

We need to move on to Chapter 8-

Sinusoidal Steady State Response- ~~the~~ where
we will learn how to represent ^{how} circuits ~~by their~~
respond to sinusoidal forcing functions that have
been on forever and will remain on forever (hence:
Steady State) This is extremely useful,
because it is the foundation for concepts to
come in future courses, for example:

Fourier Series: Any periodic function can

be represented by an (infinite) sum of sines

and cosines, each with different amplitudes, and all

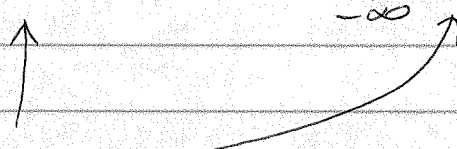
with frequencies that are multiples of the ~~the~~

fundamental frequency.

$$f_{\text{per}}(t) = \sum_{i=1}^{\infty} A_i \cos(2\pi f_{0,i} t) + \sum_{i=1}^{\infty} B_i \sin(2\pi f_{0,i} t)$$

Or the Fourier Transform:

Any (bounded, etc.) function can be represented by ~~an infinite~~ summation (i.e. integral) of an infinite set of sinusoids with different frequencies and phases:

$$f(t) = \int_{-\infty}^{\infty} A(\omega) \cos(\omega t) d\omega + \int_{-\infty}^{\infty} B(\omega) \sin(\omega t) d\omega$$


If you can figure out what $A(\omega)$ and $B(\omega)$ are, then you know what $f(t)$ is going to be.

You will spend lots of time in later classes studying these things, we're going to start with the basics, for now.

We are going to start by analyzing ckt's driven by a single frequency sinusoid, and work our way up.

First, we are going to represent sinusoids with complex numbers:

$$\text{Euler's Relation: } e^{j\theta} = \cos\theta + j\sin\theta$$

$$\text{or } \cos\theta = \text{Re}(e^{j\theta})$$

$$\sin\theta = \text{Im}(e^{j\theta})$$

A general sinusoid in the Time Domain

$$\text{is } v(t) = V_A \cos(\omega t + \phi)$$

$$= V_A \text{Re}(e^{j(\omega t + \phi)})$$

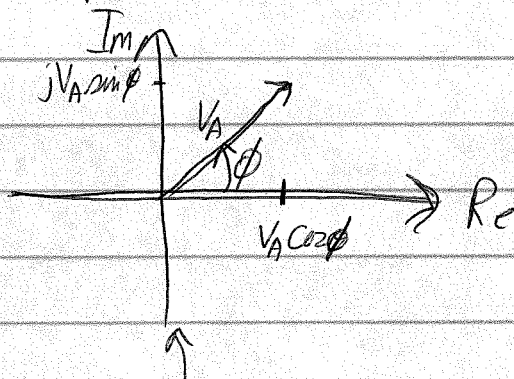
$$= V_A \text{Re}[e^{j\omega t} e^{j\phi}]$$

$$= \text{Re}[V_A e^{j\phi} e^{j\omega t}]$$

$$= \text{Re}[\hat{V}_A e^{j\omega t}]$$

\tilde{V}_A is the phasor representation (not Star Trek Phaser) of a complex # $\tilde{V}_A = V_A e^{j\phi} = V_A [\cos\phi + j\sin\phi]$

This can be represented on a 2D coordinate system:



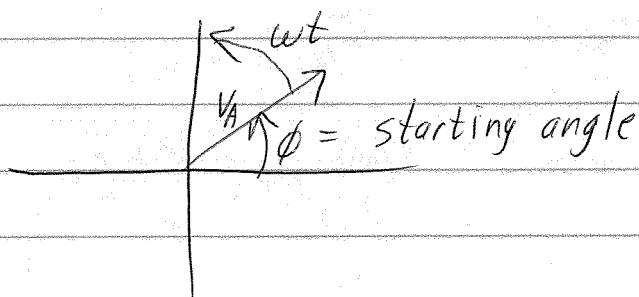
called the Phasor Diagram

Put the $e^{j\omega t}$ back in: $(V_A e^{j\phi}) e^{j\omega t}$ leads to

$$V_A e^{j(\omega t + \phi)} = V_A e^{j\delta(t)}$$

\uparrow constant (wrt time) amplitude \uparrow Phase increases

or a rotation of the vector around the origin:



At any time, t , the Real part is just the projection of the vector onto the horizontal axis.

Reverse the steps: ↖ rotating phasor

$$v(t) = \text{Re}[\tilde{V} e^{j\omega t}] \quad \text{phasor at } t=0, \text{ initial phase } \phi.$$

$$= \text{Re}[V_A e^{j\phi} e^{j\omega t}]$$

$$= V_A \text{Re}[\cos(\omega t + \phi) + j \sin(\omega t + \phi)]$$

$$= V_A \cos(\omega t + \phi) \text{ as we started with.}$$

We need to practice going from one end to the other without the intervening steps:

$$v(t) = 10 \cos(377t - 17^\circ) \Rightarrow \tilde{V} = 10 e^{j(-17^\circ)}$$

usually want this in radians:

$$\frac{-17^\circ \pi}{180^\circ} = -0.297$$

$$\tilde{V} = 10 e^{-j0.297}$$

Another way of writing the phasor is Magnitude \angle Phase

$$\tilde{V} = V_A e^{j\phi} = V_A \angle \phi \quad \text{or} \quad \tilde{V} = 10 \angle -17^\circ$$



saves writing $e^{j(\phi)}$ each time

Do exercises 8-1 + 8-2.

Also, recall that ω ^{radian frequency} = $2\pi f$ ^{frequency (Hz)} (cycles/sec)

$f(\text{Hz})$	ω
10	62.8
60	377
100	628
50	314

Recall that we multiply and ~~add~~ divide complex numbers by multiplying (dividing) amplitudes and adding (subtracting) phases:

$$(Ae^{j\phi})(Be^{j\sigma}) = AB e^{j(\phi+\sigma)}$$

Adding + Subtracting requires rectangular form:

$$\tilde{A} = \tilde{A}e^{j\phi} = A\cos\phi + jA\sin\phi \quad \tilde{B} = Be^{j\sigma} = B\cos\sigma + jB\sin\sigma$$

$$\tilde{A} + \tilde{B} = (A\cos\phi + B\cos\sigma) + j(A\sin\phi + B\sin\sigma)$$

$$= C + jD = Ee^{j\theta} \quad \text{where } E = \sqrt{C^2 + D^2} \\ + \theta = \tan^{-1}\left(\frac{D}{C}\right)$$

We also use phasors to represent currents:

$$i(t) = \text{Re}[\hat{I} e^{j\omega t}]$$

Properties of Phasors:

Sum of phasors OF THE SAME FREQUENCY

is the sum of the phasors:

$$v(t) = v_1(t) + v_2(t) + \dots + v_N(t) \quad \text{all same } f.$$

$$= \text{Re}[\tilde{V}_1 e^{j\omega t}] + \text{Re}[\tilde{V}_2 e^{j\omega t}] + \dots + \text{Re}[\tilde{V}_N e^{j\omega t}]$$

$$= \text{Re}[(\tilde{V}_1 + \tilde{V}_2 + \dots + \tilde{V}_N) e^{j\omega t}]$$

$$= \text{Re}[\tilde{V} e^{j\omega t}]$$

Time Derivative:

$$\frac{d}{dt} [\text{Re}(\tilde{V} e^{j\omega t})] = \text{Re}(\tilde{V} \frac{d}{dt} (e^{j\omega t}))$$

$$= \text{Re}(j\omega \tilde{V} e^{j\omega t})$$

$$\neq \text{Re}(j\omega \tilde{V})$$

$$= \text{Re}(j\omega \tilde{V} e^{j\omega t})$$

Recall

from $\tilde{V} = V_A e^{j\phi}$

$$j\omega \tilde{V} = \cancel{e^{j\frac{\pi}{2}}} \omega V_A e^{j\phi}$$

$$= \omega V_A e^{j(\phi + \pi/2)}$$

$$= \omega V_A e^{j(\phi + 90^\circ)}$$

derivative gives $\omega V_A + 90^\circ$

Integral (Inverse Derivative)

$$\int \tilde{V}(t) dt = \text{Re} \left(\frac{1}{j\omega} \tilde{V} e^{j\omega t} \right)$$

$$= \text{Re} \left(-j \frac{1}{\omega} \tilde{V} e^{j\omega t} \right)$$

$$= \text{Re} \left(\underbrace{\frac{1}{\omega} V_A e^{j(\phi - 90^\circ)}}_{\tilde{V}_I} e^{j\omega t} \right)$$

Integral gives $\frac{V_A}{\omega} + -90^\circ$

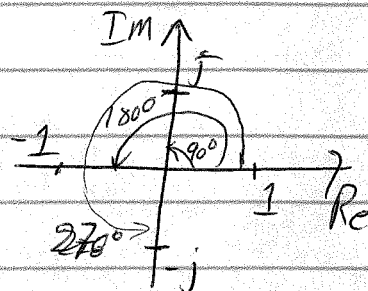
$$j^0 = 1 = e^{j0}$$

$$j^1 = j = e^{j\pi/2} = e^{j90^\circ}$$

$$j^2 = -1 = e^{j\pi} = e^{j180^\circ}$$

$$j^3 = -j = e^{j3\pi/2} = e^{j270^\circ}$$

$$j^4 = 1 = e^{j2\pi} = e^{j360^\circ}$$



$$j^{-1} = \frac{1}{j} = -j = e^{-j\pi/2} = e^{-j90^\circ}$$

$$j^{-2} = \frac{1}{j^2} = -1$$

Example:

$$i_1(t) = 50 \cos(100t) \text{ mA}$$

$$i_2(t) = 20 \cos(100t + 60^\circ) \text{ mA}$$

$$\omega = 100 \text{ s}^{-1}$$

$$\underline{\underline{I_1}} = 50 \angle 0^\circ$$

$$\underline{\underline{I_2}} = 20 \angle 60^\circ$$

$$= 50 \cos(0^\circ) + j 50 \sin(0^\circ)$$

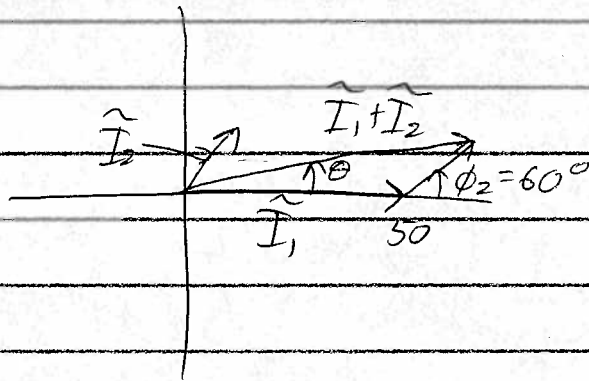
$$= 50 + j 0$$

$$= 10 + j 17.32$$

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$$\begin{aligned}
 i_1(t) + i_2(t) &= \operatorname{Re}(\tilde{I}_1 + \tilde{I}_2) e^{j\omega t} \\
 &= \operatorname{Re}((50 + j0 + 10 + j17.32) e^{j\omega t}) \\
 &= \operatorname{Re}((60 + j17.32) e^{j\omega t}) \\
 &= \operatorname{Re}(62.45 \angle 16.10^\circ e^{j\omega t}) \\
 &= 62.45 \cos(100t + 16.10^\circ)
 \end{aligned}$$

Pictorially:



$$\begin{aligned}
 \frac{d}{dt}(i_1(t)) &= \operatorname{Re}(((100)(50) \angle 90^\circ) e^{j\omega t}) \\
 &= \operatorname{Re}(5000 e^{j(\omega t + 90^\circ)}) \text{ mA/s}
 \end{aligned}$$

We can show that KCL+KVL still hold
for phasors:

KCL: The algebraic sum of phasor currents leaving any node is zero.

KVL: The algebraic sum of phasor voltage drops around a loop is zero.

And Device Constraints:

$$v_R = R i_R$$

Resistors:

$$\tilde{V}_R = R \tilde{I}_R$$

$$i_C = C \frac{d}{dt} v_C$$

Capacitors:

$$i_C(t) = C \frac{dv_C}{dt}$$

$$\begin{aligned} \tilde{I}_C &= C j\omega \tilde{V}_C \\ &= j\omega C \tilde{V}_C \end{aligned}$$

$$v_L = L \frac{d}{dt} i_L$$

or $\tilde{V}_C = \frac{1}{j\omega C} \tilde{I}_C$

Inductors: $v_L = L \frac{di_L}{dt}$

$$\tilde{V}_L = L(j\omega \tilde{I}_L)$$

$$\tilde{V}_L = j\omega L \tilde{I}_L$$

These all look like Ohm's Law:

$$\tilde{V} = \tilde{Z} \tilde{I} \quad \text{where } \tilde{Z} \text{ is called the } \underline{\text{impedance}}$$

$$\tilde{Z}_R = R \quad (\text{purely real}) \quad \text{does not change with frequency}$$

$$\tilde{Z}_C = \frac{1}{j\omega C} = -j \frac{1}{\omega C} \quad (\text{purely imaginary})$$

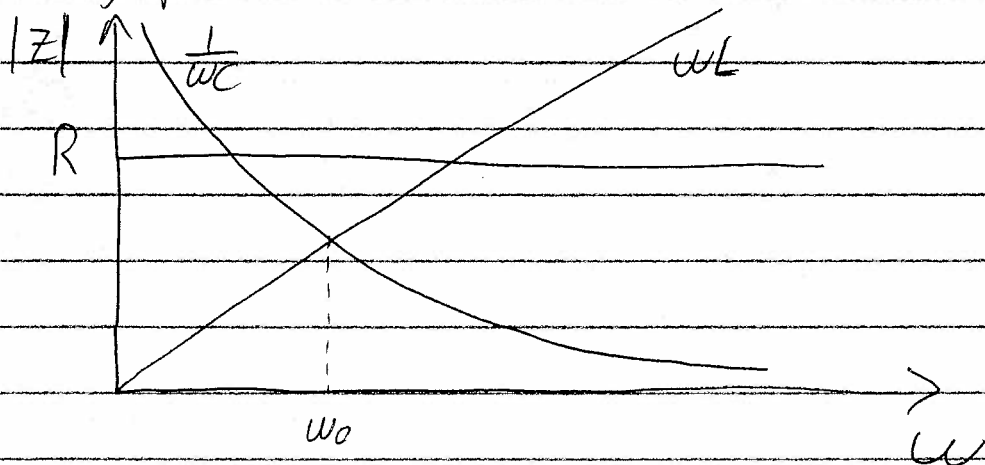
$$\tilde{Z}_L = j\omega L \quad (\text{purely imaginary}) \quad \left. \begin{array}{l} \text{change} \\ \text{with} \\ \text{frequency} \\ \omega. \end{array} \right\}$$

In general, \tilde{Z} can be a complex number:

$$\tilde{Z} = R + jX$$

impedance = $\begin{array}{l} \uparrow \text{resistance} \quad \uparrow \text{reactance} \\ \text{(reactance } > 0 \Rightarrow \text{inductive)} \\ \text{(reactance } < 0 \Rightarrow \text{capacitive)} \end{array}$

Text shows graph:



ω_0 is when $|Z_c| = |Z_L|$ or

$$\frac{1}{\omega_0 C} = \omega_0 L$$

$$\frac{1}{LC} = \omega_0^2$$

$$\omega_0 = \frac{1}{\sqrt{LC}} \quad \text{as we saw before.}$$

Note behavior at low + high frequencies:

$$\tilde{Z}_c = \frac{1}{j\omega C} \quad |\tilde{Z}_c| = \frac{1}{\omega C}$$

$$\lim_{\omega \rightarrow 0} |\tilde{Z}_c| = \infty \quad (\text{an open ckt at DC})$$

$$\lim_{\omega \rightarrow \infty} |\tilde{Z}_c| = 0 \quad (\text{a short ckt at High F})$$

$$\tilde{Z}_L = j\omega L \quad |\tilde{Z}_L| = \omega L$$

$$\lim_{\omega \rightarrow 0} |\tilde{Z}_L| = 0 \quad (\text{a short ckt at DC})$$

$$\lim_{\omega \rightarrow \infty} |\tilde{Z}_L| = \infty \quad (\text{an open ckt at high F})$$