

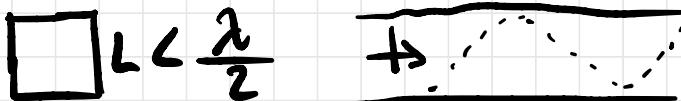

2.0 Pipes, Resonators, and Filters

Husker & Frey Ch. 10

- Series Connected Duct Elements

- automotive muffler
- acoustic filters

- Plane Wave Approach - Low frequencies



- Lumped elements

- small compared to a wavelength
- geometrical details don't matter

- Impedances

- Duct area changes + branches

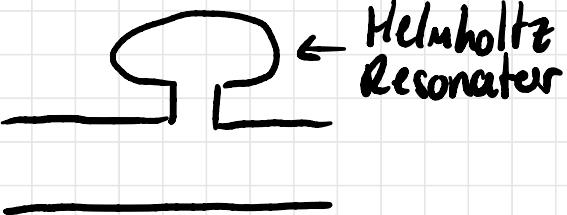
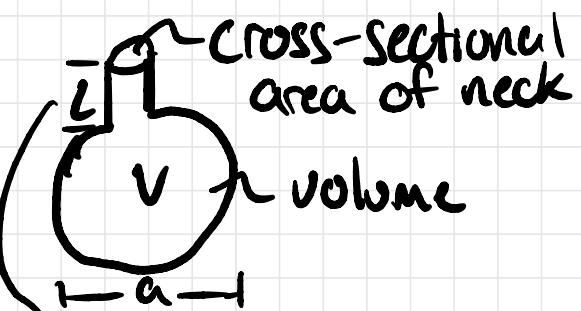
- Acoustic filters

- low pass
- high pass
- bandstop/band pass

In all cases:
low freq approach.

$\frac{JD}{D} \rightarrow$ $D < \frac{\lambda}{2}$
Plane wave propagation

2.1 Helmholtz Resonators



neck length, L - fluid in neck acts like a solid body moving back n forth
in L
Lumped elements Pressure is assumed to be instantaneously constant

- Shape is not important
- incompressible in neck
- fluid in neck acts as inertial element
- fluid trapped in reservoir acts like a stiffness

frictional losses due to viscosity as air moves back and forth in the neck

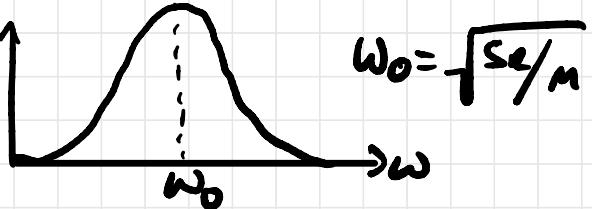
- Thermoviscous boundary layers form in the neck

- Radiation from the mouth to infinity
(radiative loss)



SDOF

α



2.2 Mass Modeling

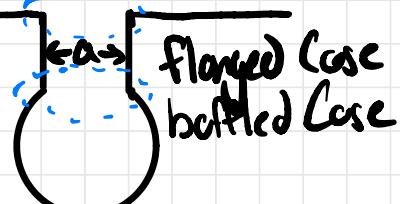
$$m = \rho_0 S_a L' \sim \text{effective length}$$

$$L' = L + \text{end corrections}$$

$$\frac{1}{L} \sim \text{end correction}$$



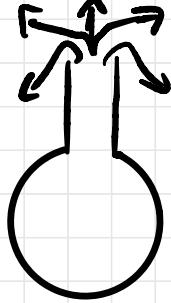
End Corrections



$$\text{E.C.} \approx \frac{8}{3\pi} a \approx 0.85a$$

Added to both inner and outer ends

$$\text{total end correction } L' = L + 1.7a$$

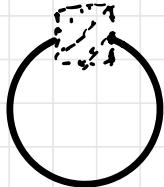


unflanged Case:

external: $0.6a$ - outer end-correction
 $.85a$ - inner end-correction

$$\text{Total} = 0.6a + 0.85a = 1.45a$$

$$L' = L + 1.45a$$



... O neck length ...
still have end-corrections

$$L' = 0.6a + 0.85a$$

True even when $L = 0$

2.2.2 Stiffness

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displacement of the piston (plunger fluid)
... positive into the reservoir

$P = B S \sim$ bulk modulus of the fluid
 \sim condensation

$$S = \frac{P - P_0}{P} \underset{\text{instantaneous density}}{=} \frac{\Delta P}{P_0}$$

Initial Density

$$\rho_0 = \frac{m_{\text{air}}}{V}$$

Final Density

$$\rho = \frac{m_{\text{air}}}{V + \Delta V}$$

$$S = \frac{\rho - \rho_0}{\rho_0} = \frac{\cancel{m_{\text{air}}}}{V + \Delta V} - \frac{\cancel{m_{\text{air}}}}{V} = \frac{\cancel{m_{\text{air}}}}{V} \left(\frac{V}{V + \Delta V} - 1 \right)$$

$$S = \frac{-\Delta V}{V(1 + \Delta V/V)} \xrightarrow{\theta} \text{due to small changes and linear acoustics assumption}$$

$$S \approx -\frac{\Delta V}{V} \approx +\frac{S_a \xi}{V}$$

cross-sectional area
 displacement of fluid in neck

$$P = B_S \quad B = P_0 C^2 = \gamma P_0 \sim \text{ambient pressure}$$

$$\approx 1.4 \times 10^5 \text{ Pa}$$

$$C = \frac{\gamma P_0}{P_0} \approx \sqrt{\frac{P_0 C^2}{P_0}} \sim \text{stiffness}$$

$P_0 C$ - impedance

$P_0 C^2$ - stiffness

$$P = B_S = P_0 C^2 \cdot \frac{S_a}{V} \cdot \xi$$

? pressure inside reservoir

- force per unit area resisting the inward motion of the piston

$$f = S_a P = \left(\frac{P_0 C^2 S_a^2}{V} \right) \xi$$

$$\text{force} = ? \cdot \text{displacement} \Rightarrow \frac{\text{force}}{\text{displacement}} = \text{Mechanical Stiffness of air trapped in reservoir}$$

$$S_r = \frac{P_0 C^2 S_a^2}{V}$$



Sound Radiation



2.2.3 Resistance

- visco thermal boundary layer forms in the neck
- Shearing of the fluid \rightarrow energy loss
- heat transfer from compressed fluid in the res. to the walls

- heat transfer back and forth from fluid to walls
- irreversible \rightarrow energy loss

- Sound Radiation Loss

$$R_m = R_w + R_r$$

S radiation to infinity
 Viscous and
Eternal losses

R_m is usually small

Formulas - $h + f$ Ch. 10

Now know all 3...

 Mechanical Impedance
 $R_m = \frac{f}{m} \approx \frac{1}{\omega}$

2.2.4 Forced Response



\hat{P}_{ext} at resonator
mouth is unaffected
by the presence of
the resonator

External Driving Force:
 $f = S_a \hat{P}_x e^{j\omega t}$

EOM

$$\frac{M \partial^2 \xi}{\partial t^2} + R_m \frac{\partial \xi}{\partial t} + S_r \xi = S_a \tilde{P} e^{j\omega t}$$

Convenient to solve for $j\omega \hat{\xi}$ = complex velocity

Mechanical Impedance...

= ratio of driving force to velocity

$$= S_a \tilde{P} / j\omega \hat{\xi}$$

$$Z_m = R_m + j\omega \left(m - \frac{S_r}{\omega} \right)$$

Acoustic Impedance = Ratio of Pressure to Volume Velocity

$$Z_a = \frac{Z_m}{S_a}$$

2.2.5 Natural Frequency

$$\text{Im}\{Z_m\} = 0$$

$$\omega_m - \frac{S_r}{\omega_0} = 0$$

$$\omega_0 = \sqrt{\frac{S_r}{m}} = C \sqrt{\frac{S_a}{V L}}$$

ω_0 decreases as neck length increases

ω_0 decreases as volume increases

ω_0 increases as neck area increases

Resonance occurs when driven at the natural freq

- large velocity in the neck
 - dissipate significant amounts of energy due to viscous boundary layer in neck

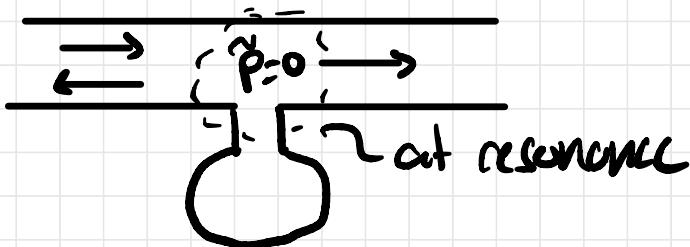
- Acoustic Absorption

May want to increase R_m



Helmholtz Resonators

- Used to absorb energy at the natural frequency
- Used to reflect energy in duct systems due to the input impedance at resonance

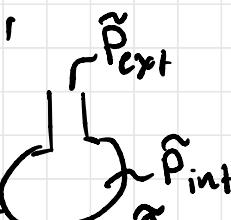


reflected at pressure release with $R=-1$

2.2.6 Pressure Amplifier

EOM in harmonic form

$$-\omega^2 M \ddot{z} + j\omega R_m \dot{z} + S_a \ddot{z} = S_a \tilde{P}_{ext}$$



$$\ddot{z} = \frac{S_a \tilde{P}_{ext}}{-\omega^2 M + j\omega R_m + S_a}$$

$$\begin{aligned}\ddot{z} &= \frac{S_a}{j} \tilde{s} \\ \tilde{P}_{int} &= B \tilde{s}\end{aligned}$$

$$\tilde{P}_{int} = \left(\frac{S_a^2 B}{j\omega V(j\omega M - jS_a + R_m)} \right) \tilde{P}_{ext}$$

if you drive at ω_0

$$\frac{\tilde{P}_{int}}{\tilde{P}_{ext}} = \frac{S_a^2 B}{j\omega V(R_m)}$$

Can perform freq analysis using a set of bichromatic resonators

2.3 Acoustic Impedances

Acoustic impedance is analogous to electrical impedance

Voltage \leftrightarrow Pressure

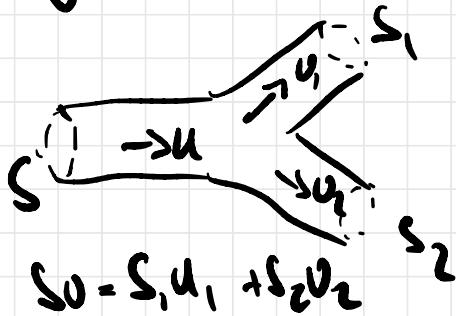
Current \leftrightarrow Volume Velocity

Inductance \leftrightarrow Mass

Capacitance \leftrightarrow Stiffness

Resistance \leftrightarrow Resistance

At a junction!



$$S_0 = S_1 u_1 + S_2 u_2$$

Acoustic Impedance

$$Z_a = \frac{P}{U} \underset{\text{sound pressure}}{\sim} \underset{\text{volume velocity}}{=} \frac{P}{Su}$$

$$\text{Acoustic Impedance} = \frac{\rho}{u} \sim \text{volume velocity}$$

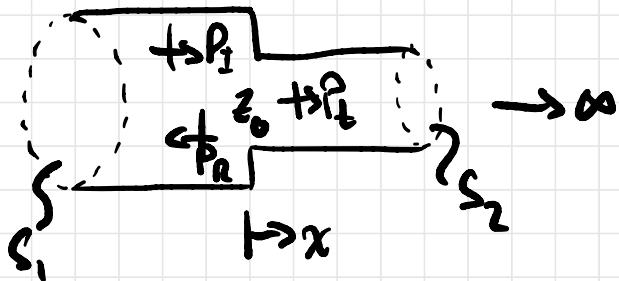
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$$\underline{\underline{\rightarrow}} |z_0|$$

$$R = \frac{Z_0 - \rho_0 c}{Z_0 + \frac{\rho_0 c}{s}}$$

$$R_{\text{II}} = |R|^2 \quad T_{\text{II}} = 1 - R_{\text{II}} = \alpha$$

2.4.2 Area transition



Z_0 = acoustic impedance "looking into" the downstream section

$$p_t = C e^{-jkx}$$

$$u_t = \frac{C}{\rho_0 c} e^{-jkx} \dots \text{particle Velocity}$$

$$U_t = \frac{S_2 \cdot C}{\rho_0 c} e^{-jkx} = S_2 \cdot u_t$$

$$Z_0 = \frac{P_L}{U_L} \Big|_{x=0} = \frac{C}{S_2 C} \cdot P_{oL} = \frac{P_{oL}}{S_2} = Z_0$$

$$R = \frac{Z_0 - \frac{P_{oL}}{S_1}}{Z_0 + \frac{P_{oL}}{S_1}} = \frac{\frac{P_{oL}}{S_2} - \frac{P_{oL}}{S_1}}{\frac{P_{oL}}{S_2} + \frac{P_{oL}}{S_1}} = \frac{\frac{1}{S_2} - \frac{1}{S_1}}{\frac{1}{S_2} + \frac{1}{S_1}}$$

$$R = \frac{1 - \left(\frac{S_2}{S_1}\right)}{1 + \left(\frac{S_2}{S_1}\right)}$$



$$R_{\pi} = |R|^2$$

$$T_{\pi} = \frac{4 \left(\frac{S_2}{S_1}\right)}{\left(1 + \left(\frac{S_2}{S_1}\right)\right)^2}$$

Special Cases

$\frac{S_2}{S_1} \rightarrow 1$ $R \rightarrow 0$ $T \rightarrow 1$... no area change

$\frac{S_2}{S_1} \rightarrow 0$ $R \rightarrow 1$ $T \rightarrow 0$... rigid termination

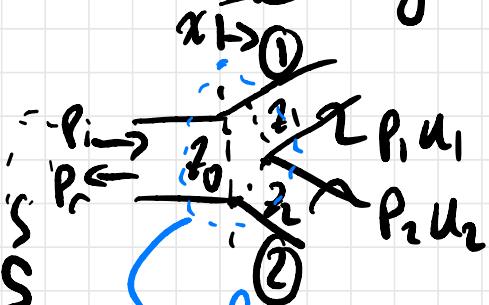
$\frac{S_2}{S_1} \rightarrow \infty$ $R \rightarrow -1$... Pressure Release Case

Magnitude of R

$$\frac{S_2}{S_1} = 10 \text{ or } \frac{1}{10} \dots |R| \text{ is the same}$$

if $\frac{S_2}{S_1} < 1 \dots$ contraction so R is positive
~inphase reflection

2.4.5 Splitting duet Case



Assume junction is small compared to a wavelength

- point junction

$$\text{at } x=0 \quad P_i + P_r = P_1 = P_2 \quad (1)$$

$$V_i + V_r = V_1 + V_2 \quad (2)$$

Divide (2) by (1)

$$\frac{V_i + V_r}{P_i + P_r} = \frac{V_1 + V_2}{P_1} = \frac{U_1 + U_2}{P_2}$$

$$\left(\frac{1}{Z_0}\right)_{x=0} = \frac{1}{Z_1} \Big|_{x=0} + \frac{1}{Z_2} \Big|_{x=0}$$

parallel addition of impedances

- smallest impedance is the most important

$$Z_0 = \frac{Z_1 Z_2}{Z_1 + Z_2} \Rightarrow R = \frac{Z_0 - \frac{P_{oc}}{S}}{Z_0 + \frac{P_{oc}}{S}}$$

Easy to extend to many branches

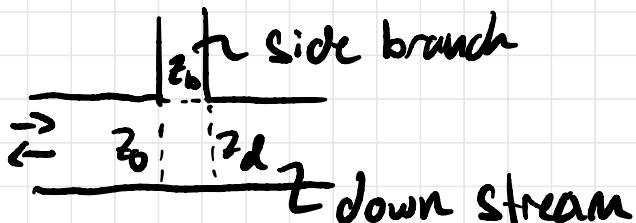


$$\frac{1}{Z_0} = \frac{1}{Z_1} + \frac{1}{Z_2} \dots + \frac{1}{Z_n}$$

n = num of branches

Junction must be small compared to a wavelength

Side branches can be treated as junctions



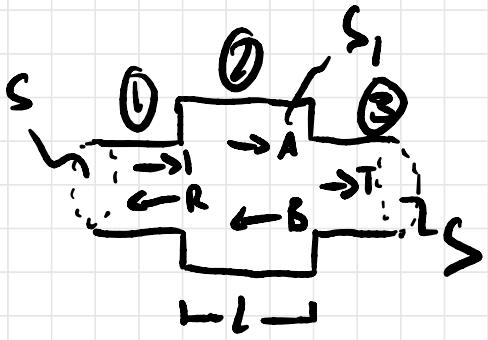
$$\frac{1}{Z_0} = \frac{1}{Z_d} + \frac{1}{Z_b}$$

* made no assumptions about the impedances

2.5 Acoustic Filters

- used in Automotive mufflers, HVAC systems, compressors, musical instruments
- low pass, band pass, band stop, high pass
- Combine expansions, contractions, orifices, side branches
- All dimensions, $D < \lambda/2$

2.5.1 - Low Pass



4 unknowns, apply 2 b.c.'s at each junction

- pressure
- volume velocity

$$\textcircled{1} \quad P_1 = e^{-j\omega x} + R e^{j\omega x}$$

$$V_1 = \frac{S}{P_o C} [e^{-j\omega x} - R e^{j\omega x}]$$

$$\textcircled{2} \quad P_2 = A e^{-j\omega x} + B e^{j\omega x}$$

$$V_2 = \frac{S_1}{P_o C} [A e^{-j\omega x} - B e^{j\omega x}]$$

$$\textcircled{3} \quad P_3 = T e^{-j\omega x}$$

$$V_3 = \frac{S}{P_o C} T e^{-j\omega x}$$

interested in $R + T$

$$\text{at } x=0 \quad 1+R = A+B$$

$$\frac{S}{P_o C} - \frac{S R}{P_o C} = \frac{S_1 A}{P_o C} - \frac{S_1 B}{P_o C}$$

$$\text{at } x=L$$

$$A e^{-j\omega L} + B e^{j\omega L} = T e^{-j\omega L}$$

$$\frac{S_1}{P_o C} A e^{-j\omega L} - \frac{S_1}{P_o C} B e^{j\omega L} = \frac{S}{P_o C} T e^{-j\omega L}$$

$$\begin{bmatrix} 4 \times 4 \\ & & & \end{bmatrix} \begin{bmatrix} A \\ B \\ R \\ T \end{bmatrix} = \begin{bmatrix} & & & \end{bmatrix}$$

solve for R and T

Power Transmission Coefficient

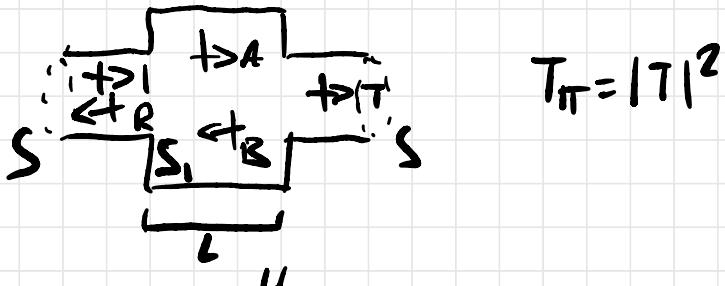
$T_{II} = |T|^2$ (since the upstream and downstream cross-sectional areas are the same)

$$T_{II} = \frac{4}{4\cos^2 kL + \left(\frac{s_1}{s} + \frac{s}{s_1}\right)^2 \sin^2 kL}$$

2.5 Acoustic Filters

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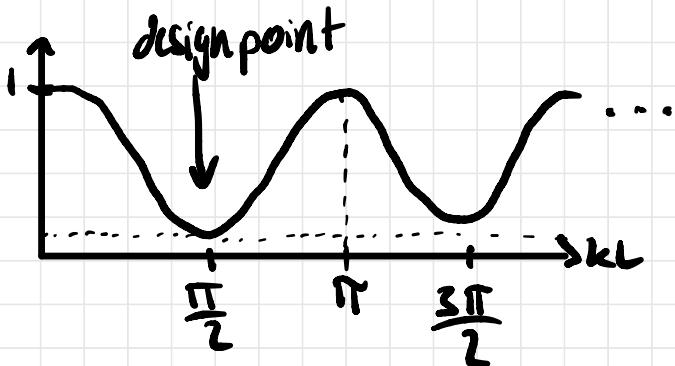
2.5.1 Low pass filter



$$T_{\pi} = \frac{4}{4\cos^2 kL + \left(\frac{S_1}{S} + \frac{S}{S_1}\right) \sin^2 kL}$$

$$KL \ll 1 \Rightarrow T_{\pi} \approx 1$$

Low freq's pass without attenuation



$$\frac{2\pi L}{\lambda} = \frac{\pi}{2} \Rightarrow \frac{L}{\lambda} = \frac{1}{4} \dots \text{for Max attenuation}$$

$$\text{if } kL = \frac{\pi}{2} \quad \cos > 0$$

$$T_{\Pi} = \frac{u}{S_1 + \frac{S_1}{S}} \Rightarrow T_{\Pi_{\min}} = \left[\frac{2 \left(\frac{S_1}{S} \right)^2}{1 + \left(\frac{S_1}{S} \right)^2} \right]^2$$

depth of minimum is controlled by the expansion ratio

if expansion is large

$$\frac{S_1}{S} \ll 1 \quad T_{\Pi_{\min}} = 4 \left(\frac{S_1}{S} \right)^2$$

Subject to the constraint that we should be operating in the plane wave region

Notes:

i) "Attenuation"

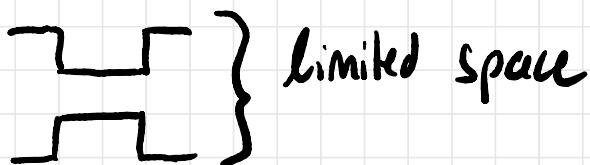
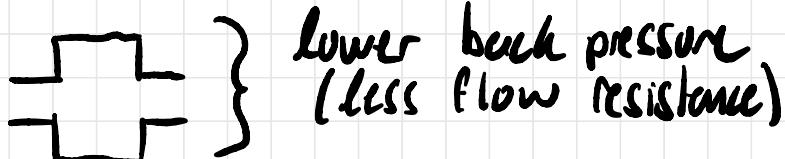
produced by reflection

(reactive systems - no energy dissipation)

ii) S_1 can be larger or smaller and create the same effect

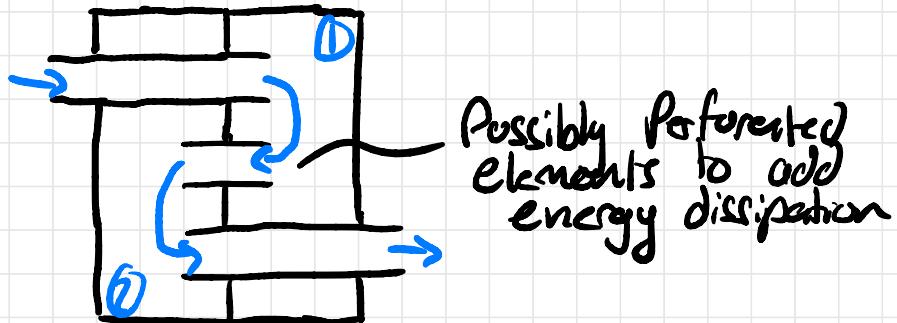
$$\frac{S_1}{S} = 10 \text{ or } \frac{S_1}{S} = \frac{1}{10} \quad T_{\Pi} \text{ is the same}$$

in both cases (expansion or contraction)

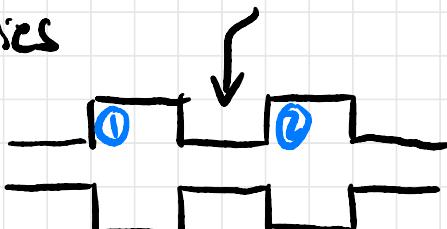


Real automotive mufflers are more complicated
- greater attenuation over range of freq's

Triple Pass -



Series



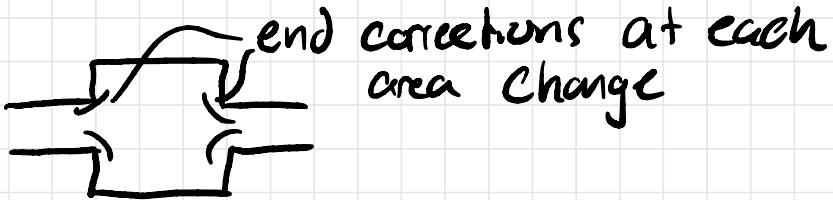
if all dimensions are small compared to a wavelength, only plane waves propagate so plane wave analysis works and the details of geometry are not important

Higher Frequencies

- Finite Element Software
- Boundary Elements
- Modal expansion

Flow - also complicates

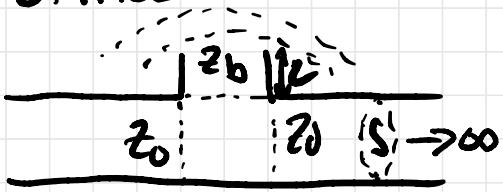
- sound speed in up and downstream sections is different



2.5.2 High Pass Filter

— aperture inside of the duct

- Allows only HF components to propagate into downstream section past the orifice



$$\frac{1}{Z_0} = \frac{1}{Z_b} + \frac{1}{Z_d}$$
$$Z_d = \frac{\rho c}{S}$$

Z_b = radiation acoustic impedance of the aperture

~ model as an unflanged piston

$\text{Im}\{Z_b\}$ = mass reactance of fluid in the neck + mass reactance of an unflanged piston (exterior end correction) + + internal end correction (flanged opening)

$\text{Re}\{Z_b\}$ = resistance = radiation away to infinity
+ viscothermal losses in the neck

Very small at low freq's

$$Z_b \approx \frac{\rho_0 C k^2}{4\pi r} + j \left(\frac{\rho_0 L' w}{\pi r^2} \right)$$

↑
aperture radius

~ Mass-like impedance

L' - effective neck length

$$L' = L + \underbrace{1.5a}_{\text{end corrections}}$$

~ inner flanged
outer unflanged

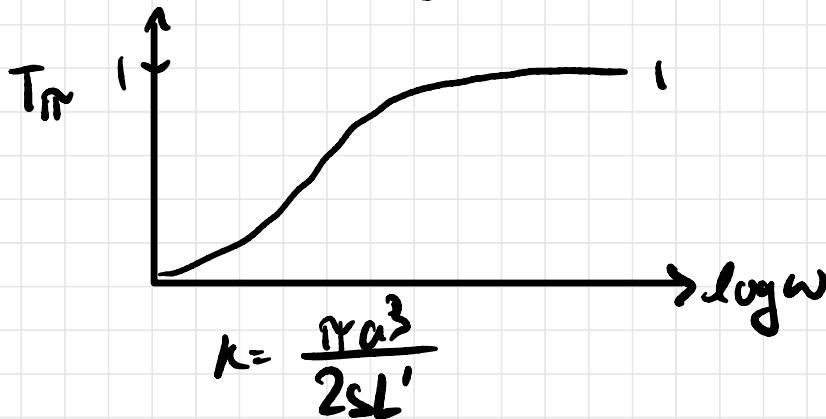
at low freq's...

$$Z_0 \approx Z_b \rightarrow 0 \text{ as } \omega \rightarrow 0$$

$Z_0 = f_L \rightarrow 0 \text{ as } P \rightarrow 0$... creates pressure release b.c. at the junction

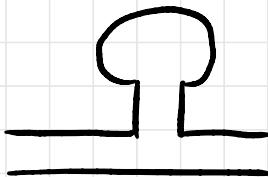
$R \rightarrow -1$ as $\omega \rightarrow 0$

As $\omega \rightarrow HF$ $\frac{1}{Z_0} \rightarrow 0$ $Z_0 \approx Z_b$



2.5.3 BandStop filters

- attenuate a particular frequency
- side branch



Helmholtz
Resonator



$\frac{1}{4}$ Wavelength
Resonator

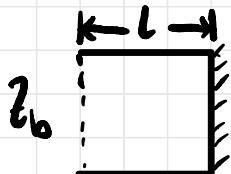
Reactive
Devices

input impedance looking into the resonator

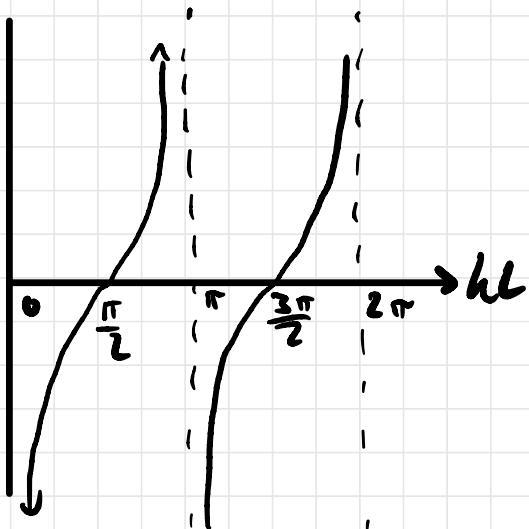
$\rightarrow 0$ at tuning frequency

$Z_b \rightarrow 0, Z_0 \rightarrow 0, R \rightarrow -1$

Quarter Wave Resonator



$$Z_0 = -j \frac{P_0 C}{S} \cot kL$$

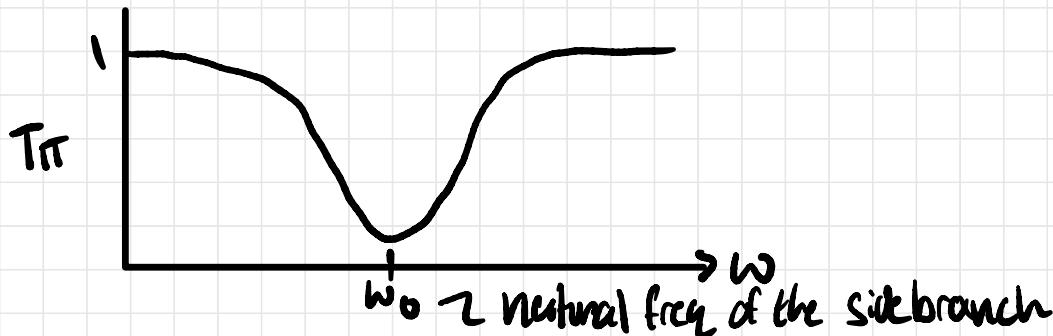


$$\text{when } hL = \frac{\pi}{2}$$

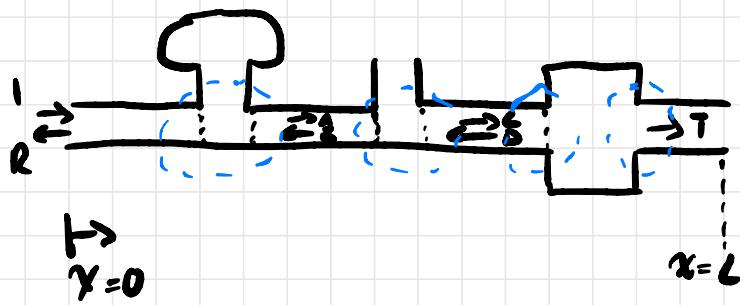
$$\frac{2\pi}{\lambda} L = \frac{\pi}{2} \Rightarrow \frac{L}{\lambda} = \frac{1}{4}$$

input impedance of a rigidly terminated tube
 $\rightarrow 0$ at freq when $\frac{L}{\lambda} = \frac{1}{4}$

$$\frac{1}{Z_0} = \frac{1}{Z_b} + \frac{1}{Z_d} \Rightarrow R \rightarrow -1 + T\pi \rightarrow 0$$



2.5.4 Combinations



6 unknowns

6 boundary conditions (3×2)

Create system of equations by substituting plane wave solutions into b.c.'s

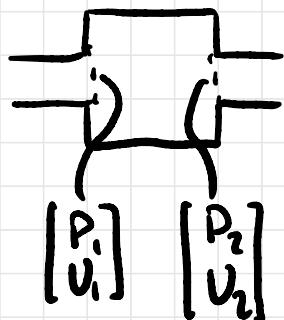
$$\begin{bmatrix} 6 \times 6 \\ \text{coeff} \end{bmatrix} \begin{bmatrix} R \\ A \\ \vdots \\ T \end{bmatrix} = \begin{bmatrix} \text{forcing vector} \end{bmatrix}$$

Solve for Constants as you wish

Modify System of equations each time you add a new element

2.6 Transfer Matrix Approach

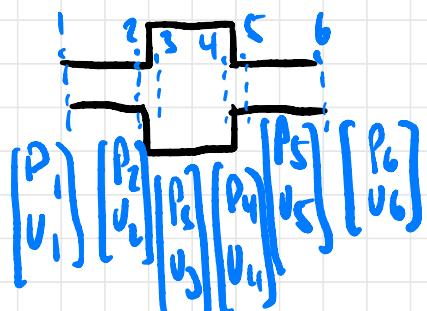
2.6.1 General Approach



- Relate P and V on 2-sides
of an element by a
 T -matrix $[2 \times 2]$

- flexible assembly of complicated
systems

$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix} = \begin{bmatrix} T_{11} & T_{12} \\ T_{21} & T_{22} \end{bmatrix} \begin{bmatrix} P_2 \\ V_2 \end{bmatrix}$$



$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix} = \begin{bmatrix} T_1 \end{bmatrix} \begin{bmatrix} P_2 \\ V_2 \end{bmatrix}$$

$$\begin{bmatrix} P_2 \\ V_2 \end{bmatrix} = \begin{bmatrix} T_2 \end{bmatrix} \begin{bmatrix} P_3 \\ V_3 \end{bmatrix} \text{ and so on}$$

$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix} = \underbrace{\begin{bmatrix} T_1 \end{bmatrix}}_{[2 \times 2]} \underbrace{\begin{bmatrix} T_2 \end{bmatrix}}_{[2 \times 2]} \underbrace{\begin{bmatrix} T_3 \end{bmatrix}}_{[2 \times 2]} \underbrace{\begin{bmatrix} T_4 \end{bmatrix}}_{[2 \times 2]} \underbrace{\begin{bmatrix} T_5 \end{bmatrix}}_{[2 \times 2]} \begin{bmatrix} P_6 \\ V_6 \end{bmatrix} = \underbrace{\begin{bmatrix} T_6 \end{bmatrix}}_{[2 \times 2]} \begin{bmatrix} P_6 \\ V_6 \end{bmatrix}$$

$$P_1 = 1 + R$$

$$U_1 = \frac{S}{P_{0C}} (1 - R)$$

$$P_0 = T$$

$$U_0 = \frac{S}{P_{0C}} T$$

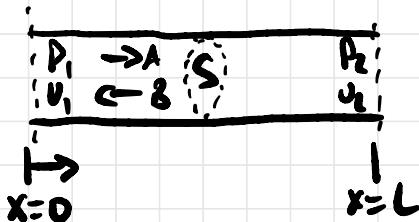
Multiply out to obtain 2 equations and 2 unknowns

Powerful approach to modeling

- T-matrix are simple
- multiply together to obtain complete system matrix
- lends itself to automated model assembly

2.6.3 Transfer Matrices for Simple Elements

2.6.3.1 Straight Pipe Section



$$P_1 = A + B$$

$$U_1 = \frac{S}{P_{0C}} (A - B)$$

$$P_2 = A e^{-j\omega L} + B e^{j\omega L}$$

$$U_2 = \frac{S}{P_{0C}} (A e^{j\omega L} - B e^{-j\omega L})$$

$$A = \frac{1}{2} \left(P_2 + \frac{P_{0c}}{S} U_2 \right) e^{-j\delta_{KL}}$$

$$e^{j\delta_{KL}} = \cos \delta_{KL} - j \sin \delta_{KL}$$

$$B = \frac{1}{2} \left(P_2 - \frac{P_{0c}}{S} U_2 \right) e^{j\delta_{KL}}$$

$$P_1 = (\cosh \delta_{KL}) P_2 + j \left(\frac{P_{0c}}{S} \sin \delta_{KL} \right) U_2$$

$$U_1 = \left(j \frac{S}{P_{0c}} \sin \delta_{KL} \right) P_2 + (\cosh \delta_{KL}) U_2$$

$$\begin{bmatrix} P_1 \\ U_1 \end{bmatrix} = \begin{bmatrix} \cosh \delta_{KL} & j \frac{P_{0c}}{S} \sin \delta_{KL} \\ j \frac{S}{P_{0c}} \sin \delta_{KL} & \cosh \delta_{KL} \end{bmatrix} \begin{bmatrix} P_2 \\ U_2 \end{bmatrix}$$

T-Matrix for a pipe of
length L $x=0 \Rightarrow x=L$

... produces same result

$$\overbrace{x'_1 \quad x'_{1+L}}$$

Transfer matrix is independent of absolute position or origin

2.6.5.2 Side branch

Assumptions:

- Lumped Element
- Point Junction
- Pressure is uniform throughout the junction

$$\frac{P_1}{U_1} \left| -Z_b \right| \frac{P_b = \frac{P_b}{U_b}}{U_b}$$

$$P_1 = P_2 = P_b$$

$$P_b = Z_b U_b$$

$$P_2 = Z_b U_b \Leftarrow U_b = U_1 - U_2$$

$$P_2 = Z_b (U_1 - U_2)$$

$$U_1 Z_b = P_2 + U_2 Z_b$$

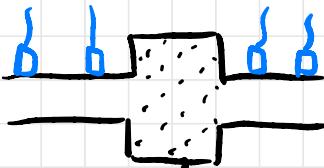
$$U_1 = \frac{P_2}{Z_b} + U_2$$

$$\begin{bmatrix} P_1 \\ U_1 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ \frac{1}{Z_b} & 1 \end{bmatrix} \begin{bmatrix} P_2 \\ U_2 \end{bmatrix}$$

? transfer matrix across a side branch
of impedance Z_b

Side branch: orifice/aperture, Helmholtz resonator,
 $\frac{1}{4}$ wave resonators

2.6.5.3 Unknown Systems



4-microphones
2-terminations (rigid + absorbing)

{black box - unknown internal structure}

$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix}_1 = [T] \begin{bmatrix} P_2 \\ V_2 \end{bmatrix}_1, \quad \text{Measurement with rigid termination}$$

- repeat the measurement with a different termination

$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix}_2 = [T] \begin{bmatrix} P_2 \\ V_2 \end{bmatrix}_2$$

$$\begin{bmatrix} P_{1,1} \\ V_{1,1} \\ P_{1,2} \\ V_{1,2} \end{bmatrix} = \begin{bmatrix} [T] & 0 \\ 0 & [T] \end{bmatrix} \begin{bmatrix} P_{2,1} \\ V_{2,1} \\ P_{2,2} \\ V_{2,2} \end{bmatrix}$$

4 eqn's & 4 unknowns

solve for $T_{11}, T_{12}, T_{21}, T_{22}$

Assumption: properties of unknown system don't depend on the downstream termination condition

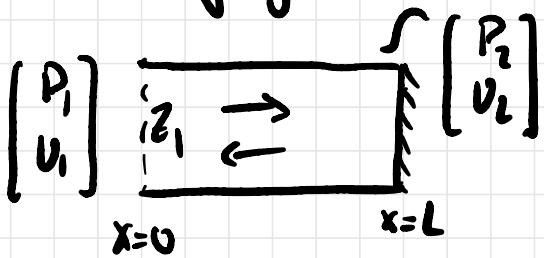
Conduct this experiment numerically - with boundary or finite element model of the system

Two-Load Method

- Mungal - Duct Acoustics
- Song + Bolton (JASA 2000)

2.6.4 Examples

2.6.4.1 Rigidly terminated tube-section



What is Z_1 ?

$$\begin{bmatrix} P_1 \\ U_1 \end{bmatrix} = \begin{bmatrix} \cosh kL & j \frac{P_{oc}}{S} \sinh kL \\ j \frac{S}{P_{oc}} \sinh kL & \cosh kL \end{bmatrix} \begin{bmatrix} P_2 \\ U_2 \end{bmatrix}$$

$$P_1 = \cosh kL P_2$$

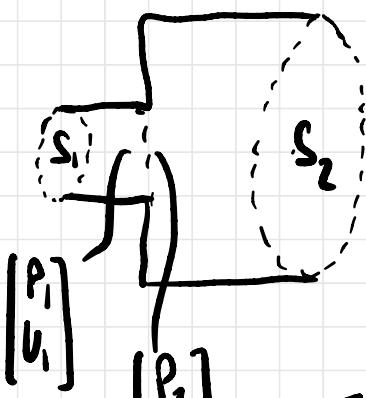
$$U_1 = j \frac{S}{P_{oc}} \sinh kL P_2 \Rightarrow Z_1 = \frac{\cosh kL}{\sinh kL} \cdot \frac{P_{oc}}{jS}$$

$$Z_1 = -j \frac{P_{oc}}{S} \coth kL$$

$$\begin{bmatrix} P_1 \\ Z_1 \\ J_1 \end{bmatrix}_{x=0} \quad \begin{bmatrix} P_2 \\ Z_2 \\ V_2 \end{bmatrix}_{x=L}, \quad \text{known but arbitrary}$$

$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix} = [T, \text{straight pipe}] \begin{bmatrix} P_2 \\ V_2 \end{bmatrix} \quad V_2 = \frac{P_2}{Z_L}$$

2.6.4.2 Anechoically Terminated Area Transition



$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix} \quad \begin{bmatrix} P_2 \\ V_2 \end{bmatrix} \Rightarrow T = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \dots \text{T-matrix for area transition}$$

$$(I+R) \begin{bmatrix} P_1 \\ V_1 \end{bmatrix} = \begin{bmatrix} P_2 \\ V_2 \end{bmatrix} T$$

$$\frac{S_1}{S_2} (I-R)$$

$$T = \frac{2}{1 + \frac{S_2}{S_1}}$$

$$R = \frac{1 - \frac{S_2}{S_1}}{1 + \frac{S_2}{S_1}} \quad T_{\pi\pi} + R_{\pi\pi} = 1$$

if $s_2 < s_1 \dots \frac{s_2}{s_1} < 1$ (contraction)

$T > 1 \rightarrow 2$ when $\frac{s_2}{s_1} \ll 1$

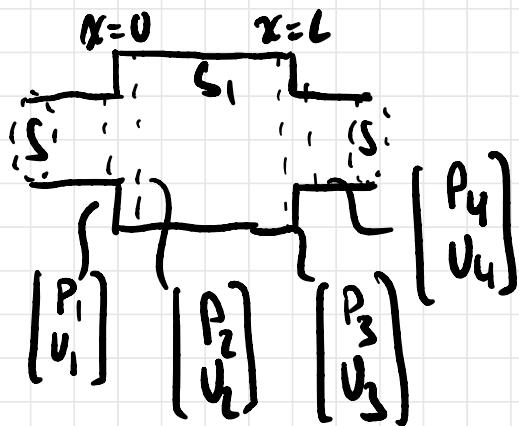
analogous to

$$\begin{array}{c} \text{air} \quad | \quad \text{water} \\ \downarrow \\ R \sim 1 \quad \frac{2P_1}{2P_2} \end{array}$$

$T > 1$, but T_{rr} is very small

$$\begin{array}{c} \downarrow \frac{P_{oc}}{s_1} \quad \downarrow \frac{P_{oc}}{s_2} \\ \hline I \end{array} \quad \text{I mostly sees a rigid termination}$$

2.6.4.3 Expansion Muffler



$$\begin{bmatrix} P_2 \\ U_2 \end{bmatrix} = [T]_{SP} \begin{bmatrix} P_3 \\ U_3 \end{bmatrix}$$

^ Straight Pipe

$$\begin{bmatrix} P_1 \\ U_1 \end{bmatrix} = \begin{bmatrix} P_2 \\ U_2 \end{bmatrix} \quad \text{and} \quad \begin{bmatrix} P_3 \\ U_3 \end{bmatrix} = \begin{bmatrix} P_4 \\ U_4 \end{bmatrix}$$

$$\begin{bmatrix} P_1 \\ U_1 \end{bmatrix} = [T]_{SP} \begin{bmatrix} P_4 \\ U_4 \end{bmatrix}$$

$$\begin{bmatrix} P_1 \\ U_1 \end{bmatrix} = \begin{bmatrix} \cosh L & j \frac{\rho_{oc}}{S_1} \sinh L \\ j \frac{S_1}{\rho_{oc}} \sinh L & \cosh L \end{bmatrix} \begin{bmatrix} P_4 \\ U_4 \end{bmatrix}$$

$$P_1 = 1 + R$$

$$U_1 = \frac{S}{\rho_{oc}} (1 - R)$$

$$P_4 = T e^{-j \delta L}$$

$$U_4 = \frac{S}{\rho_{oc}} T e^{-j \delta L}$$

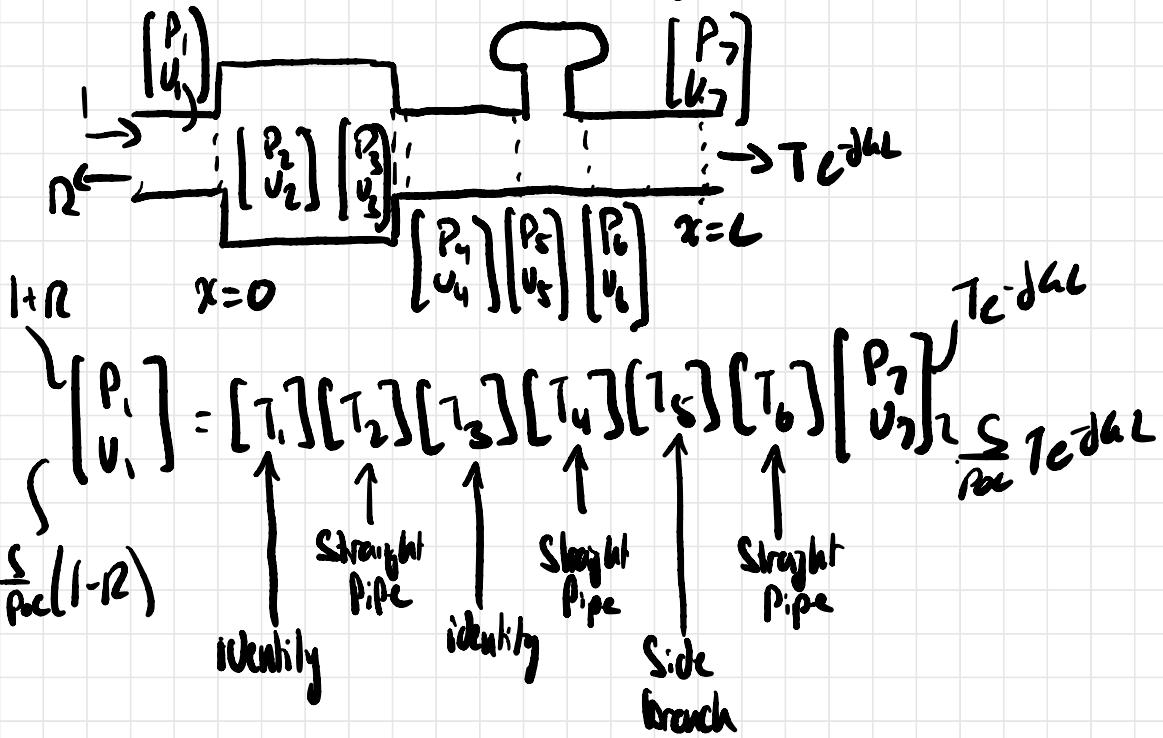
$$T = \frac{2e^{j \delta L}}{2 \cosh L + j \left(\frac{S_1}{S} + \frac{S_1}{S} \right) \sinh L}$$

$$T_{rr} = |T|^2 = \frac{4}{4 \cos^2 \delta L + \left(\frac{S_1}{S} + \frac{S_1}{S} \right)^2 \sin^2 \delta L}$$

$$R = \frac{J \left(\frac{S}{S_1} - \frac{S_1}{S} \right) \sinh kL}{2 \cosh kL + J \left(\frac{S}{S_1} + \frac{S_1}{S} \right) \sinh kL} \quad \dots \quad R_{\text{TR}} = |R|^2$$

$$T_{\text{IR}} + R_{\text{IR}} = 1 \quad \dots \quad \text{Energy Conservation}$$

2.6.4.4 Complicated System



$$T_5 = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

Solve for $R + j$ or the input impedance

Helmholtz Resonator

$$Z_m = j(\omega_m - \frac{\rho_r}{\omega}) \quad \dots \text{mechanical impedance}$$

Acoustic impedance

$$Z_a = \frac{Z_m}{S_n^2}$$

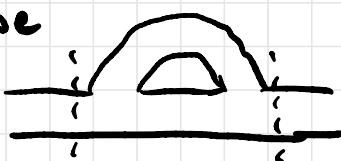
S_n = neck area

6.5 further Complications

T-matrix approach suits series connections



Quincke Tube



parallel connections are not so easy)

Solve System separately to create a
2x2 T-matrix for the subsystems

- Flow complicates the situation

- speed of sound is different in
the up- and down-stream sections



soundfield is super-imposed on the moving flow

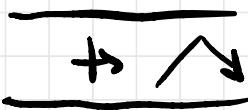
downstream $C+V = C_{\text{eff}}$ $k_d = \frac{\omega}{C+V}$

upstream $C-V = C_{\text{eff}}$ $k_u = \frac{\omega}{C-V}$

$$\rho = \underbrace{A e^{-j k_d x}}_{\text{downstream}} + \underbrace{B e^{j k_u x}}_{\text{upstream}}$$

Normally Assume flow velocity is 0 if
Mach number < 0.1 $\frac{V}{C} < 0.1$

2.7 Summary



- Plane Waves
- fundamentally a low freq approach
- lumped elements, geometric features are small compared to a wavelength
- point junctions -assume pressure is constant
- precise geometrical details don't matter
- Helmholtz resonator

- $\frac{1}{4}$ wave resonator



$$\text{Im}\{Z_m\} = 0, \text{ at natural frequency}$$

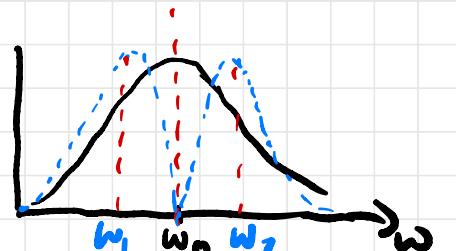
- Pressure release boundary condition
so nothing transmits beyond point

Creates Band-Stop Filter



$$\omega_0 = \sqrt{\zeta_h/m}$$

lisp



Can add tuned mass-damper



leads to above 2DOF response ...



Modifies room response

How to find T , T_{12} ...

$$\begin{bmatrix} P_1 \\ V_1 \end{bmatrix} = \begin{bmatrix} T_{11} & T_{12} \\ T_{21} & T_{22} \end{bmatrix} \begin{bmatrix} P_2 \\ V_2 \end{bmatrix} \xrightarrow[R]{\xleftarrow{Q}} \begin{bmatrix} P_1 \\ V_1 \end{bmatrix} \xrightarrow{x=L} T_C \begin{bmatrix} P_2 \\ V_2 \end{bmatrix}$$

$$P_1 = T_{11} P_2 + T_{12} V_2 \quad ①$$

$$V_1 = T_{21} P_2 + T_{22} V_2 \quad ②$$

$$\xrightarrow{x=0">>$$

$$\xrightarrow{x=L">>$$

$$P_1 = 1 + R = T_{11} (T_C e^{-\frac{d}{\lambda L}}) + T_{12} \left(\frac{S}{P_{OC}} T_C e^{-\frac{d}{\lambda L}} \right) \quad \textcircled{1}$$

$$V_1 = \frac{S}{P_{OC}} (1 - R) = T_{21} (T_C e^{-\frac{d}{\lambda L}}) + T_{22} \left(\frac{S}{P_{OC}} T_C e^{-\frac{d}{\lambda L}} \right) \quad \textcircled{2}$$

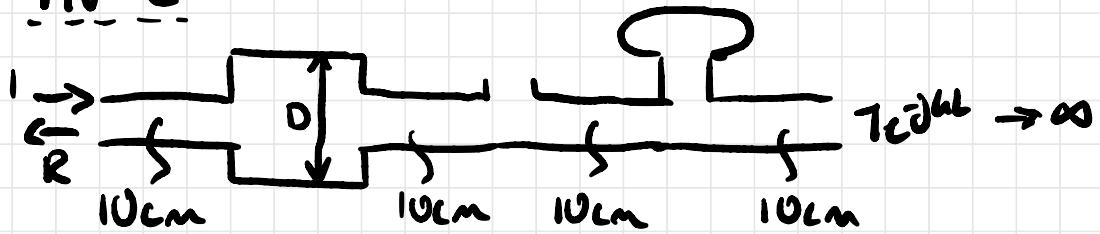
$$2 = \left[T_{11} + \left(\frac{S}{P_{OC}} \right) T_{12} + \frac{P_{OC}}{S} T_{21} + T_{22} \right] T_C e^{-\frac{d}{\lambda L}}$$

$$T = \frac{2 e^{\frac{d}{\lambda L}}}{T_{11} + \frac{S}{P_{OC}} T_{12} + \frac{P_{OC}}{S} T_{21} + T_{22}}$$

$$TL = 10 \log \frac{1}{|T|^2}$$

Derive Similar Result for R can be derived

HW 2.



$$R_\pi + T_\pi = 1$$

Group 4
Masters
you
him