

114-1 電工實驗（通信專題）

OFDM

Chia-Yi Yeh (葉佳宜)

ycyyeh@ntu.edu.tw

Department of Electrical Engineering
National Taiwan University

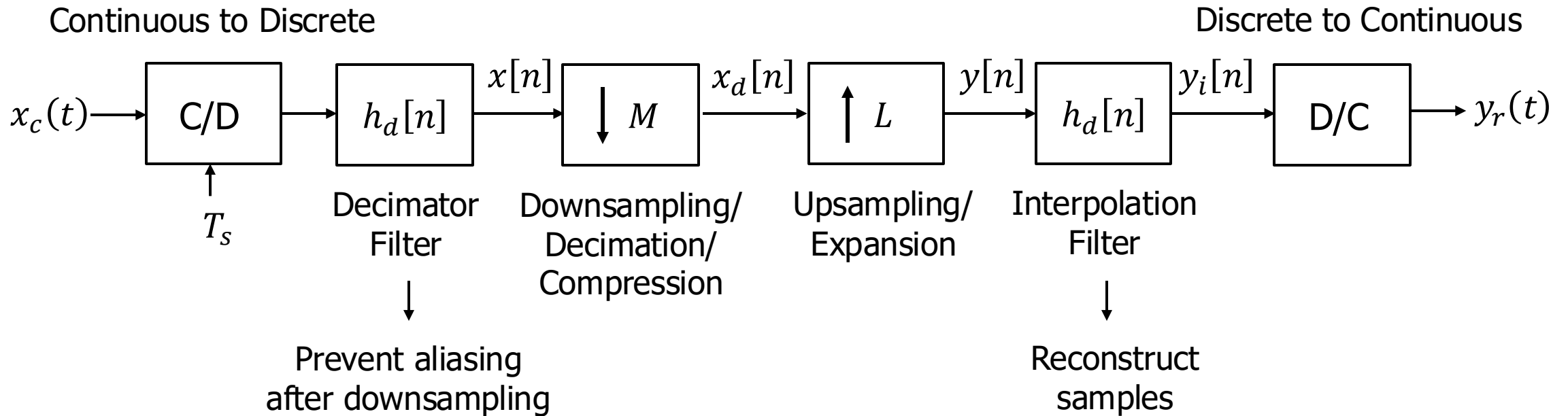
Review

Up/Down-Sampling
Filtering

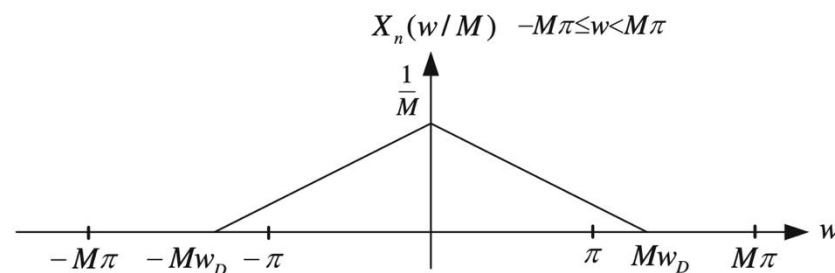
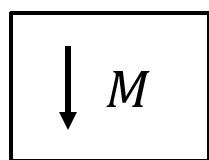
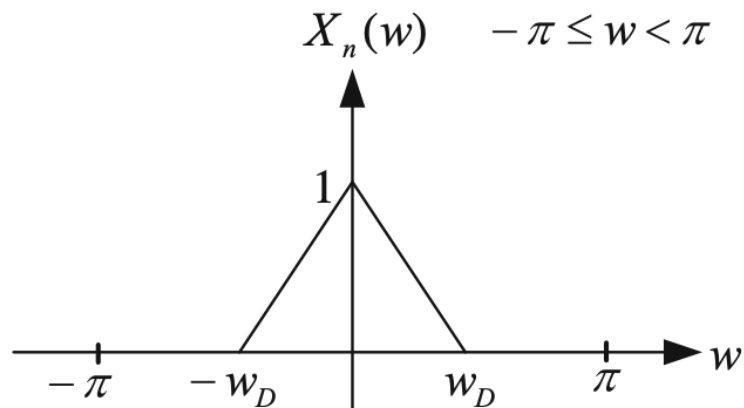
Up/Down-Sampling

Why change sampling frequency?

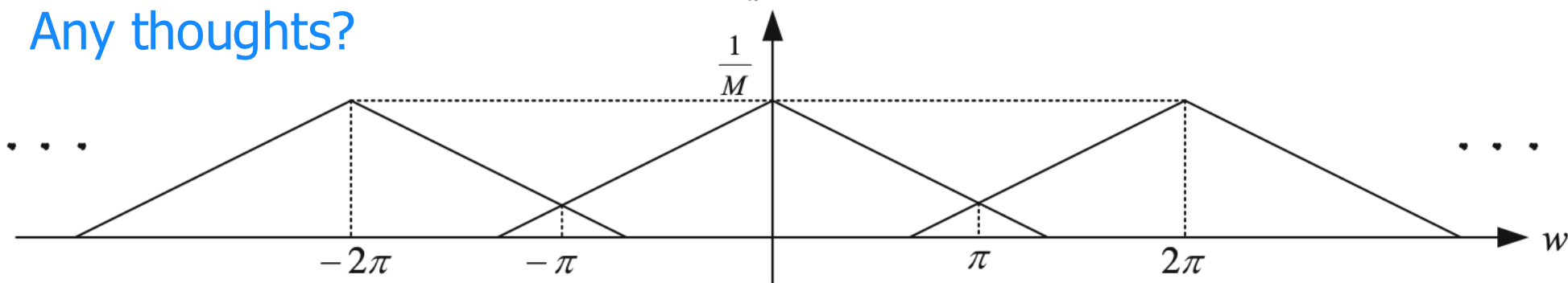
Compatibility between different devices



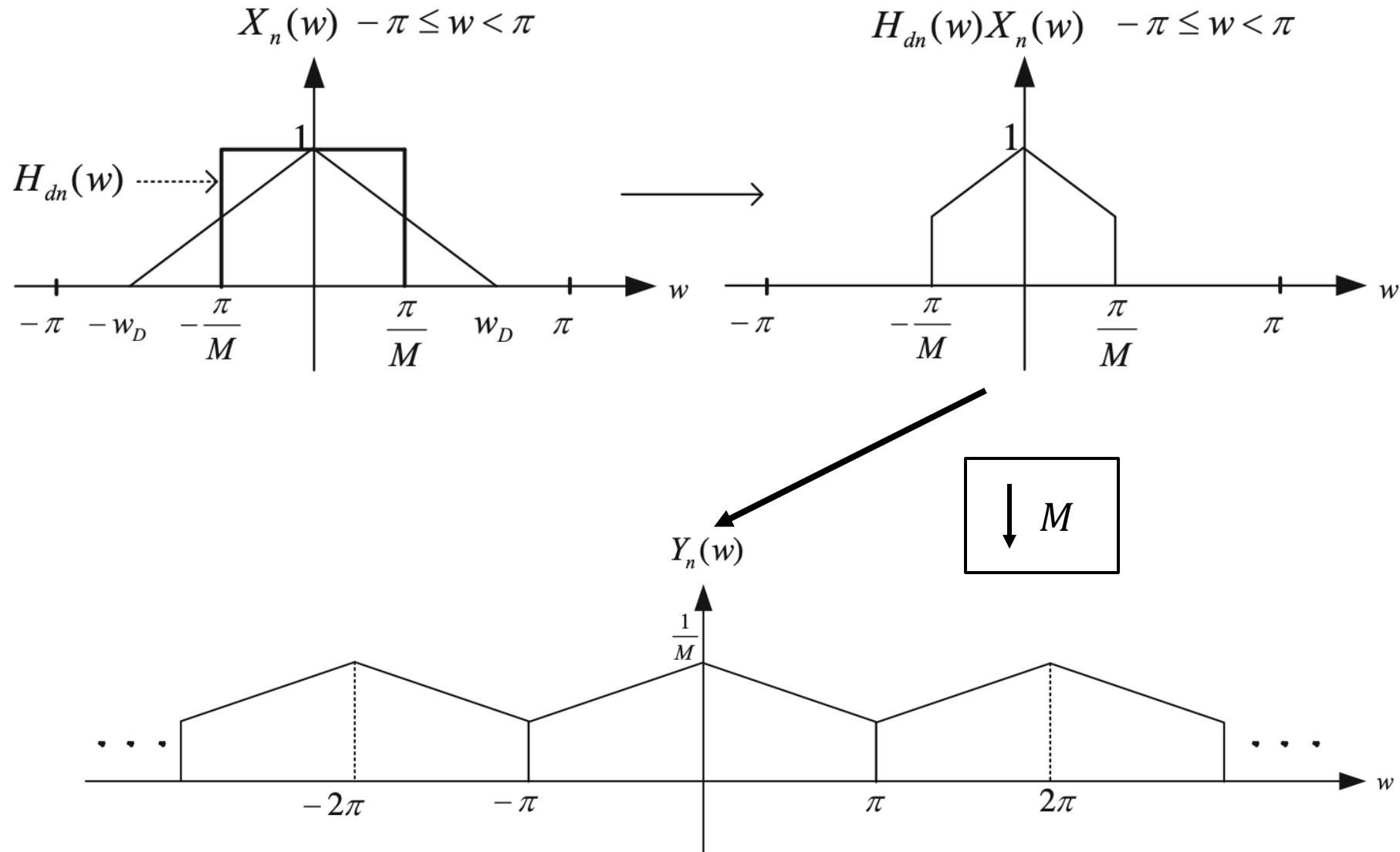
Aliasing in Downsampling



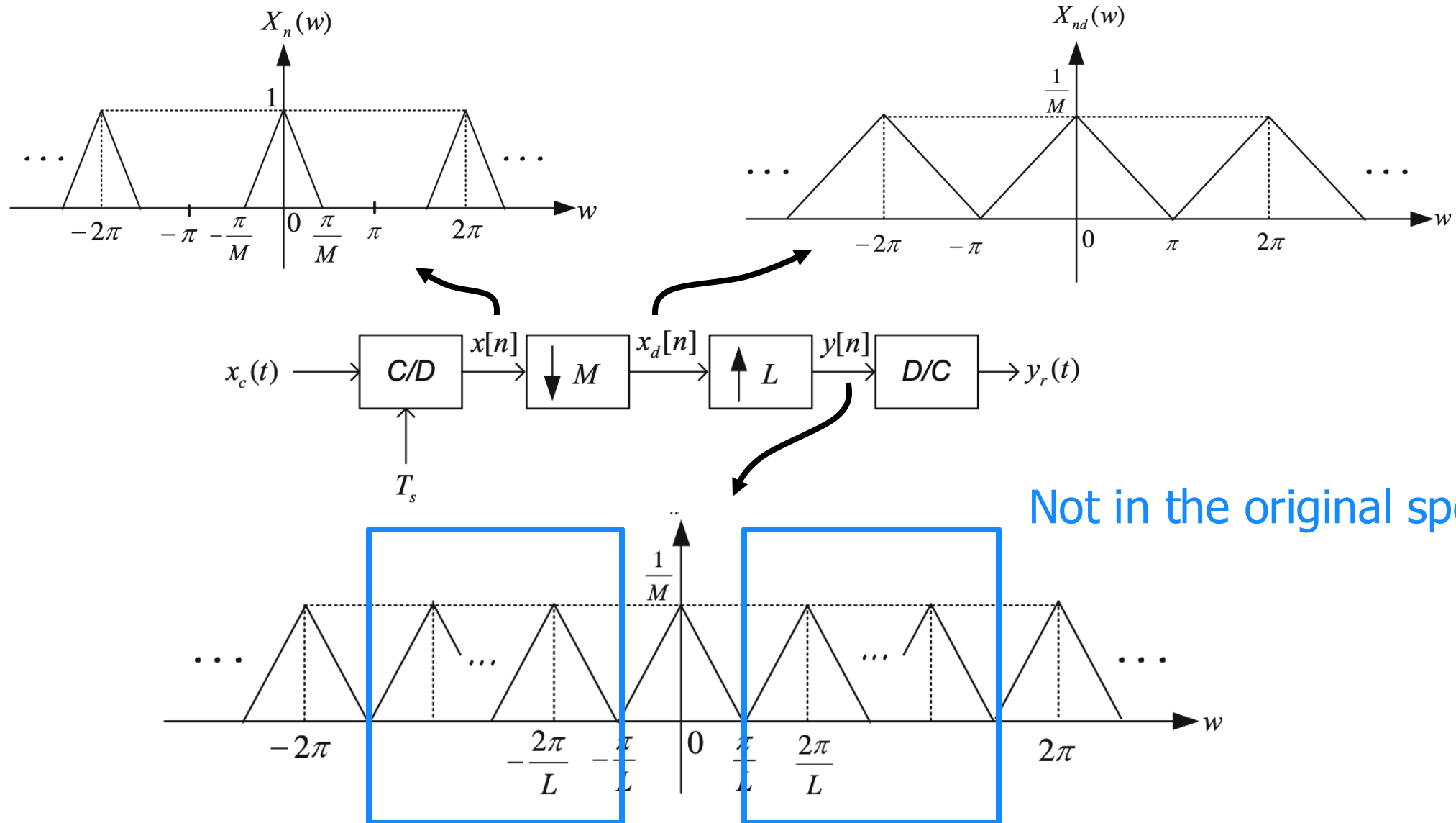
Any thoughts?



Prevent Aliasing: Lowpass filter before downsampling

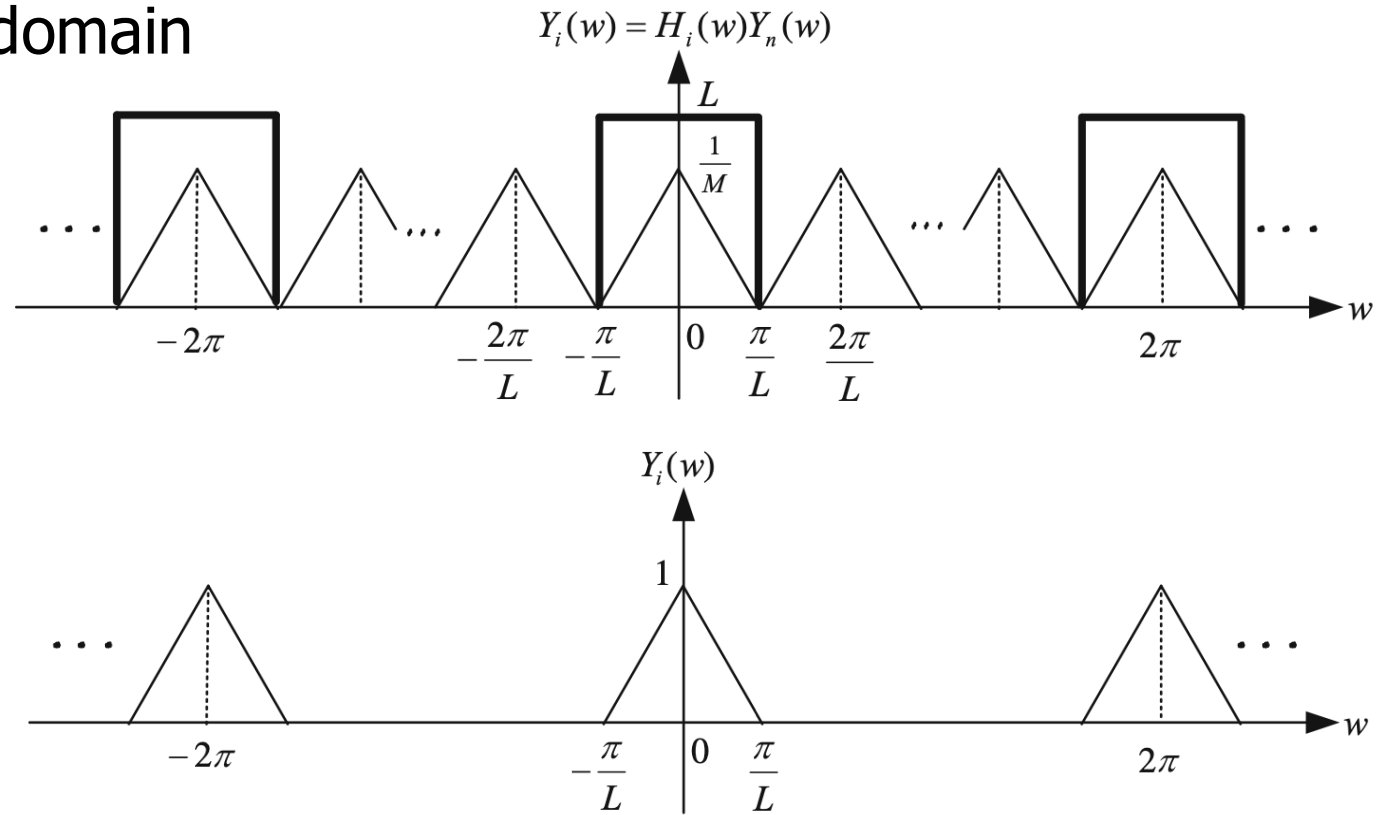


Upsampling: From Frequency Domain Perspective



Signal Reconstruction for Upsampling

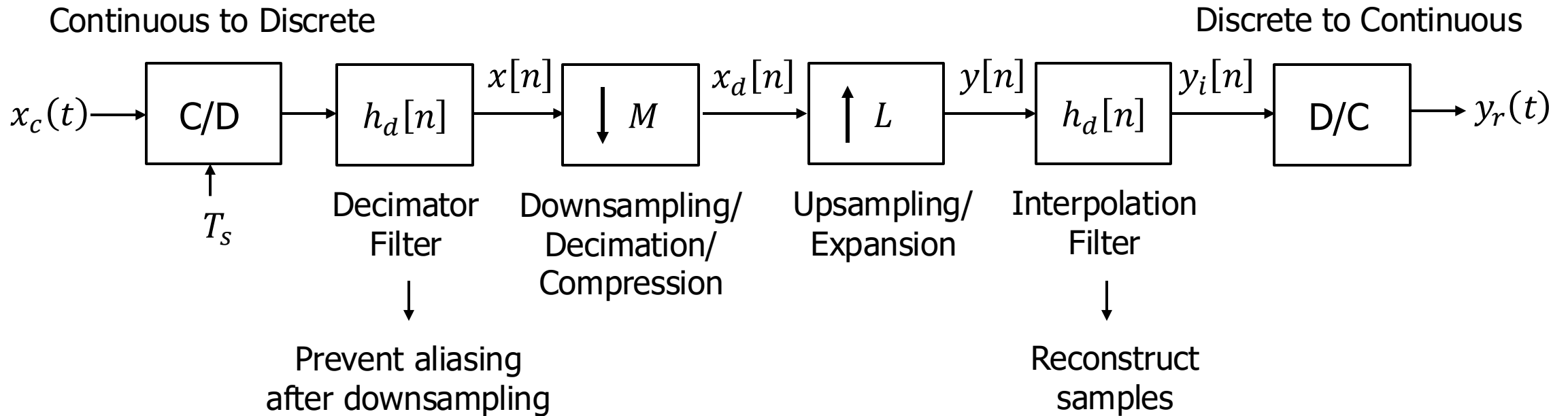
Filter from the frequency domain



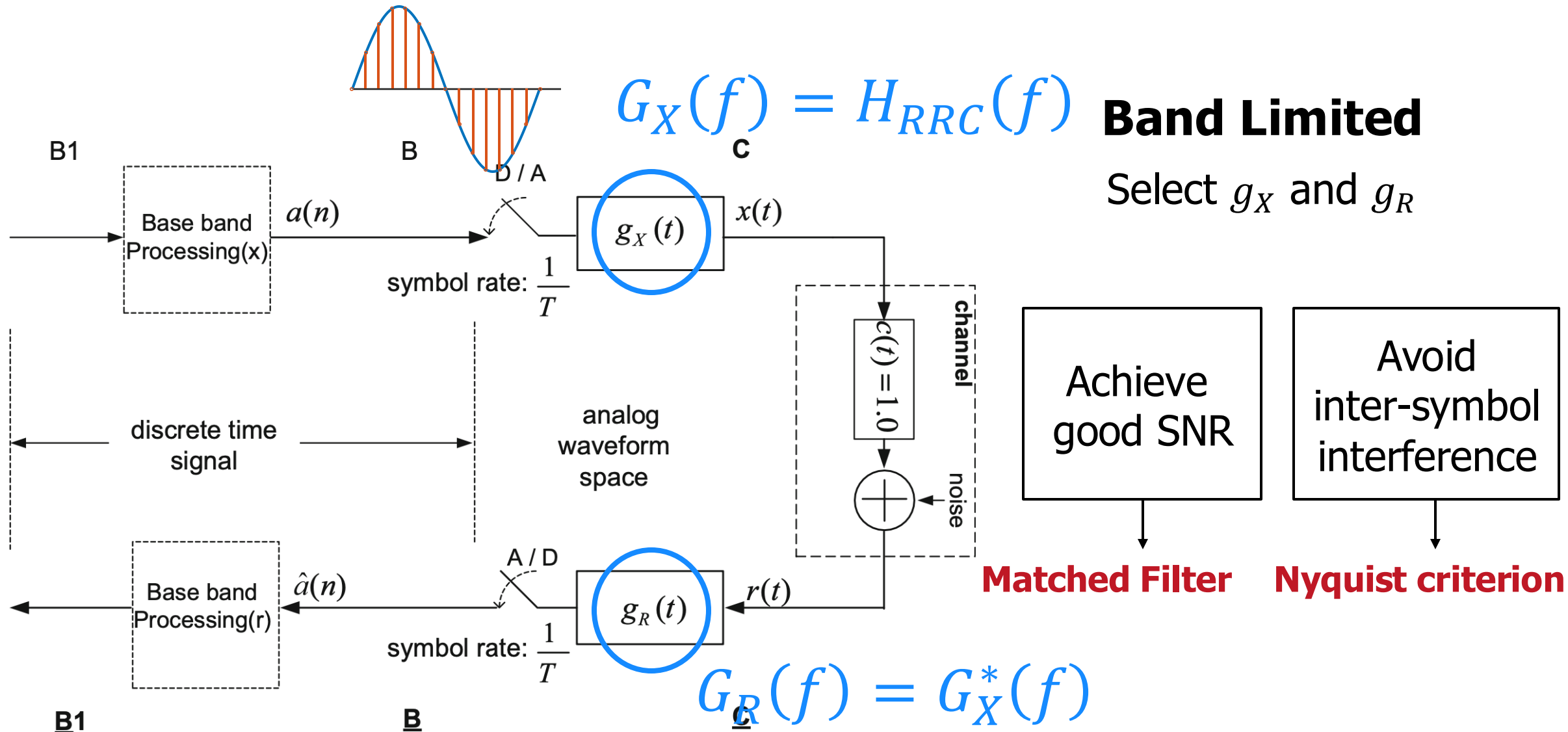
Up/Down-Sampling

Why change sampling frequency?

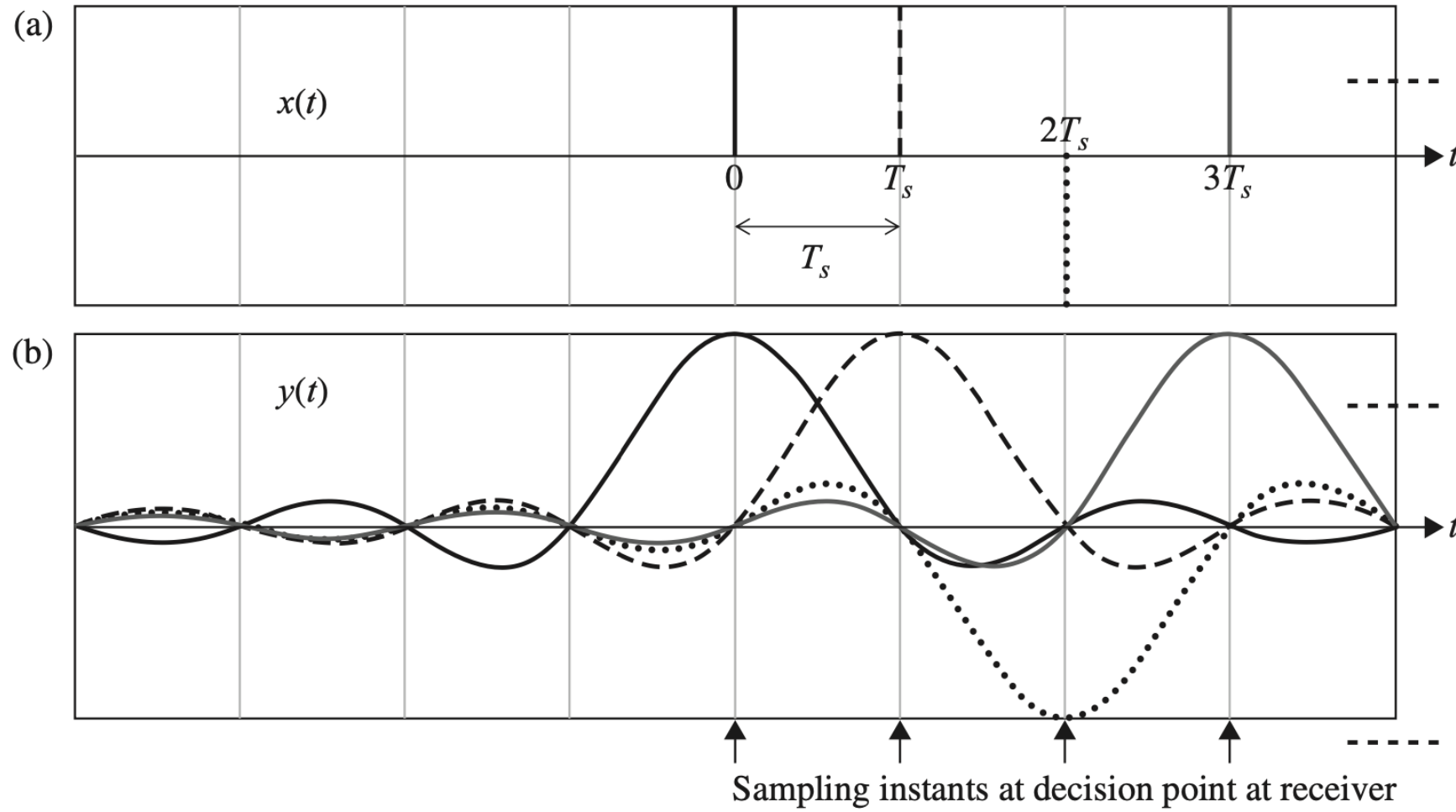
Compatibility between different devices



How to select the transmit and receive filter pair?



Nyquist Criterion for Zero ISI – Time Domain



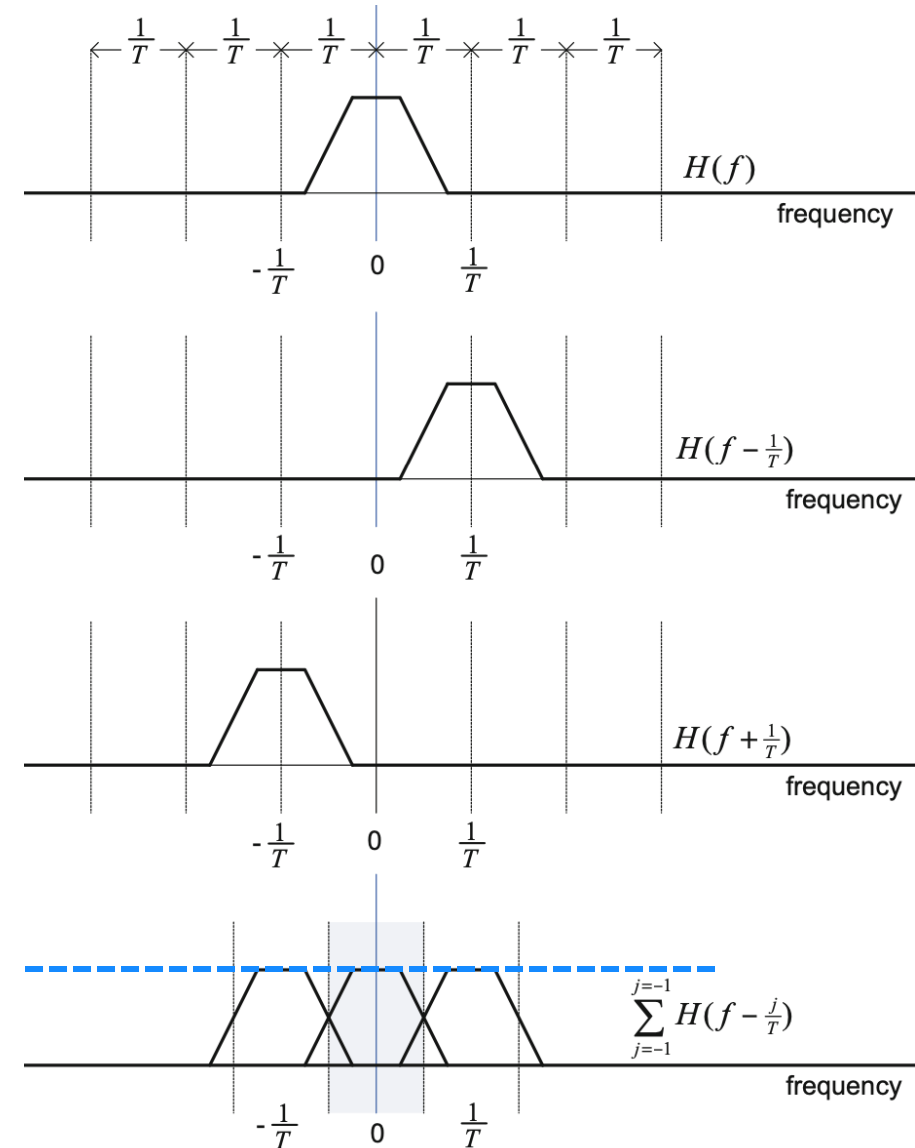
All other symbols happen to be zero

Frequency Domain Expression of Nyquist Criterion

$$\sum_{k=-\infty}^{+\infty} H\left(f - \frac{k}{T}\right) = T$$

The sum of all periodic repetitions of H is a constant

A lowpass satisfying this condition is also called a Nyquist lowpass



邑恆：在 Digital communications by John G. Proakis 第五版中的 605 頁有提供完整的推導。這本書在 Lab 331 中有，應該可以提供給同學做參考。

Raised Cosine Filtering

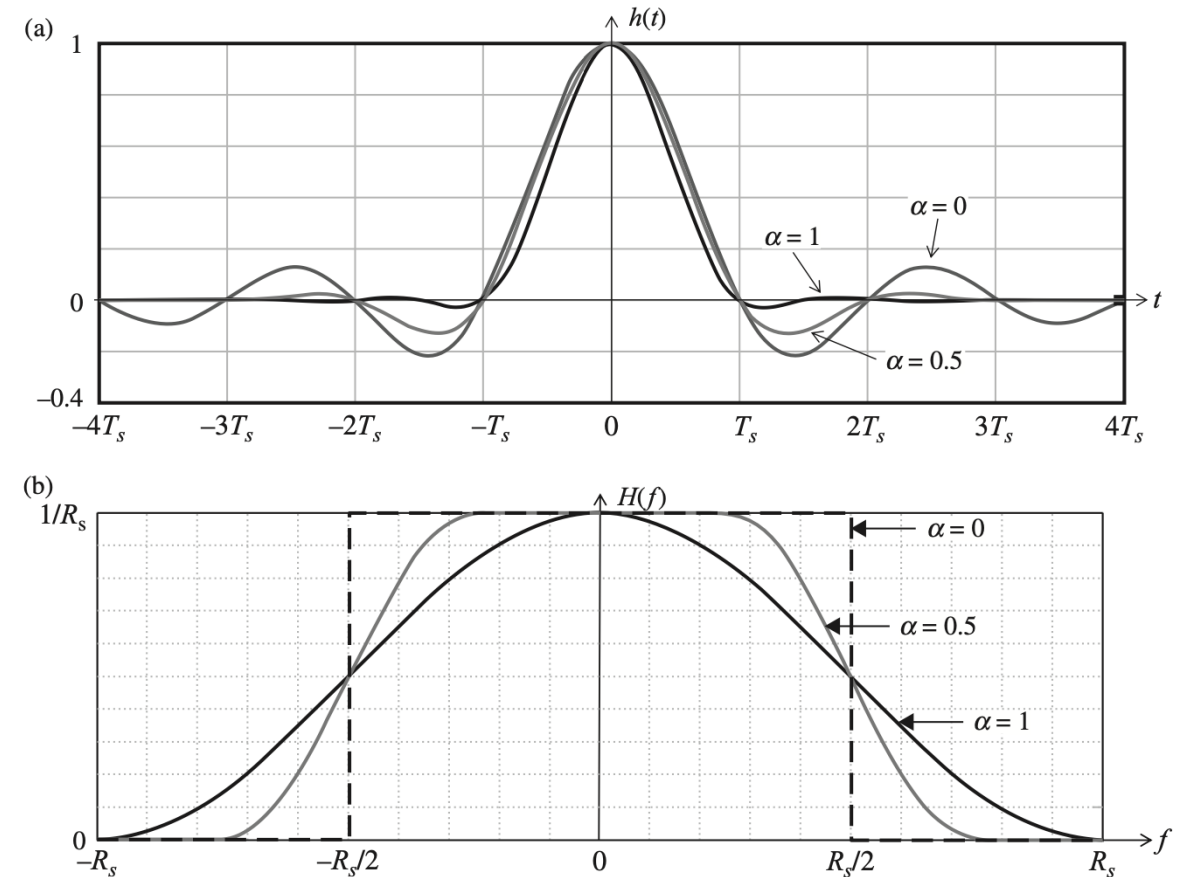
$$H(f) = \frac{1}{R_s} \times \begin{cases} 1, & |f| \leq f_1 \\ \frac{1}{2} \left[1 + \cos \left(\pi \frac{|f| - f_1}{f_2 - f_1} \right) \right], & f_1 \leq |f| \leq f_2 \\ 0, & |f| \geq f_2 \end{cases}$$

$$f_1 = (1 - \alpha)R_s/2; \quad f_2 = (1 + \alpha)R_s/2; \quad 0 \leq \alpha \leq 1$$

α : roll-off factor

gradual roll-off makes the raised cosine filter characteristic easier to approximate than the ideal Nyquist filter

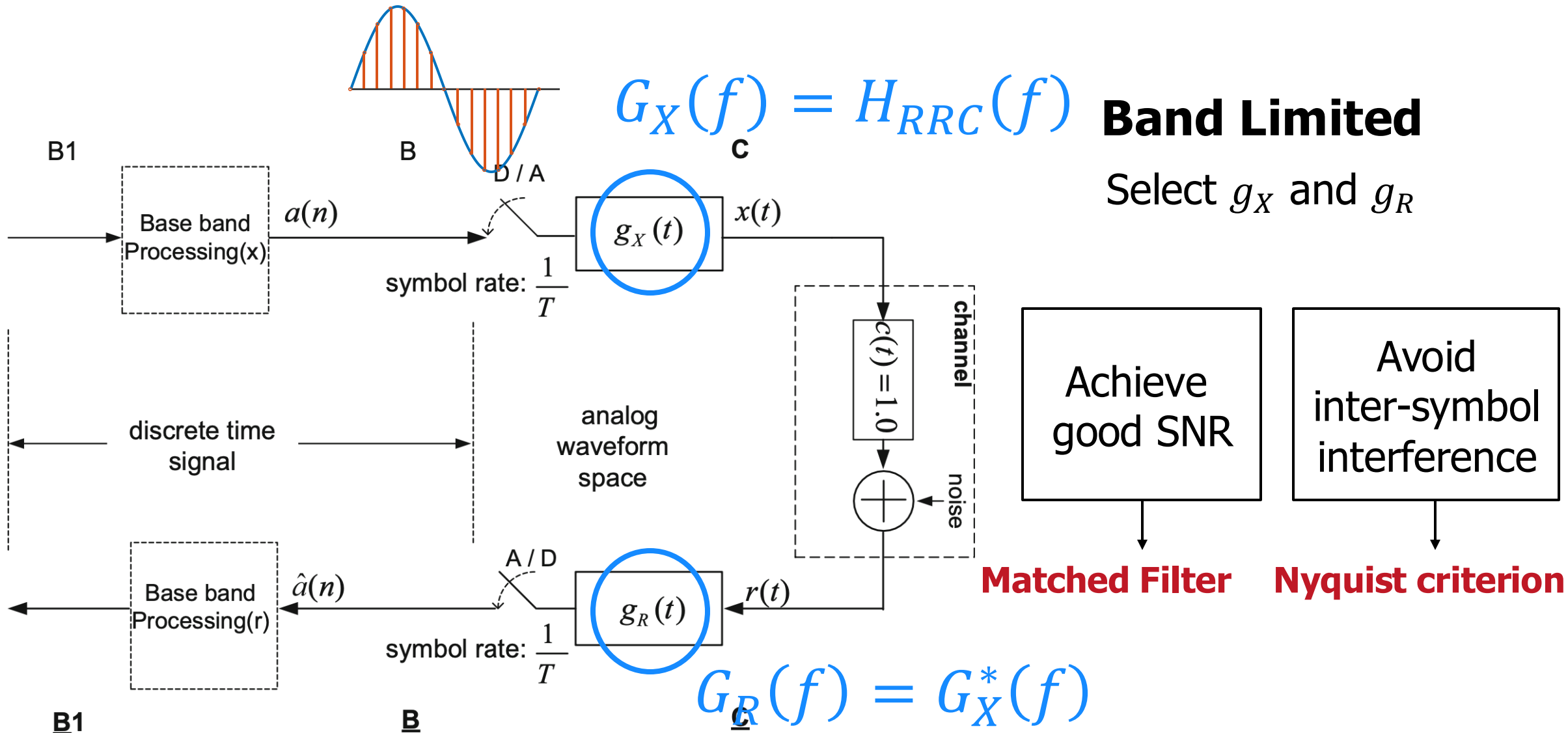
$\alpha = 0$: Nyquist Filtering



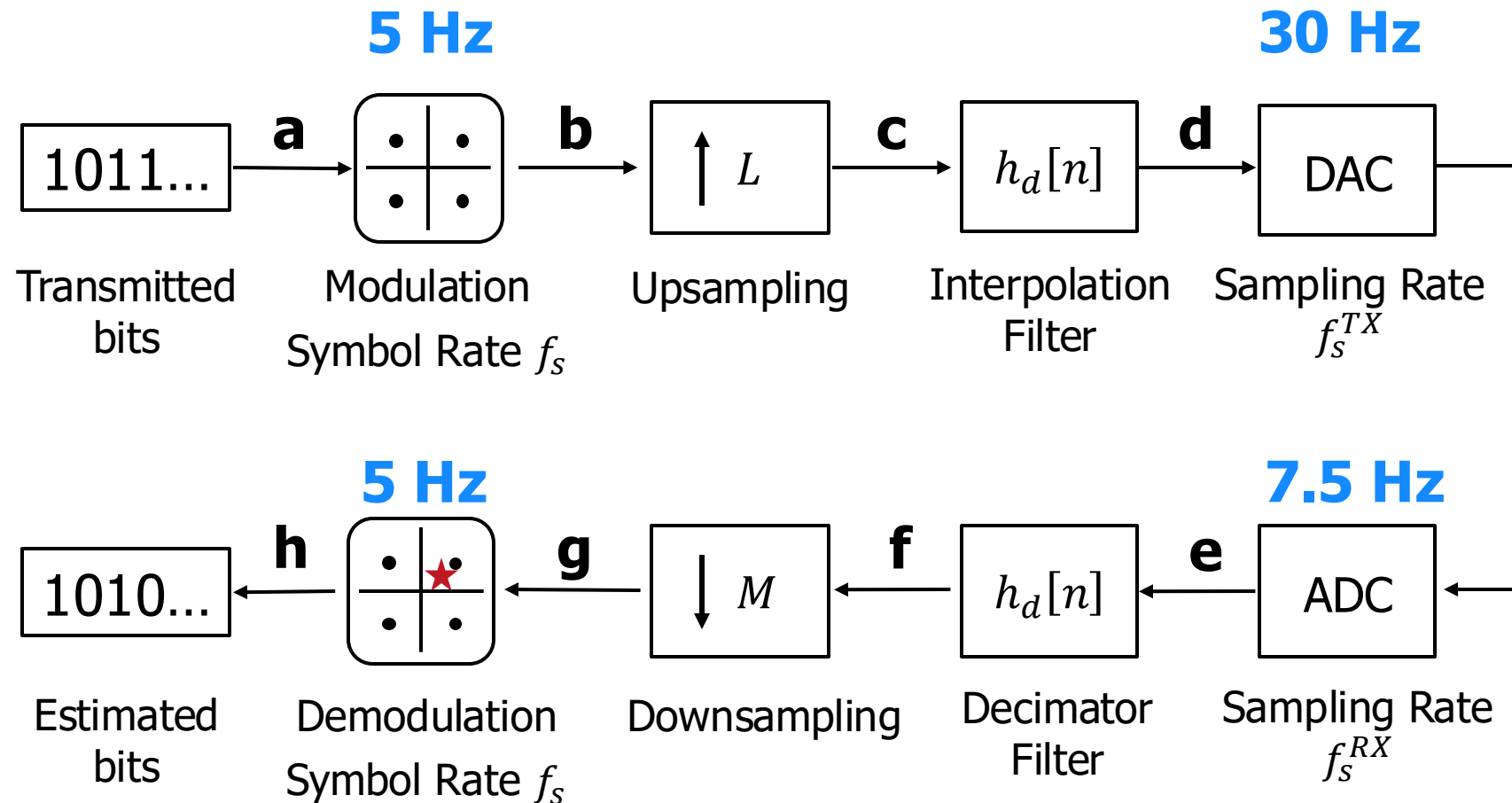
Does it satisfy the Nyquist Criterion?

Yes!

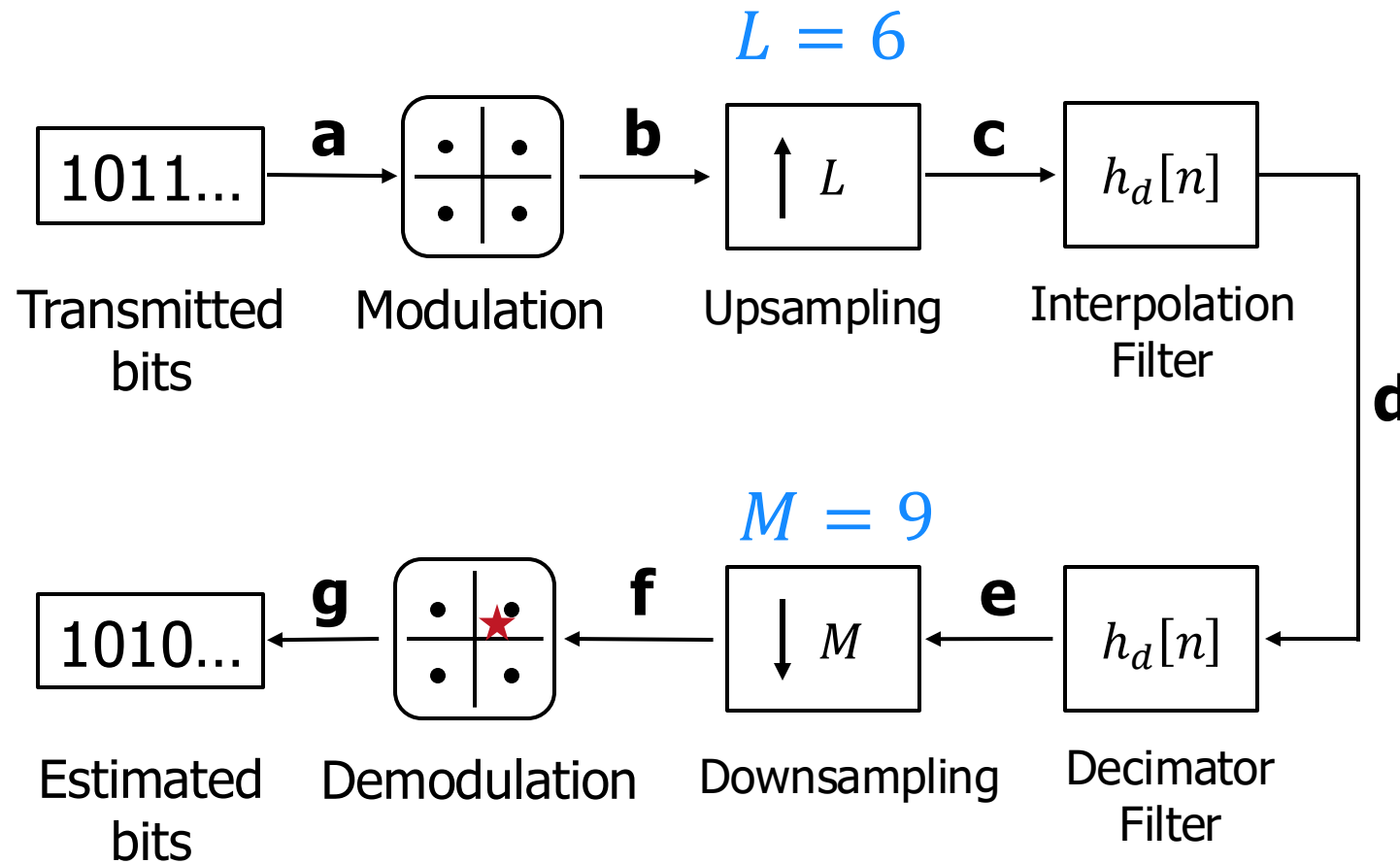
How to select the transmit and receive filter pair?



Lab 2 – Part 1 (Sampling rate adaptation)



Lab 2 – Part 2 (Aliasing)



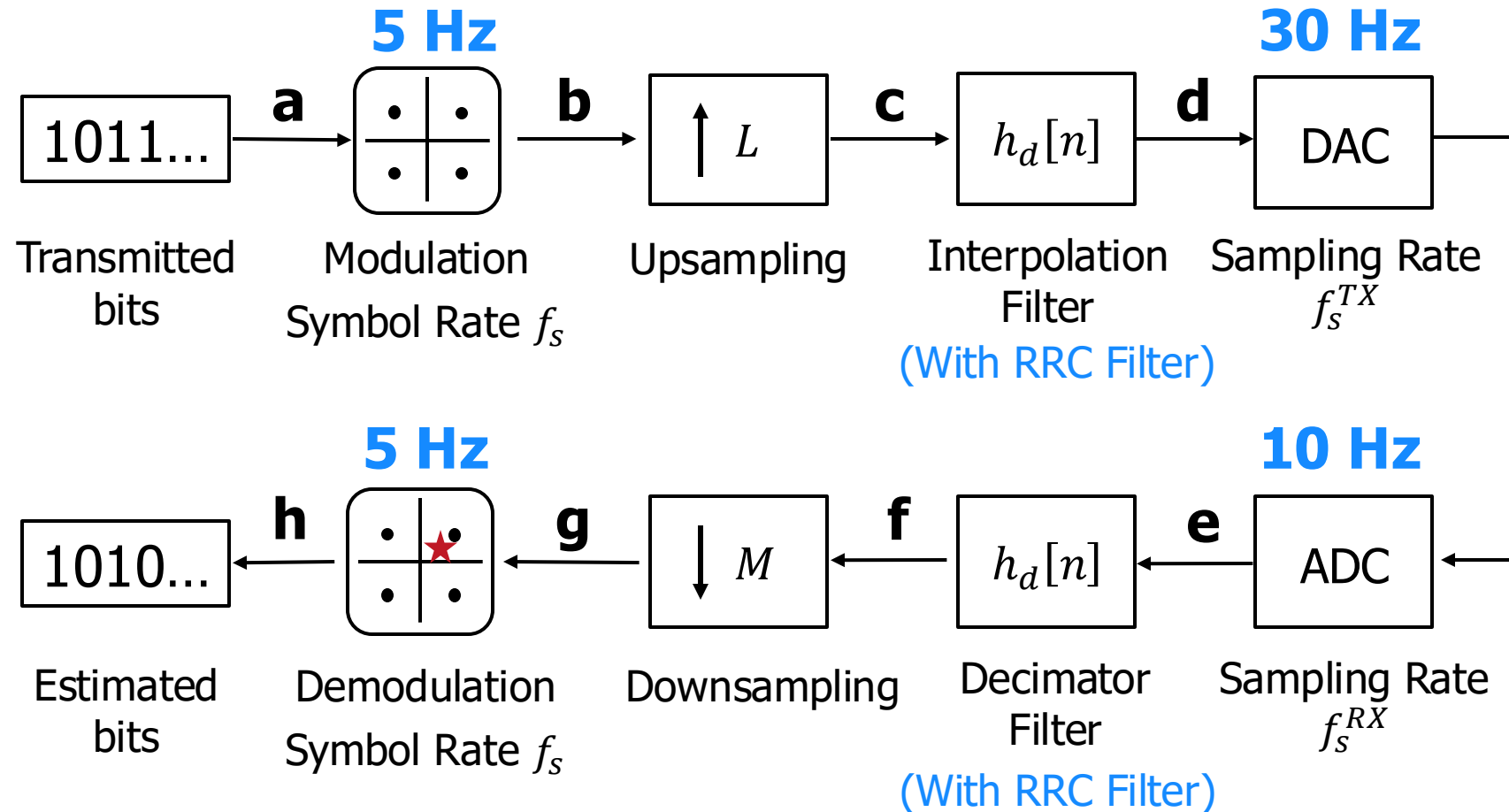
Transmitted bits:

[1,0,1,0,1,0,1,0,1,0,1,0,1,0,1,0];

[1,1,1,0,0,0,1,1,1,0,0,0,1,1,1,0,0,0];

Lab 2 – Part 3 (RRC filter)

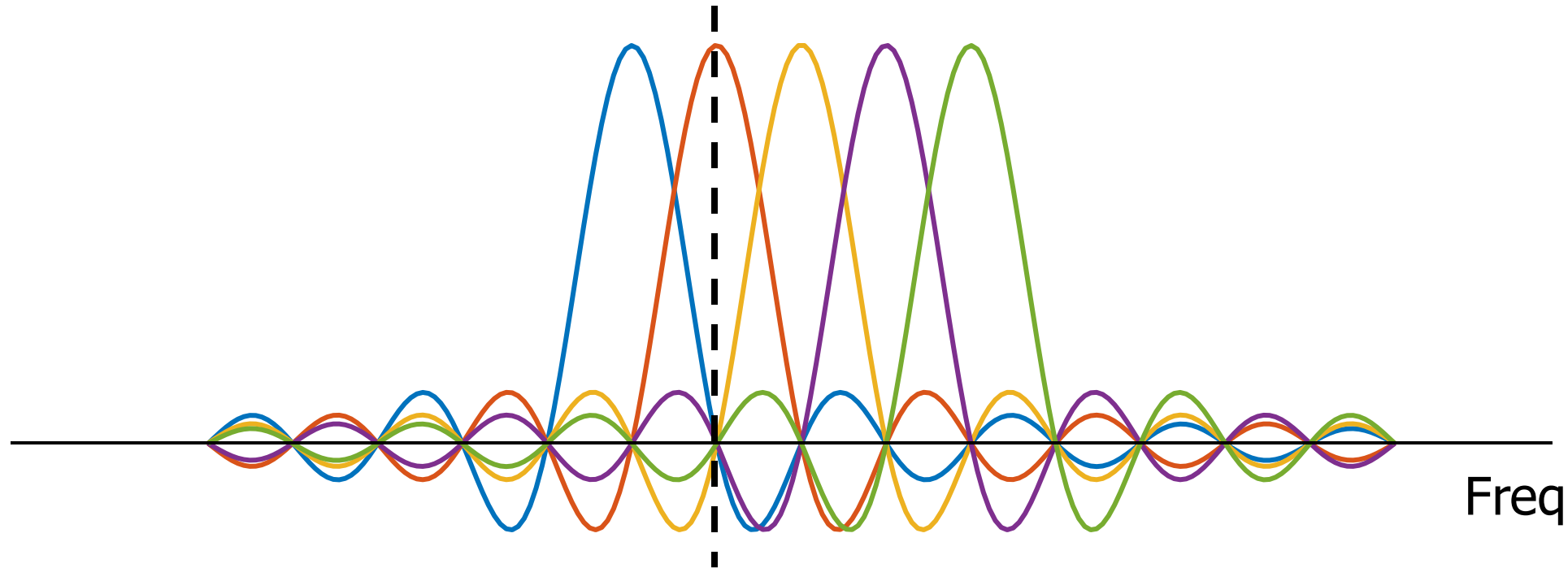
(continue the simulation in Part 1)



OFDM

Orthogonal Frequency-Division Multiplexing

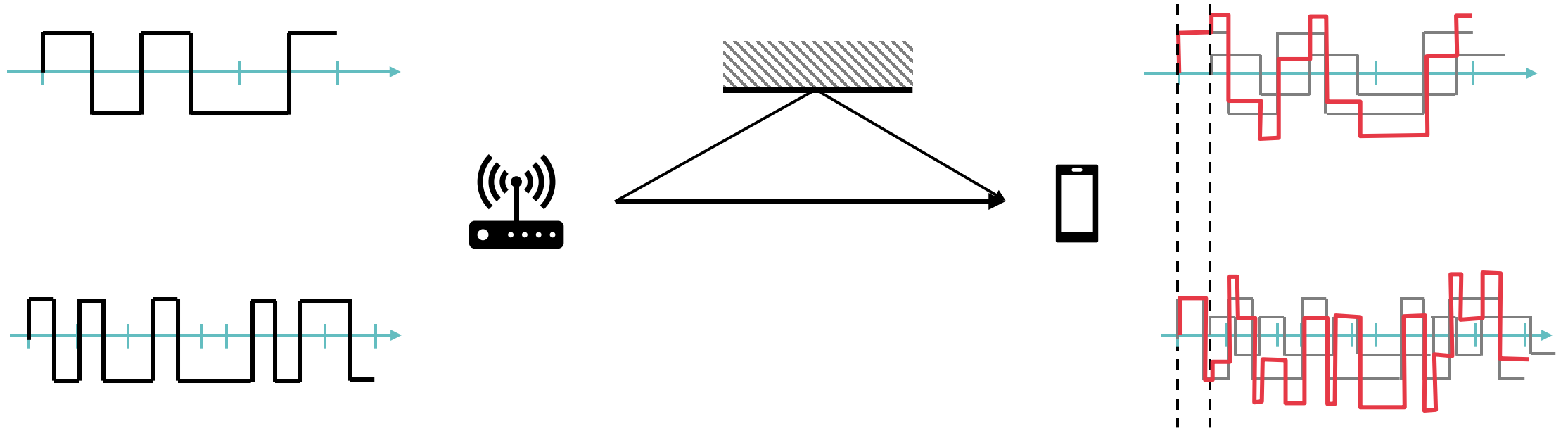
OFDM: Orthogonal Frequency-Division Multiplexing



Spectra of different modulated carriers overlap
But each carrier is in the spectral nulls of all other carriers

Wi-Fi (IEEE 802.11a, g, n, ac, ax), LTE

Motivation of OFDM

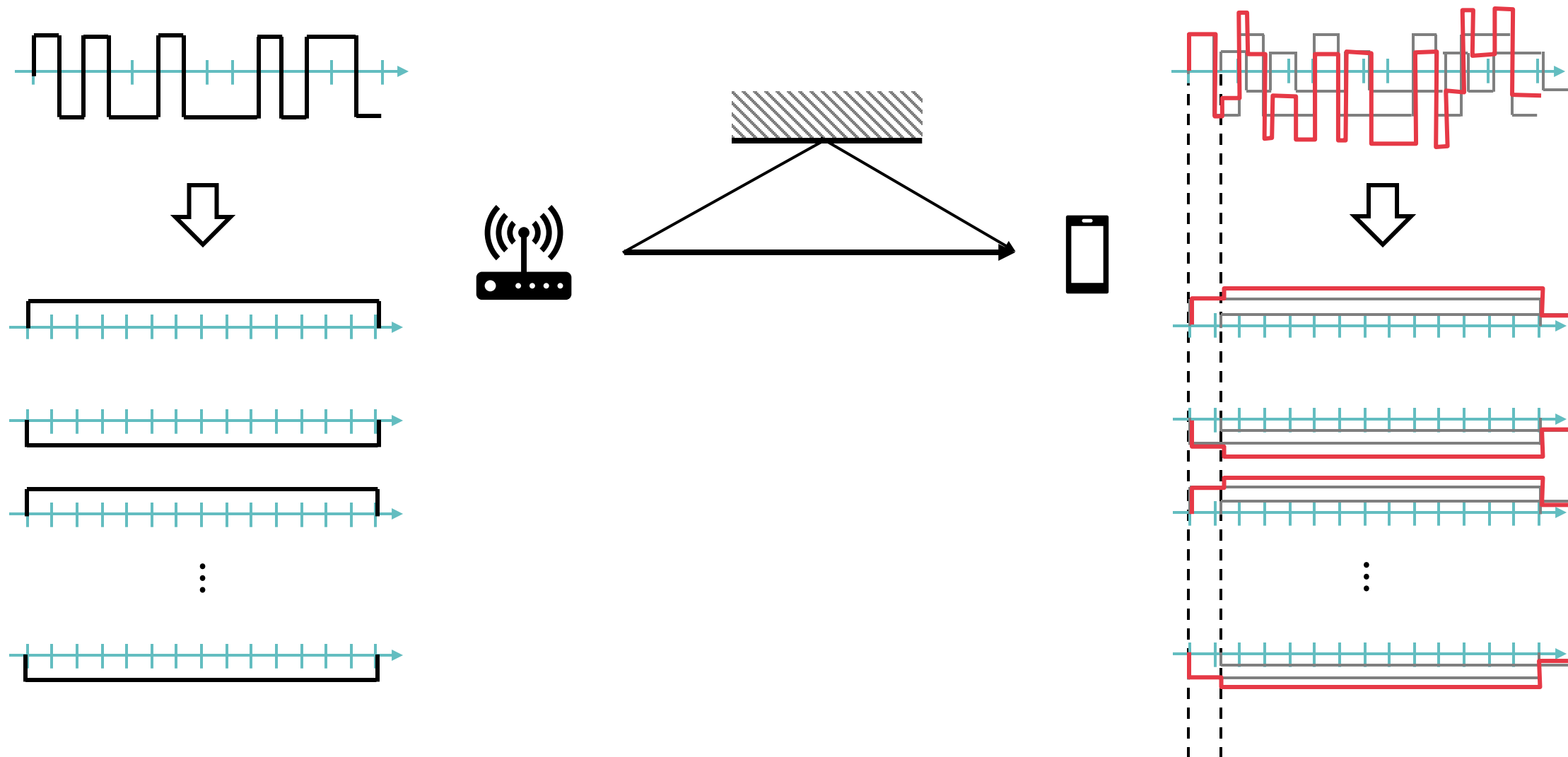


Higher data rate

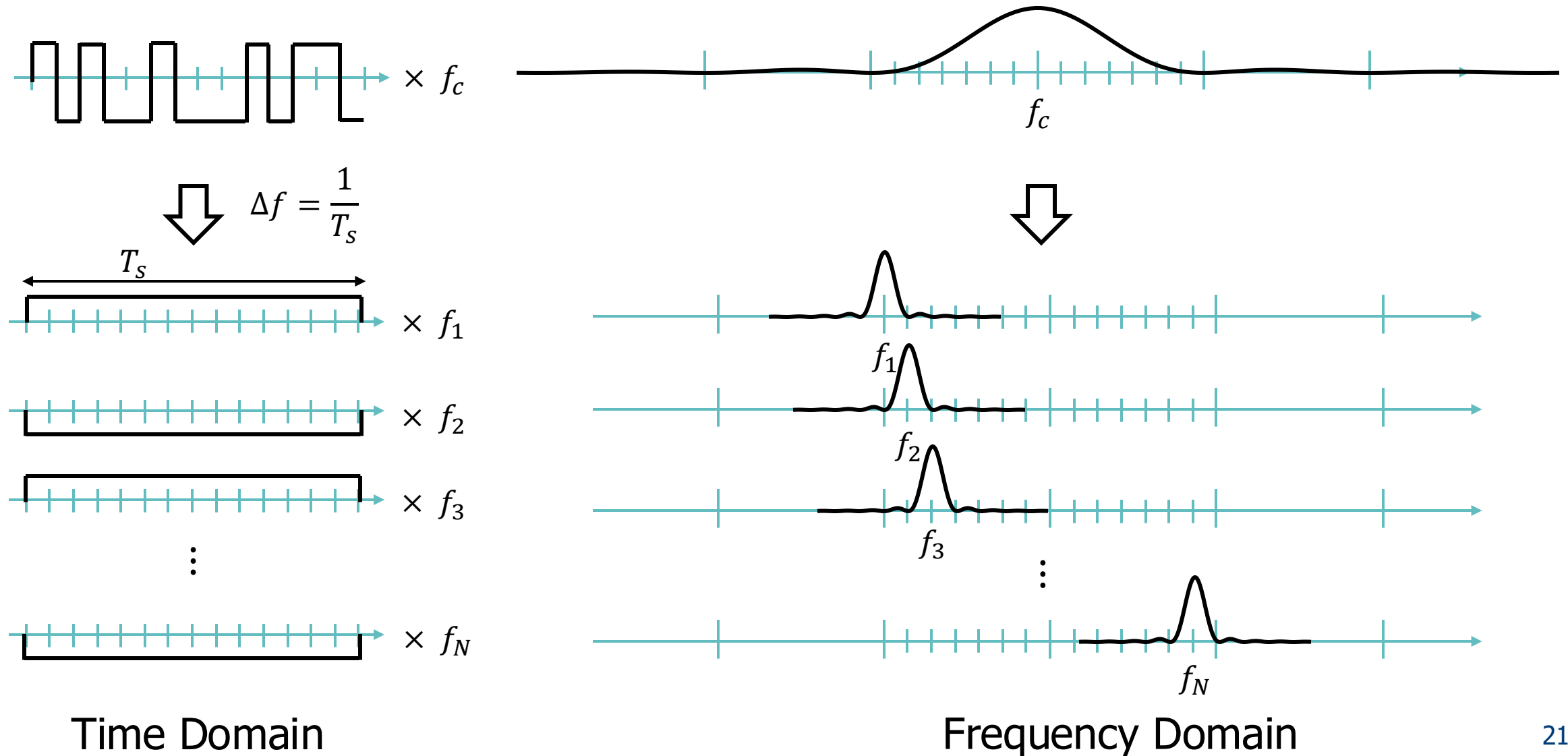
⇒ Shorter symbol time

⇒ Inter-symbol interference more significant

Motivation of OFDM



OFDM: Orthogonal Frequency-Division Multiplexing

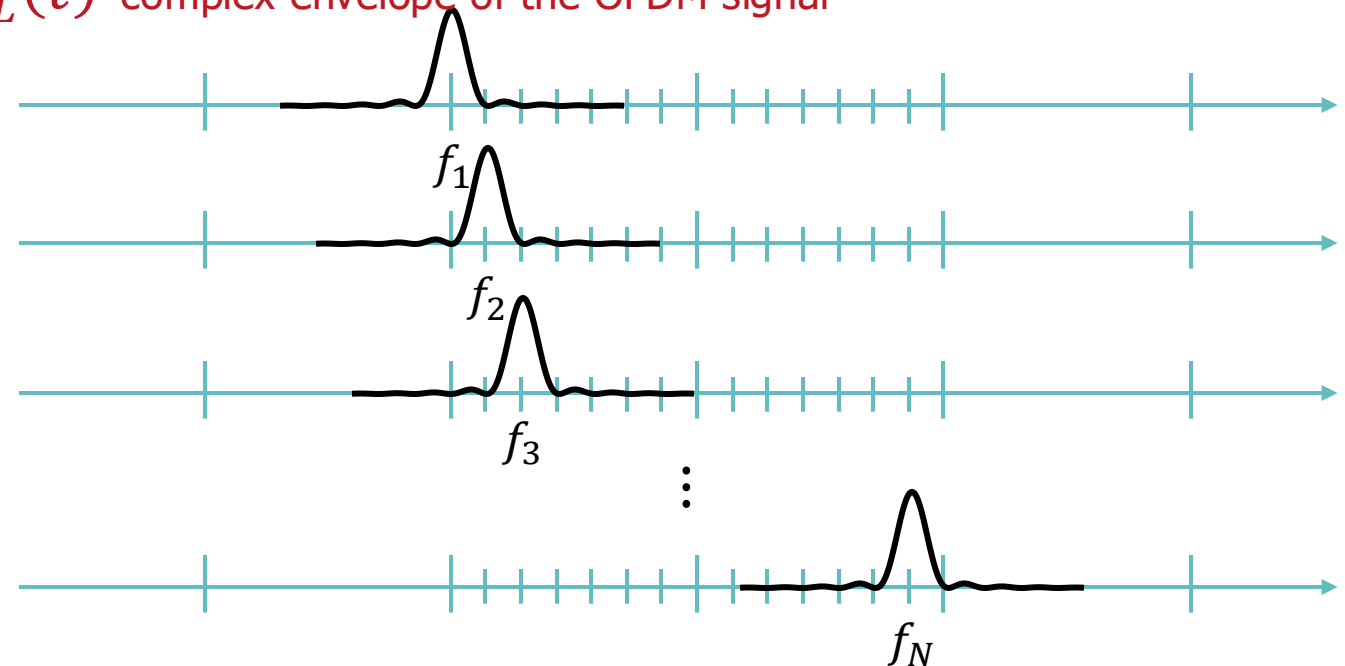
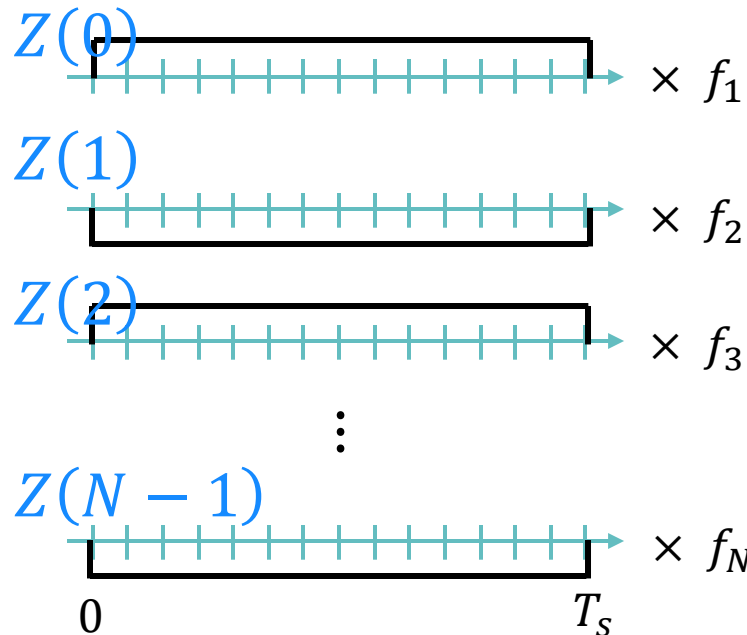


OFDM – Analog View

$$s_k(t) = \text{Re}\{Z(k)e^{j[2\pi(f_c+k\Delta f)t]}\}, \quad 0 \leq t < T_s, \quad \Delta f = 1/T_s$$

Modulated Subcarrier

$$s_{\text{OFDM signal}}(t) = \frac{1}{N} \sum_{k=0}^{N-1} s_k(t) = \text{Re} \left\{ \underbrace{\left\{ \frac{1}{N} \sum_{k=0}^{N-1} Z(k)e^{j2\pi k\Delta f t} \right\}}_{s_L(t) \text{ complex envelope of the OFDM signal}} e^{j2\pi f_c t} \right\}, \quad 0 \leq t < T_s$$



Complex envelope of the k -th subcarrier, constant during a symbol interval

OFDM – Digital View

$$s_k(t) = \text{Re}\{Z(k)e^{j[2\pi(f_c+k\Delta f)t]}\}, \quad 0 \leq t < T_s, \quad \Delta f = 1/T_s$$

Modulated Subcarrier

$$\underset{\text{OFDM signal}}{s(t)} = \frac{1}{N} \sum_{k=0}^{N-1} s_k(t) = \text{Re} \left\{ \underbrace{\left\{ \frac{1}{N} \sum_{k=0}^{N-1} Z(k)e^{j2\pi k\Delta f t} \right\}}_{s_L(t) \text{ complex envelope of the OFDM signal}} e^{j2\pi f_c t} \right\}, \quad 0 \leq t < T_s$$

Sampling time $\frac{T_s}{N}$ (sample at $t = \frac{nT_s}{N}, n = 0, 1, \dots, N-1$)

$$s_L(n) = \frac{1}{N} \sum_{k=0}^{N-1} Z(k)e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$

$$= \text{IDFT}\{z_k(t)\} \quad \longrightarrow \quad \text{Generate time sequence } s_L(n) \text{ from frequency samples}$$

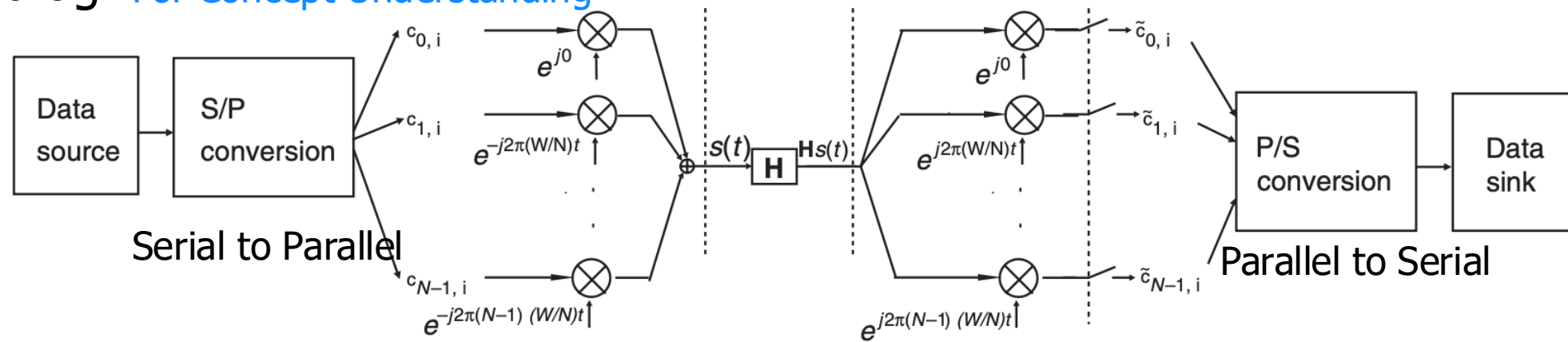
$$\Delta f = \frac{1}{T_s}, \quad \Delta t = \frac{T_s}{N}$$

$$Z(k) = \sum_{n=0}^{N-1} s_L(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$

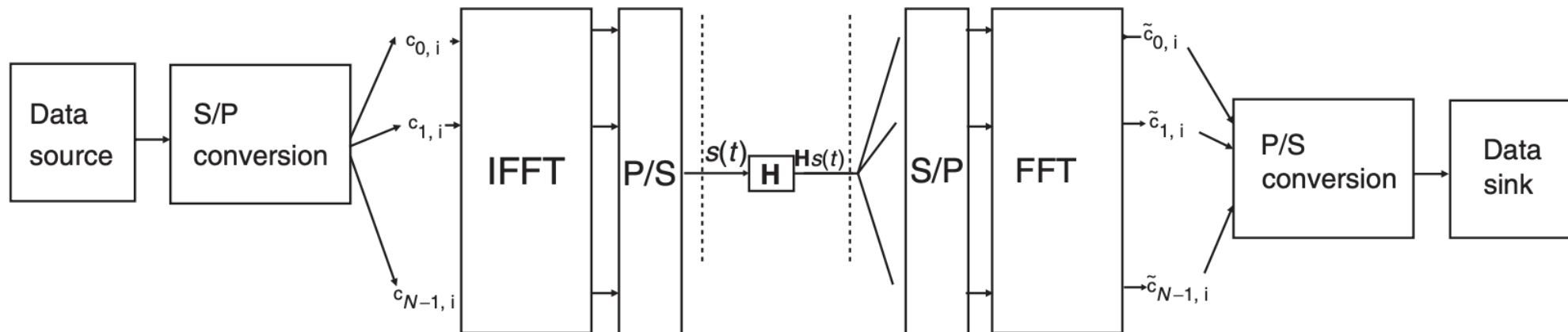
$$= \text{DFT}\{s_L(n)\}$$

OFDM Representation

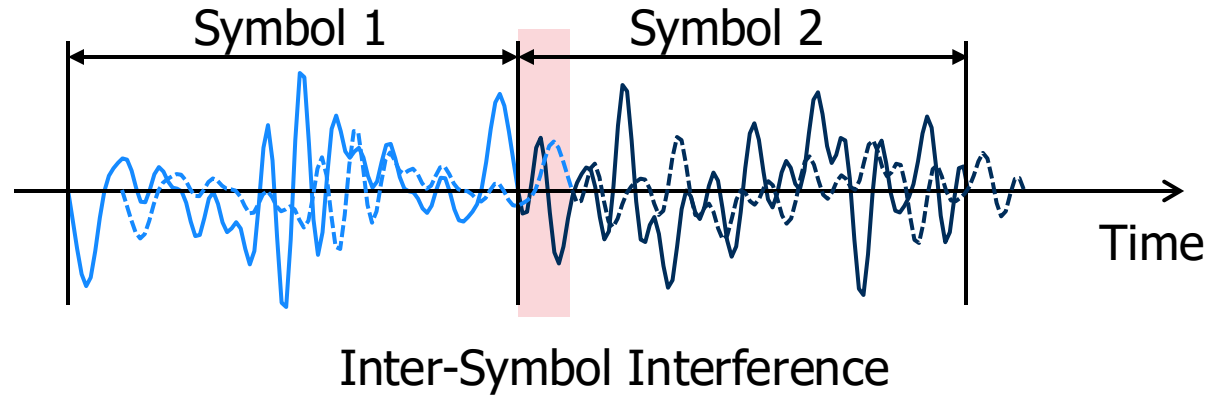
Analog For Concept Understanding



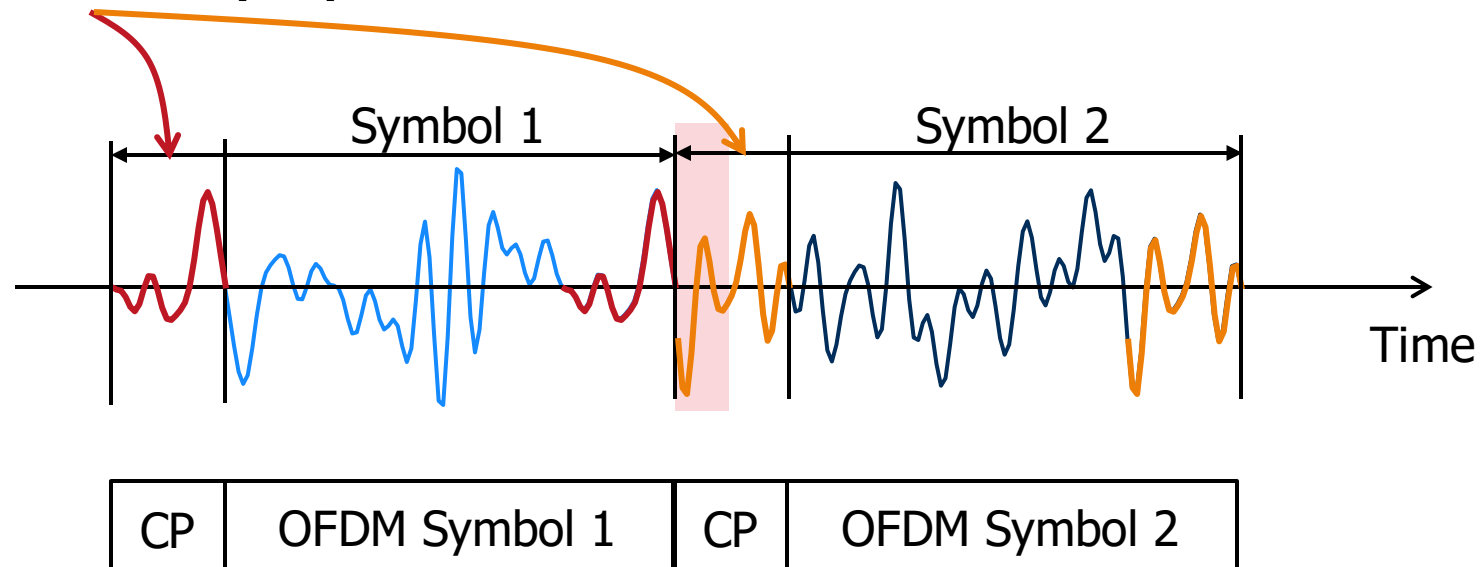
Digital Actual implementation



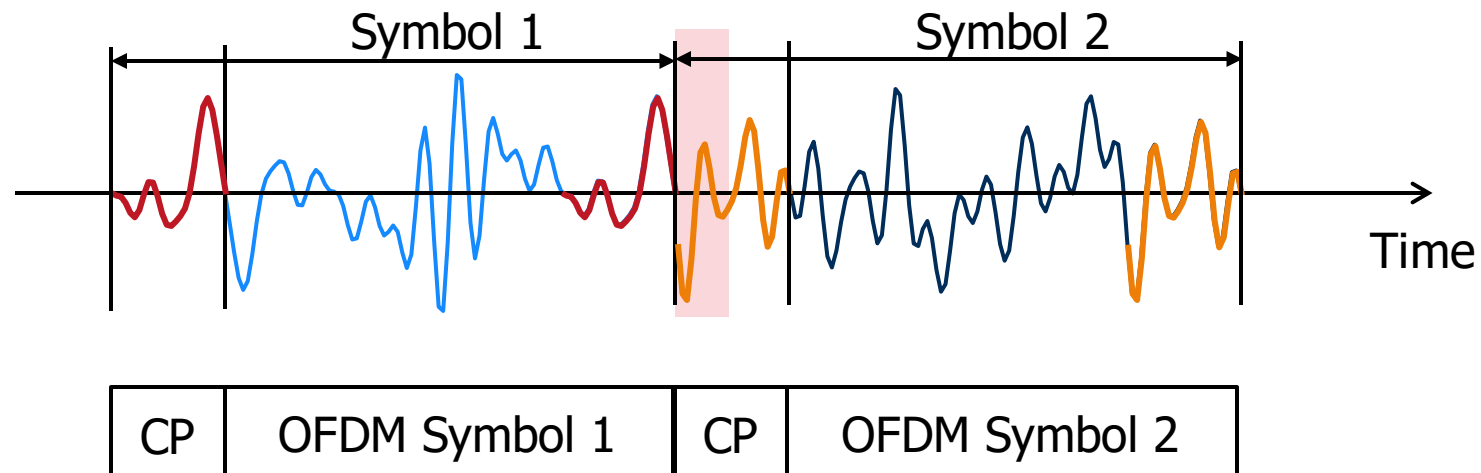
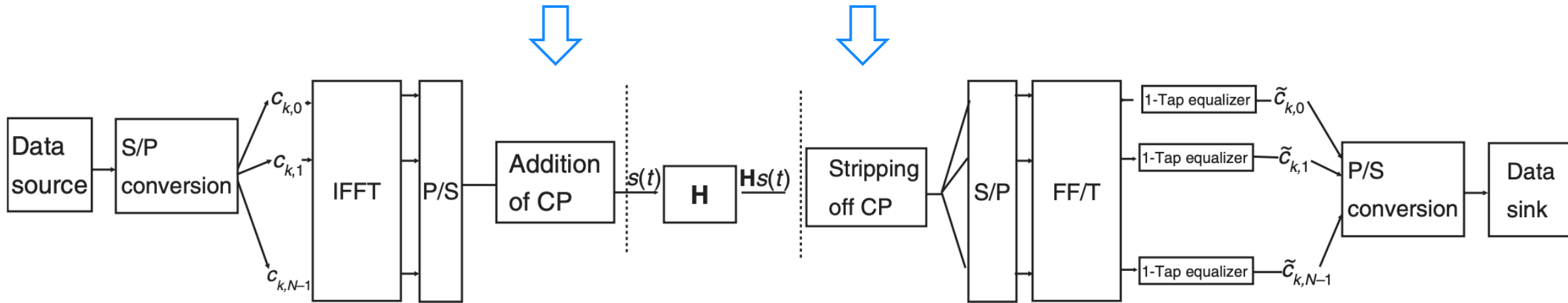
Cyclic Prefix to Combat Delay Dispersion



Cyclic Prefix (CP)



OFDM Implementation with CP



Reference - OFDM

- Andreas F. Molisch. *Wireless Communications*. John Wiley & Sons, 2011. [[NTU Library Link](#)]
 - Chap. 19: Orthogonal Frequency Division Multiplexing (OFDM)
- Sung-Moon Michael Yang. *Modern Digital Radio Communication Signals and Systems*. Springer, 2020. [[NTU Library Link](#)]
 - Chap. 5: OFDM Signals and Systems
- Tri T. Ha. *Theory and Design of Digital Communication Systems*. Cambridge University Press, 2010.
 - 6.18 Orthogonal frequency division multiplexing (OFDM)
 - 7.23 OFDM demodulation
- YouTube Playlist: Orthogonal Frequency Division Multiplexing (OFDM) by Iain Explains Signals, Systems, and Digital Comms [[Link](#)]

Paper Debate

Debate Format

20 minutes	Defense Team
10 minutes	Offense Team
5 minutes	Preparation time
10 minutes	Follow up arguments
5 minutes	Questions and comments from class

Timing will be strictly enforced!

Paper 1: OPTICS: Human Activity-Aware Integrated Optical Wireless Communication and Sensing

<https://forms.gle/2jFMuYXXQSGVzj9s6>



Only the audience vote!

Presenters: Upload your slides
(With 分工表)