

Graph Signal Processed

Saturday, January 21, 2023 6:14 PM

Key ingredients of GSP :-
the mathematics and formulation.

concept of Shift in DSP :- then developing
a corresponding notion of shift in GSP.

time signals \rightarrow DSP
Signal samples \rightarrow GSP
indexed by nodes
of a Graph

At high level DSP and GSP study -

- ① Signals and their representation
- ② Systems that process signals
filters.
- ③ Signal transforms
 Z transform
Fourier transform
- ④ Sampling of signals
etc.

s_n , $n = 0, 1, \dots, N-1$.

We restrict ourselves $\rightarrow N$ samples.

finite impulse response FIR. +

$s = \{ s_n : n = 0, 1, \dots, N-1 \}$ \leftarrow time sample.
 $\downarrow Z$ transform

\downarrow Z transform

$S(z) \leftarrow$ ordered set of time samples.

$z^{-1} \leftarrow$ Shift / delay

$$\begin{aligned} S(z) &= \sum_{n=0}^{N-1} s_n z^{-n} \\ &= s_0 z^{-0} + s_1 z^{-1} + \dots + s_{N-1} z^{-(N-1)} \\ &\Rightarrow s_0 + s_1 z^{-1} + \dots + s_{N-1} z^{-(N-1)} \end{aligned}$$

Formal polynomial representation of the signal
that is useful in studying how signals are
processed by filters.

Given $S(z)$ we can recover the signal.

Discrete FT :-

DFT of the signal s is $\hat{s} = \{\hat{s}_k : k = 0, \dots, N-1\}$

$$\hat{s}_k = \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} s_n e^{-j \frac{2\pi}{N} kn}$$

$$\frac{2\pi k}{N} = \omega_k \leftarrow \text{discrete frequencies.}$$

$$k = 0, 1, \dots, N-1$$

$$\left\{ \begin{array}{l} x_k[n] = \frac{1}{\sqrt{N}} e^{-j \frac{2\pi k n}{N}} \\ \text{--- signals} \\ \text{--- Spectral components.} \end{array} \right. \quad \begin{array}{l} n=0,1,\dots,N-1 \\ k=0 \\ N-1 \end{array}$$

Inverse DFT :-

$$S_n = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} \hat{s}_k e^{j \frac{2\pi k n}{N}}$$

$$n=0,1,\dots,N-1$$

If $H(z)$ is an FIR filter, it can be represented

as,

$$h(z) = \sum_{m=0}^{N-1} h_m z^{-m}$$

$$S_{\text{out}}(z) = h(z) \cdot S_{\text{in}}(z)$$

$S_{\text{out}}(z)$ could result in polynomial in z^{-1} of degree greater than $N-1$, \leftarrow Need to consider boundary condition.

Leveraging the periodicity,

$$S_n = S_{n \bmod N}$$

Shift filter :- $h_{Shift}(z) = z^{-1}$

applying it to a signal $S_n = (S_0, S_1, \dots, S_{N-1})$
gives an output

$$\begin{aligned} S_{out} &= h_{Shift} \cdot S_{in} \\ &= (S_{N-1}, S_0, S_1, \dots, S_{N-2}) \end{aligned}$$

Shift Invariance :-

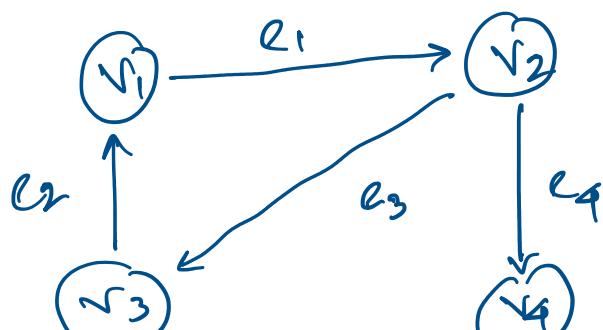
$$z^{-1} \cdot h(z) = h(z) \cdot z^{-1}$$

series combination of the filters are commutative.

$$h * S_{out} = (S_{N-1}, S_0, S_1, \dots, S_{N-2})$$

$$(h_0, h_1, \dots, h_N)$$

∇ = Incidence matrix



$$\nabla = \begin{bmatrix} -1 & 1 & 0 & 0 \\ 1 & 0 & -1 & 0 \\ 0 & -1 & 1 & 0 \\ 0 & -1 & 0 & 1 \end{bmatrix}$$

$$\nabla f(v_{ij}) = f(v_j) - f(v_i)$$

$$\begin{bmatrix} f(2) - f(1) \\ f(1) - f(3) \\ f(3) - f(2) \\ f(4) - f(2) \end{bmatrix}$$

$$\nabla^T \nabla = L \quad \text{Laplacian}$$

$$\begin{bmatrix} -1 & 1 & 0 & 0 \\ 1 & 0 & -1 & -1 \\ 0 & -1 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix} \quad \begin{bmatrix} -1 & 1 & 0 & 0 \\ 1 & 0 & -1 & 0 \\ 0 & -1 & 1 & 0 \\ 0 & -1 & 0 & 1 \end{bmatrix}$$

$$\Delta = \begin{bmatrix} 2 & -1 & -1 & 0 \\ -1 & 3 & -1 & -1 \\ -1 & -1 & 2 & 0 \\ 0 & -1 & 0 & 1 \end{bmatrix}$$

$$= D - A$$

D = degree matrix

A = Adjacency matrix

Graph Lasso Loss Function Derivation

Friday, March 10, 2023 12:47 PM

The Graph Lasso Loss function is written by

$$\hat{\Theta} = \underset{\Theta}{\operatorname{argmin}} \operatorname{tr}(S(\Theta)) - \log |\Theta| + \lambda \|\Theta\|_1$$

one should always demean the data before applying graphical Lasso.

Reason Graph lasso is all about covariance. the concept of which is 'de-meaned'.

Mathematical proof of the Loss function.

Consider the density of the Multivariate Gaussian

$$f(x) = \frac{1}{(2\pi)^{d/2} |\Sigma|^{1/2}} \exp\left(-\frac{1}{2} (x-\mu)^T \Sigma^{-1} (x-\mu)\right)$$

If we convert it to likelihood, $z_i \sim N(\mu, \Sigma)$

$$\hat{\Theta} = \underset{\Theta}{\operatorname{argmax}} \log f(x_1, \dots, x_n | \Theta)$$

$$= \underset{\Theta}{\operatorname{argmax}} \log \prod_{i=1}^n f(x_i | \Theta)$$

$$= \underset{\Theta}{\operatorname{argmax}} \sum_{i=1}^n \log f(x_i | \Theta)$$

$$= \underset{\Theta}{\operatorname{argmax}} \sum_{i=1}^n \log \left[\frac{1}{(2\pi)^{d/2} |\Sigma|^{1/2}} \exp\left(-\frac{1}{2} (x_i - \mu)^T \Theta (x_i - \mu)\right) \right]$$

$$= \underset{\Theta}{\operatorname{argmax}} \sum_{i=1}^n \log \left[\frac{1}{(2\pi)^{d/2} |\Sigma|^{1/2}} \exp \left(-\frac{1}{2}(x_i - \mu)^\top \Theta^{-1} (x_i - \mu) \right) \right]$$

$$= \underset{\Theta}{\operatorname{argmax}} \sum_{i=1}^n \frac{1}{2} \log |\Sigma| + \left[-\frac{1}{2} (x_i - \mu)^\top \Theta (x_i - \mu) \right]$$

$$= \underset{\Theta}{\operatorname{argmax}} \frac{n}{2} \log |\Theta| - \frac{1}{2} \sum_{i=1}^n (x_i - \mu)^\top \Theta (x_i - \mu)$$

$$\hat{\Sigma} = \frac{1}{n} \sum_{i=1}^n (x_i - \bar{x})(x_i - \bar{x})^\top \quad \text{sample cov.}$$

$$= \underset{\Theta}{\operatorname{argmax}} \frac{n}{2} \log |\Theta| - \frac{n}{2} \operatorname{trace}(x^\top \Theta x)$$

$$- \frac{n}{2} \operatorname{trace}(x^\top x \Theta)$$

$$- \frac{n}{2} \operatorname{trace}(S \Theta)$$



Optimization Essentials

Tuesday, March 14, 2023 2:38 PM

In most ML scenarios, we care about some performance measure P , (test set)

We optimize P indirectly

We reduce a different cost function $J(\theta)$ and hope that optimizing $J(\theta)$ will help us to improve P .

$$J(\theta) = \underset{\substack{x, y \sim \hat{P}_{\text{data}} \\ \text{training set}}}{\mathbb{E}} L(f(x|\theta), y)$$

But we are interested to minimize the cost function for the entire population.

like generalization.

But we don't know $P_{\text{data}}(x,y)$ we have
 $\hat{P}_{\text{data}}(x,y) \leftarrow \text{samples}$.

then we replace the true dist by empirical

distribution. and minimize the Empirical Risk.

Small batch sizes can offer a regularizing effect

- due to the noise they add to the Learning process.

generalization error is often best for batch size 1.

- Small learning rate is required if the batch size is very small.
 - High variance of the gradient estimate-

Q Why minibatches are sampled randomly?

- for computing unbiased estimates it is important it is required that the samples are indep
- subsequent gradient estimates are also indep from each other.

In the context of NN training,

- Ⓐ We don't care about finding the exact min.
- Ⓑ Only want its value to be reduced sufficiently to obtain a Good generalization error.

SGD :- It is possible to obtain an unbiased estimate of the gradient by taking the average gradient on a minibatch of m examples drawn iid from the data generating dist.

Shifts in GSP

Saturday, January 21, 2023 9:44 PM

Graph Signals:- Signals whose samples are indexed by the nodes of arbitrary graphs.

$$S = (s_0, s_1, \dots, s_{N-1})$$

$$S = [s_0 \ s_1 \ \dots \ s_{N-1}]^T \in \mathbb{C}^N$$

filter h is represented by Matrix H

$$S_{\text{out}} = H \cdot S_{\text{in}}$$

→ experiment if possible.
for one example

Not only summary, But
What's new in that domain?

Experimental result if it
is not good.

Criticize the paper →

- reproducibility →
- experimental details.

Presentation :-

very basic + topics in the paper.

Appendix slides.

Survey → Break down
on topics.

How it is useful ?

Clear Basics first. ⇒



check the references
all references.

$$Z^T$$

A diagram showing a person holding a book labeled 'A'. To the right is a head labeled 'v'. Below them is a large arrow pointing down to a line labeled 'c n x n v.'. A smaller arrow points from the 'v' to the 'n x n' part. At the end of the arrow is a circle containing a '1'. To the right of the arrow is a circle containing a '0'. Below the arrow is a circle containing a '2'.

$$A v = c \underset{1}{\cancel{n \times n}} v.$$
$$\approx 0$$

List of Preparations.

Monday, January 30, 2023 7:49 PM

Proof

- ① In a symmetric matrix, all the eigen values are real and eigen vectors are also real and form an ortho normal Basis.
 - ② For real symmetric matrix, the AM is equal to geometric multiplicity.
 - ③ A symmetric matrix is always diagonalizable.
 - ④ If $\lambda_i \neq \lambda_j$ then $s_i \cdot s_j^* = 0$
 - ⑤ Why does Laplacian matrix always have a 0 eigen value
 - ⑥ If the characteristic poly $P_A(z)$ and the minimal poly $m_A(z)$ of A are equal then every filter commuting with A is polynomial in A.

$$H = \mathcal{U}(A)$$

GSP paper page - 5.

Q

⑦ $\deg(e(z)) = \deg(p_A(z)) \leq N-1$

Shift invariant filters are polynomials with degree at most $\deg(m_A(z))$.

⑧ How to do symmetric normalization of Laplacian?

⑨ The roles of time shift in DSF.

all the details of the shift operator.

⑩ Proof of convolution.

⑪ Discrete FT

⑫ IFT

⑬ What is FIR.

⑭ 6-7 page of GSP paper.

⑮ Positive semi definite matrix has all eigen values real and non-negative and a full set of orthogonal

vectors can be obtained

$$L = U \Lambda U^T \rightarrow LU$$

decomp

$U \rightarrow GFT$ matrix.

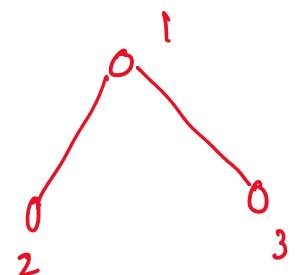
$$U^T = U^{-1}$$

SVD

eigen decom..

Proof of Laplacian is positive semi-definite

$$L = D - A = B B^T$$



(18) Study about SVD, eigen decom,

LU decomposition,

Rayleigh quotient.

$$A = \begin{bmatrix} 0 & 1 & 1 \\ 1 & 0 & 0 \\ 1 & 0 & 0 \end{bmatrix}$$

(16) polynomial approximation $D = \begin{bmatrix} 2 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$

with chebyshev's filter.

$$L = \begin{bmatrix} 2 & -1 & -1 \\ -1 & 1 & 0 \\ -1 & 0 & 1 \end{bmatrix}$$

$$x^T L x \rightarrow L = U \Lambda U^T$$

$$U^T L U = \Lambda.$$

$$\overbrace{T^T(U_k)}^{\cong} = \lambda_k$$

Q17 Why faithfulness assumption is required for obtaining Graph from data ??

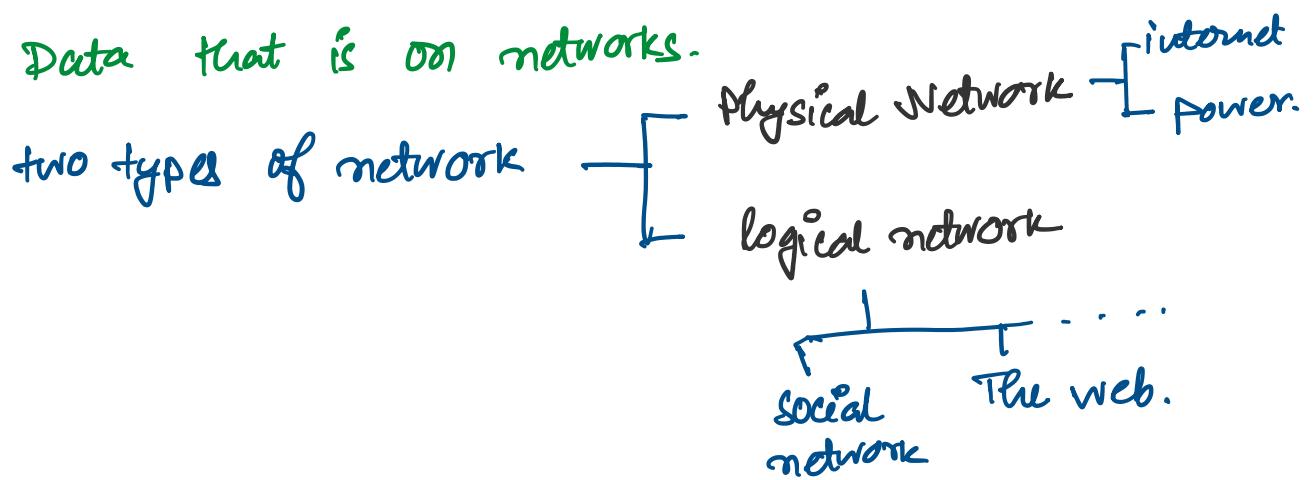
Graph Signal Processing Lecture.

Wednesday, February 1, 2023

2:32 PM

- ① Graph Frequency
 - ② Graph Transforms and Filters.
 - ③ Sampling of Graph Signals.
 - ④ Learning graphs from data

Data that is on networks.



How do we take into the structure of these datasets (information encoded in the social network who is connected to whom) to make sense of the information that is available in the network.

Example :-

Relative positions of sensors , temperature

Question :- Does temperature vary smoothly ?

do you see anomalies? or diff behaviour
that is not expected.

Social network :- Information about the connection (age, income)

connection (age, income)

PyGSP

the graph Laplacian

$$L = D - A$$

how do we go from
this notation to frequency notation?

Degree- Adjacency.

$$L = D - A = U \Lambda U^t$$

Eigen vectors of $L = U = \{U_k\}_{k=1:n}$

Eigen values of L

Diagonal terms of Λ .

Eigen pair System

(λ_k, U_k)

= Fourier basis of the
graph signals.

this is called Graph Fourier Transform.

Fourier may not be happy about this. \rightarrow
But there is an analogy. between the
Fourier transform frequencies with the
graph eigen values and vectors.

Why GFT?

Because it captures the notion of frequency

Because it captures the notion of frequency

Finding eigen values and eigen vectors.

$$\lambda_k = \min_{\substack{u \perp u_1, \dots, u_{k-1}}} \frac{u^T L u}{u^T u}$$

Solving an optimization which involves the Rayleigh quotient. So, if we have

Selected eigenvectors 1 through $k-1$, we are looking for an eigenvector that is orthogonal to all the prev. ones and minimizing the Rayleigh coefficient.

L = Laplacian \leftarrow derived from the graph.

$$u_i = 1$$

$$\lambda_1 = 0$$

$$u^T L u = \sum_{i \neq j} w_{ij} (u_i - u_j)^2$$

u_i, u_j are the values of the eigenvectors at two nodes i and j .

Interpretation:- Basis vectors ordered by increasing variation on the graph.

If $\lambda = 0$ then all the eigenvectors are

$$\begin{bmatrix} 1 \\ 1 \\ \vdots \\ 1 \end{bmatrix}$$

If $\lambda = 0$ then all the eigenvectors are

$$\begin{bmatrix} 1 \\ 1 \\ \vdots \\ 1 \end{bmatrix}$$

$u^T L u \leftarrow$ measures the variation of a signal in a graph.

minimum variation signal is 1.

there is no variation, all the signals are same.

Interpret

then when find the next eigen vectors while they are orthogonal to 1 ...

we are finding signals that have increased variation while orthogonal to the previous one.

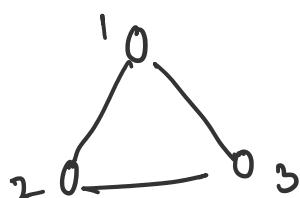
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FB example.

open questions about the good definition of frequency? (Find those open questions)

low frequency - low variation on the graph.

high frequency - high / rapid variation on the graph



$$L = \begin{bmatrix} 2 & -1 & -1 \\ -1 & 2 & -1 \\ -1 & -1 & 2 \end{bmatrix}$$

$$\begin{vmatrix} 2-\lambda & -1 & -1 \\ -1 & 2-\lambda & -1 \\ -1 & -1 & 2-\lambda \end{vmatrix} = 0 \quad \lambda = 0, 3, 3$$

Learning Graph.

Sunday, February 26, 2023 1:51 PM

Stationary Graph signal.

A graph signal x is stationary w.r.t. the shift s iff $x = Hw$ where $H = \sum_{i=0}^{l-1} h_i s^i$ and w is white.

The covariance matrix C_x of the stationary signal x is a polynomial on s .

$$\begin{aligned} C_x &= \mathbb{E}[x x^T] = \mathbb{E}[Hw(Hw)^T] \\ &= H \mathbb{E}[ww^T] H^T \\ &= HH^T = H^2 \\ &= \left(\sum_{i=0}^{l-1} h_i s^i \right)^2 \end{aligned}$$

$$C_x s = s C_x \quad \text{and} \quad \text{eigenvects}(C_x) = \text{eigenvects}(s)$$

Learning a Graph from Nodal Observations.

Given a collection $x := [x_1, \dots, x_p] \in \mathbb{R}^{N \times p}$ of graph signal observations supported on the unknown graph $G(V, E, w)$ find an optimal s .

\uparrow
shift.

Most classical approaches focus on pairwise similarities

$$\text{sim}(i, j) = f(x_i, x_j)$$

(i) In classical statistics the $\text{sim}(i, j)$ is

① In classical statistics the $\text{Sim}(i,j)$ is defined as the correlation coefficient between the nodes.

$$\text{Sim}(i,j) = \rho_{ij} = \frac{\text{cov}[x_i, x_j]}{\sqrt{\text{var}[x_i] \text{var}[x_j]}}$$

then If we really want to know if $\text{Sim}(i,j) = 0$ or not, we can do it by performing statistical tests.

$$\hat{\rho}_{ij} = \frac{\hat{c}_{ij}}{\sqrt{\hat{c}_{ii} \hat{c}_{jj}}}$$

$$\text{Edge exist if } 0.5 \log\left(\frac{1 + \hat{\rho}_{ij}}{1 - \hat{\rho}_{ij}}\right) > \frac{z_{\alpha/2}}{\sqrt{p-3}}$$

α = false alarm rate.

The main idea is to obtain the covariance matrix and then sparsify it.

$$y = ax$$

$$\hat{y} = \frac{\text{cov}(x, y)}{\text{var}(x)} x$$

$$\text{So, } y - \hat{y} = ax - \frac{\text{cov}(x, y)}{\text{var}(x)} x.$$

$$= \left[a - \frac{\text{cov}(x, y)}{\text{var}(x)} \right] x$$

Idea of the partial correlation :-

For any 2 variables x_1 and x_2 , we try to predict x_1 and x_2 with all other variables except x_1 and x_2 . Then obtain the residual.

If the residual is uncorrelated, or correlated we can say if there is any direct influence from x_1 to x_2 or not.

For Partial correlation we look into the precision matrix we try to sparsify the precision matrix.

Learning Graph From Smooth Signals.

Given observations $\mathbf{x} = [x_1, \dots, x_p] \in \mathbb{R}^{N \times p}$, identify a graph G such that signals in \mathbf{x} are smooth on G .

Dirichlet energy on the Graph G with Laplacean L is given by $E = \frac{1}{2} \mathbf{x}^T L \mathbf{x}$.

Distributed energy on the support -

Search of the shift $S = L$ such that

$$TV(x) = x^T L x \stackrel{P}{\text{is small.}}$$

$$TV(x) = \sum_{P=1}^P x_P^T L x_P.$$

$$L^* = \underset{L}{\operatorname{argmin}} \left\{ \sum_{P=1}^P x_P^T L x_P + \frac{P}{2} \|L\|_F^2 \right\}$$

$$\text{subject to } \operatorname{trace}(L) = N \quad ?$$

$$L \mathbf{1} = \mathbf{0}$$

$$L_{ij} = L_{ji} \leq 0 \quad i \neq j$$

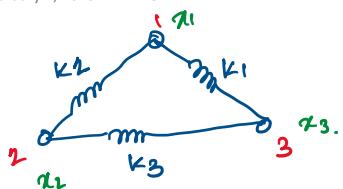
Some questions about the GroN DAG paper.

Monday, February 27, 2023 12:46 AM

- ① What different Acyclicity constraint they can take account on?
- ② Why faithfulness assumption is not required here? — Resit
- ③ to EMRE. Why adding Edges doesn't Reduce the Negative log likelihood?

Spectral Graph Theory (power systems perspective)

Thursday, February 2, 2023 11:54 AM



$$x_3 = \frac{1}{d_3} \sum_{j \in NEE} x_j$$

$$= \frac{1}{d_3} (x_1 + x_2)$$

$$= \frac{x_1 + x_2}{d_3}$$

what is d_i ?

d_i : degree of node(i)

$$x_1 = 10 \quad x_2 = 20$$

$$x_3 = 15$$

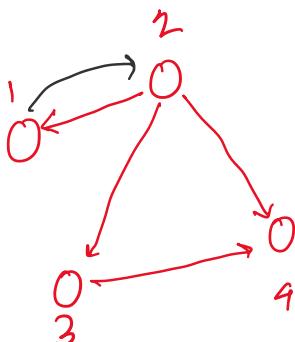
Q Aim it to check, If Graph FT can provide any information about the cascading failure.

$$G = (V, E) \quad V = \{v_1, \dots, v_n\}$$

$$A_G = \{a_{ij}\}$$

$$k = 1$$

	1	2	3	4
1	0	0	0	0
2	1	0	1	1
3	0	0	0	1
4	0	0	0	0
	1	2	3	4



$$A^2 = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix}$$

$$A^3 = \begin{bmatrix} 1 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 1 & 0 & 1 & 1 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \end{bmatrix}$$

$$A^B = \begin{bmatrix} 0 & v & v & v & \downarrow & L \\ 0 & 1 & 0 & 1 \\ 1 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}$$

$$\operatorname{tr}(I - B)^{-1} = d$$

$$\operatorname{tr} \sum_{k=1}^{\infty} B^k = 0$$

for acyclic Graphs.

$$\operatorname{tr}(I - B)^{-1} = \operatorname{tr} \sum_{k=0}^{\infty} B^k$$

$$= \operatorname{tr} \left[(I) + \sum_{k=1}^{\infty} B^k \right]$$

$$= d + 0$$

$$= d$$

If $\operatorname{eig}(B) < 1$ then this results hold true

Is automatically satisfied when B is DAG.

Fault: entries of B^k can easily exceed machine precision for even small values of d .

- lacks numerical stability -

$$\operatorname{tr}(e^B) \neq d$$

$$\text{no cycles} \Rightarrow (B^k)_{ii} = 0 \quad \forall k \geq 1$$

$$\begin{aligned}\text{tr}(e^B) &= \text{tr}\left(I + \frac{B}{1!} + \frac{B^2}{2!} + \dots\right)^{\infty} \\ &= d + \text{tr} \sum_{i=1}^{\infty} \frac{B^i}{i!} = 0.\end{aligned}$$

\rightarrow matrix exponential is well-defined for all square matrices.

nodes $d \uparrow \rightarrow$ # edges $\uparrow \Rightarrow$ # possible closed walks \uparrow

$\text{tr}(I - B)^{-1} \rightarrow$ rapidly becomes ill conditioned.

By reweighting the # of lengths - k closed walks by $\frac{k!}{k!}$ this becomes easier to manage.

Fault discrete space

Weighted Adjacency

If B is replaced by w (Weighted Adjacency)
this $\boxed{\text{tr } e^B = d}$ doesn't work

need some alternate criterion.

How even If w is nonnegative weight matrix this can be proved.

$$h(w) = \text{tr}(e^{w \odot w}) - d = 0$$

$$\Rightarrow h(w) = e^{(w \odot w)} \odot z(w)$$

\rightarrow counting weighted closed walks each edge weight is w_{ij}^2 .

$$h(w) > h(w')$$

meaning \rightarrow (a) w has more cycles than w'

(b) cycles in w are more heavily weighted than in w' .

Equality constrained Program : (ECP)

Plan Tomorrow.

Tuesday, February 28, 2023 12:05 AM

- ① Read the causal direction paper for GSP
- ② Is there any way to incorporate GSP techniques from causal discovery perspective?
- ③ Drawbacks in the Gaussian case.
GSP doesn't fall the functional assumption.
But it has a smoothness assumption.
- ④ Is there any relation between smoothness and causality?
- ⑤ How to incorporate this optimization procedure

— this optimization procedure
for cyclic causal discovery?

Constraint B sparse

and B needs to be
stable

$$\text{cig}(B) < 1$$

can I encode this
somehow with the continuous
optimization?

then // Gaussian - Non Gaussian
problem will be solved.

and // Row permutation
problem will also
be solved.

But can I obtain the
relationship from the
logistic regression?



Q

What does it mean by
regression in the case
for cyclic SEM.?