

VII. Other Time Frequency Distributions (II)

The trend of time-frequency analysis in recent years:

- (1) S transform and its generalization
- (2) Time-frequency reassignment and the synchrosqueezing transform
- (3) Time-variant signal expansion (Compressive sensing)
- (4) Improvement for the Hilbert-Huang transform

VII-A S Transform

(Modification from the Gabor transform)

$$S_x(t, f) = |f| \int_{-\infty}^{\infty} x(\tau) \exp\left[-\pi(t-\tau)^2 f^2\right] \exp(-j2\pi f\tau) d\tau$$

closely related to the wavelet transform

advantages and disadvantages

[Ref] R. G. Stockwell, L. Mansinha, and R. P. Lowe, “Localization of the complex spectrum: the S transform,” *IEEE Trans. Signal Processing*, vol. 44, no. 4, pp. 998–1001, Apr. 1996.

S transform 和 Gabor transform 相似。

但是 Gaussian window 的寬度會隨著 f 而改變

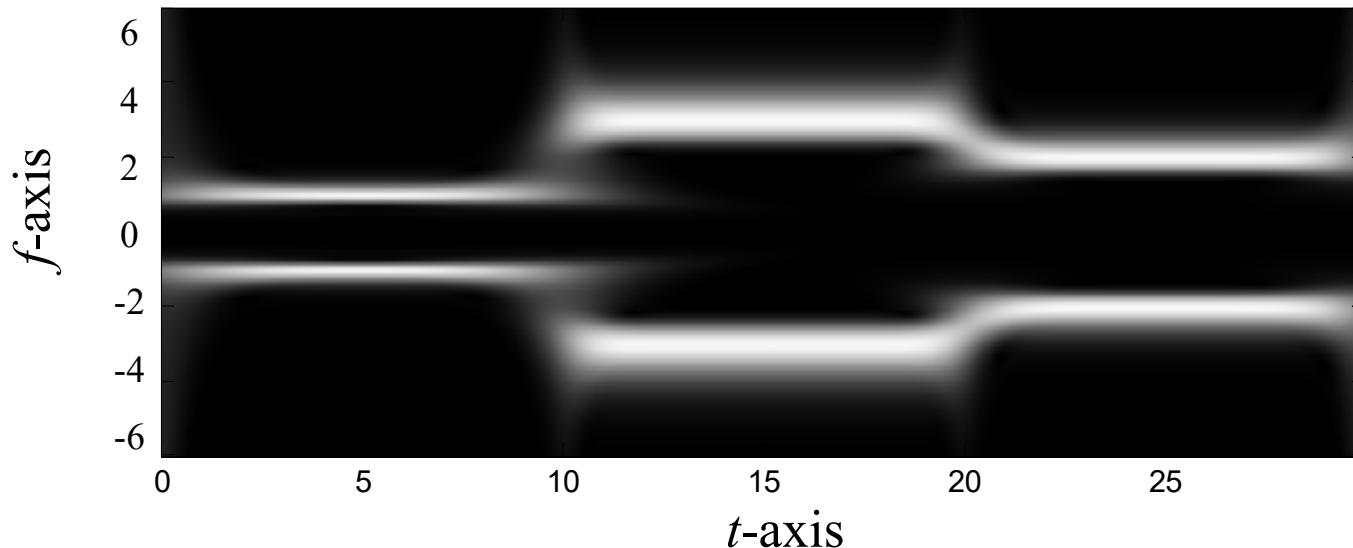
$$w(t) = \exp[-\pi t^2]$$

$$w(t) = |f| \exp[-\pi t^2 f^2]$$

低頻： worse time resolution, better frequency resolution

高頻： better time resolution, worse frequency resolution

The result of the S transform (compared with page 95)



- General form

$$S_x(t, f) = |s(f)| \int_{-\infty}^{\infty} x(\tau) \exp\left[-\pi(t-\tau)^2 s^2(f)\right] \exp(-j2\pi f\tau) d\tau$$

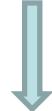
$s(f)$ increases with f

C. R. Pinnegar and L. Mansinha, “The S-transform with windows of arbitrary and varying shape,” *Geophysics*, vol. 68, pp. 381-385, 2003.

Fast algorithm of the S transform

When f is fixed, the S transform can be expressed as a convolution form:

$$S_x(t, f) = |s(f)| \int_{-\infty}^{\infty} x(\tau) \exp\left[-\pi(t-\tau)^2 s^2(f)\right] \exp(-j2\pi f\tau) d\tau$$



$$S_x(t, f) = |s(f)| \left(x(t) \exp(-j2\pi f t) *_{\substack{\text{convolution} \\ \text{along } t\text{-axis}}} \exp\left[-\pi t^2 s^2(f)\right] \right)$$

(for every fixed f)

Remember: $g(t) * h(t) = \int g(\tau)h(t-\tau)d\tau$

Q: Can we use the FFT-based method on page 121 to implement the S transform?

VII-B Generalized Spectrogram

[Ref] P. Boggiatto, G. De Donno, and A. Oliaro, "Two window spectrogram and their integrals," *Advances and Applications*, vol. 205, pp. 251-268, 2009.

Generalized spectrogram: $SP_{x,w_1,w_2}(t,f) = G_{x,w_1}(t,f)G_{x,w_2}^*(t,f)$

$$G_{x,w_1}(t,f) = \int_{-\infty}^{\infty} w_1(t-\tau)x(\tau)e^{-j2\pi f\tau}d\tau$$

$$G_{x,w_2}(t,f) = \int_{-\infty}^{\infty} w_2(t-\tau)x(\tau)e^{-j2\pi f\tau}d\tau$$

Original spectrogram: $w_1(t) = w_2(t)$

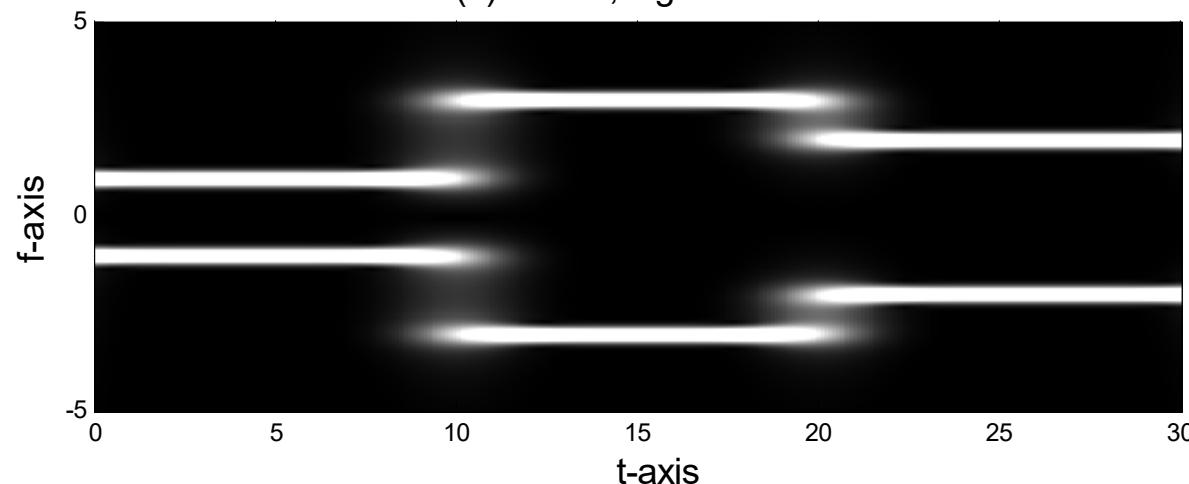
To achieve better clarity, $w_1(t)$ can be chosen as a **wider window**,
 $w_2(t)$ can be chosen as a **narrower window**.

$x(t) = \cos(2\pi t)$ when $t < 10$,

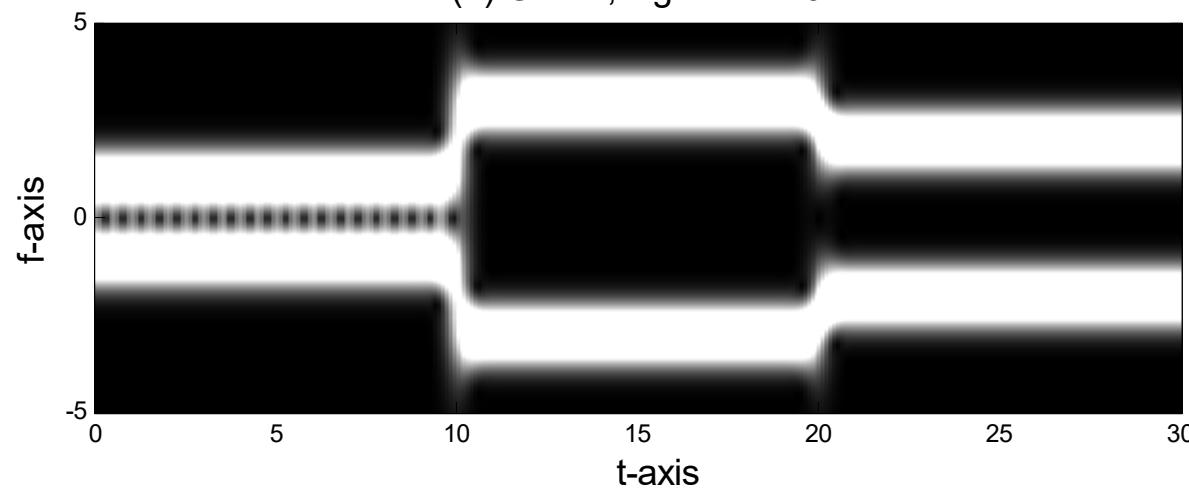
$x(t) = \cos(6\pi t)$ when $10 \leq t < 20$,

$x(t) = \cos(4\pi t)$ when $t \geq 20$

(a) Gabor, sigma = 0.1



(b) Gabor, sigma = 1.6

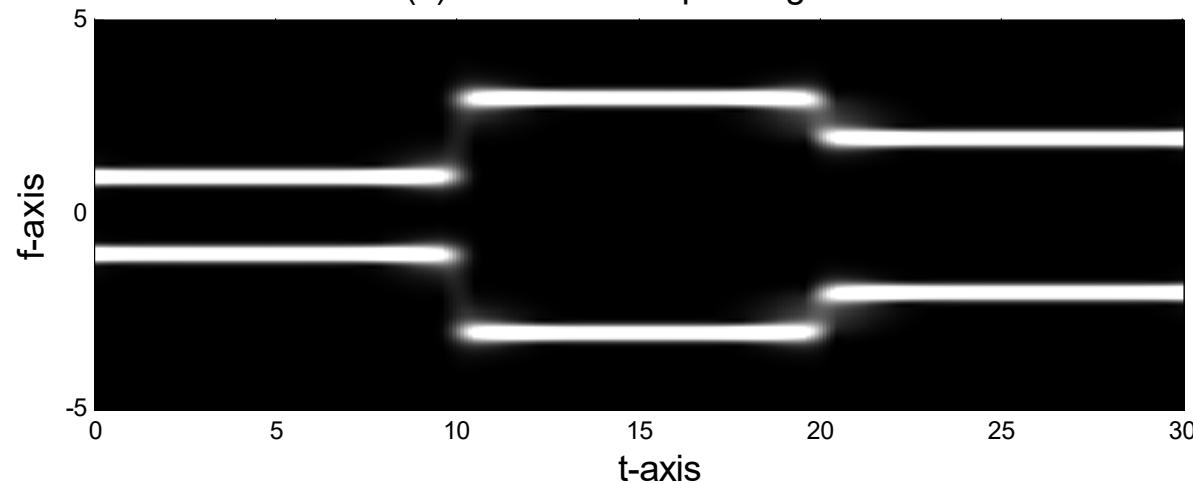


$x(t) = \cos(2\pi t)$ when $t < 10$,

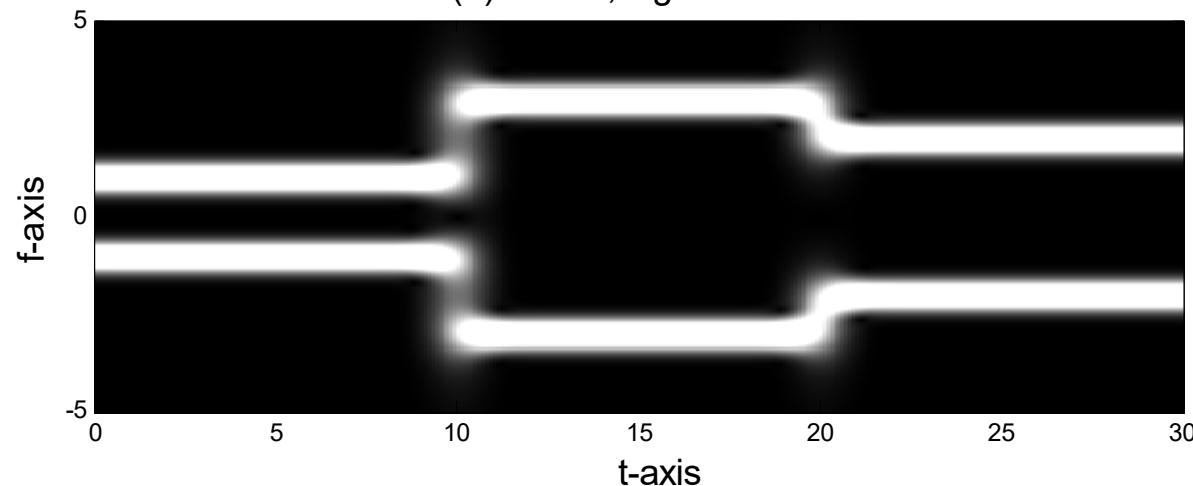
$x(t) = \cos(6\pi t)$ when $10 \leq t < 20$,

$x(t) = \cos(4\pi t)$ when $t \geq 20$

(c) Generalized spectrogram



(d) Gabor, sigma = 0.4



Generalized spectrogram: $SP_{x,w_1,w_2}(t,f) = G_{x,w_1}(t,f)G_{x,w_2}^*(t,f)$

Further Generalization for the spectrogram:

$$SP_{x,w_1,w_2}(t,f) = G_{x,w_1}^\alpha(t,f)\overline{G_{x,w_2}^\beta(t,f)}$$

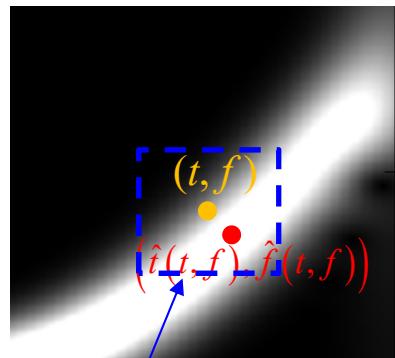
or

$$SP_{x,w_1,w_2}(t,f) = |G_{x,w_1}(t,f)|^\alpha |G_{x,w_2}(t,f)|^\beta$$

VII-C Reassignment Method

After computing the time-frequency distribution, we can use the following way to **make the energy even more concentrated**.

(1) First, estimate the offset.



$$\hat{t}(t, f) = \frac{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} u \cdot \varphi(u - t, v - f) \cdot X(u, v) dudv}{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \varphi(u - t, v - f) \cdot X(u, v) dudv}$$

$$\hat{f}(t, f) = \frac{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} v \cdot \varphi(u - t, v - f) \cdot X(u, v) dudv}{\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \varphi(u - t, v - f) \cdot X(u, v) dudv}$$

$$\varphi(u - t, v - f)$$

$X(t, f)$: time-frequency analysis (STFT, WDF...) of $x(t)$,

$$\varphi(u, v) = 1 \text{ when } |u|, |v| < B$$

$$\varphi(u, v) = 0 \text{ otherwise}$$

(2) Then, shift the time frequency distribution at (t, f) to $(\hat{t}(t, f), \hat{f}(t, f))$

(2) Then, shift the time frequency distribution at (t, f) to $(\hat{t}(t, f), \hat{f}(t, f))$

$$\hat{X}(t, f) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} X(t_1, f_1) \delta(t - \hat{t}(t_1, f_1)) \delta(f - \hat{f}(t_1, f_1)) dt_1 df_1$$

References

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- [2] F. Auger, P. Flandrin, Y.T. Lin, S. McLaughlin, S. Meignen, T. Oberlin, and H.T. Wu, “Time-frequency reassignment and synchrosqueezing: An overview,” *IEEE Signal Processing Magazine*, vol. 30, issue 6, pp. 32-41, 2013.

PS: 感謝 2017 年修課的盧德晏同學

Synchrosqueezing Transform

$$\hat{X}(t, f) = \int_{-\infty}^{\infty} X(t, f_1) \delta(f - \hat{f}(t, f_1)) df_1$$

It only moves along f -axis.

(simplification form of the reassignment method)

C. Li and M. Liang, “A generalized synchrosqueezing transform for enhancing signal time–frequency representation,” *Signal Processing*, vol. 92, issue 9, 2264-2274, 2012.

T. Oberlin, S. Meignen, and V. Perrier. “The Fourier-based synchrosqueezing transform,” in *IEEE Int. Conf. Acoustics, Speech and Signal Processing*, pp. 315-319, 2014.

VII-D Basis Expansion Time-Frequency Analysis

就如同

- Fourier series: $\varphi_m(t) = \exp(j2\pi f_m t)$, $x(t) \approx \sum_{m=1}^M a_m \exp(j2\pi f_m t)$

$$a_m = \frac{\langle x(t), \varphi_m(t) \rangle}{\langle \varphi_m(t), \varphi_m(t) \rangle} = \frac{1}{T} \int_0^T x(t) \exp(-j2\pi f_m t) dt \quad f_m = m/T$$

部分的 Time-Frequency Analysis 也是意圖要將 signal 表示成如下的型態

$$x(t) \approx \sum_{m=1}^M a_m \varphi_m(t)$$

並且要求在 M 固定的情形下，

$$\text{approximation error} = \int_{-\infty}^{\infty} \left| x(t) - \sum_{m=1}^M a_m \varphi_m(t) \right|^2 dt \quad \text{為最小}$$

將 $\varphi_m(t)$ 一般化，不同的 basis 之間不只是有 frequency 的差異

(1) Three Parameter Atoms

$$x(t) \approx \sum a_{t_0, f_0, \sigma} \varphi_{t_0, f_0, \sigma}(t)$$

$$\varphi_{t_0, f_0, \sigma}(t) = \frac{2^{1/4}}{\sigma^{1/2}} \exp(j2\pi f_0 t) \exp\left(-\frac{\pi(t-t_0)^2}{\sigma^2}\right)$$

3 parameters: t_0 controls the central time
 f_0 controls the frequency
 σ controls the scaling factor

[Ref] S. G. Mallat and Z. Zhang, “Matching pursuits with time-frequency dictionaries,” *IEEE Trans. Signal Processing*, vol. 41, no. 12, pp. 3397-3415, Dec. 1993.

Since $\varphi_{t_0, f_0, \sigma}(t)$ are not orthogonal, $a_{t_0, f_0, \sigma}$ should be determined by a **matching pursuit process**.

(2) Four Parameter Atoms (Chirplet)

$$x(t) \approx \sum a_{t_0, f_0, \sigma, \eta} \varphi_{t_0, f_0, \sigma, \eta}(t)$$

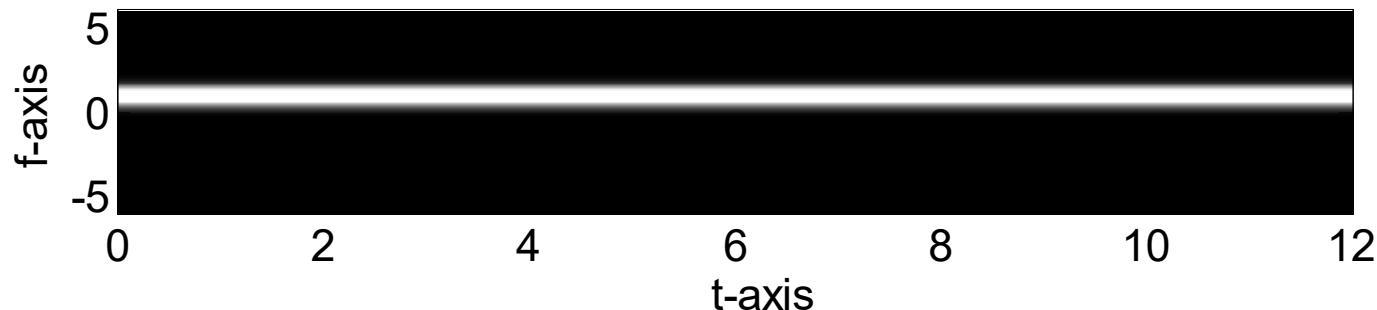
$$\varphi_{t_0, f_0, \sigma}(t) = \frac{2^{1/4}}{\sigma^{1/2}} \exp(j2\pi(f_0 t + \frac{\eta}{2} t^2) - \frac{\pi(t-t_0)^2}{\sigma^2})$$

- 4 parameters:
 - t_0 controls the central time
 - f_0 controls the initial frequency
 - σ controls the scaling factor
 - η controls the chirp rate

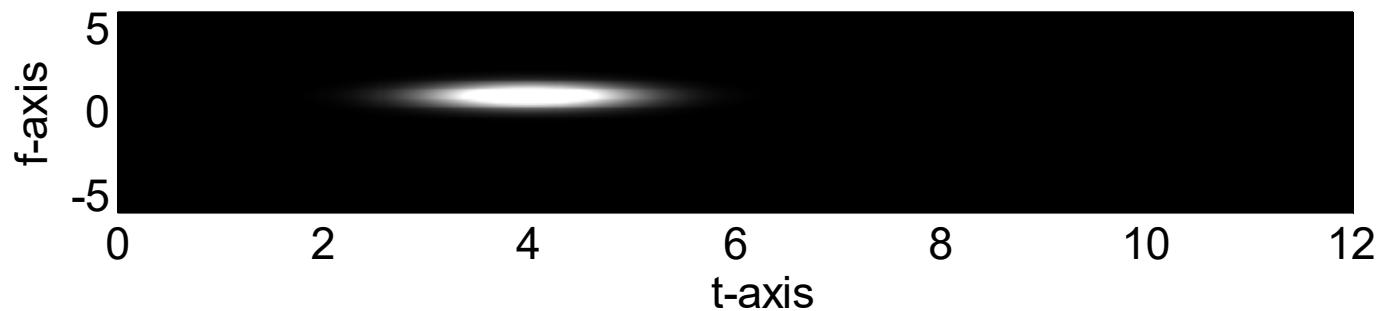
[Ref] A. Bultan, "A four-parameter atomic decomposition of chirplets," *IEEE Trans. Signal Processing*, vol. 47, no. 3, pp. 731–745, Mar. 1999.

[Ref] C. Capus, and K. Brown. "Short-time fractional Fourier methods for the time-frequency representation of chirp signals," *J. Acoust. Soc. Am.* vol. 113, issue 6, pp. 3253-3263, 2003.

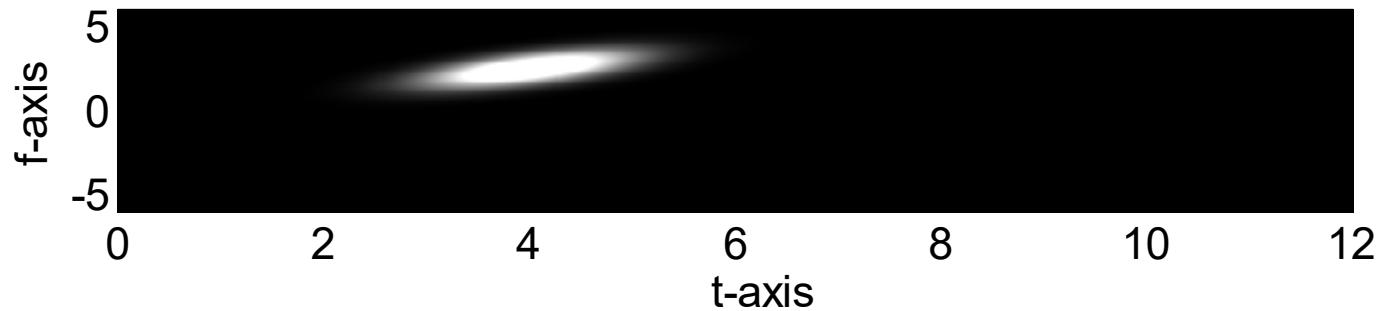
(a) STFT of a Fourier basis



(b) STFT of a 3-parameter atom



(c) STFT of a chirplet (4-parameter atom)



(3) Prolate Spheroidal Wave Function (PSWF)

$$x(t) \cong \sum_{n,T,\Omega,t_0,f_0} a_{n,T,\Omega,t_0,f_0} \psi_{n,T,\Omega}(t-t_0) \exp(j2\pi f_0 t)$$

where $\psi_{n,T,\Omega}(t)$ is the prolate spheroidal wave function

[Ref] D. Slepian and H. O. Pollak, “Prolate spheroidal wave functions, Fourier analysis and uncertainty-I,” *Bell Syst. Tech. J.*, vol. 40, pp. 43-63, 1961.

Note (i) No function has finite supports in both time and frequency domains.

(ii) The PSWF satisfies $x(t) \cong 0$ for $|t| > T$

(iii) For all functions that satisfy $x(t) \cong 0$ for $|t| > T$

The PSWF can maximize
$$\frac{\int_{-\Omega}^{\Omega} |X(f)|^2 df}{\int_{-\infty}^{\infty} |X(f)|^2 df}$$

Concept of the prolate spheroidal wave function (PSWF):

- FT: $X(f) = \int_{-\infty}^{\infty} \exp(-j2\pi f t) x(t) dt$, $x, f \in (-\infty, \infty)$.

energy preservation property (Parseval's property)

$$\int_{-\infty}^{\infty} |X(f)|^2 df = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

- finite Fourier transform (fi-FT):

$$X_{fi}(f) = \int_{-T}^{T} \exp(-j2\pi f t) x(t) dt$$

space interval: $t \in [-T, T]$, frequency interval: $f \in [-\Omega, \Omega]$

$$0 < \text{energy preservation ratio} = \frac{\int_{-\Omega}^{\Omega} |X_{fi}(f)|^2 df}{\int_{-T}^{T} |x(t)|^2 dt} < 1$$

The PWSF $\psi_{0,T,\Omega}(t)$ can maximize $\frac{\int_{-\Omega}^{\Omega} |X_{fi}(f)|^2 df}{\int_{-T}^{T} |x(t)|^2 dt}$

The PWSF $\psi_{0,T,\Omega}(t)$ can maximize $\frac{\int_{-\Omega}^{\Omega} |X_{f_i}(f)|^2 df}{\int_{-T}^T |x(t)|^2 dt}$

Among the functions orthogonal to $\psi_{0,T,\Omega}$

$\psi_{1,T,\Omega}(t)$ can maximize $\frac{\int_{-\Omega}^{\Omega} |X_{f_i}(f)|^2 df}{\int_{-T}^T |x(t)|^2 dt}$

Among the functions orthogonal to $\psi_{0,T,\Omega}$ and $\psi_{1,T,\Omega}$

$\psi_{2,T,\Omega}(t)$ can maximize $\frac{\int_{-\Omega}^{\Omega} |X_{f_i}(f)|^2 df}{\int_{-T}^T |x(t)|^2 dt}$

and so on.

- Prolate spheroidal wave functions (PSWFs) are the continuous functions that satisfy: $\int_{-T}^T K_{F,\Omega}(t_1, t) \psi_{n,T,\Omega}(t) dt = \lambda_{n,T,\Omega} \psi_{n,T,\Omega}(t_1)$,

where $K_{F,\Omega}(t_1, t) = \frac{\sin[2\pi\Omega(t_1 - t)]}{\pi(t_1 - t)}$

PSWFs are orthonormal and can be sorted according to the values of $\lambda_{n,T,\Omega}$'s:

$$\int_{-T}^T \psi_{m,T,\Omega}(t) \psi_{n,T,\Omega}(t) dt = \delta_{m,n}$$

$$1 > \lambda_{0,T,\Omega} > \lambda_{1,T,\Omega} > \lambda_{2,T,\Omega} > \dots > 0. \quad (\text{All of } \lambda_{n,T,\Omega} \text{'s are real})$$

Different from orthogonal basis expansion, which applies a complete and orthogonal basis set, compressive sensing is to use an over-complete and non-orthogonal basis set to expand a signal.

Example:

Fourier series expansion is an orthogonal basis expansion method:

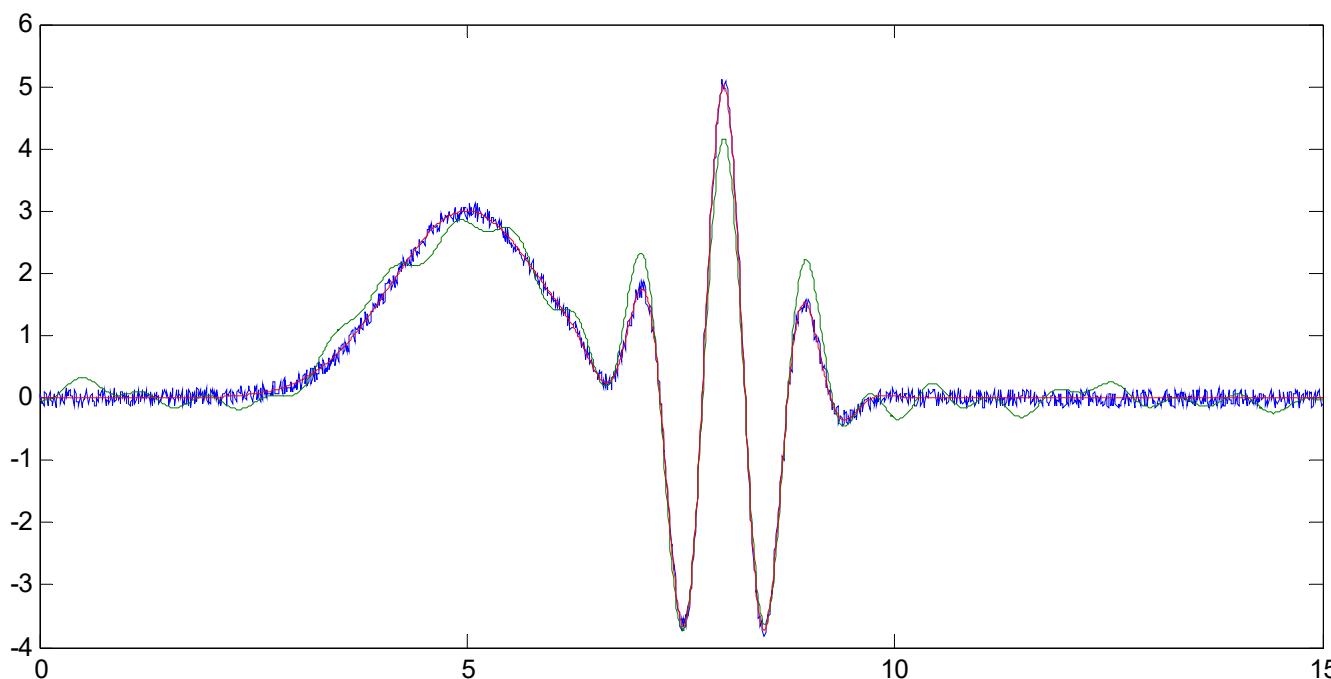
$$x(t) \approx \sum_{m=1}^M a_m \exp(j2\pi f_m t)$$

$$\int \exp(j2\pi f_m t) \overline{\exp(j2\pi f_n t)} dt = 0 \quad \text{if } f_m \neq f_n$$

Three-parameter atom expansion, Four-parameter atom (chirplet) expansion, and PSWF expansion are over-complete and non-orthogonal basis expansion methods.

$$x(t) \approx \sum a_{t_0, f_0, \sigma} \varphi_{t_0, f_0, \sigma}(t)$$

$\varphi_{t_0, f_0, \sigma}(t)$ do not form a complete and orthogonal set.



For example, in the above figure, the **blue line** is the original signal

- When using three-parameter atoms, the expansion result is the **red line**

$$x(t) = 3e^{-0.2\pi(t-5)^2} + 2.5e^{-0.4\pi(t-8)^2+j2\pi t} + 2.5e^{-0.4\pi(t-8)^2-j2\pi t}$$

Only 3 terms are used and the normalized root square error is 0.39%

- When using Fourier series, if **31 terms** are used, the expansion result is the **green line** and the normalized root square error is 3.22%

The problems that compressive sensing deals with:

Suppose that $b_0(t), b_1(t), b_2(t), b_3(t) \dots$ form an over-complete and non-orthogonal basis set.

(Problem 1) We want to minimize $\|c\|_0$ ($\|\cdot\|_0$ is L_0 norm, $\|c\|_0$ 意指 c_m 的值不為 0 的個數) such that

Sparse: smaller L_0 norm

$$x(t) = \sum_m c_m b_m(t)$$

(Problem 2) We want to minimize $\|c\|_0$ such that

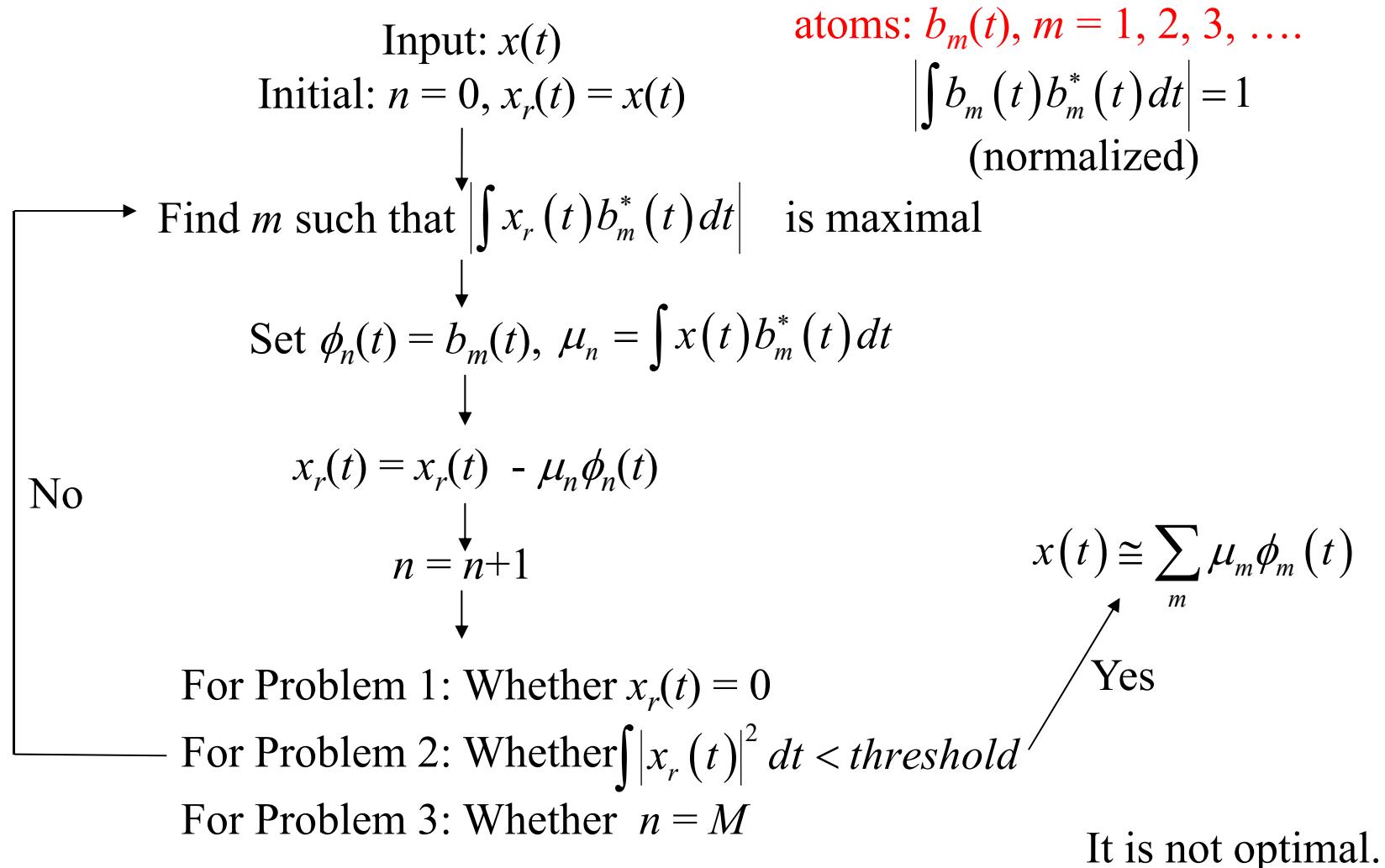
$$\int \left(x(t) - \sum_m c_m b_m(t) \right)^2 dt < \text{threshold}$$

(Problem 3) When $\|c\|_0$ is fixed to M , we want to minimize

$$\int \left(x(t) - \sum_{m=1}^M c_m b_m(t) \right)^2 dt$$

Question: How do we solve the optimization problems on page 231?

Method 1: Matching Pursuit (Greedy Algorithm)



Method 2: Basis Pursuit

Change the L_0 norm into the L_1 norm

$$\|c\|_1 = |c_0| + |c_1| + |c_2| + \dots$$

(Problem 1) We want to minimize $\|c\|_1$ such that

$$x(t) = \sum_m c_m b_m(t)$$

(Problem 2) We want to minimize $\|c\|_1$ such that

$$\int \left(x(t) - \sum_m c_m b_m(t) \right)^2 dt < \text{threshold}$$

(Problem 3) When $\|c\|_1 \leq M$, we want to minimize

$$\int \left(x(t) - \sum_m c_m b_m(t) \right)^2 dt$$

Definitions of Generalized Norms

Norm (L_α norm): $\|x[n]\|_\alpha = \sqrt[\alpha]{\sum_{n=0}^{N-1} |x[n]|^\alpha}$

$$\lim_{\alpha \rightarrow 0} (L_\alpha \text{ norm})^\alpha = K \quad \text{where } K \text{ is the number of points such that } x[n] \neq 0$$

(Physical meaning: The number of nonzero points)

L_1 norm: $\|x[n]\|_1 = \sum_{n=0}^{N-1} |x[n]|$

(Physical meaning: Sum of Amplitudes)

L_2 norm: $\|x[n]\|_2 = \sqrt{\sum_{n=0}^{N-1} |x[n]|^2}$

(Physical meaning: Distance)

Matching Pursuit: Zero order norm $\lim_{\alpha \rightarrow 0} (L_\alpha \text{ norm})^\alpha$

Basis Pursuit: First order norm L_1 norm

The L_α norm is convex if $\alpha \geq 1$.

[Compressive Sensing 參考文獻]

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- W. He and T. Qu, "Audio lossless coding/decoding method using basis pursuit algorithm," IEEE Int. Conf. Acoustics, Speech and Signal Processing, pp. 552-555, May 2013. (使用 basis pursuit 來做信號壓縮)