

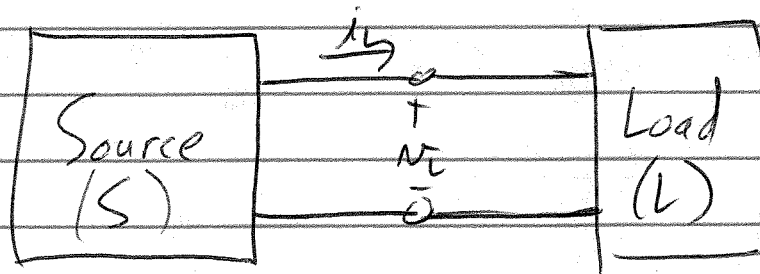
Thevenin + Norton Equivalents

We have already talked about these
as ^{Practical} Equivalents $V + I$ srcs, you even had
a question about them on the exam.

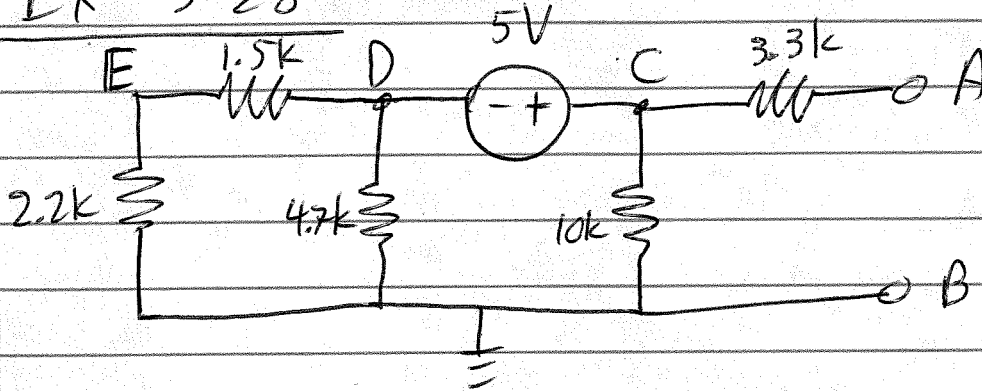
Thévenin proved for V src, and Norton
for I src, that:

If a src ckt with 2 terminals is linear,
then the "interface signals" $v + i$ do not
change if src. is replaced with ~~its~~ its

• Thevenin or Norton equivalent.



Ex 3-28



Find Thev + Norton Equivalents between A+B.
Several possible methods

1) Source transformations?

$$\underbrace{(2.2k + 1.5k)}_{\text{awkward}} // 4.7k + 5V = \text{Thev}(V_{\text{src}})$$

$$\text{Change to } I_{\text{src}} \Rightarrow \underbrace{R_{\text{eq}} // 10k}_{\text{awkward}}$$

$$\text{Change to } V_{\text{src}} \Rightarrow R'_{\text{eq}} + 3.3k = R_s$$

Instead try

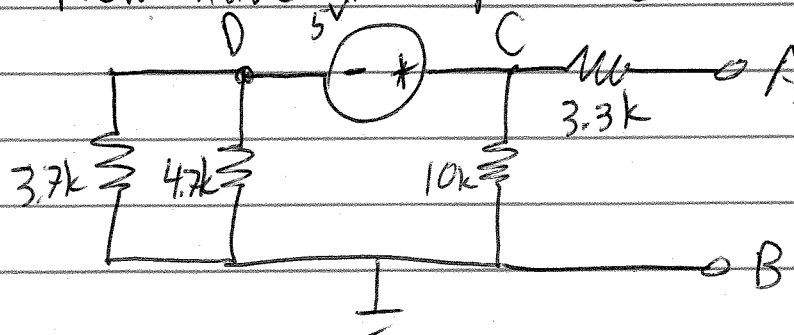
2) Nodal analysis for $V_{\text{oc}} + I_{\text{sc}}$:

A-B open ckt to get V_{oc}

A-B short ckt to get I_{sc}

Combine $1.5k + 2.2k = 3.7k$ + eliminate node E

Now have 1 supernode:



114

$$\frac{V_D}{3.7k} + \frac{V_D}{4.7k} + \frac{V_C}{10k} + \frac{V_C - V_A}{3.3k} = 0$$

$$V_C - V_D = 5V$$

For open ckt, $V_A = V_C$, so last term = 0, $V_{oc} = V_C$

For short ckt, $V_A = 0$, so last term = $\frac{V_C}{3.3k}$, $I_{sc} = \frac{V_C}{3.3k}$

$$O.C.: \left(\frac{1}{3.7k} + \frac{1}{4.7k}\right)V_D + \left(\frac{1}{10k}\right)V_C = 0$$
$$V_D = V_C - 5V$$

$$\frac{4.7k + 3.7k}{(3.7k)(4.7k)}(V_C - 5V) + \frac{1}{10k}V_C = 0$$

$$\frac{0.483 \times 10^{-3} V_C}{\Omega} + \frac{0.1 \times 10^{-3} V_C}{\Omega} = \frac{(0.483 \times 10^{-3}) 5V}{\Omega}$$

$$0.583 \times 10^{-3} V_C = 2.42 \text{ mA}$$

$$\underline{V_C = 4.15 \text{ V} = V_{oc}}$$

$$S.C.: 0.483 \times 10^{-3} V_C - 2.42 \text{ mA} + \left(\frac{1}{10k} + \frac{1}{3.3k}\right)V_C = 0$$

$$0.483 \times 10^{-3} V_C + 0.403 \times 10^{-3} V_C = 2.42 \text{ mA}$$

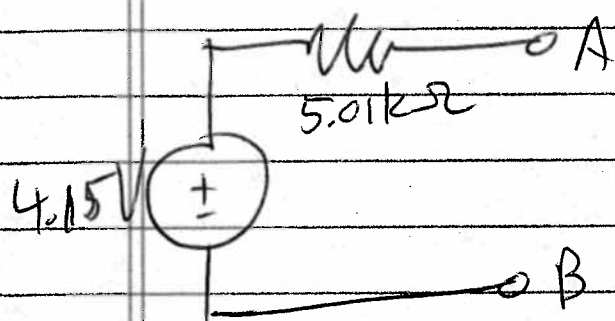
$$0.886 \times 10^{-3} V_C =$$

$$\underline{V_C = 2.73 \text{ V}}$$

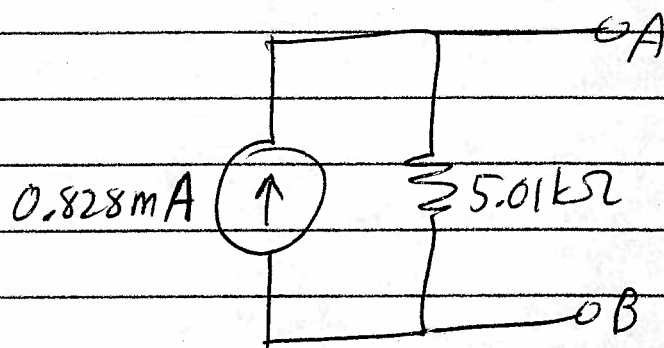
$$I_{sc} = \frac{V_C}{3.3k\Omega} = 0.828 \text{ mA}$$

$$R_s = \frac{V_{oc}}{I_{sc}} = \frac{4.15 \text{ V}}{0.828 \text{ mA}} = \underline{5.01 \text{ k}\Omega}$$

~~Ther~~ Thev. Eq.



(115)
Norton Eq

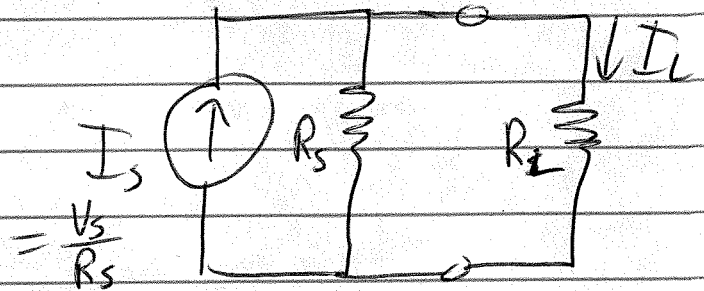
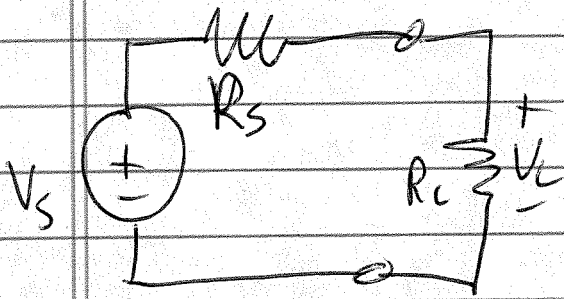


More exercises and examples in text,
pages 116-125. Read them, understand them.
do them

I want to address 2 things a
little out of order from the text,
so I am swapping them.

I want to first look at questions of
Power, then come back to applications
to Nonlinear Loads.

Take our 2 types of Eq. Ckts with Loads.



Let $R_L = \chi R_s$, $\chi \in [0, \infty)$

This can express every situation:

- $\chi = 0$, short ckt
- $0 < \chi < 1$, R_L less than R_s
- $\chi = 1$, $R_L = R_s$
- $1 < \chi < \infty$, R_L greater than R_s
- $\chi = \infty$, open ckt.

Division: $V_L = \frac{R_L}{R_s + R_L} V_s$

$$= \frac{\chi R_s}{R_s + \chi R_s} V_s$$

$$= \frac{R_s \cdot \chi}{R_s(1 + \chi)} V_s$$

$$I_L = \frac{R_s}{R_s + R_L} I_s$$

$$= \frac{R_s(1)}{R_s(1 + \chi)} I_s$$

$$= \frac{1}{1 + \chi} I_s$$

~~$V_L = \frac{R_L}{R_s + R_L} V_s$~~

Power $P_L = \frac{V_L^2}{R_L} = \frac{\chi^2}{(1 + \chi)^2} V_s^2 \frac{1}{\chi R_s}$

$$P_L = \frac{\chi}{(1 + \chi)^2} \frac{V_s^2}{R_s}$$

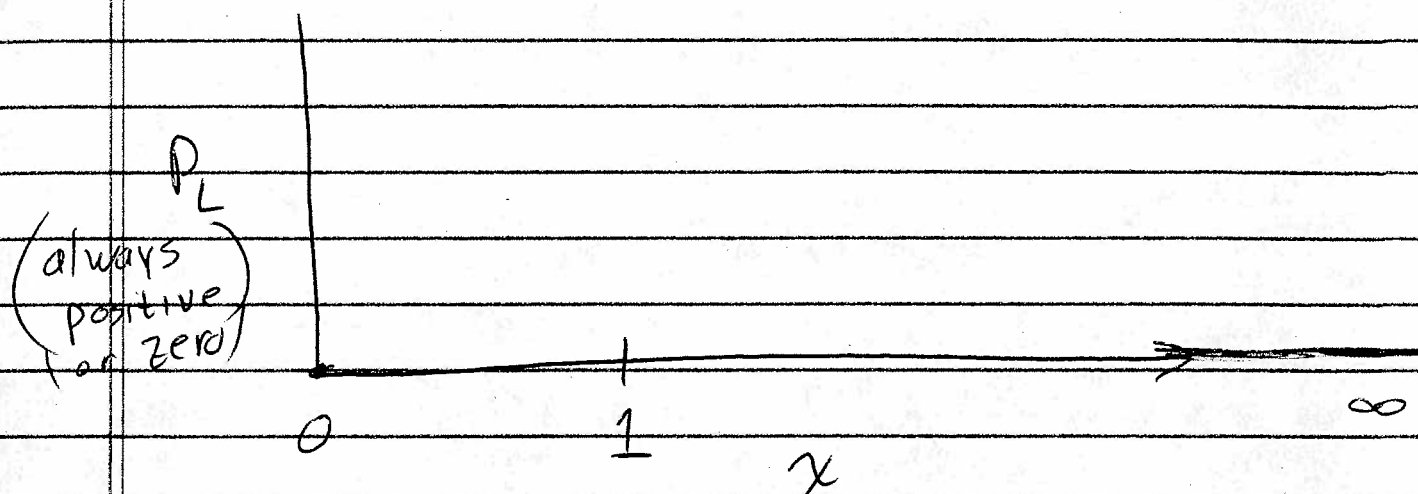
~~$P_L = I_L^2 R_L = \left(\frac{1}{1 + \chi}\right)^2 I_s^2 \chi R_s$~~

but $I_s = \frac{V_s}{R_s}$, so

$$P_L = \frac{1}{(1 + \chi)^2} \frac{V_s^2}{R_s^2} \chi R_s$$

Same!! $\rightarrow P_L = \frac{\chi}{(1 + \chi)^2} \frac{V_s^2}{R_s}$

Plot vs x :



limits: $R_L \rightarrow 0 \Rightarrow x \rightarrow 0$ (short ckt)
 $P_L \Rightarrow \frac{\infty}{(1+0)^2} \frac{V_s^2}{R_s} = 0$

$R_L \rightarrow \infty \Rightarrow x \rightarrow \infty$ (open ckt)
 $P_L = \frac{\infty}{(1+\infty)^2} \frac{V_s^2}{R_s} = \frac{1}{\infty} \frac{V_s^2}{R_s} = 0$

Now, we have something that is always positive or zero, starts at zero and ends at zero,
 (at $x=0$) (at $x=\infty$)

Any infinities? (Any places where the denominator $= 0$?) No. \Rightarrow must be bounded.

So P_L must come up to a maximum and back to zero as $x \rightarrow 0 \rightarrow \infty$.

Where does maximum occur?

$\frac{dP_L}{dx} = 0$ (I will not expect you to ever derive this again, use the results!)

$$\frac{d}{dx} \left(\frac{x}{(1+x)^2} \frac{V_s^2}{R_s} \right) = \frac{V_s^2}{R_s} \left[\frac{(1+x)^2(1) - x^2(1+x)(1)}{(1+x)^4} \right]$$

$$\neq \frac{V_s^2}{R_s} \left[\frac{(1+x)(1-2x)}{(1+x)^4} \right]$$

$$= \frac{V_s^2}{R_s} \left[\frac{(1+x)\{(1+x) - 2x\}}{(1+x)^4} \right]$$

$$= \frac{V_s^2}{R_s} \left[\frac{1-x}{(1+x)^3} \right]$$

When does this = 0? When numerator = 0, or

$$\text{When } 1-x=0$$

$$\underline{\underline{x=1}}$$

Maximum of P_L when $x=1$ or when $R_L = 1R_s$

$$\underline{\underline{R_L = R_s}}$$

Go back to sketch.

$$\text{Value at max? } P_{\max} = \frac{1}{(1+1)^2} \frac{V_s^2}{R_s} = \frac{1}{4} \frac{V_s^2}{R_s}$$

For Norton:

$$P_L = \frac{x}{(1+x)^2} R_s I_s^2$$

$$\frac{dP_L}{dx} = R_s I_s^2 (-2(1+x)(1+x)^3)$$

$$= R_s I_s^2 \cancel{-2(1+x)} \\ \frac{dP_L}{dx} = R_s I_s^2 \left[\frac{(1)(1+x)^2 - 2(1+x)(1)}{(1+x)^4} \right]$$

$$= R_s I_s^2 \cancel{(1+x)(1+x-2)} \\ = R_s I_s^2 \left[\frac{(1+x)(1+x-2)}{(1+x)^4} \right]$$

$$= R_s I_s^2 \left[\frac{x-1}{(1+x)^3} \right]$$

= 0 when $x=1 \Rightarrow$ when $R_L = 1R_s = R_s$

Same sketch

Value at max? $P_L = \frac{1}{(1+1)^2} R_s I_s^2 = \frac{1}{4} R_s I_s^2$

$$\text{let } I_s = \frac{V_s}{R_s}, \quad P_L = \frac{1}{4} R_s \left(\frac{V_s}{R_s} \right)^2 \\ = \frac{1}{4} \frac{V_s^2}{R_s}$$

↑

Same as for
Thevenin ! !

(120)

So, from point of view of load, the Voltage, Current, and Power dissipated are the same, so I cannot tell the difference between Thev. + Norton Equivalents.

Maximum Power Dissipated In Load

When $R_L = R_S$! ! ! ~~***~~

Let's look at 2 more things before we do more with this statement:

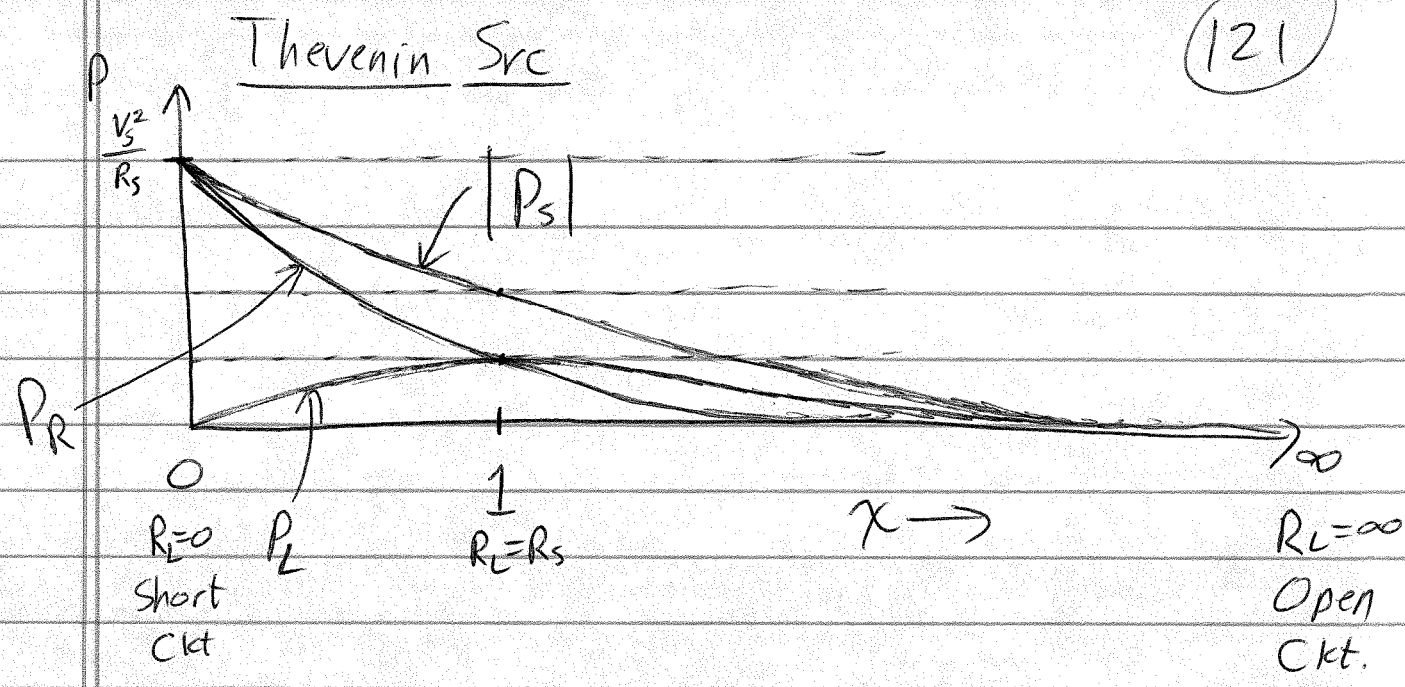
Look at power dissipated in R_S and power supplied by V_S or I_S for $R_L = x R_S$:

$$P_R = \frac{\left(\frac{R_S}{R_S + x R_S}\right)^2 V_S^2}{R_S} = \frac{R_S}{(R_S + x R_S)^2} V_S^2 = \frac{1}{(1+x)^2} \frac{V_S^2}{R_S}$$

$$\text{Graph: } P_R(0) = \frac{V_S^2}{R_S}, P_R(1) = \frac{1}{4} \frac{V_S^2}{R_S}, P_R(\infty) = 0$$

$$P_S = -I V_S = -\left(\frac{V_S}{R_S + x R_S}\right) V_S = -\frac{1}{(1+x)} \frac{V_S^2}{R_S}$$

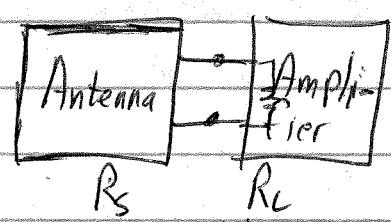
$$\text{Graph: } P_S(0) = -\frac{V_S^2}{R_S}, P_S(1) = -\frac{1}{2} \frac{V_S^2}{R_S}, P_S(\infty) = 0$$



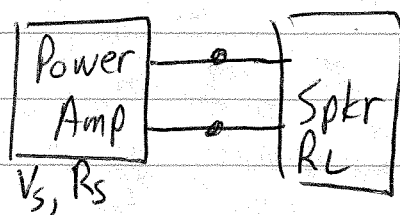
So, P_s supplies the most power, and all of that power is dissipated inside the box, when it is shorted out. (why batteries get hot when shorted out.)

Where do you want to operate?

It depends:

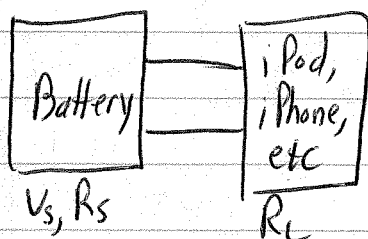


Probably want max power transfer, so want $R_L = R_S$



Want max power,
so want $R_L = R_s$
If $R_L \neq R_s$, no damage to
speakers, less power is
delivered to either, \Rightarrow not as
loud.

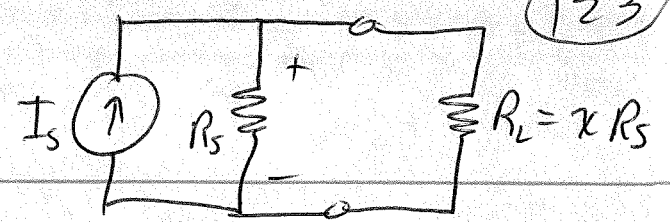
But if $R_L < R_s$ the power
supplied (P_s) and dissipated
internally (P_R) go up.



Don't want max power, as

that would give minimum
battery life. Want minimum
power delivered, for max.
battery life. $R_L \rightarrow \infty$

Now, Norton Equiv.



$$P_L = \frac{\chi}{(1+\chi)^2} R_s I_s^2$$

$\chi=0: P_L=0, \chi=1: P_L = \frac{1}{4} R_s I_s^2, \chi \rightarrow \infty: P_L=0$

$$P_R = R_s \left[\frac{\chi R_s}{R_s + \chi R_s} I_s \right]^2 = R_s \left[\frac{\chi}{1+\chi} I_s \right]^2 = \frac{\chi^2}{(1+\chi)^2} R_s I_s^2$$

$\chi=0: P_R=0$

$\chi=1: P_R = \frac{1}{4} R_s I_s^2$

$\chi=\infty: P_R = R_s I_s^2$

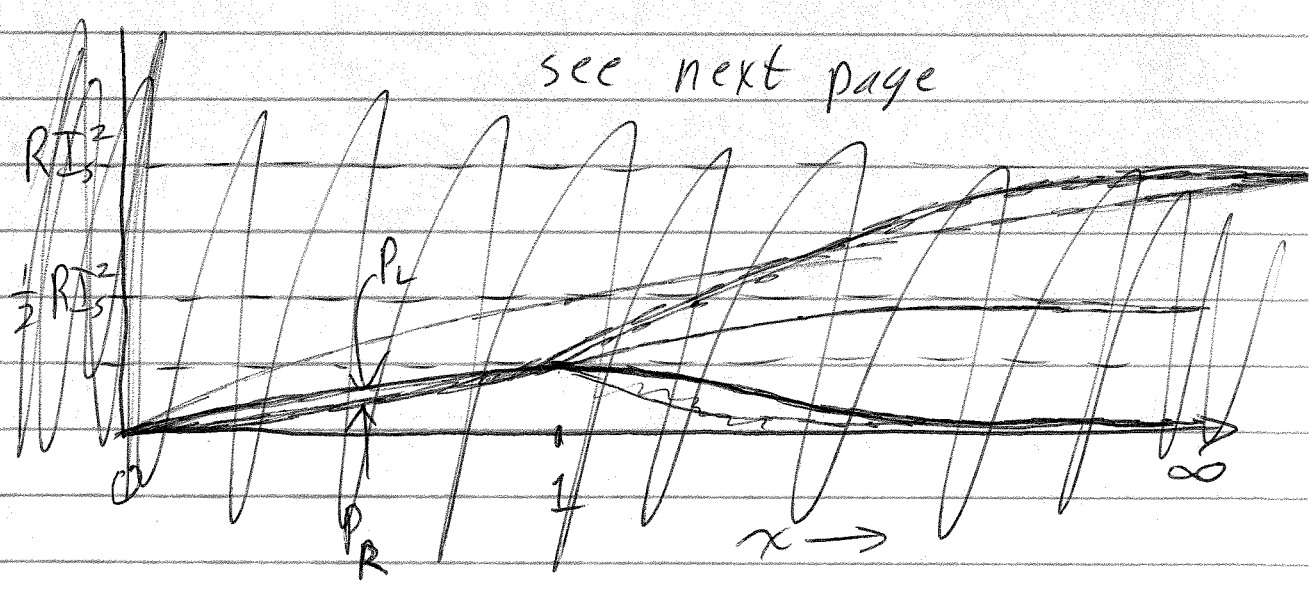
$$P_L = -I_s \left(\frac{(R_s)(\chi R_s)}{R_s + \chi R_s} I_s \right) = -I_s^2 \left(\frac{\chi}{1+\chi} R_s \right) = -\frac{\chi}{1+\chi} R_s I_s^2$$

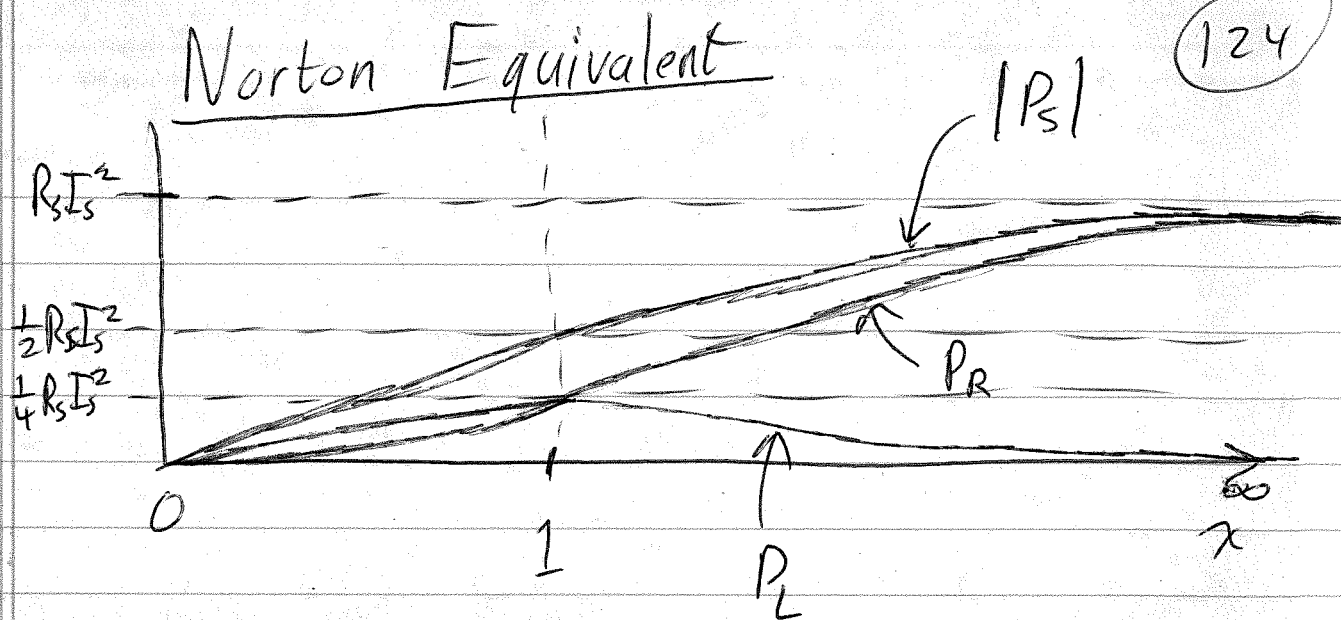
$\chi=0: P_L=0$

$\chi=1: P_L = -\frac{1}{2} R_s I_s^2$

$\chi=\infty: P_L = -R_s I_s^2$

see next page





Now, if the "Black Box" really behaves like a Current Source, it dissipates its maximum ~~power~~ power internally When It Is Open Circuited not short ckt'd.!

Very few things really behave in this way. Laboratory devices that are unprotected current sources have big warning labels on them, "Do not disconnect load with power on" or "Do not operate without load connected."

So, most devices (batteries, power supplies, etc.) really behave like ~~the~~ Practical Voltage Sources, or Thevenin Equivalents.

Sometimes using an equivalent Practical Current Source as a Model can make analysis easier.