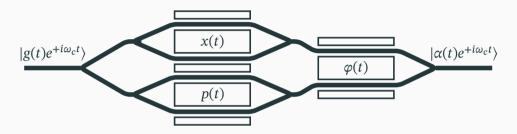


Figure 12: Drawing of an integrated in-phase and quadrature modulator.

$$|g(t)e^{+i\omega_c t}\rangle \to |\alpha(t)e^{+i\omega_c t}\rangle \qquad \alpha(t) = g(t)[x(t) + iq(t)]$$
 (3)

Interaction Hamiltonian of the electro-optical phase modulator

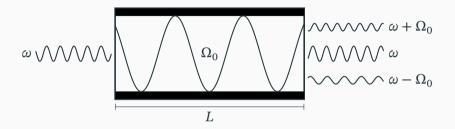


Figure 13: Traveling-wave electro-optical phase modulator of length *L*.

$$\hat{H}_{\rm int} = \int \frac{\mathrm{d}\omega}{2\pi} \int \frac{\mathrm{d}\Omega}{2\pi} g(\omega, \Omega) \hat{a}^{\dagger}(\omega) \hat{a}(\Omega) \hat{a}(\omega - \Omega) + \text{H.c.}$$
 (4)

Single-mode frequency-conversion operator

Simplifying assumptions:

$$\hat{a}(\Omega) \rightarrow \beta(\Omega)$$

$$\cdot \beta(\Omega) = \beta_0 2\pi \delta^{(1)} \delta^{(1)}(\Omega - \Omega_0)$$

•
$$\theta e^{i\varphi}/2 = \beta_0^* g(\omega, \Omega)/T$$

Evolution operator:

$$\hat{U}_{\text{int}} = e^{-\frac{1}{2}\theta \hat{T}_{+}(\Omega_{0})e^{-i\varphi} + \text{H.c.}}$$
 (5)

Upconversion operator:

$$\hat{T}_{+}(\Omega_{0}) = \int \frac{d\omega}{2\pi} \hat{a}^{\dagger}(\omega + \Omega_{0}) \hat{a}(\omega) \quad (6)$$

$$[\hat{T}_{+}(\Omega_0), \hat{a}^{\dagger}(\omega)] = \hat{a}^{\dagger}(\omega + \Omega_0) \quad (7)$$

Creation operator transform:

$$\hat{U}^{\dagger}\hat{a}(\omega)\hat{U} = \sum_{m \in \mathbb{Z}} J_m(\theta)\hat{a}(\omega + m\omega_l)e^{-im\theta}$$
(8)

Coherent state receiver

Signal processing pipeline for symbol decoding

$$\operatorname{Re}\left[\beta(t)e^{+i\omega_{c}t}\right] \longrightarrow \underbrace{\left(\sum_{l=1}^{\infty} u(t)\right)}_{\operatorname{cos}(\omega_{l}t + \vartheta)} u(t)$$

Figure 14: Block diagram of single-quadrature downconversion.

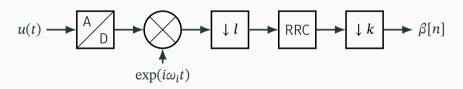


Figure 15: Block diagram of the analog-to-digital conversion and symbol-decoding.

Single-quadrature downconversion in the frequency domain

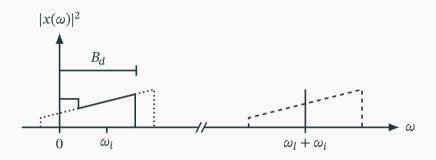


Figure 16: Power spectrum illustrating single-quadrature downconversion.

$$u(t) = \operatorname{Re} \int_{0}^{+B_{d}/2} \frac{\mathrm{d}\omega}{2\pi} \left[\beta(\omega + \omega_{l}) e^{+i\omega t} + \beta(\omega - \omega_{l})^{*} e^{-i\omega t} \right] e^{-i\vartheta} \tag{9}$$

Single-quadrature homodyne detection

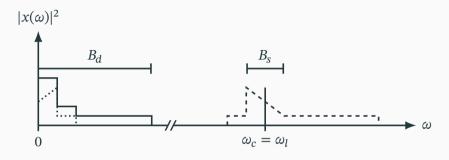


Figure 17: Power spectrum illustrating homodyne detection.

$$u(t) = \operatorname{Re} \int_{0}^{+B_{d}/2} \frac{d\omega}{2\pi} \left[\beta(\omega) e^{+i\omega t} + \beta(\omega)^{*} e^{-i\omega t} \right] e^{-i\vartheta}$$
 (10)

Single-quadrature heterodyne detection

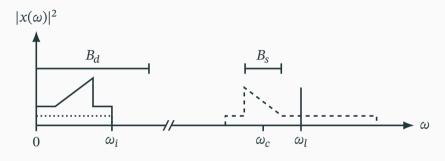


Figure 18: Power spectrum illustrating heterodyne detection.

$$u(t) = \operatorname{Re} \int_{0}^{+B_{d}/2} \frac{\mathrm{d}\omega}{2\pi} \left[\beta(\omega + \omega_{i})e^{+i\omega t} + \beta(\omega - \omega_{i})^{*}e^{-i\omega t} \right] e^{-i\vartheta}$$
 (11)

Comparison homo- and heterodyne detection

Table 2: Comparison of coherent receiver designs and their implications

	Homodyne (single)	Homodyne (dual)	Heterodyne
Balanced detectors	1	2	1
Quadratures	1	2	2
Intermediate frequency	$\omega_i = 0$		$\omega_i \neq 0$
Optical complexity	Low	High	Low
Signal bandwidth	High	High	Low
LO synchronization	Frequency and phase	Frequency	Bandwidth
SNR	High	Low	Low

Balanced detector as single-quadrature downconverter

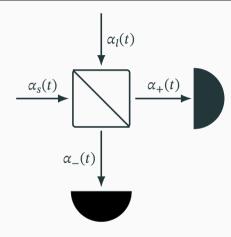


Figure 19: Electro-optical setup for balanced detection.

Differential photocurrent signal

$$i(t) \propto \text{Re} \int_{-R_{c}/2}^{+B_{d}/2} \frac{d\omega}{2\pi} \beta(\omega + \omega_{l}) e^{+i(\omega t + \vartheta)}$$
 (12)

Phase-rotated quadrature operator

$$\hat{X}(t) = \int_{-\infty}^{+\infty} \frac{\mathrm{d}\omega}{2\pi} \left[\hat{a}(\omega) e^{-i(\omega t + \theta)} + \text{H.c.} \right]$$
 (13)

Frequency-converted annihilation operator

$$\hat{U}^{\dagger}\hat{a}(\omega)\hat{U} = \sum_{m} J_{m}(\theta)\hat{a}(\omega + m\omega_{l})e^{-im\theta} \quad (14)$$

Conclusion and outlook

Summary

- Introduction to CV-QKD as a mechanism for secure key distribution.
- · Raised awareness to dependence of transmissions.
- Introduction to a software-defined coherent transmission system.
- · Introduction to concepts and methods from communication engineering.
- Extension of coherent transmission system to quantum regime.

Additional topics covered in the thesis

- Overview and comparison of DV- and CV-QKD protocols including post processing.
- Derivation of continuous-mode quantum theory of light rooted in quantum field theory.
- Summary of quantum models for the most important (electro-)optical components as building blocks for communication system.

Outlook

- · Noise model for measurements.
- Comparison of image band and squeezing attack.
- Further transfer of telecommunication methods, e.g., orthogonal frequency-division multiplexing (OFDM).
- · Properties of frequency-squeezed states.
- Equivalence of tensor-product with continuous-time coherent state transmission.

Maxwell equation of motion in the radiation gauge and momentum space

Free equation of motion in radiation gauge:

$$\partial_t^2 \mathbf{A} = \nabla^2 \mathbf{A} \qquad A_0 = 0 \qquad \partial_i A^i = 0 \tag{15}$$

Four-dimensional Fourier transform:

$$\mathbf{A}(t,\mathbf{x}) = \int \frac{\mathrm{d}^4 p}{(2\pi)^4} \mathbf{A}(p_0,\mathbf{p}) e^{-ip_0 t + i\mathbf{p} \cdot \mathbf{x}}$$
(16)

Equation of motion in momentum space:

$$0 = (p_0 - \|\mathbf{p}\|)(p_0 + \|\mathbf{p}\|) \qquad \omega(\mathbf{p}) = \|\mathbf{p}\|$$
 (17)

Plane-wave and polarization basis expansion in the radiation gauge

General plane-wave expansion in radiation gauge:

$$\mathbf{A}(t, \mathbf{x}) = \int \frac{\mathrm{d}^4 p}{(2\pi)^3} \delta^{(1)} \left(p_0^2 - \omega(\mathbf{p})^2 \right) \mathbf{A}(p_0, \mathbf{p}) e^{-ip_0 t + i\mathbf{p} \cdot \mathbf{x}}$$

$$= \int \frac{\mathrm{d}^4 p}{(2\pi)^3 \sqrt{2\omega(\mathbf{p})}} \mathbf{A}(\omega(\mathbf{p}), \mathbf{p}) e^{-ip_0 t + i\mathbf{p} \cdot \mathbf{x}} + \text{h.c.}$$
(18)

Polarization basis expansion

$$\mathbf{A}(t, \mathbf{x}) = \sum_{\lambda=1,2} a_{\lambda}(\mathbf{p}) \hat{\mathbf{e}}_{\lambda}(\mathbf{p})$$
(19)

Polarization basis transverse to momentum:

$$\mathbf{p} \cdot \hat{\mathbf{e}}_{\lambda}(\mathbf{p}) = 0 \qquad \hat{\mathbf{e}}_{\lambda}(\mathbf{p})\hat{\mathbf{e}}_{\sigma}(\mathbf{p}) = \delta_{\lambda,\sigma} \qquad \sum_{\lambda=1,2} \hat{\mathbf{e}}_{\lambda}(\mathbf{p})^{i}\hat{\mathbf{e}}_{\lambda}(\mathbf{p})^{j} = \delta^{ij} - \frac{p^{i}p^{j}}{\mathbf{p}^{2}}$$
(20)

Canonical quantization in the Coulomb gauge

Conjugate momentum density:

$$\Pi_i = \partial_t A_i = E_i = -E^i \tag{21}$$

Equal-time commutation relations:

$$[\hat{A}_i(t,\mathbf{x}),\hat{E}_j(t,\mathbf{y})] = -i\delta_{\perp,ij}^{(3)}(\mathbf{x} - \mathbf{y})$$
(22)

$$[\hat{A}_{i}(t, \mathbf{x}), \hat{A}_{j}(t, \mathbf{y})] = [\hat{E}_{i}(t, \mathbf{x}), \hat{E}_{j}(t, \mathbf{y})] = 0$$
(23)

Transverse delta distribution:

$$\delta_{\perp,ij}^{(3)}(\mathbf{x} - \mathbf{y}) = \left(\delta_{ij} - \frac{\partial_i \partial_j}{\partial^2}\right) \delta^{(3)}(\mathbf{x} - \mathbf{y}) = \int \frac{\mathrm{d}^3 p}{(2\pi)^3} \left(\delta_{ij} - \frac{p_i p_j}{\mathbf{p}^2}\right) e^{i\mathbf{p} \cdot \mathbf{x}}$$
(24)

Maxwell field operators

Negative-frequency Maxwell field operator

$$\mathbf{A}^{(-)}(t,\mathbf{x}) = \sum_{\lambda=1,2} \int \frac{\mathrm{d}^3 p}{(2\pi)^3 \sqrt{2\omega(\mathbf{p})}} \hat{a}_{\lambda}(\mathbf{p}) \hat{\mathbf{e}}_{\lambda}(\mathbf{p}) e^{-i\omega(\mathbf{p})t + i\mathbf{p}\cdot\mathbf{x}}$$
(25)

Negative-frequency electric field operator

$$\mathbf{E}^{(-)}(t,\mathbf{x}) = -i\sum_{\lambda=1,2} \int \frac{\mathrm{d}^3 p}{(2\pi)^3 \sqrt{2\omega(\mathbf{p})}} \omega(\mathbf{p}) \hat{a}_{\lambda}(\mathbf{p}) \hat{\mathbf{e}}_{\lambda}(\mathbf{p}) e^{-i\omega(\mathbf{p})t + i\mathbf{p} \cdot \mathbf{x}}$$
(26)

Hamiltonian and momentum operator

$$\hat{H} = \sum_{\lambda=1,2} \int \frac{\mathrm{d}^3 p}{(2\pi)^3} \omega(\mathbf{p}) \hat{a}_{\lambda}^{\dagger}(\mathbf{p}) \hat{a}_{\lambda}(\mathbf{p}) \qquad \hat{\mathbf{P}} = \sum_{\lambda=1,2} \int \frac{\mathrm{d}^3 p}{(2\pi)^3} \mathbf{p} \hat{a}_{\lambda}^{\dagger}(\mathbf{p}) \hat{a}_{\lambda}(\mathbf{p})$$
(27)

Axiomatic particle state construction

Momentum state

Vacuum state

Algebraic commutation relations:

Poincare invariance:

$$[\hat{N}, \hat{a}^{\dagger}(\mathbf{p})] = \hat{a}^{\dagger}(\mathbf{p}) \tag{31}$$

$$\hat{U}(a,\Lambda)|0\rangle = |0\rangle$$

$$[\hat{\mathbf{P}}, \hat{a}^{\dagger}(\mathbf{p})] = \mathbf{p}\hat{a}^{\dagger}(\mathbf{p}) \tag{32}$$

Implies zero energy and momentum: Implies eigenstates:

$$\hat{H}|0\rangle = 0|0\rangle$$

$$\hat{H}|0\rangle = 0|0\rangle$$
 $\hat{\mathbf{P}}|0\rangle = \mathbf{0}|0\rangle$

$$\hat{N}\hat{a}^{\dagger}(\mathbf{p})|0\rangle = 1\hat{a}^{\dagger}(\mathbf{p})|0\rangle \tag{33}$$

(34)

Implies annihilation operator destroying vacuum:

$$\hat{a}(\mathbf{p})|0\rangle = 0$$

$$\langle 0|\hat{a}(\mathbf{p})\hat{a}^{\dagger}(\mathbf{p})|0\rangle = (2\pi)^3 \delta^{(3)}(\mathbf{p} - \mathbf{q})$$
 (3)

 $\hat{\mathbf{P}}\hat{a}^{\dagger}(\mathbf{p})|0\rangle = \mathbf{p}\hat{a}^{\dagger}(\mathbf{p})|0\rangle$

Smeared single-particle state

Interpretation as quantum operators as distributions:

$$\hat{A}^{(+)}[f]|0\rangle = \int d^4x f(t, \mathbf{x}) \hat{A}^{(+)}(t, \mathbf{x}) = \int \frac{d^3p}{(2\pi)^3} \frac{f(\omega(\mathbf{p}), \mathbf{p})}{\sqrt{2\omega(\mathbf{p})}}$$
(36)

Normalizable iff:

$$\langle 0|\hat{A}^{(-)}[f]\hat{A}^{(+)}[f]|0\rangle = \int \frac{\mathrm{d}^3 p}{(2\pi)^3} \left| \frac{f(\omega(\mathbf{p}), \mathbf{p})}{\sqrt{2\omega(\mathbf{p})}} \right|^2 = 1$$
 (37)

Implies eigenstates:

$$\hat{N}\hat{A}^{(+)}[f]|0\rangle = 1\hat{A}^{(+)}[f]|0\rangle \qquad \hat{H}\hat{A}^{(+)}[f]|0\rangle = \int \frac{\mathrm{d}^{3}p}{(2\pi)^{3}}\omega(\mathbf{p}) \left| \frac{f(\omega(\mathbf{p}),\mathbf{p})}{\sqrt{2\omega(\mathbf{p})}} \right|^{2} \hat{A}^{(+)}[f]|0\rangle$$
(38)

Generalized number and coherent state

Number state

$$|n_f\rangle = \frac{1}{\sqrt{n!}}\hat{A}^{(+)}[f]^n|0\rangle$$

Mean energy:

$$n \int \frac{\mathrm{d}^3 p}{(2\pi)^3} \omega(\mathbf{p}) \left| \frac{f(\omega(\mathbf{p}), \mathbf{p})}{\sqrt{2\omega(\mathbf{p})}} \right|^2$$

Electric field at (t, \mathbf{x}) :

$$0 \pm \frac{1}{2} \int \frac{\mathrm{d}^3 p}{(2\pi)^3} \omega(\mathbf{p}) + n|\Psi(t, \mathbf{x})|^2$$
 (41)

Coherent state

(39)
$$|\alpha\rangle = \exp\left\{\hat{A}^{(+)}[\alpha] - \text{H.c.}\right\}|0\rangle$$

Mean energy:

(40)
$$\int \frac{\mathrm{d}^3 p}{(2\pi)^3} \omega(\mathbf{p}) \left| \frac{\alpha(\omega(\mathbf{p}), \mathbf{p})}{\sqrt{2\omega(\mathbf{p})}} \right|^2 \qquad (46)$$

(42)

Electric field at (t, \mathbf{x}) :

$$0 \pm \frac{1}{2} \int \frac{\mathrm{d}^3 p}{(2\pi)^3} \omega(\mathbf{p}) \tag{44}$$

Connection to continuous-mode quantum optics

Central assumption

No relative motion, i.e., no Lorentz boosts.

Typical approximations:

- Neglect polarization degrees of freedom
- Neglect transverse momentum distribution
- Neglect relativistic Lorentz factor $1/\sqrt{2\omega}$

Phase-rotated quadrature operator:

$$\hat{X}(t) = \frac{1}{\sqrt{2}} \int \frac{d\omega}{2\pi} \left\{ \hat{a}(\omega) e^{-i(\omega t + \vartheta)} + \text{H.c.} \right\}$$
(45)

Maxwell field at (t, \mathbf{x}) :

$$\hat{A}(t, \mathbf{x}) = \int \frac{d\omega}{2\pi\sqrt{2\omega}} \left\{ \hat{a}(\omega)e^{-i\omega(t-x)} + \text{H.c.} \right\}$$
(46)