

## EE152: HW4

### Fourier series

1. Let  $x[n]$  and  $y[n]$  be periodic signals with common period  $N$ , and let

$$z[n] = \sum_{r=\langle N \rangle} x[r]y[n-r]$$

be their period convolution.

- (a) Show that  $z[n]$  is also periodic with period  $N$ .  
(b) Verify that if  $a_k$ ,  $b_k$  and  $c_k$  are the Fourier coefficients of  $x[n]$ ,  $y[n]$  and  $z[n]$  respectively, then

$$c_k = Na_k b_k$$

- (c) Let

$$x[n] = \sin\left(\frac{3\pi n}{4}\right)$$

and

$$y[n] = \begin{cases} 1, & 0 \leq n \leq 3 \\ 0, & 4 \leq n \leq 7 \end{cases}$$

be two signals that are periodic with period 8. Find the Fourier series representation for the periodic convolution of these signals.

**Solution:**

- (a)  $z[n] = \sum_{r=\langle N \rangle} x[r]y[n-r]$  so we can get  $z[n+N] = \sum_{r=\langle N \rangle} x[r]y[n+N-r]$ , since  $y$  is a periodic signal with period  $N$ ,  $y[n+N-r] = y[n-r]$  holds. In this way, we can easily get

$$z[n+N] = \sum_{r=\langle N \rangle} x[r]y[n+N-r] = \sum_{r=\langle N \rangle} x[r]y[n-r] = z[n]$$

, so that  $z[n]$  is also periodic with period  $N$ .

(b)

$$\begin{aligned}
\sum_{r=\langle N \rangle} x[r]y[n-r] &= \sum_{r=\langle N \rangle} \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 r} \sum_{l=\langle N \rangle} b_l e^{jl\omega_0(n-r)} \\
&= \sum_{r=\langle N \rangle} \sum_{k=\langle N \rangle} a_k b_l \sum_{l=\langle N \rangle} e^{jl\omega_0(n-r)} e^{jk\omega_0 r} \\
&= \sum_{r=\langle N \rangle} \sum_{k=\langle N \rangle} a_k b_l \underbrace{\sum_{l=\langle N \rangle} e^{j(k-l)\omega_0 r}}_{N\delta[k-l]} e^{jl\omega_0 n} \\
&= \sum_{r=\langle N \rangle} \sum_{k=\langle N \rangle} N a_k b_l \delta[k-l] e^{jl\omega_0 n} \\
&= \sum_{k=\langle N \rangle} N a_k b_k e^{jl\omega_0 n}
\end{aligned}$$

In this way we proved  $c_k = N a_k b_k$ .

(c)  $x[n] = \sin(\frac{3\pi n}{4}) = \frac{e^{j\frac{3\pi n}{4}} - e^{-j\frac{3\pi n}{4}}}{2j}$ , here the period of this signal is 8 so we can easily get  $a_3 = -a_{-3} = \frac{1}{2j}$ , here  $-a_{-3} = -a_5$  as a Fourier Series of a discrete time signal.

So we only need to evaluate  $b_3$  and  $b_5$  for signal  $y[n]$ .

$$b_k = \frac{1}{8} \sum_{n=\langle N \rangle} y[n] e^{-jk\frac{2\pi n}{8}}$$

so these two coefficients can be easily get by this formula.

$$\begin{aligned}
b_3 &= \frac{1}{8} \sum_{n=0}^3 e^{-j3\frac{2\pi n}{8}} = \frac{1}{4} \frac{1}{1 - e^{-j\frac{3\pi}{4}}} \\
b_5 &= \frac{1}{8} \sum_{n=0}^3 e^{-j5\frac{2\pi n}{8}} = \frac{1}{4} \frac{1}{1 - e^{-j\frac{5\pi}{4}}}
\end{aligned}$$

From  $0 \leq k \leq 7$ ,  $c_3 = 8a_3b_3$  and  $c_5 = 8a_5b_5$ .

2. Let  $x[n]$  be a periodic signal with period  $N = 8$  and Fourier series coefficients  $a_k = -a_{k-4}$ . A signal

$$y[n] = \left(\frac{1 + (-1)^n}{2}\right)x[n-1]$$

with period  $N = 8$  is generated. Denoting the Fourier series coefficients of  $y[n]$  by  $b_k$ , find a function  $f[k]$  such that

$$b_k = f[k]a_k$$

.

**Solution:**  $a_k = -a_{k-4} \longrightarrow -a_k = a_{k-4} \longrightarrow \underbrace{-a_{k+4} = a_k = -a_{k-4}}_{\text{will be used later}} \text{ so } a_{k+4} = a_{k-4} \longrightarrow$

$$a_k = a_{k+8}$$

$$\begin{aligned} y[n] &= \left[ \frac{1 + (-1)^n}{2} \right] x[n-1] \\ &= \left[ \frac{1}{2} + \frac{1}{2} \cos(n\pi) \right] x[n-1] \\ &= \left[ \frac{1}{2} + \frac{1}{4} (e^{j\pi n} + e^{-j\pi n}) \right] x[n-1] \end{aligned}$$

(1)

$$\frac{1}{2} x[n-1] \xrightarrow{FS} \frac{1}{2} a_k e^{-jk \frac{2\pi}{8}}$$

(2)

$$\begin{aligned} \frac{1}{4} e^{j\pi n} x[n-1] &\xrightarrow{FS} \frac{1}{4} a_{k-4} e^{-j(k-4) \frac{2\pi}{8}} \\ &= \frac{1}{4} a_k (e^{-j \frac{2\pi k}{8}}) \end{aligned}$$

(3)

$$\begin{aligned} \frac{1}{4} e^{-j\pi n} x[n-1] &\xrightarrow{FS} \frac{1}{4} a_{k+4} e^{-j(k+4) \frac{2\pi}{8}} \\ &= \frac{1}{4} a_k (e^{-j \frac{2\pi k}{8}}) \end{aligned}$$

From (1) (2) (3) we have  $y[n] \xrightarrow{FS} a_k e^{-j \frac{2\pi k}{8}}$ , so  $f[k] = e^{-j \frac{2\pi k}{8}}$

3. [Hint] (Another form of Fourier series)

$$a_0 = \frac{2}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u(t) dt \quad \omega = \frac{2\pi}{T}$$

$$a_n = \frac{2}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u(t) \cos(n\omega t) dt$$

$$b_n = \frac{2}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u(t) \sin(n\omega t) dt$$

$$u(t) = \frac{1}{2} a_0 + \sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t)$$

A voltage  $u(t)$  is applied to a resistor of one ohm.

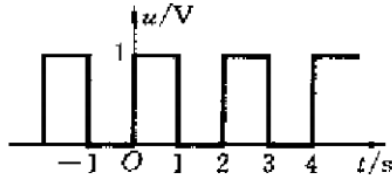


Figure 1: Voltage  $u(t)$

- (a) Find the trigonometric Fourier series for  $u(t)$ .
- (b) Use the result of (a) and  $u(\frac{1}{2})$  to find the sum of the following infinite series

$$S = 1 - \frac{1}{3} + \frac{1}{5} - \frac{1}{7} + \dots$$

- (c) Find the average power of the above resistors.
- (d) Use the result of (c) to find the sum of the following infinite series

$$S = 1 + \frac{1}{3^2} + \frac{1}{5^2} + \frac{1}{7^2} + \dots$$

**Solution:**

(a)

$$a_0 = \frac{2}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u(t) dt = \int_0^1 1 dt = 1$$

$$a_n = \frac{2}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u(t) \cos(n\omega t) dt = \int_0^1 \cos(n\omega t) dt = 0, \quad n = 1, 2, 3, \dots$$

$$b_n = \frac{2}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u(t) \sin(n\omega t) dt = \int_0^1 \sin(n\omega t) dt = \frac{1 - \cos(n\omega)}{n\omega}, \quad n = 1, 2, 3, \dots$$

$$\begin{aligned} u(t) &= \frac{1}{2} a_0 + \sum_{n=1}^{\infty} a_n \cos(n\pi t) + \sum_{n=1}^{\infty} b_n \sin(n\pi t) \\ &= \frac{1}{2} + \sum_{n=1}^{\infty} \frac{1 - \cos(n\pi)}{n\pi} \sin(n\pi t), \quad n = 1, 2, 3, \dots \end{aligned}$$

(b)

$$u\left(\frac{1}{2}\right) = 1 = \frac{1}{2} + \sum_{n=1}^{\infty} \frac{1 - (-1)^n}{n\pi} \sin\left(\frac{n\pi}{2}\right)$$

so we have:

$$\sum_{n=1}^{\infty} \frac{1 - (-1)^n}{n\pi} \sin\left(\frac{n\pi}{2}\right) = \frac{1}{2}$$

which means

$$2(1 - \frac{1}{3} + \frac{1}{5} - \frac{1}{7} + \dots) = \frac{\pi}{2}$$

in the end

$$S = 1 - \frac{1}{3} + \frac{1}{5} - \frac{1}{7} + \dots = \frac{\pi}{4}$$

(c)

$$P = \frac{1}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u^2(t) dt = \frac{1}{2} \int_0^1 1 dt = \frac{1}{2}$$

(d) From Parseval's relation:

$$p = \frac{1}{2} = (\frac{1}{2})^2 + \frac{1}{2} \sum_{n=1}^{\infty} [\frac{1 - \cos(n\pi)}{n\pi}]^2 = \frac{1}{4} + \frac{1}{2} \sum_{n=1}^{\infty} [\frac{1 - (-1)^n}{n\pi}]^2$$

which means that

$$\sum_{n=1}^{\infty} [\frac{1 - (-1)^n}{n}]^2 = \frac{1}{2} \pi^2$$

$\Rightarrow$

$$4(1 + \frac{1}{3^2} + \frac{1}{5^2} + \frac{1}{7^2} + \dots) = \frac{1}{2} \pi^2$$

in the end

$$S = 1 + \frac{1}{3^2} + \frac{1}{5^2} + \frac{1}{7^2} + \dots = \frac{\pi^2}{8}$$

### Continuous time Fourier transform

1. A causal and stable LTI system S has the frequency response

$$H(j\omega) = \frac{j\omega + 4}{6 - \omega^2 + 5j\omega}$$

(a) Determine a differential equation relating the input  $x(t)$  and the output  $y(t)$  of the system.

(b) Determine the impulse response  $h(t)$ .

**Solution:**

(a) We have  $\frac{Y(j\omega)}{X(j\omega)} = \frac{j\omega + 4}{6 - \omega^2 + 5j\omega}$  Cross-multiplying and taking the inverse Fourier transform, we obtain

$$\frac{d^2 y(t)}{dt^2} + 5 \frac{dy(t)}{dt} + 6y(t) = \frac{dx(t)}{dt} + 4x(t)$$

(b)  $H(j\omega) = \frac{j\omega + 4}{(2 + j\omega)(3 + j\omega)}$

**Step1:**

$$H(j\omega) = \frac{k_1}{2 + j\omega} + \frac{k_2}{3 + j\omega}$$

**Step2:**

$$k_1 = H(j\omega) * (2 + j\omega)|_{2+j\omega=0} = \frac{j\omega + 4}{3 + j\omega}|_{\omega=2j} = 2$$

$$k_2 = H(j\omega) * (3 + j\omega)|_{3+j\omega=0} = \frac{j\omega + 4}{2 + j\omega}|_{\omega=3j} = -1$$

$$\text{So } H(j\omega) = \frac{2}{2+j\omega} - \frac{1}{3+j\omega} \longrightarrow h(t) = 2e^{-2t} - e^{-3t}$$

2. Find the Fourier transform of the following signal:



Figure 2: Signals

**Solution:**

(a)

$$\begin{aligned} F_1(j\omega) &= \int_{-\infty}^{\infty} f_1(t) e^{-j\omega t} dt \\ &= \int_0^{\tau} e^{-j\omega t} dt \\ &= \frac{1 - e^{-j\omega\tau}}{j\omega} \\ &= \frac{(e^{\frac{j\omega\tau}{2}} - e^{-\frac{j\omega\tau}{2}})}{j\omega} e^{-\frac{j\omega\tau}{2}} \\ &= \tau \frac{e^{\frac{j\omega\tau}{2}} - e^{-\frac{j\omega\tau}{2}}}{2 \frac{\omega\tau}{2}} e^{-\frac{j\omega\tau}{2}} \\ &= \tau \frac{\sin(\frac{j\omega\tau}{2})}{\frac{j\omega\tau}{2}} e^{-\frac{j\omega\tau}{2}} \\ &= \tau \text{Sa}(\frac{j\omega\tau}{2}) e^{-\frac{j\omega\tau}{2}} \end{aligned}$$

(b)

$$\begin{aligned} F_2(j\omega) &= \int_{-\infty}^{\infty} f_2(t)e^{-j\omega t} dt \\ &= \int_0^{\tau} \frac{1}{\tau} te^{-j\omega t} dt \\ &= \frac{1}{\tau} \frac{1}{-j\omega} \int_0^{\tau} td(e^{-j\omega t}) \\ &= -\frac{1}{j\omega\tau} [te^{-j\omega t}|_0^{\tau} - \int_0^{\tau} e^{-j\omega t} dt] \\ &= -\frac{1}{j\omega\tau} [\tau e^{-j\omega\tau} + \frac{1}{j\omega} (e^{-j\omega\tau} - 1)] \\ &= \frac{j\omega\tau e^{-j\omega\tau} + e^{-j\omega\tau} - 1}{\omega^2\tau} \end{aligned}$$