

**ECE 203**

**Circuits I**

**Power revisited**

**Lecture 13-2**

# Calculating Power for Sinusoidal Signals

$$v(t) = V_M \cos(\omega t + \theta_v)$$

$$i(t) = I_M \cos(\omega t + \theta_i)$$

$$p(t) = v(t)i(t)$$

Power delivered or absorbed  
varies with time

$$= V_M I_M \cos(\omega t + \theta_v) \cos(\omega t + \theta_i)$$

Instantaneous power at any time  $t$  can be calculated using this expression.

# Power calculations continued

From previous expression, using trig identity:

$$p(t) = \frac{V_M I_M}{2} [\cos(\theta_v - \theta_i) + \cos(2\omega t + \theta_v + \theta_i)]$$

Can also calculate instantaneous power using this expression.

# Average Power ( $P$ )

- To calculate average power, integrate  $p(t)$  over one cycle and divide by period

$$\begin{aligned} P &= \frac{1}{T} \int_{t_0}^{t_0+T} p(t) dt = \frac{1}{T} \int_{t_0}^{t_0+T} [V_M \cos(\omega t + \theta_v)] [I_M \cos(\omega t + \theta_i)] dt \\ &= \frac{1}{2} V_M I_M \cos(\theta_v - \theta_i) \end{aligned}$$

- Can use this to calculate power to entire circuit, or to individual circuit elements
- Recall that passive sign convention says:
  - $P > 0$ , power is being absorbed
  - $P < 0$ , power is being supplied

# Average Power: Special Cases

- **Purely resistive circuit:**  $\theta_v = \theta_i$

$$P = \frac{1}{2} V_M I_M$$

The power dissipated in a resistor is

$$P = \frac{1}{2} V_M I_M = \frac{V_M^2}{2 R} = \frac{1}{2} I_M^2 R$$

- **Purely reactive circuit:**  $\theta_v - \theta_i = \pm 90^\circ$

$$P = 0$$

- Capacitors and inductors are lossless elements and absorb no average power
- A purely reactive network operates in a mode in which it stores energy over one part of the cycle and releases it over another part

# ***Average Power Summary***

Circuit Element	Average Power
V or I source	$P = \frac{1}{2} V_M I_M \cos(\theta_v - \theta_i)$
Resistor	$P = \frac{1}{2} V_M I_M = \frac{1}{2} I_M^2 R$
Capacitor or Inductor	$P = 0$

**Does the expression for the resistor power look identical to that for DC circuits?**

# Example

- Go to example 13-2.1

# Maximum Power Transfer in ac Circuits

Power delivered to a load:

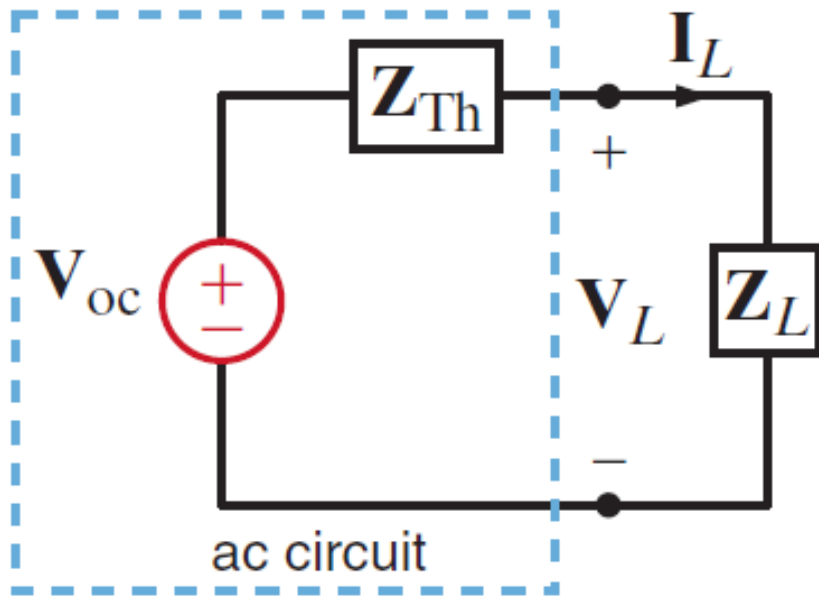
$$P_L = \frac{1}{2} V_L I_L \cos(\theta_{v_L} - \theta_{i_L})$$

$$\theta_{v_L} - \theta_{i_L} = \theta_{Z_L}$$

$$Z_L = R_L + jX_L$$

$$\text{So, } \cos \theta_{Z_L} = \frac{R_L}{(R_L^2 + X_L^2)^{1/2}}$$





$$I_L = \frac{V_{oc}}{Z_{Th} + Z_L}$$

$$V_L = \frac{V_{oc} Z_L}{Z_{Th} + Z_L}$$

$$Z_{Th} = R_{Th} + jX_{Th}$$

$$Z_L = R_L + jX_L$$

Magnitude of current and voltage at the load:

$$I_L = \frac{V_{oc}}{[(R_{Th} + R_L)^2 + (X_{Th} + X_L)^2]^{1/2}}$$

$$V_L = \frac{V_{oc}(R_L^2 + X_L^2)^{1/2}}{[(R_{Th} + R_L)^2 + (X_{Th} + X_L)^2]^{1/2}}$$

# Finding Max Power Transfer to Load

Magnitude of the power to the load:

$$P_L = \frac{1}{2} \frac{V_{oc}^2 R_L}{(R_{Th} + R_L)^2 + (X_{Th} + X_L)^2}$$

$R_{Th}$  and  $R_L$  are real and positive;  $X_{Th}$  and  $X_L$  are imaginary and can be positive or negative.

So  $P_L$  will be max when  $X_L = -X_{Th}$

# Max power transfer continued

$$P_L = \frac{1}{2} \frac{V_{oc}^2 R_L}{(R_L + R_{Th})^2}$$

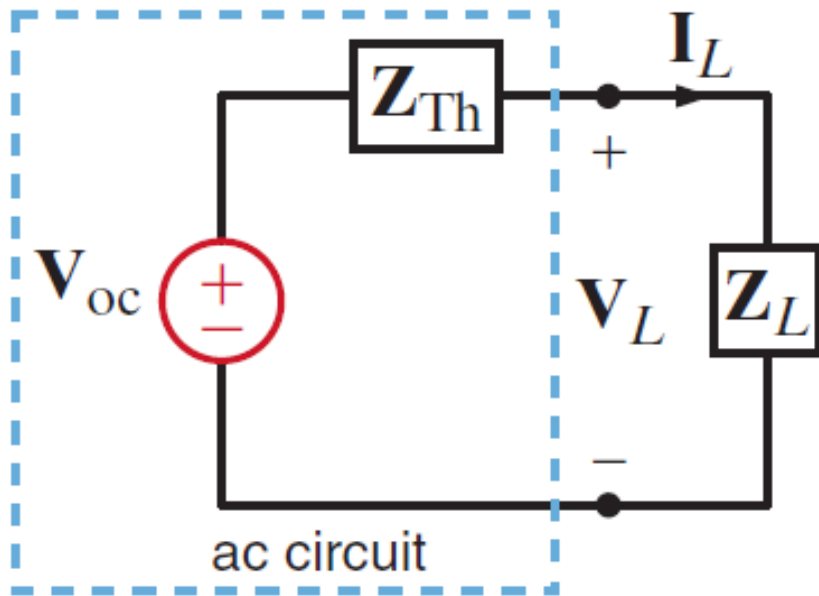
Just like when we had a purely resistive circuit, max power transfer when  $R_L = R_{Th}$ . So:

$$P_{L,Max} = \frac{1}{8} \frac{V_{oc}^2}{R_{Th}}$$

# Impedance for Max Power Transfer

$$\mathbf{Z}_L = R_L + jX_L = R_{Th} - jX_{Th} = \mathbf{Z}_{Th}^*$$

So, max power transfer occurs when the load is equal to the complex conjugate of the Thevenin impedance of the driving circuit.



## Maximum Average Power Transfer

To find the load for maximum power transfer:

- Step 1.** Remove the load  $Z_L$  and find the Thévenin equivalent for the remainder of the circuit.
- Step 2.** Construct the circuit shown in Fig. 9.6.
- Step 3.** Select  $Z_L = Z_{Th}^* = R_{Th} - jX_{Th}$ , and then  $I_L = V_{oc}/2 R_{Th}$  and the maximum average power transfer  $= \frac{1}{2} I_L^2 R_{Th} = V_{oc}^2/8 R_{Th}$ .

# Example

- Go to example 13-2.2

# Effective values of time-varying signals

We have already discussed how circuits react to dc and sinusoidal ac signals.

Can have a variety of different voltage or current sources exciting a circuit: square waves, triangle waves, ...

Would like to talk about the “effective” voltage or current of these signals

# Effective values of time-varying signals

Average power:

$$P = \frac{1}{T} \int_{t_o}^{t_o+T} i^2(t) R dt$$

Or:  $P = I_{eff}^2 R$

So:  $I_{eff} = \left[ \frac{1}{T} \int_{t_o}^{t_o+T} i^2(t) dt \right]^{1/2}$



# *Effective or RMS Values*

- Root-mean-square value (formula reads like the name: rms)

$$I_{rms} = \sqrt{\frac{1}{T} \int_{t_0}^{t_0+T} i^2(t) dt}$$

and

$$V_{rms} = \sqrt{\frac{1}{T} \int_{t_0}^{t_0+T} v^2(t) dt}$$

- For a sinusoid:  $I_{rms} = I_M/\sqrt{2}$ 
  - For example, AC household outlets are around 120 Volts-rms

# Why RMS Values?

- The effective/rms current allows us to write average power expressions like those used in dc circuits (i.e.,  $P=I^2R$ ); that relation is really the basis for defining the rms value
- The average power ( $P$ ) is


$$P_{source} = \frac{1}{2} V_M I_M \cos(\theta_v - \theta_i) = V_{rms} I_{rms} \cos(\theta_v - \theta_i)$$

$$P_{resistor} = \frac{1}{2} V_M I_M = V_{rms} I_{rms} = \frac{V_{rms}^2}{R} = I_{rms}^2 R$$

- Go to example 13-2.3

# Power Factor

- Power factor ( $0 \leq pf \leq 1$ )


$$pf = \frac{\text{average power}}{\text{apparent power}} = \frac{P}{V_{rms} I_{rms}}$$
$$= \frac{V_{rms} I_{rms} \cos(\theta_v - \theta_i)}{V_{rms} I_{rms}} = \cos(\theta_v - \theta_i) = \cos(\theta_{Z_L})$$

- A low power factor requires more rms current for the same load power which results in greater utility transmission losses in the power lines, therefore utilities penalize customers with a low  $pf$

# ***Power Factor Angle $\theta_{Z_L}$***

- *power factor angle* is  $\theta_v - \theta_i = \theta_{Z_L}$  (the phase angle of the load impedance)
- *power factor (pf) special cases*
  - purely resistive load:  $\theta_{Z_L} = 0^\circ \Rightarrow pf=1$
  - purely reactive load:  $\theta_{Z_L} = \pm 90^\circ \Rightarrow pf=0$

Power Factor Angle	I/V Lag/Lead	Load Equivalent
$-90^\circ < \theta_{Z_L} < 0^\circ$	Leading	Equivalent RC
$0^\circ < \theta_{Z_L} < 90^\circ$	Lagging	Equivalent RL

- Go to example 13-2.4