

LDO Regulator for RF and Analog Circuits - Ultra-Low **Noise and High PSRR**

450 mA

NCP161

The NCP161 is a linear regulator capable of supplying 450 mA output current. Designed to meet the requirements of RF and analog circuits, the NCP161 device provides low noise, high PSRR, low quiescent current, and very good load/line transients. The device is designed to work with a 1 µF input and a 1 µF output ceramic capacitor. It is available in two thickness ultra-small 0.35P, 0.64 mm x 0.64 mm Chip Scale Package (CSP) and XDFN4 0.65P, 1 mm x 1 mm.

Features

- Operating Input Voltage Range: 1.9 V to 5.5 V
- Available in Fixed Voltage Option: 1.8 V to 5.14 V
- ±2% Accuracy Over Load/Temperature
- Ultra Low Quiescent Current Typ. 18 μA
- Standby Current: Typ. 0.1 μA
- Very Low Dropout: 150 mV at 450 mA
- Ultra High PSRR: Typ. 98 dB at 20 mA, f = 1 kHz
- Ultra Low Noise: 10 μV_{RMS}
- Stable with a 1 µF Small Case Size Ceramic Capacitors
- Available in -WLCSP4 0.64 mm x 0.64 mm x 0.4 mm
 - -WLCSP4 0.64 mm x 0.64 mm x 0.33 mm
 - -XDFN4 1 mm x 1 mm x 0.4 mm
 - -SOT23-5 2.9 mm x 2.8 mm x 1.2 mm
- These Devices are Pb-Free and are RoHS Compliant

Typical Applications

- Battery-powered Equipment
- Wireless LAN Devices
- Smartphones, Tablets
- Cameras, DVRs, STB and Camcorders

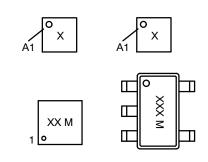






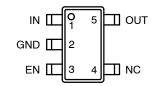
WLCSP4

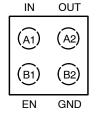
MARKING DIAGRAMS

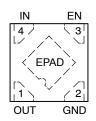


X, XX, XXX = Specific Device Code = Date Code

PIN CONNECTIONS (Top Views)







ORDERING INFORMATION

See detailed ordering and shipping information on page 17 of this data sheet.

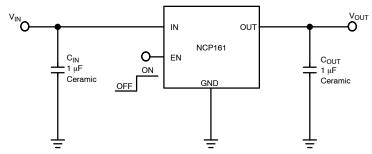


Figure 1. Typical Application Schematics

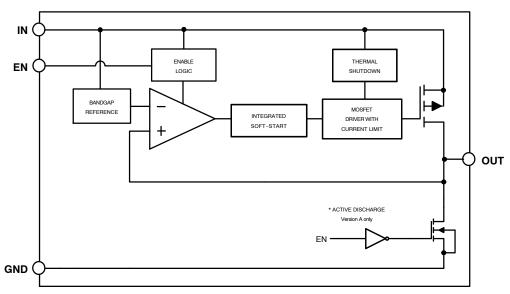


Figure 2. Simplified Schematic Block Diagram

PIN FUNCTION DESCRIPTION

Pin No. CSP4	Pin No. SOT23-5	Pin No. XDFN4	Pin Name	Description
A1	1	4	IN	Input voltage supply pin
A2	5	1	OUT	Regulated output voltage. The output should be bypassed with small 1 μF ceramic capacitor.
B1	3	3	EN	Chip enable: Applying V_{EN} < 0.4 V disables the regulator, Pulling V_{EN} > 1.2 V enables the LDO.
B2	2	2	GND	Common ground connection
_	_	EPAD	EPAD	Expose pad should be tied to ground plane for better power dissipation

ABSOLUTE MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Input Voltage (Note 1)	V _{IN}	-0.3 V to 6	V
Output Voltage	V _{OUT}	-0.3 to V _{IN} + 0.3, max. 6	V
Chip Enable Input	V_{CE}	-0.3 to 6	V
Output Short Circuit Duration	t _{SC}	unlimited	s
Maximum Junction Temperature	TJ	150	°C
Storage Temperature	T _{STG}	-55 to 150	°C
ESD Capability, Human Body Model (Note 2)	ESD _{HBM}	2000	V
ESD Capability, Machine Model (Note 2)	ESD _{MM}	200	V

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Refer to ELECTRICAL CHARACTERISTIS and APPLICATION INFORMATION for Safe Operating Area.

- This device series incorporates ESD protection and is tested by the following methods: ESD Human Body Model tested per EIA/JESD22–A114

 - ESD Machine Model tested per EIA/JESD22-A115
 Latchup Current Maximum Rating tested per JEDEC standard: JESD78.

THERMAL CHARACTERISTICS

Rating			Value	Unit
Thermal Characteristics, CSP4 (Note 3)	Thermal Resistance, Junction-to-Air		108	
Thermal Characteristics, XDFN4 (Note 3)	Thermal Resistance, Junction-to-Air	$R_{ hetaJA}$	198.1	°C/W
Thermal Characteristics, SOT23-5 (Note 3)	Thermal Resistance, Junction-to-Air		218	

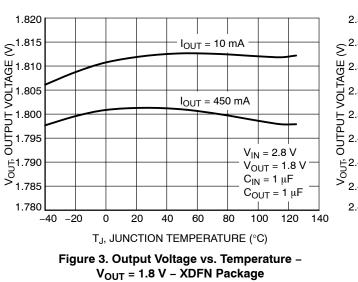
^{3.} Measured according to JEDEC board specification. Detailed description of the board can be found in JESD51-7

Parameter	arameter Test Conditions		Symbol	Min	Тур	Max	Unit
Operating Input Voltage			V_{IN}	1.9		5.5	V
Output Voltage Accuracy	$V_{IN} = V_{OUT(NOM)} + 1 V$ $0 \text{ mA} \le I_{OUT} \le 450 \text{ mA}$	WLCSP4, XDFN4	V _{OUT}	-2		+2	%
	V _{IN} = V _{OUT(NOM)} + 1 V SOT23-5		001	-2		+2	
Line Regulation	V _{OUT(NOM)} + 1 V	$V_{OUT(NOM)} + 1 V \le V_{IN} \le 5.5 V$			0.02		%/V
Load Regulation	I _{OUT} = 1 mA to 450 mA	WLCSP4, XDFN4	Line _{Reg}		0.001		2// 4
	WLCSP4, XDFN4	SOT23-5	Load _{Reg}		0.005	0.008	%/mA
Dropout Voltage (Note 5)	I _{OUT} = 450 mA	V _{OUT(NOM)} = 1.8 V			300	450	
	WLCSP4, XDFN4	V _{OUT(NOM)} = 1.85 V			290	393	
		V _{OUT(NOM)} = 2.5 V			190	315	1
		V _{OUT(NOM)} = 2.8 V			175	290	1
		V _{OUT(NOM)} = 2.85 V			170	290	
		V _{OUT(NOM)} = 3.0 V	V_{DO}		165	275	mV
		V _{OUT(NOM)} = 3.3 V			160	260	1
		V _{OUT(NOM)} = 3.5 V			150	255	1
		V _{OUT(NOM)} = 4.5 V			120	210	
		V _{OUT(NOM)} = 5.0 V			105	190	
		V _{OUT(NOM)} = 5.14 V			105	185	
Dropout Voltage (Note 5)	I _{OUT} = 450 mA	V _{OUT(NOM)} = 1.8 V			365	480	
	SOT23-5	V _{OUT(NOM)} = 2.8 V	\ \ \		260	345	mV
		V _{OUT(NOM)} = 3.0 V	V _{DO}		240	330	
		V _{OUT(NOM)} = 3.3 V			225	305	
Output Current Limit	V _{OUT} = 90% \	V _{OUT} = 90% V _{OUT(NOM)}		450	700		A
Short Circuit Current	V _{OUT} =	0 V	I _{SC}		690		mA
Quiescent Current	I _{OUT} = 0) mA	IQ		18	23	μΑ
Shutdown Current	$V_{EN} \le 0.4 \text{ V, } $	/ _{IN} = 4.8 V	I _{DIS}		0.01	1	μΑ
EN Pin Threshold Voltage	EN Input Vo	Itage "H"	V_{ENH}	1.2			\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \
	EN Input Vo	ltage "L"	V_{ENL}			0.4	٧
EN Pull Down Current	V _{EN} = 4	.8 V	I _{EN}		0.2	0.5	μΑ
Turn-On Time	C _{OUT} = 1 μF, From as V _{OUT} = 95% \	ssertion of V _{EN} to V _{OUT(NOM)}			120		μs
Power Supply Rejection Ratio	I _{OUT} = 20 mA	f = 100 Hz f = 1 kHz f = 10 kHz f = 100 kHz	PSRR		91 98 82 48		dB
Output Voltage Noise	f = 10 Hz to 100 kHz	I _{OUT} = 1 mA I _{OUT} = 250 mA	V _N		14 10		μV_{RMS}
Thermal Shutdown Threshold	Temperatu	re rising	T _{SDH}		160		°C
	Temperatur	e falling	T_{SDL}		140		°C
Active output discharge resistance	V _{EN} < 0.4 V, Ve	ersion A only	R _{DIS}		280		Ω
Line transient (Note 6)	V _{IN} = (V _{OUT(NOM)} + 1 V) to (V _{OUT(NOM)} + 1.6 V) in 30 μs, I _{OUT} = 1 mA V _{IN} = (V _{OUT(NOM)} + 1.6 V) to (V _{OUT(NOM)} + 1 V) in 30 μs, I _{OUT} = 1 mA		Tues	-1			\/
			Tran _{LINE}			+1	mV
Load transient (Note 6)	I _{OUT} = 1 mA to 45	50 mA in 10 μs	_	-40			
	I _{OUT} = 450 mA to 1mA in 10 μs		Tran _{LOAD}			+40	mV

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

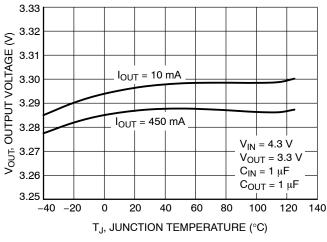
^{4.} Performance guaranteed over the indicated operating temperature range by design and/or characterization. Production tested at T_A = 25°C. Low duty cycle pulse techniques are used during the testing to maintain the junction temperature as close to ambient as possible.

Dropout voltage is characterized when V_{OUT} falls 100 mV below V_{OUT(NOM)}.
 Guaranteed by design.



2.520 (2.515 BB 2.510 2.505 2.500 2.495 2.495 $I_{OUT} = 10 \text{ mA}$ $I_{OUT} = 450 \text{ mA}$ $V_{IN} = 3.5 V$ 2.490 $V_{OUT} = 2.5 V$ C_{IN} = 1 μF 2.485 $C_{OUT} = 1 \mu F$ 2.480 -40 -20 0 20 40 60 80 100 120 140 T_J, JUNCTION TEMPERATURE (°C)

Figure 4. Output Voltage vs. Temperature – V_{OUT} = 2.5 V – XDFN Package



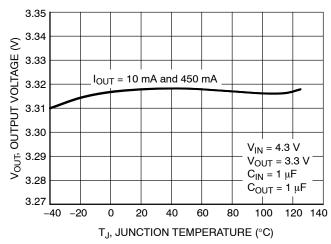
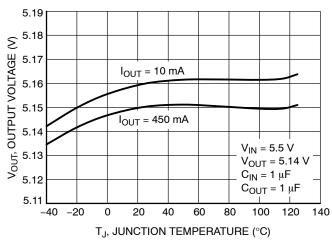


Figure 5. Output Voltage vs. Temperature – V_{OUT} = 3.3 V – XDFN Package

Figure 6. Output Voltage vs. Temperature – V_{OUT} = 3.3 V – CSP Package



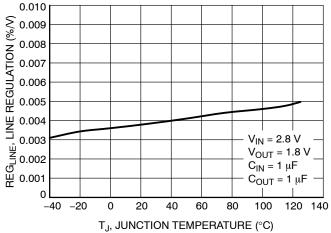


Figure 7. Output Voltage vs. Temperature – V_{OUT} = 5.14 V – XDFN Package

Figure 8. Line Regulation vs. Temperature – V_{OUT} = 1.8 V

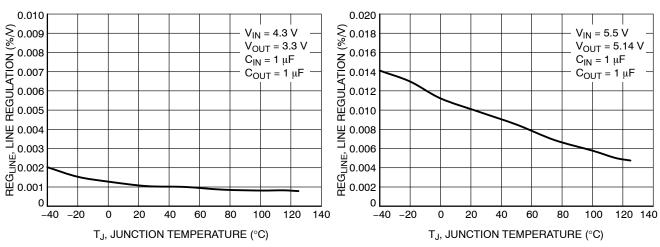


Figure 9. Line Regulation vs. Temperature - $V_{OUT} = 3.3 V$

Figure 10. Line Regulation vs. Temperature - $V_{OUT} = 5.14 V$

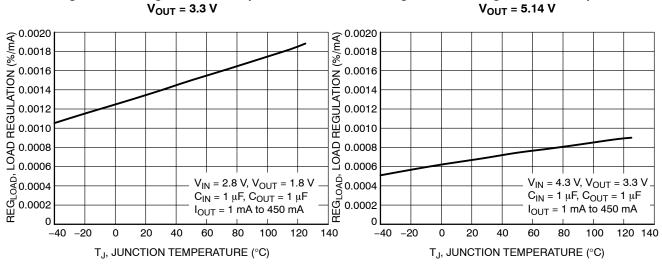


Figure 11. Load Regulation vs. Temperature - $V_{OUT} = 1.8 V (WLCSP4)$

Figure 12. Load Regulation vs. Temperature -V_{OUT} = 3.3 V (WLCSP4)

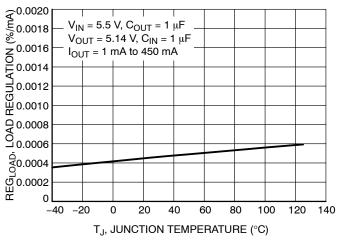


Figure 13. Load Regulation vs. Temperature -**V_{OUT}** = 5.14 **V** (WLCSP4)

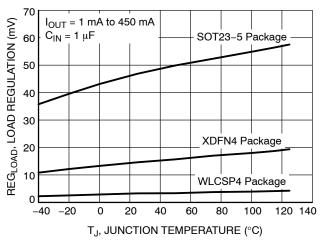


Figure 14. Load Regulation vs. Temperature

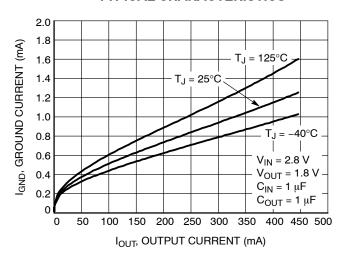


Figure 15. Ground Current vs. Load Current – $V_{OUT} = 1.8 \text{ V}$

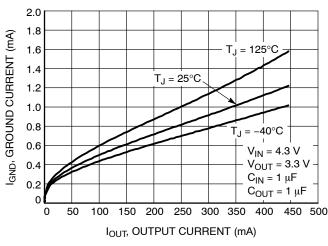


Figure 16. Ground Current vs. Load Current – $V_{OUT} = 3.3 \text{ V}$

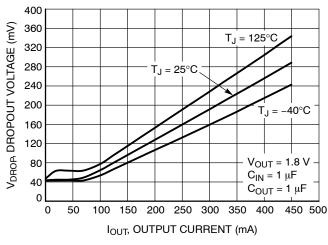


Figure 18. Dropout Voltage vs. Load Current – $V_{OUT} = 1.8 \text{ V}$

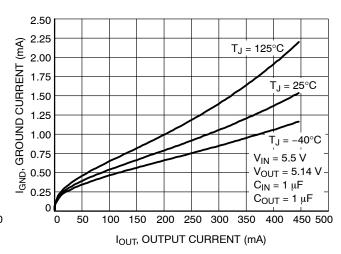


Figure 17. Ground Current vs. Load Current – $V_{OUT} = 5.14 \text{ V}$

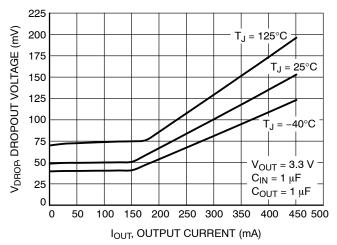


Figure 19. Dropout Voltage vs. Load Current – $V_{OUT} = 3.3 \text{ V}$

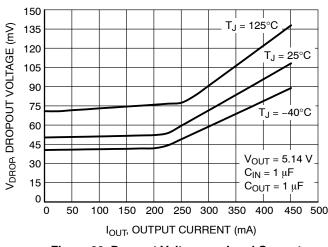


Figure 20. Dropout Voltage vs. Load Current – V_{OUT} = 5.14 V

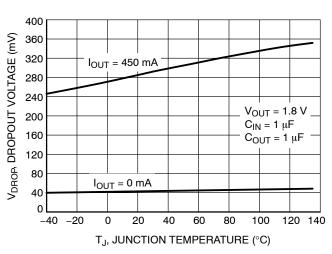


Figure 21. Dropout Voltage vs. Temperature– V_{OUT} = 1.8 V

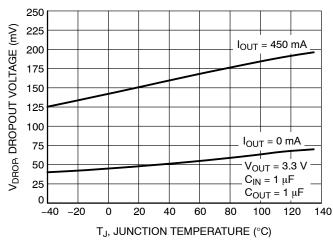


Figure 22. Dropout Voltage vs. Temperature– $V_{OUT} = 3.3 \text{ V}$

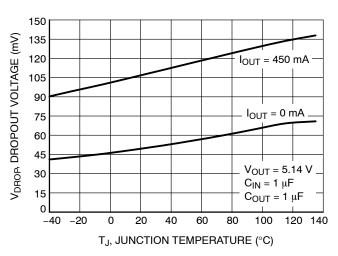


Figure 23. Dropout Voltage vs. Temperature– V_{OUT} = 5.14 V

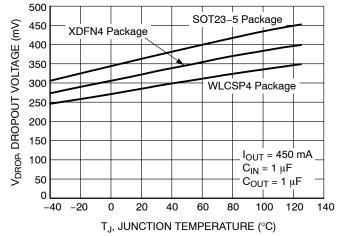
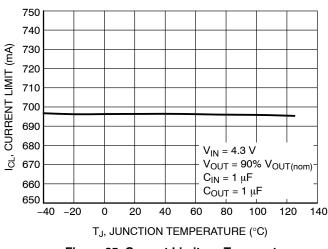


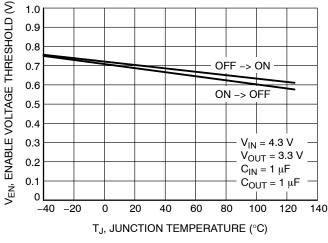
Figure 24. Dropout Voltage vs. Temperature $V_{OUT} = 1.8 \text{ V}$



700 690 I_{SC}, SHORT CIRCUIT CURRENT 680 670 660 650 640 $V_{IN} = 4.3 V$ 630 V_{OUT} = 0 V (Short) 620 $C_{IN} = 1 \mu F$ 610 $C_{OUT} = 1 \mu F$ 600 -40 -20 20 40 60 80 100 120 140 T_J, JUNCTION TEMPERATURE (°C)

Figure 25. Current Limit vs. Temperature

Figure 26. Short Circuit Current vs. Temperature



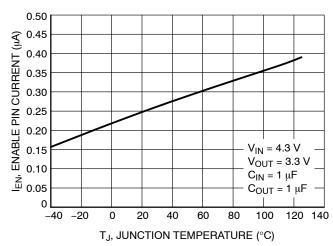
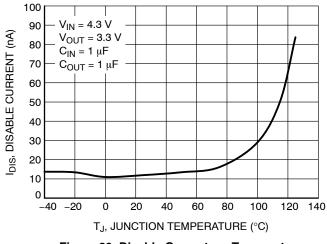


Figure 27. Enable Threshold Voltage vs. Temperature

Figure 28. Enable Current Temperature



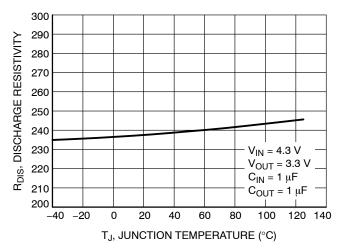
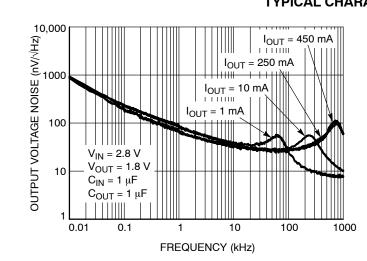


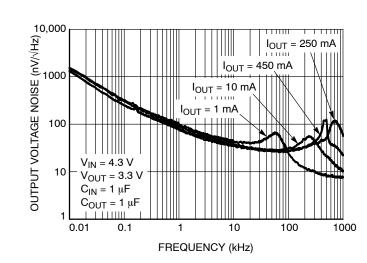
Figure 29. Disable Current vs. Temperature

Figure 30. Discharge Resistivity vs. Temperature



	RMS Output Noise (μV)			
l _{OUT}	10 Hz – 100 kHz	100 Hz – 100 kHz		
1 mA	14.62	14.10		
10 mA	11.12	10.48		
250 mA	10.37	9.82		
450 mA	10.22	9.62		

Figure 31. Output Voltage Noise Spectral Density – V_{OUT} = 1.8 V



	RMS Output Noise (μV)			
l _{out}	10 Hz – 100 kHz	100 Hz – 100 kHz		
1 mA	16.9	15.79		
10 mA	12.64	11.13		
250 mA	11.96	10.64		
450 mA	11.50	10.40		

Figure 32. Output Voltage Noise Spectral Density – V_{OUT} = 3.3 V

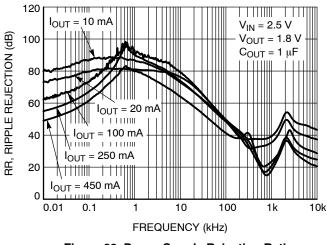


Figure 33. Power Supply Rejection Ratio, V_{OUT} = 1.8 V

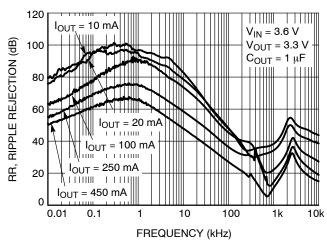


Figure 34. Power Supply Rejection Ratio, V_{OUT} = 3.3 V

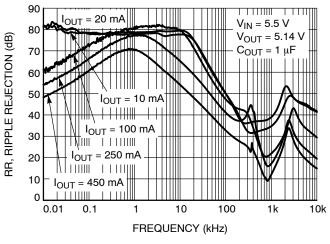


Figure 35. Power Supply Rejection Ratio, V_{OUT} = 5.14 V

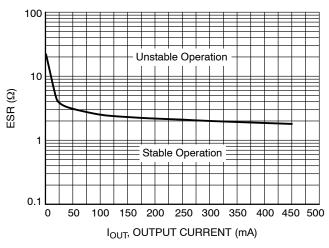


Figure 36. Stability vs. ESR

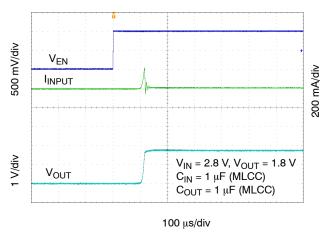


Figure 37. Enable Turn-on Response – C_{OUT} = 1 μ F, I_{OUT} = 10 mA

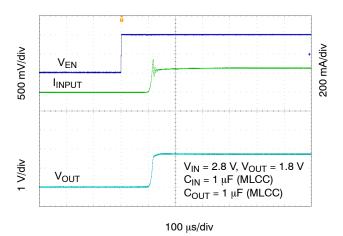


Figure 38. Enable Turn-on Response – C_{OUT} = 1 μ F, I_{OUT} = 250 mA

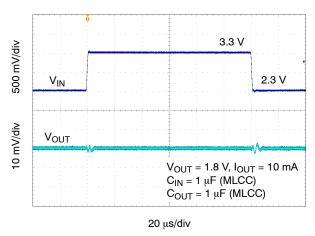


Figure 39. Line Transient Response – $V_{OUT} = 1.8 \text{ V}$

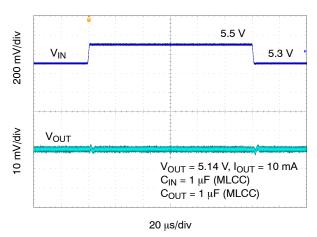


Figure 41. Line Transient Response – V_{OUT} = 5.14 V

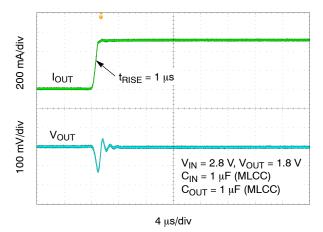


Figure 43. Load Transient Response – 1 mA to 450 mA – V_{OUT} = 1.8 V

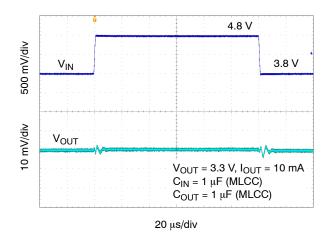


Figure 40. Line Transient Response – $V_{OUT} = 3.3 \text{ V}$

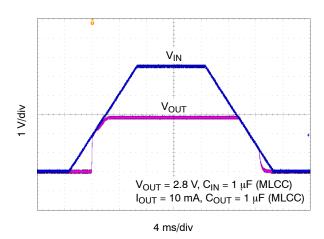


Figure 42. Turn-on/off - Slow Rising V_{IN}

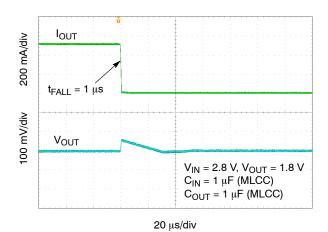


Figure 44. Load Transient Response – 450 mA to 1 mA – V_{OUT} = 1.8 V

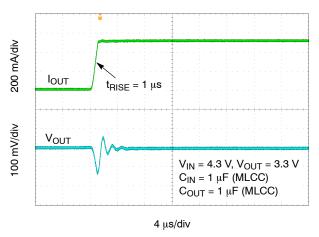


Figure 45. Load Transient Response – 1 mA to 450 mA – V_{OUT} = 3.3 V

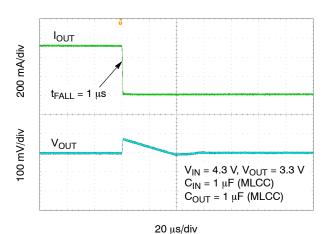


Figure 46. Load Transient Response – 450 mA to 1 mA – V_{OUT} = 3.3 V

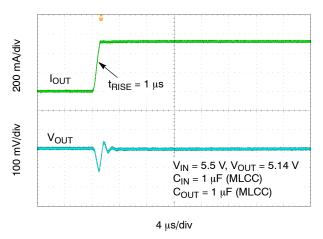


Figure 47. Load Transient Response – 1 mA to 450 mA – V_{OUT} = 5.14 V

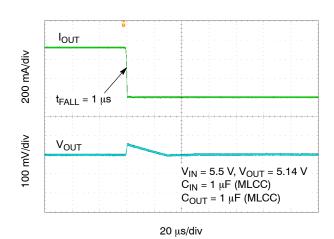


Figure 48. Load Transient Response – 450 mA to 1 mA – V_{OUT} = 5.14 V

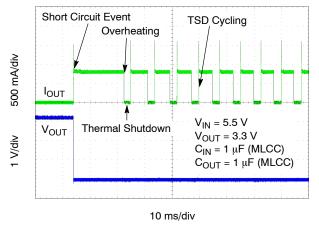


Figure 49. Short Circuit and Thermal Shutdown

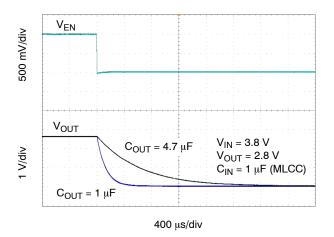


Figure 50. Enable Turn-off

APPLICATIONS INFORMATION

General

The NCP161 is an ultra-low noise 450 mA low dropout regulator designed to meet the requirements of RF applications and high performance analog circuits. The NCP161 device provides very high PSRR and excellent dynamic response. In connection with low quiescent current this device is well suitable for battery powered application such as cell phones, tablets and other. The NCP161 is fully protected in case of current overload, output short circuit and overheating.

Input Capacitor Selection (C_{IN})

Input capacitor connected as close as possible is necessary for ensure device stability. The X7R or X5R capacitor should be used for reliable performance over temperature range. The value of the input capacitor should be 1 μF or greater to ensure the best dynamic performance. This capacitor will provide a low impedance path for unwanted AC signals or noise modulated onto constant input voltage. There is no requirement for the ESR of the input capacitor but it is recommended to use ceramic capacitors for their low ESR and ESL. A good input capacitor will limit the influence of input trace inductance and source resistance during sudden load current changes.

Output Decoupling (COUT)

The NCP161 requires an output capacitor connected as close as possible to the output pin of the regulator. The recommended capacitor value is 1 μF and X7R or X5R dielectric due to its low capacitance variations over the specified temperature range. The NCP161 is designed to remain stable with minimum effective capacitance of 0.7 μF to account for changes with temperature, DC bias and package size. Especially for small package size capacitors such as 0201 the effective capacitance drops rapidly with the applied DC bias. Please refer Figure 51.

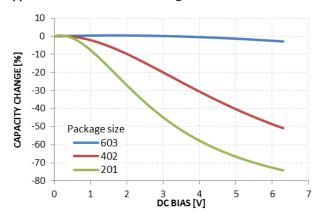


Figure 51. Capacity vs DC Bias Voltage

There is no requirement for the minimum value of Equivalent Series Resistance (ESR) for the C_{OUT} but the maximum value of ESR should be less than 2 Ω Larger output capacitors and lower ESR could improve the load

transient response or high frequency PSRR. It is not recommended to use tantalum capacitors on the output due to their large ESR. The equivalent series resistance of tantalum capacitors is also strongly dependent on the temperature, increasing at low temperature.

Enable Operation

The NCP161 uses the EN pin to enable/disable its device and to deactivate/activate the active discharge function.

If the EN pin voltage is <0.4 V the device is guaranteed to be disabled. The pass transistor is turned—off so that there is virtually no current flow between the IN and OUT. The active discharge transistor is active so that the output voltage V_{OUT} is pulled to GND through a 280 Ω resistor. In the disable state the device consumes as low as typ. 10 nA from the V_{IN} .

If the EN pin voltage >1.2 V the device is guaranteed to be enabled. The NCP161 regulates the output voltage and the active discharge transistor is turned–off.

The EN pin has internal pull-down current source with typ. value of 200 nA which assures that the device is turned-off when the EN pin is not connected. In the case where the EN function isn't required the EN should be tied directly to IN.

Output Current Limit

Output Current is internally limited within the IC to a typical 700 mA. The NCP161 will source this amount of current measured with a voltage drops on the 90% of the nominal V_{OUT} . If the Output Voltage is directly shorted to ground ($V_{OUT} = 0$ V), the short circuit protection will limit the output current to 690 mA (typ). The current limit and short circuit protection will work properly over whole temperature range and also input voltage range. There is no limitation for the short circuit duration.

Thermal Shutdown

When the die temperature exceeds the Thermal Shutdown threshold ($T_{SD}-160^{\circ}\text{C}$ typical), Thermal Shutdown event is detected and the device is disabled. The IC will remain in this state until the die temperature decreases below the Thermal Shutdown Reset threshold ($T_{SDU}-140^{\circ}\text{C}$ typical). Once the IC temperature falls below the 140°C the LDO is enabled again. The thermal shutdown feature provides the protection from a catastrophic device failure due to accidental overheating. This protection is not intended to be used as a substitute for proper heat sinking.

Power Dissipation

As power dissipated in the NCP161 increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the

ambient temperature affect the rate of junction temperature rise for the part.

The maximum power dissipation the NCP161 can handle is given by:

$$P_{D(MAX)} = \frac{\left[125^{\circ}C - T_{A}\right]}{\theta_{1A}}$$
 (eq. 1)

The power dissipated by the NCP161 for given application conditions can be calculated from the following equations:

$$P_D \approx V_{IN} \cdot I_{GND} + I_{OUT} (V_{IN} - V_{OUT})$$
 (eq. 2)

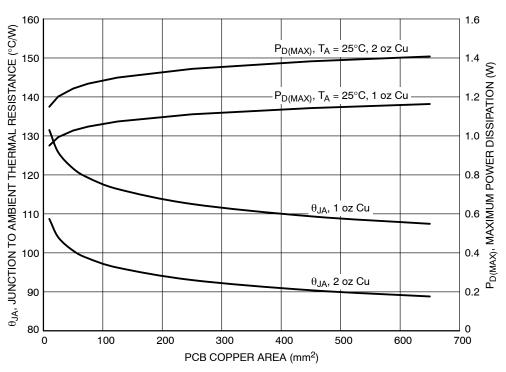


Figure 52. θ_{JA} and $P_{D \text{ (MAX)}}$ vs. Copper Area (CSP4)

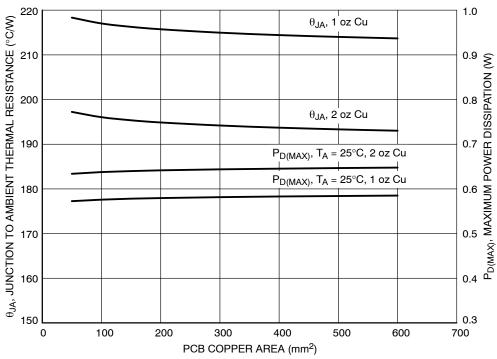


Figure 53. θ_{JA} and $P_{D~(MAX)}$ vs. Copper Area (XDFN4)

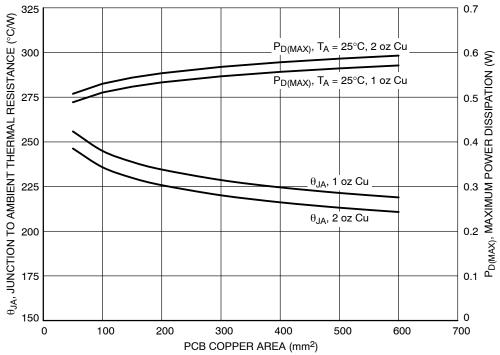


Figure 54. θ_{JA} and $P_{D\;(MAX)}$ vs. Copper Area (SOT23–5)

Reverse Current

The PMOS pass transistor has an inherent body diode which will be forward biased in the case that $V_{OUT} > V_{IN}$. Due to this fact in cases, where the extended reverse current condition can be anticipated the device may require additional external protection.

Power Supply Rejection Ratio

The NCP161 features very high Power Supply Rejection ratio. If desired the PSRR at higher frequencies in the range $100~\rm kHz-10~MHz$ can be tuned by the selection of C_{OUT} capacitor and proper PCB layout.

Turn-On Time

The turn-on time is defined as the time period from EN assertion to the point in which V_{OUT} will reach 98% of its nominal value. This time is dependent on various application conditions such as $V_{OUT(NOM)}$, C_{OUT} , T_A .

PCB Layout Recommendations

To obtain good transient performance and good regulation characteristics place $C_{\rm IN}$ and $C_{\rm OUT}$ capacitors close to the device pins and make the PCB traces wide. In order to minimize the solution size, use 0402 or 0201 capacitors with appropriate capacity. Larger copper area connected to the pins will also improve the device thermal resistance. The actual power dissipation can be calculated from the equation above (Equation 2). Expose pad can be tied to the GND pin for improvement power dissipation and lower device temperature.

ORDERING INFORMATION

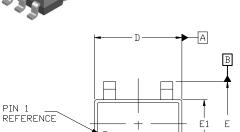
	Nominal Output					
Device	Voltage	Description	Marking	Rotation	Package	Shipping [†]
NCP161AFCS180T2G	1.8 V		Α	180°		
NCP161AFCS250T2G	2.5 V		D	180°		
NCP161AFCS270T2G	2.7 V		V	180°		
NCP161AFCS280T2G	2.8 V		Е	180°		
NCP161AFCS285T2G	2.85 V		F	180°		
NCP161AFCS300T2G	3.0 V	450 mA, Active Discharge	J	180°	WLCSP4 CASE 567KA*	5000 / Tape
NCP161AFCS320T2G	3.2 V	450 IIIA, Active Discharge	Т	180°	(Pb-Free)	& Reel
NCP161AFCS330T2G	3.3 V		K	180°	(1 5 1 100)	
NCP161AFCS350T2G	3.5 V		L	180°		
NCP161AFCS450T2G	4.5 V		Р	180°		
NCP161AFCS500T2G	5.0 V		R	180°		
NCP161AFCS514T2G	5.14 V		Q	180°		
NCP161BFCS180T2G	1.8 V		Α	270°		
NCP161BFCS250T2G	2.5 V		D	270°		
NCP161BFCS280T2G	2.8 V		Е	270°		
NCP161BFCS285T2G	2.85 V		F	270°	WILCODA	
NCP161BFCS300T2G	3.0 V	450 mA, Non-Active	J	270°	WLCSP4 CASE 567KA*	5000 / Tape & Reel
NCP161BFCS330T2G	3.3 V	Discharge	K	270°	(Pb-Free)	
NCP161BFCS350T2G	3.5 V		L	270°	(Pb-Fiee)	
NCP161BFCS450T2G	4.5 V		Р	270°		
NCP161BFCS500T2G	5.0 V		R	270°		
NCP161BFCS514T2G	5.14 V		Q	270°		
NCP161AFCT180T2G	1.8 V		Α	180°		
NCP161AFCT185T2G	1.85 V		V	180°		
NCP161AFCT250T2G	2.5 V		D	180°		
NCP161AFCT280T2G	2.8 V		Е	180°		
NCP161AFCT285T2G	2.85 V		F	180°		
NCP161AFCT290T2G	2.9 V		T	180°		
NCP161AFCT300T2G	3.0 V	450 mA, Active Discharge	J	180°	WLCSP4	
NCP161AFCT310T2G	3.1 V		6	180°	CASE 567JZ	5000 / Tape
NCP161AFCT330T2G	3.3 V		K	180°	(Pb-Free)	& Reel
NCP161AFCT350T2G	3.5 V		L	180°		
NCP161AFCT450T2G	4.5 V		Р	180°		
NCP161AFCT500T2G	5.0 V		R	180°		
NCP161AFCT514T2G	5.14 V		Q	180°		
NCP161AFCTC280T2G	2.8 V	450 mA, Active	Е	180°		
NCP161AFCTC350T2G	3.5 V	Discharge, Backside Coating	L	180°		
NCP161BFCT180T2G	1.8 V	-	Α	270°		
NCP161BFCT185T2G	1.85 V		V	270°		
NCP161BFCT250T2G	2.5 V		D	270°		
NCP161BFCT280T2G	2.8 V		Е	270°		
NCP161BFCT285T2G	2.85 V		F	270°	WLCSP4	
NCP161BFCT300T2G	3.0 V	450 mA, Non-Active Discharge	J	270°	CASE 567JZ	5000 / Tape & Reel
NCP161BFCT330T2G	3.3 V	. Discriarye	K	270°	(Pb-Free)	a neel
NCP161BFCT350T2G	3.5 V	1	L	270°		1
NCP161BFCT450T2G	4.5 V	1	Р	270°		1
NCP161BFCT500T2G	5.0 V	1	R	270°		1
NCP161BFCT514T2G	5.14 V	1	Q	270°		

^{*}UBM = 180 μm (±5 μm)

ORDERING INFORMATION (continued)

Device	Nominal Output Voltage	Description	Marking	Package	Shipping [†]
NCP161AMX180TBG	1.8 V		DN		
NCP161AMX185TBG	1.85 V	1	EY	1	
NCP161AMX250TBG	2.5 V	1	DP		
NCP161AMX280TBG	2.8 V]	DQ		
NCP161AMX285TBG	2.85 V]	DR		
NCP161AMX300TBG	3.0 V	450 m.A. Astina Disabanca	DT	XDFN4	3000 / Tape
NCP161AMX320TBG	3.2 V	450 mA, Active Discharge	DZ	(Pb-Free)	& Reel
NCP161AMX330TBG	3.3 V]	DD		
NCP161AMX350TBG	3.5 V		DU		
NCP161AMX450TBG	4.5 V		DV		
NCP161AMX500TBG	5.0 V		DX		
NCP161AMX514TBG	5.14 V		DE		
NCP161BMX180TBG	1.8 V		EN	XDFN4 (Pb-Free)	3000 / Tape & Reel
NCP161BMX250TBG	2.5 V		EP		
NCP161BMX280TBG	2.8 V		EQ		
NCP161BMX285TBG	2.85 V]	ER		
NCP161BMX300TBG	3.0 V	450 mA Non Astina Disabana	ET		
NCP161BMX330TBG	3.3 V	450 mA, Non-Active Discharge	ED		
NCP161BMX350TBG	3.5 V		EU	1	
NCP161BMX450TBG	4.5 V		EV	1	
NCP161BMX500TBG	5.0 V		EX	1	
NCP161BMX514TBG	5.14 V		EE	1	
NCP161ASN180T1G	1.8 V		JAF		
NCP161ASN250T1G	2.5 V	1	JAA	1	
NCP161ASN280T1G	2.8 V		JAC	SOT23-5L (Pb-Free)	
NCP161ASN300T1G	3.0 V	450 mA, Active Discharge	JAD		3000 / Tape & Reel
NCP161ASN330T1G	3.3 V		JAG	(1 0-1166)	5
NCP161ASN350T1G	3.5 V		JAH	1	
NCP161ASN500T1G	5.0 V	1	JAE	1	



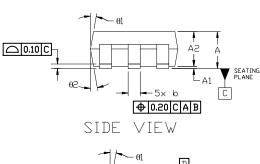


SOT-23, 5 Lead CASE 527AH **ISSUE A**

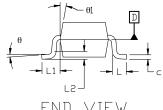
DATE 09 JUN 2021

NUTES

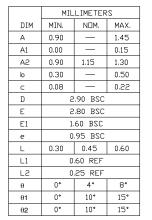
- DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 19894
- CONTROLLING DIMENSION: MILLIMETERS
- MAXIMUM LEAD THICKNESS INCLUDES LEAD FINISH THICKNESS. MINIMUM LEAD THICKNESS IS THE MINIMUM THICKNESS OF THE
- DIMENSIONS D AND E1 DO NOT INCLUDE MOLD FLASH, PROTRUSIONS, OR GATE BURRS, MOLD FLASH, PROTRUSIONS, OR GATE BURRS SHALL NOT EXCEED 0.25 PER SIDE. D AND E1 DIMENSIONS ARE DETERMINED AT DATUM D.
- DIMENSION 'b' DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION SHALL BE O. 08mm TOTAL IN EXCESS OF THE 'b' DIMENSION AT MAXIMUM MATERIAL CONDITION. MINIMUM SPACE BETWEEN PROTRUSION AND AN ADJACENT LEAD SHALL NOT BE LESS THAN 0.07mm.

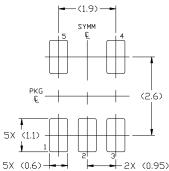


TOP VIEW



END VIEW





GENERIC MARKING DIAGRAM*



XXX = Specific Device Code = Date Code

*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot "=", may or may not be present. Some products may not follow the Generic Marking.

RECOMMENDED MOUNTING FOOTPRINT

For additional information on our Pb-Free strategy and soldering details, please download the $\square N$ Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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DESCRIPTION:	SOT-23, 5 LEAD		PAGE 1 OF 1	

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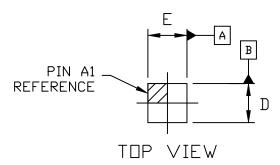




WLCSP4, 0.64x0.64x0.33

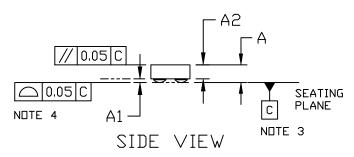
CASE 567JZ **ISSUE B**

DATE 16 MAY 2022

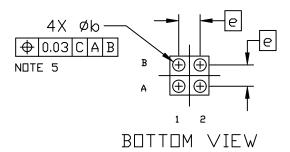


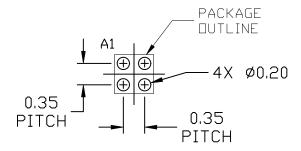
NOTES:

- DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 2009.
- 2. CONTROLLING DIMENSION: MILLIMETERS
- DATUM C, THE SEATING PLANE, IS DEFINED BY THE SPERICAL CROWNS OF THE SOLDER BUMPS.
- COPLANARITY APPLIES TO THE SPHERICAL CROWNS OF THE SOLDER BUMPS.
- DIMENSION & IS MEASURED AT THE MAXIMUM SOLDER BALL DIAMETER PARALLEL TO DATUM C.



	MILLIMETERS				
DIM	MIN.	N□M.	MAX.		
Α			0.33		
A1	0.04	0.06	0.08		
A2		0.23 REF			
b	0.180	0.200	0.220		
D	0.610	0.640	0.670		
E	0.610	0.640	0.670		
е	0.35 BSC				





GENERIC MARKING DIAGRAM*



= Specific Device Code Х

= Date Code M

For additional information on our Pb-Free strategy and soldering details, please download the IN Semiconductor Soldering and Mounting Techniques Reference Manual, SILDERRM/D.

RECOMMENDED MOUNTING FOOTPRINT *

(NSMD PAD TYPE)

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DOCUMENT NUMBER:	98AON85781F	Electronic versions are uncontrolled except when accessed directly from Printed versions are uncontrolled except when stamped "CONTROLLED (
DESCRIPTION:	WLCSP4, 0.64X0.64x0.33		PAGE 1 OF 1

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MECHANICAL CASE OUTLINE

SCALE 4:1

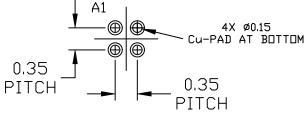
WLCSP4, 0.64x0.64 CASE 567KA ISSUE B

DATE 24 MAR 2020

NOTES:

- DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 2009.
- 2. CONTROLLING DIMENSION: MILLIMETERS
- DATUM C, THE SEATING PLANE, IS DEFINED BY THE SPERICAL CROWNS OF THE CONTACT BALLS.
- 4. COPLANARITY APPLIES TO THE SPHERICAL CROWNS OF THE CONTACT BALLS.
- 5. DIMENSION 6 IS MEASURED AT THE MAXIMUM CONTACT BALL DIAMETER PARALLEL TO DATUM C.

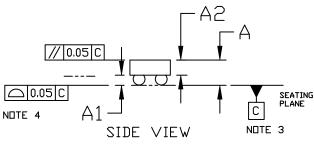
	MILLIMETERS		
DIM	MIN.	N□M.	MAX.
Α	0.355	0.405	0.455
A1	0.13	0.15	0.17
A2	0.255 REF		
b	0.167	0.187	0.207
D	0.610	0.640	0.670
E	0.610	0.640	0.670
е	0.35 BSC		

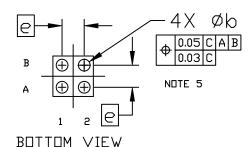


RECOMMENDED MOUNTING FOOTPRINT* (NSMD PAD TYPE)

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PIN A1 REFERENCE TOP VIEW





GENERIC MARKING DIAGRAM*



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PIN ONE

REFERENCE

2X 0.05 C

2X 0.05 C

// 0.05 C

□ 0.05 C

NOTE 4

XDFN4 1.0x1.0, 0.65P CASE 711AJ ISSUE C

В

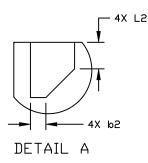
(A3)

SEATING

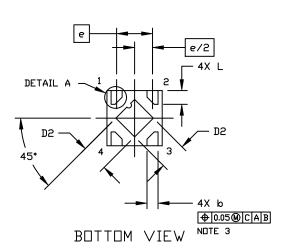
DATE 08 MAR 2022

NOTES:

- 1. DIMENSIONING AND TOLERANCING PER. ASME Y14.5M, 1994.
- 2. CONTROLLING DIMENSION: MILLIMETERS
- 3. DIMENSION & APPLIES TO THE PLATED TERMINALS AND IS MEASURED BETWEEN 0.15 AND 0.20 FROM THE TERMINAL TIPS.
- 4. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERMINALS.

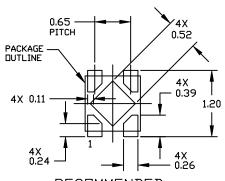


	MILLIMETERS		
DIM	MIN	NDM	MAX
Α	0.33	0.38	0.43
A1	0.00		0.05
A3	0.10 REF		
b	0.15	0.20	0.25
b2	0.02	0.07	0.12
D	0.90	1.00	1.10
D2	0.43	0.48	0.53
E	0.90	1.00	1.10
e	0.65 BSC		
L	0.20		0.30
L2	0.07		0.17



TOP VIEW

SIDE VIEW



RECOMMENDED MOUNTING FOOTPRINT

* FOR ADDITIONAL INFORMATION ON OUR PO-FRE STRATEGY AND SOLDERING DETAILS, PLEASE DOWNLOAD THE ONSEMI SOLDERING AND MOUNT TECHNIQUES REFERENCE MANUAL, SOLDERRM/D

GENERIC MARKING DIAGRAM*



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