



# Signal Processing for Astronomy

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Narrabri – 25 September 2017  
+ some slides from John Tuthill



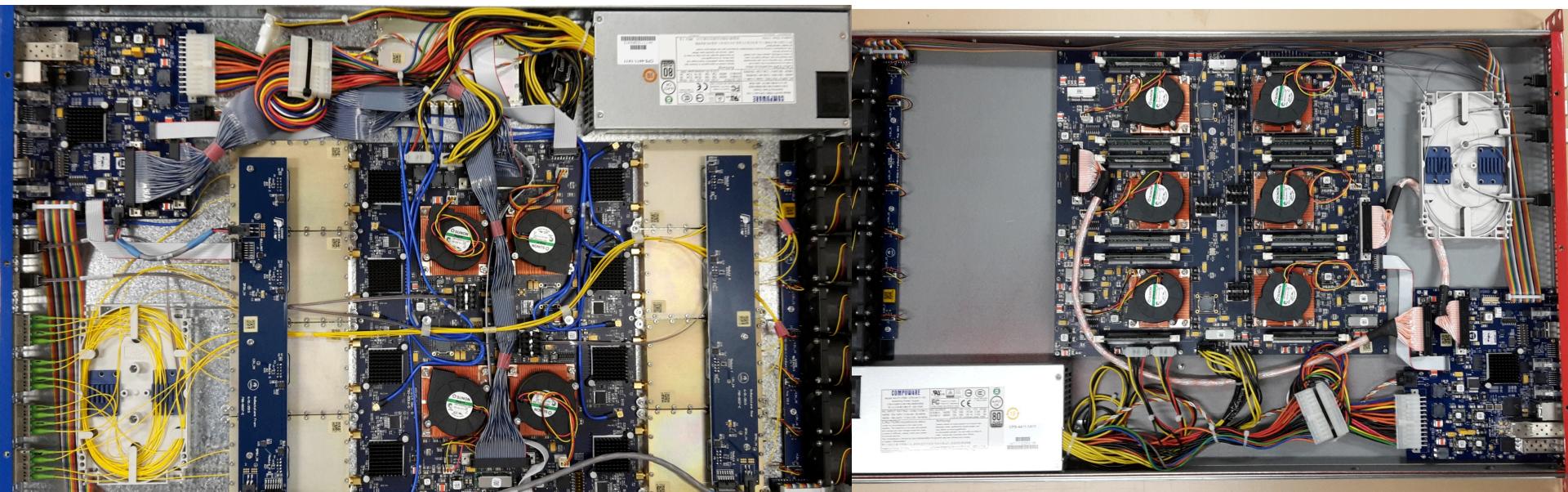
# Context

Digital Signal Processing – manipulations of discrete samples approximating some continuous function

- Beamformer
- Tied-array unit
- Fringe rotator
- Correlator

In astronomy:  
this is about backends

- well-defined core calculations
- often highly optimised
- often implemented in hardware



A large white radio telescope dish is positioned in a field of dry, yellowish-brown grass. In the foreground, several kangaroos are grazing. The sky is clear and blue.

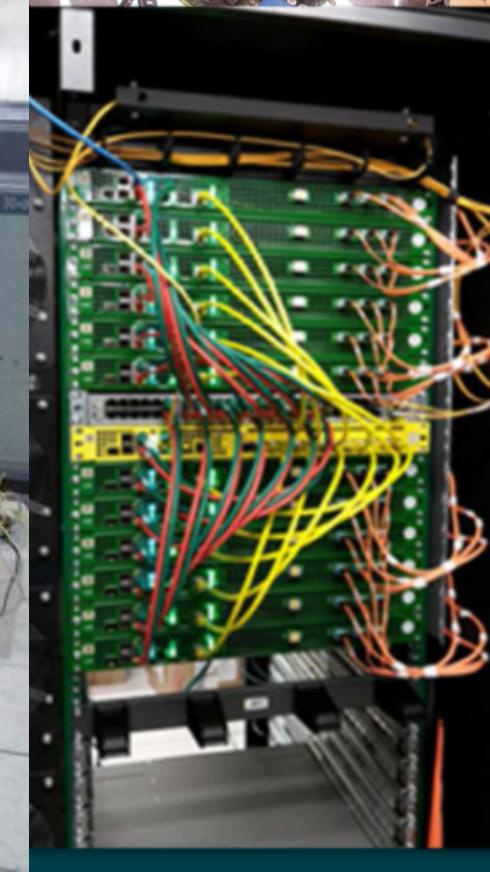
Correlator is the brain  
of radio-interferometer

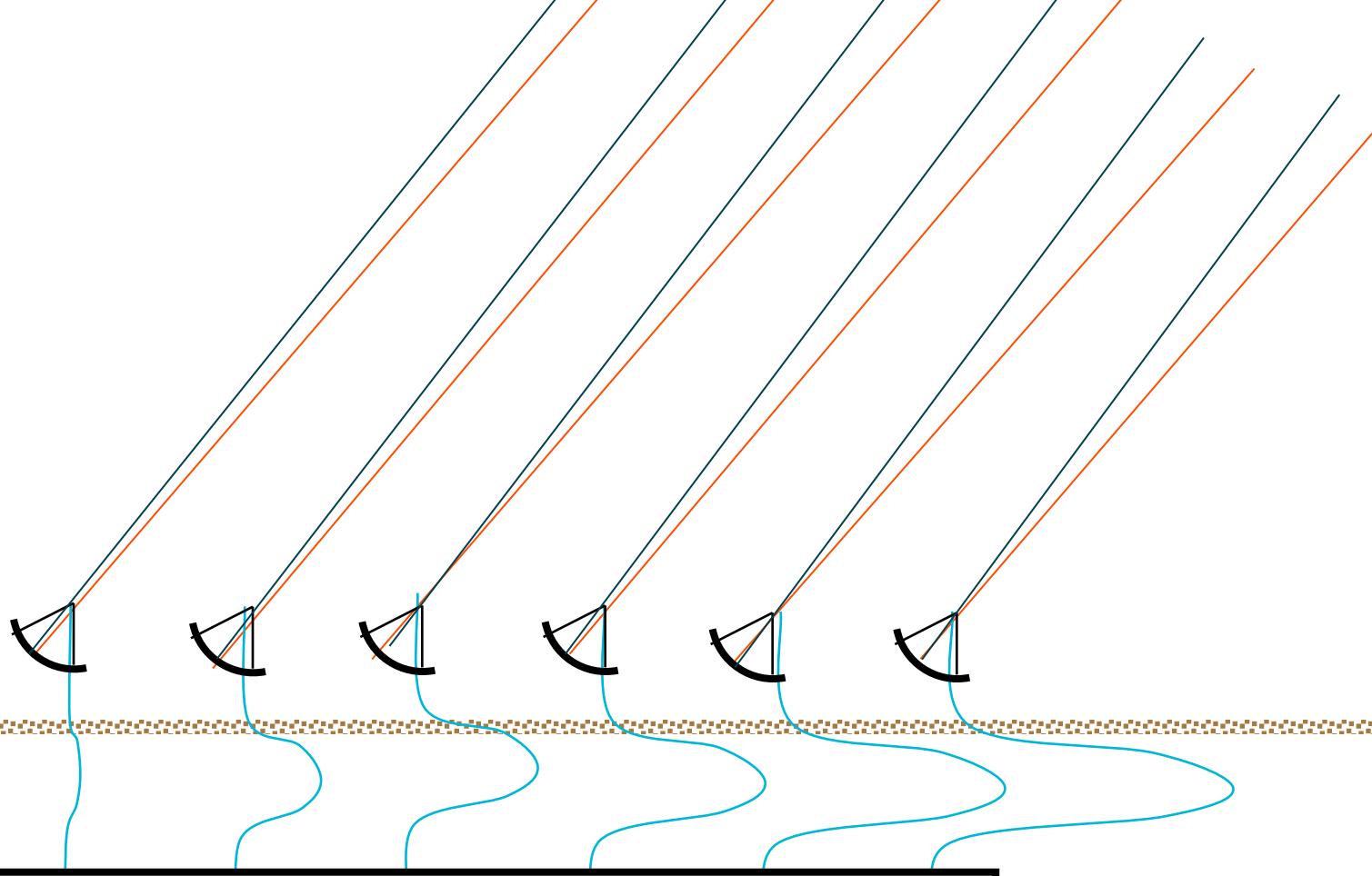
*(as it computes visibilities)*



# **”black box” for most science work**

May need to know details,  
if you start pushing the limits





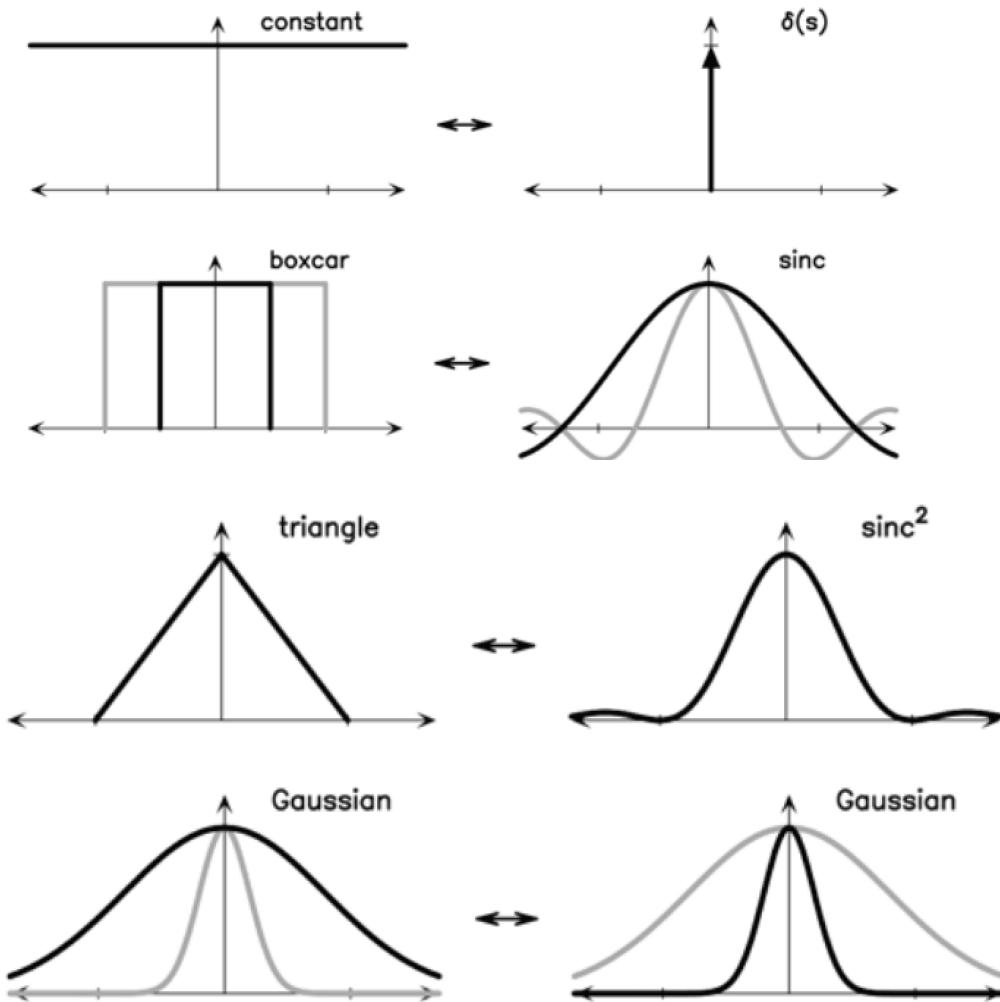
van Cittert-Zernike  
theorem

Fourier Transform

Image credit: Ron Ekers

$I(r)$

# Fourier transforms – intro



Forward:

$$F(s) \equiv \int_{-\infty}^{\infty} f(x) e^{-2\pi i s x} dx$$

Reverse:

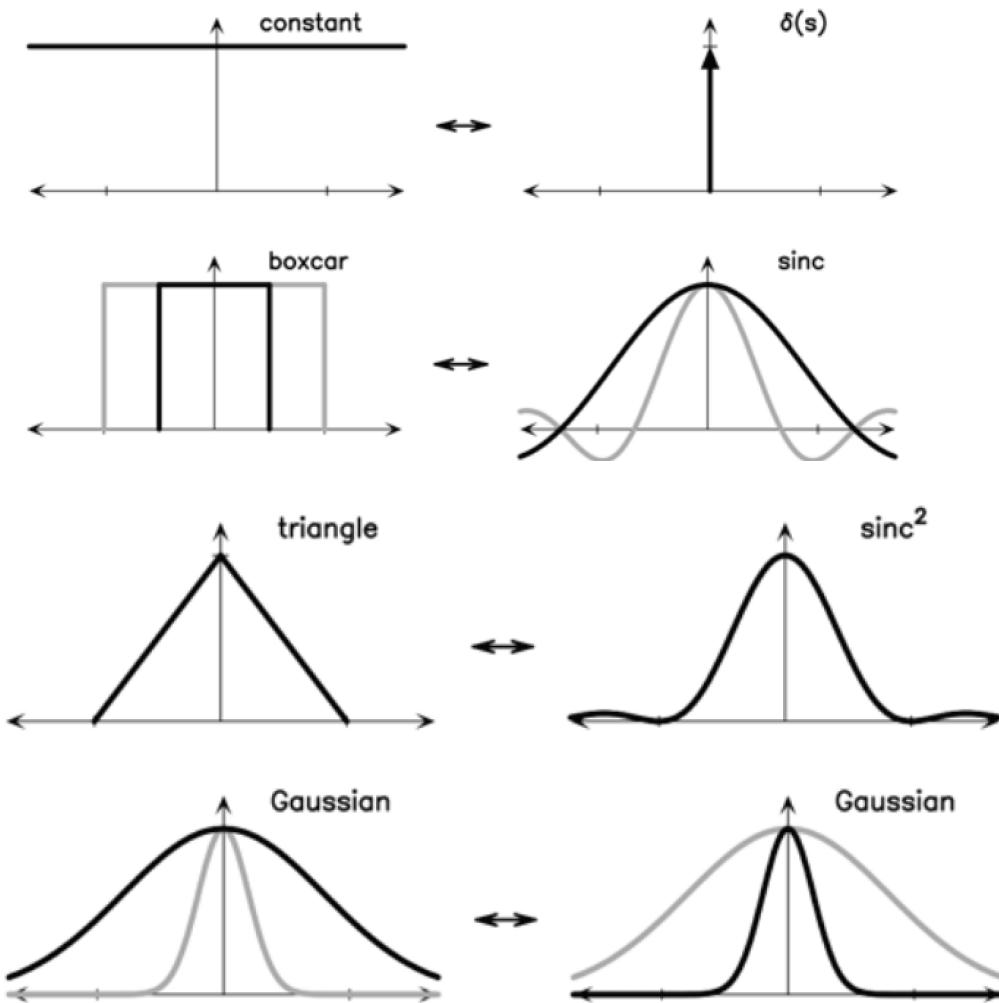
$$f(x) \equiv \int_{-\infty}^{\infty} F(s) e^{2\pi i s x} ds$$

Small  $\leftrightarrow$  Large

Convolution  $\leftrightarrow$  Product

# Fourier transforms – intro

$$j = \sqrt{-1}$$



Forward:

$$F(s) \equiv \int_{-\infty}^{\infty} f(x) e^{-2\pi j s x} dx$$

Reverse:

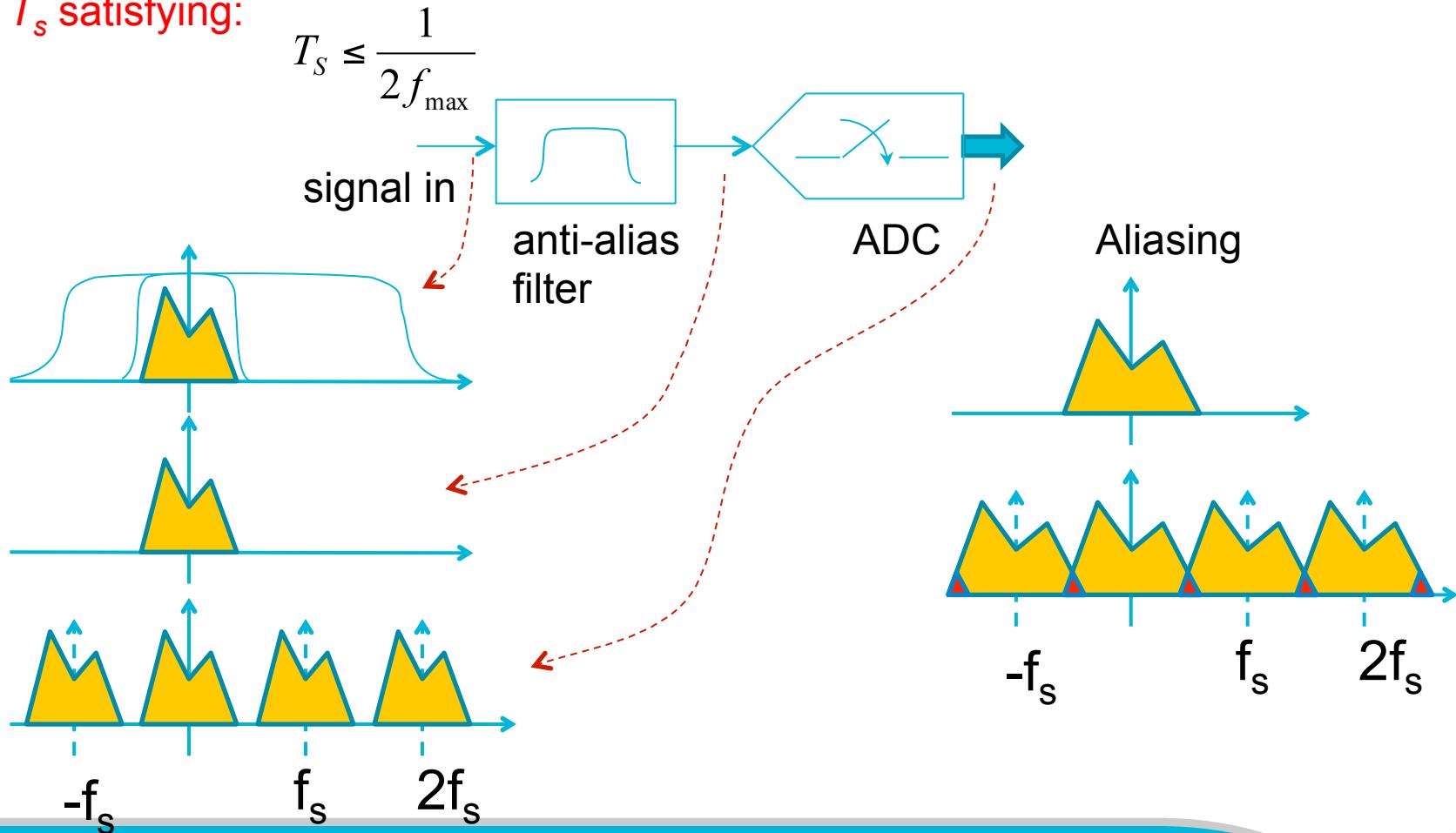
$$f(x) \equiv \int_{-\infty}^{\infty} F(s) e^{2\pi j s x} ds$$

Small  $\leftrightarrow$  Large

Convolution  $\leftrightarrow$  Product

# Sampling

*The Sampling Theorem:* A band-limited signal having no frequency components above  $f_{\max}$  can be determined uniquely by values sampled at uniform intervals of  $T_s$  satisfying:



# Simplest possible case

Assuming bandwidth is small enough & we don't care about frequency structure

$$V_{ij} = \frac{1}{N} \sum_{k=1}^N E_i(k) \times E_j^*(k - \tau)$$



Compute correlation directly  
(e.g. in software)

Code snippet from BETA-3 (3-baseline) software correlator:

```
IndexType offset1 = itsDelay1;  
IndexType offset2 = itsDelay2;  
IndexType offset3 = itsDelay3;
```

```
for (; (offset1 < size)&&(offset2 < size)&&(offset3 < size); ++offset1, ++offset2, ++offset3) {  
    itsVis12 += *(stream1 + offset1) * conj(*((stream2+offset2));  
    itsVis13 += *(stream1 + offset1) * conj(*((stream3+offset3));  
    itsVis23 += *(stream2 + offset2) * conj(*((stream3+offset3));  
}
```

# Non-monochromatic input signal

More intuitive description via continuous formalism:

$$E(t) = \int s(v) \exp\{2\pi jvt\} dv$$

*spectral representation*

Correlation between antenna 1 and 2 data streams for a given lag:

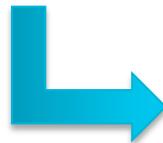
$$\gamma(\tau) = \int E_1(t) E_2^*(t - \tau) dt = \int s_1(v) s_2^*(v) \exp\{2\pi jv\tau\} dv$$

Option 1

*Power (cross-correlation) spectrum*

Correlate streams for a number of lags and Fourier-transform

Option 2



Lag or XF correlator

Fourier-transform input streams and cross-multiply

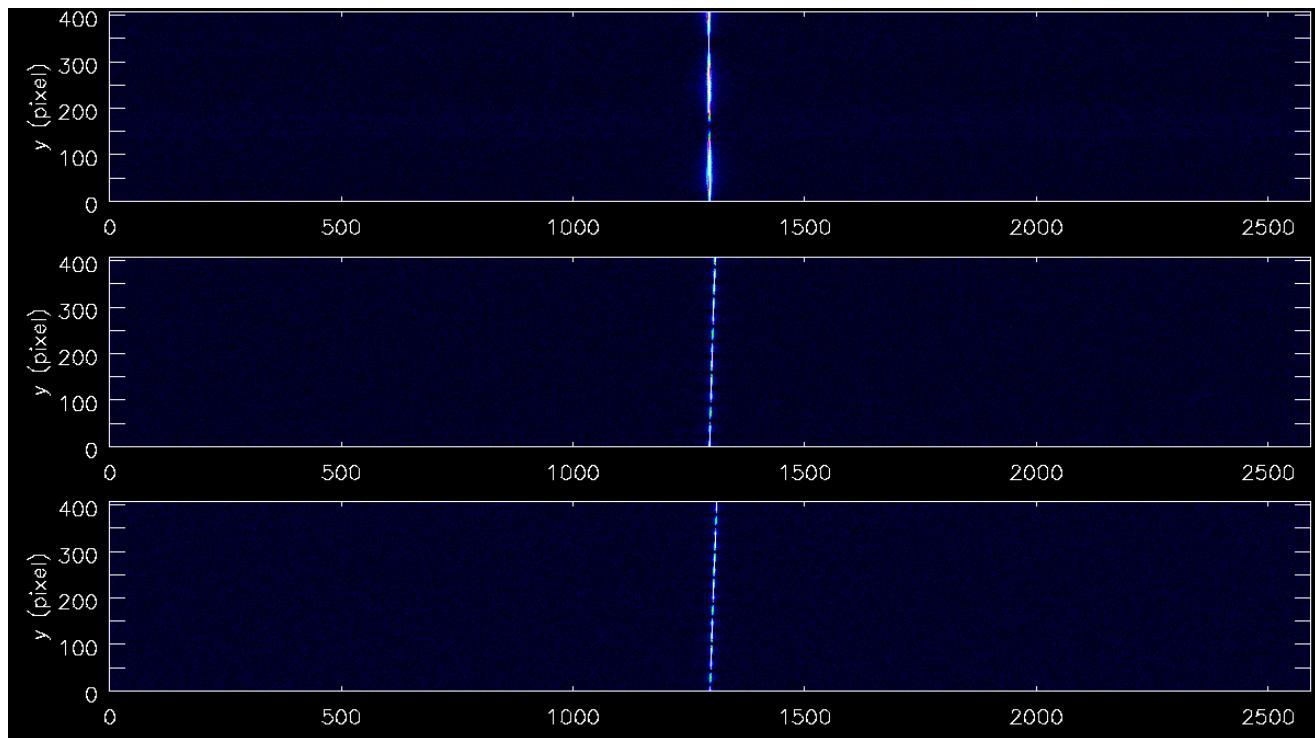
FX-correlator

# More on lag domain

Regardless of the correlator architecture, lag spectrum

$$\gamma(\tau) = \int V_{12}(v) \exp\{2\pi j v \tau\} dv \quad \text{is a useful diagnostic tool}$$

Correlator cycle



Lag (1296 pixel is the centre)

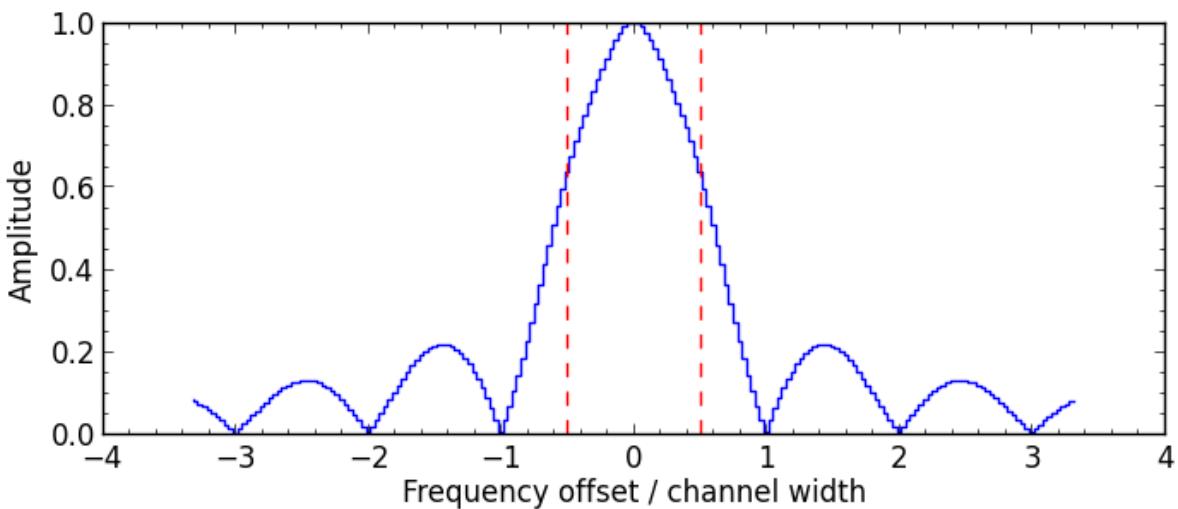
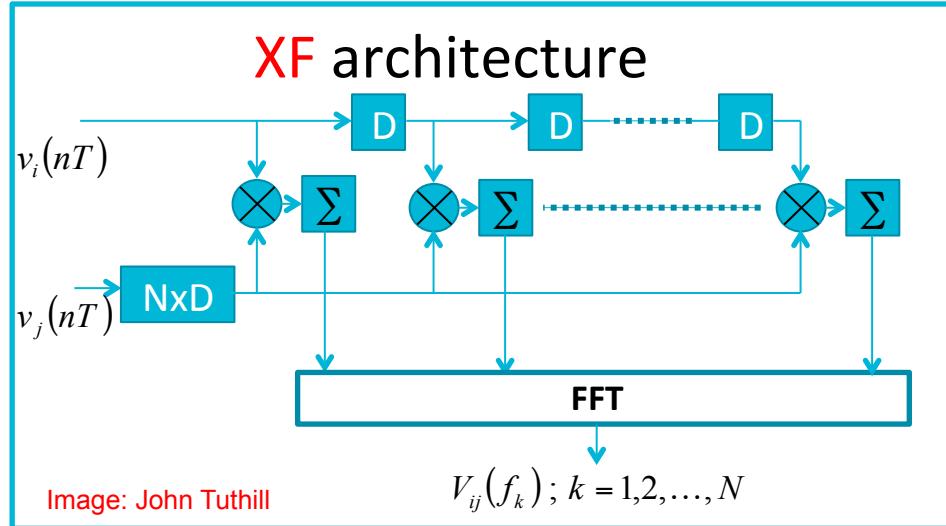
AK04-AK05

AK04-AK12

AK05-AK12

# Lag (XF) correlator

Old ATCA correlator (pre-CABB) is a good example



Fractional delays (< 1 sample)  
need to be corrected in some  
other way

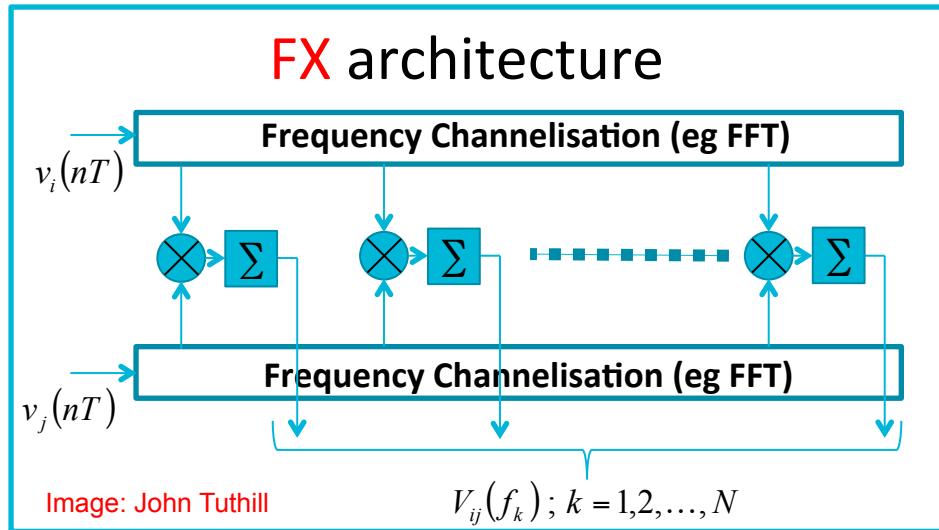
$$\gamma(\tau) = \int V_{12}(\nu) \exp\{2\pi j\nu\tau\} d\nu$$

Only finite number of lags can  
be measured



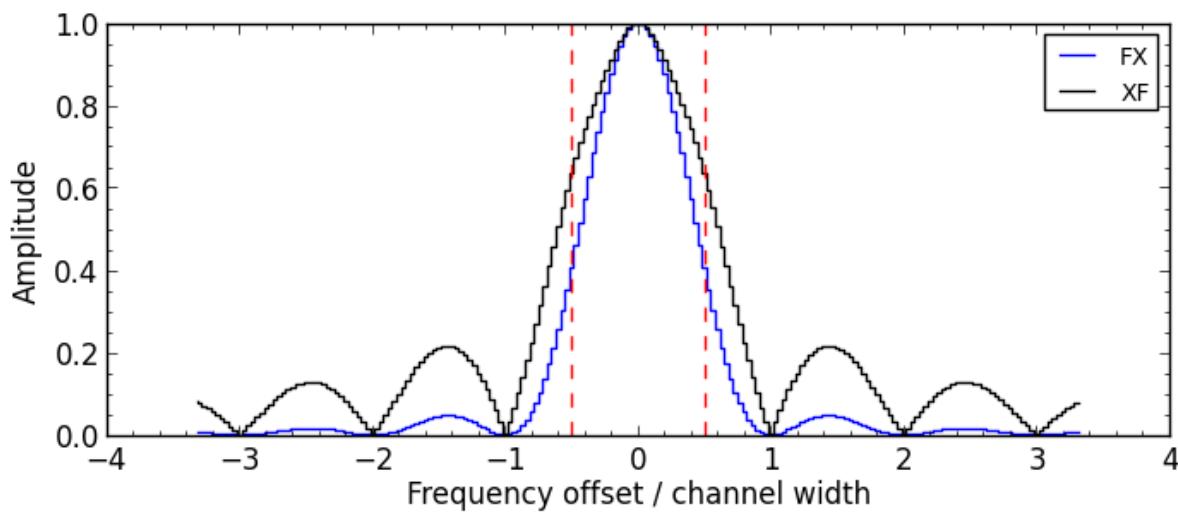
Convolution with *sinc* in the  
frequency domain

# FX correlator



Fractional delays are easy to implement

Output is the product of two Fourier Transforms, each is presented with a finite chunk of data

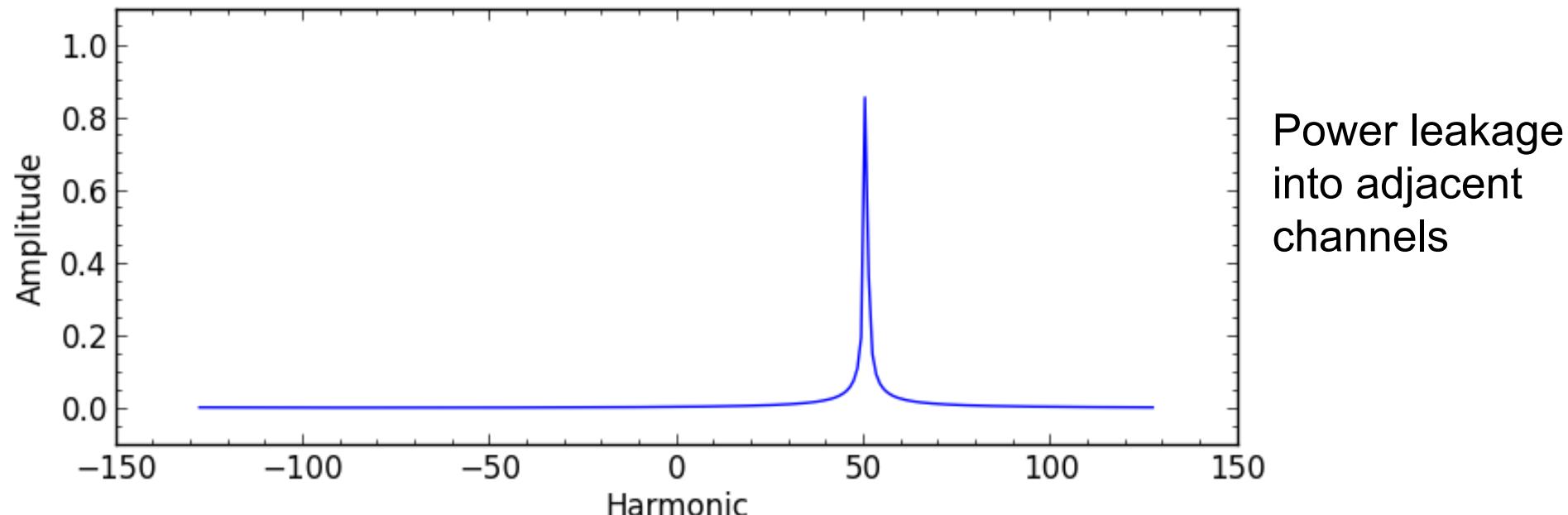


Convolution with  $\text{sinc}^2$  in the frequency domain

# Can we get a better channel response?

Let's consider FX architecture as a channeliser + cross-multiplier  
(think of a simple correlator described earlier)

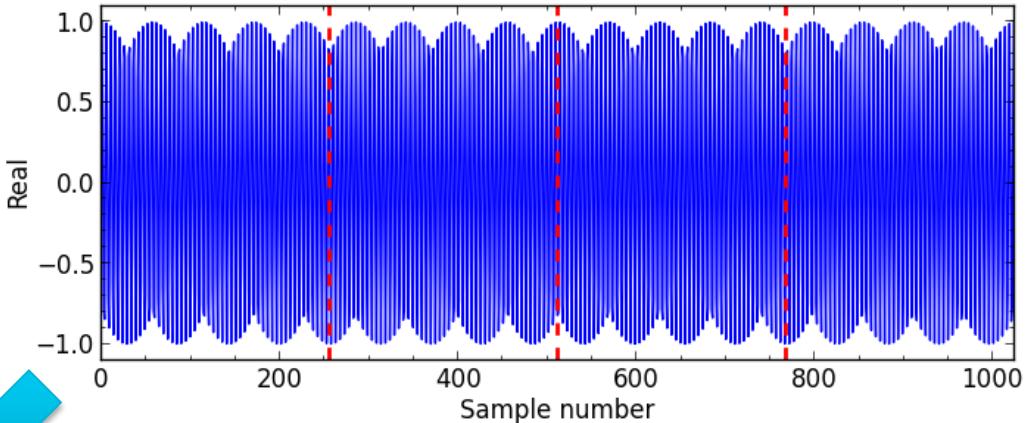
The channel response is determined by the channeliser performance  
and FFT is known to be quite bad



FFT of the sine wave with frequency of  $50.3 * f_s / 256$

# Filtering

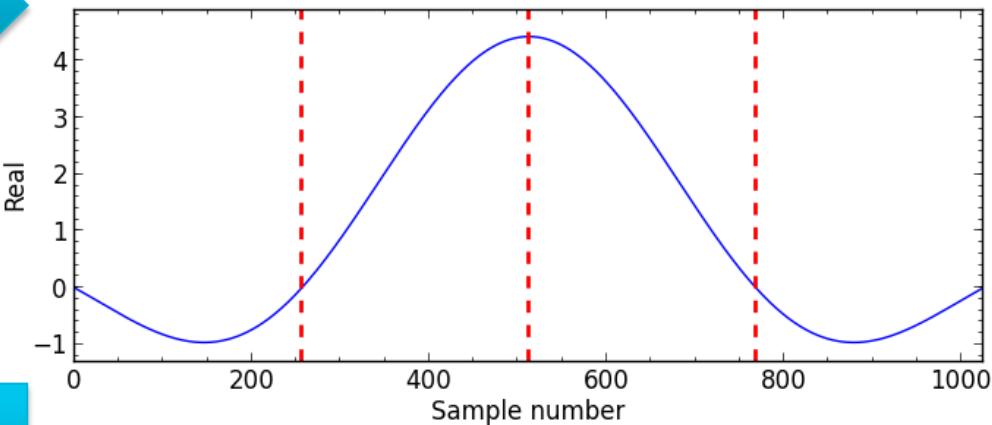
Input data samples



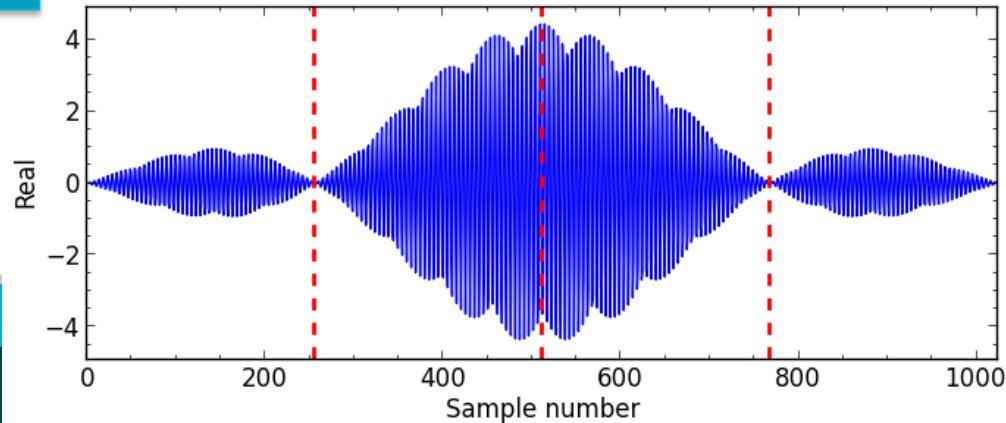
Sine wave with the frequency of  $50.3 f_s / 256$

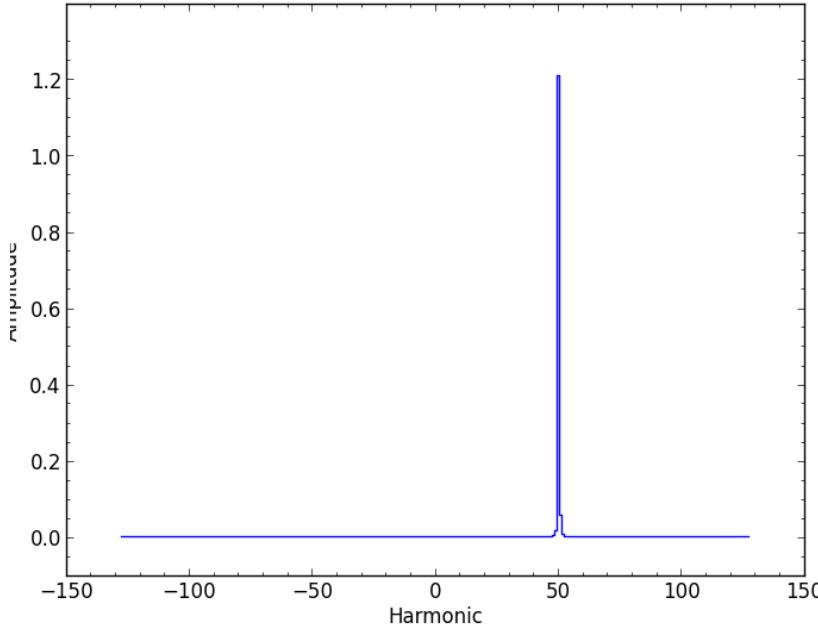
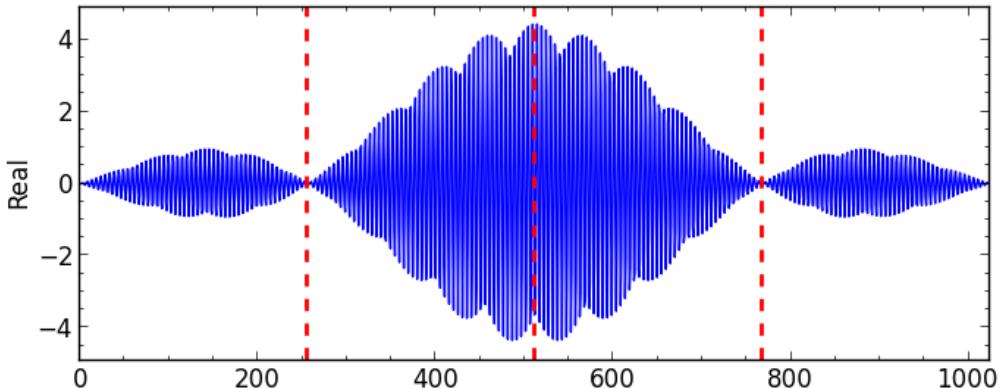
Filter

(Sinc in this particular example)



Filtered data samples





In this example:

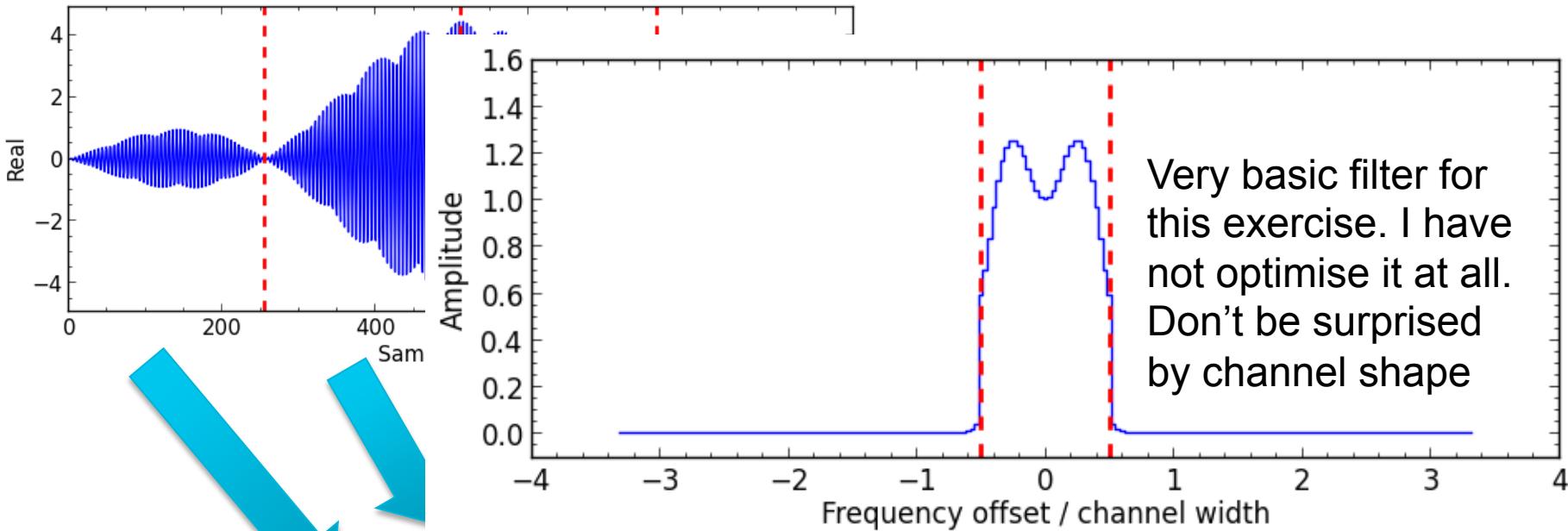
$$N = 256$$

$$M = 4$$

$$y(n) = \sum_{m=0}^{M-1} x(n + mN)h(n + mN)$$

**FFT**

polyphase sub-filters



Very basic filter for this exercise. I have not optimise it at all. Don't be surprised by channel shape

In this example:

$$N = 256$$

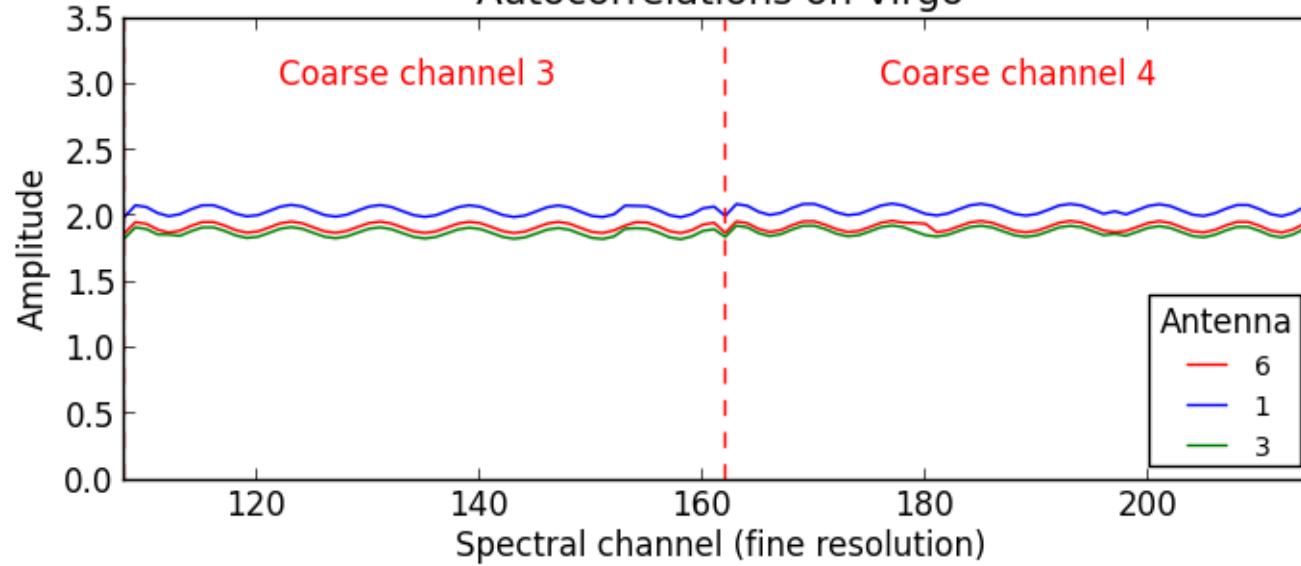
$$M = 4$$

**FFT**

$$y(n) = \sum_{m=0}^{M-1} x(n + mN)h(n + mN)$$

polyphase sub-filters

### Autocorrelations on Virgo

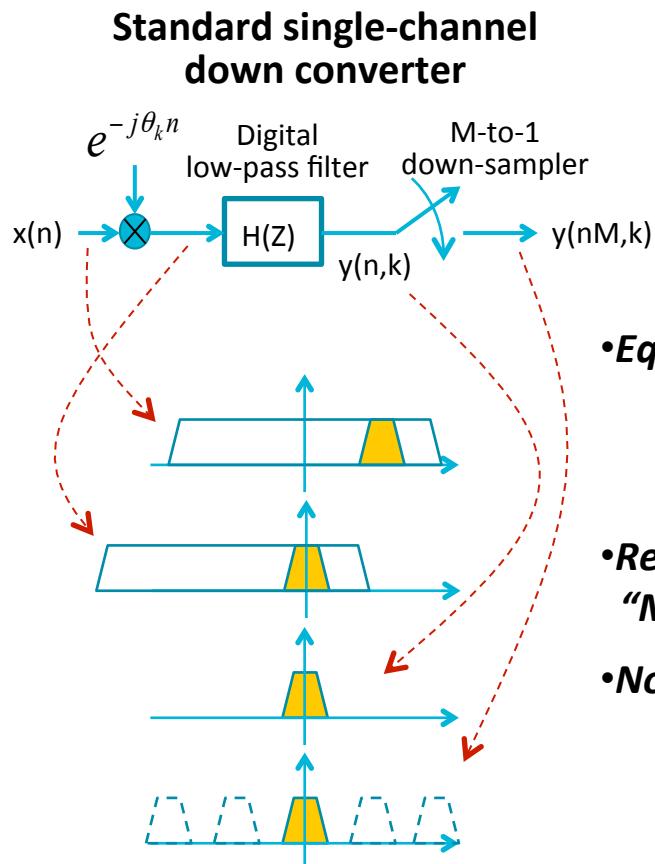


**But filter does  
cause a ripple  
(can be corrected)**

Early BETA tests

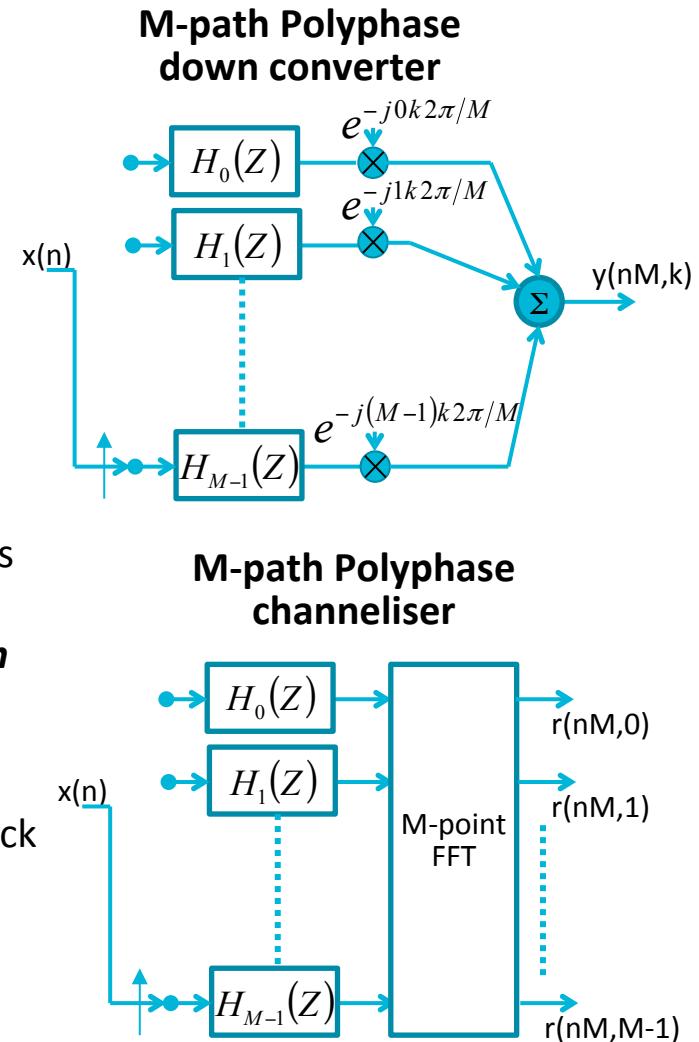


# Polyphase decomposition: engineer's view

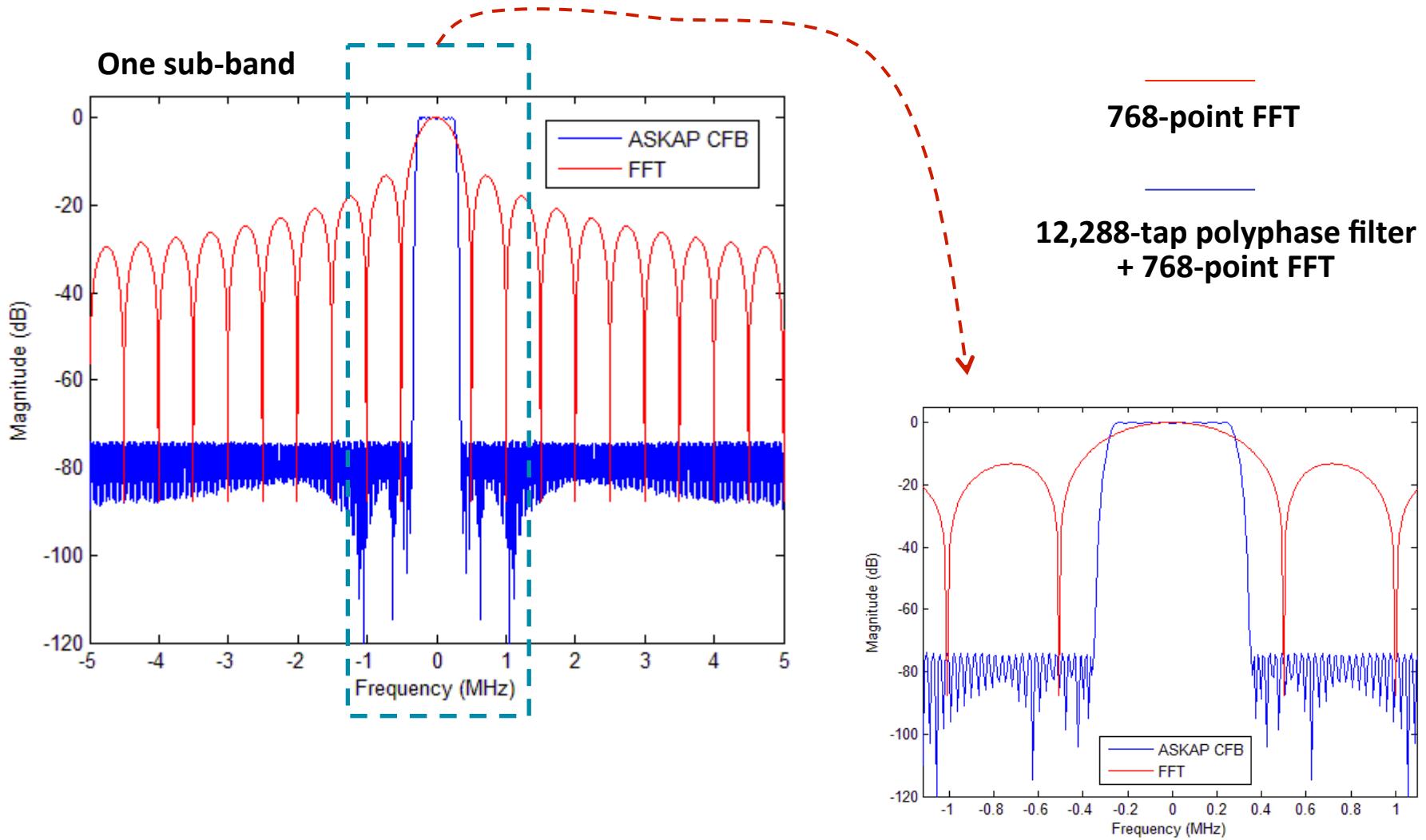


- **Equivalency Theorem**
  - Exchange mixer and low-pass filter with a band-pass filter and a mixer.
- **Re-write the band-pass filter in “M-path form”**
- **Noble Identity**
  - Move a down-sampler back through a digital filter

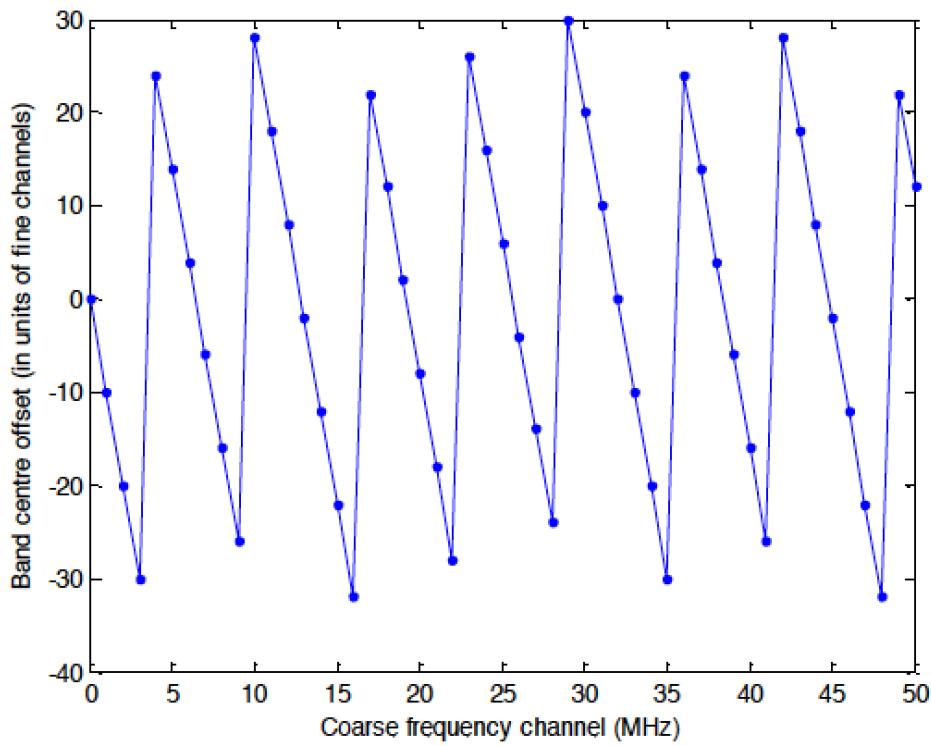
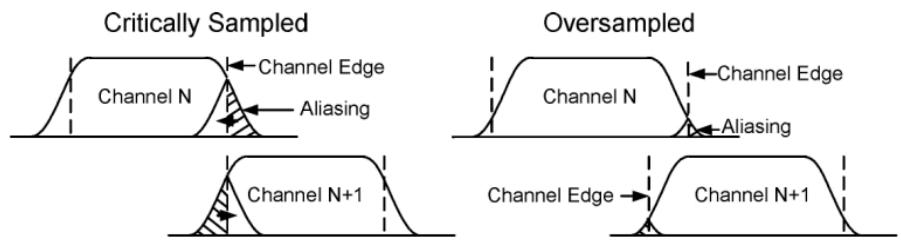
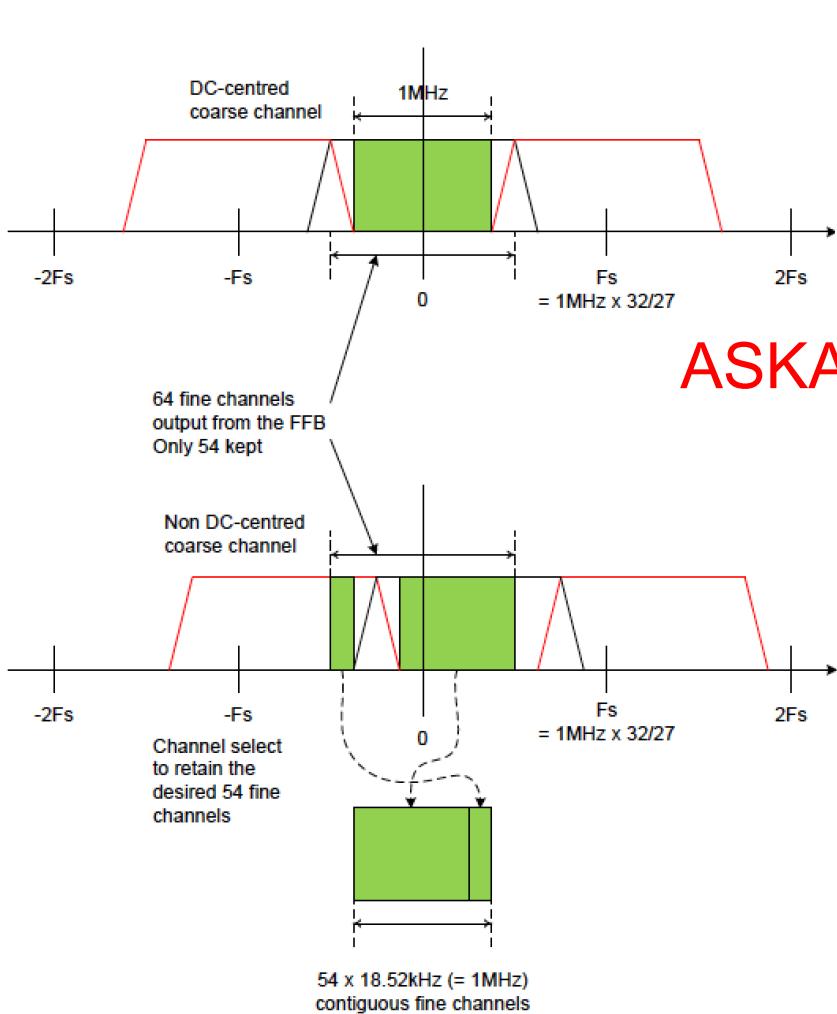
$$H(Z^M) \left( \downarrow M \right) = \left( \downarrow M \right) H(Z)$$



# Filterbanks: FFT vs Polyphase Filters



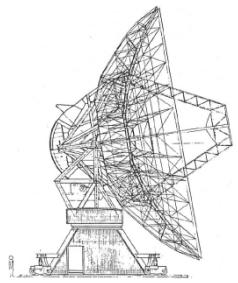
# A few words on oversampled PFBs



# ATCA + CABB see Wilson et al. (2011, MNRAS, 416, 832)



# Compact Array Broadband Backend (CABB)



Per antenna

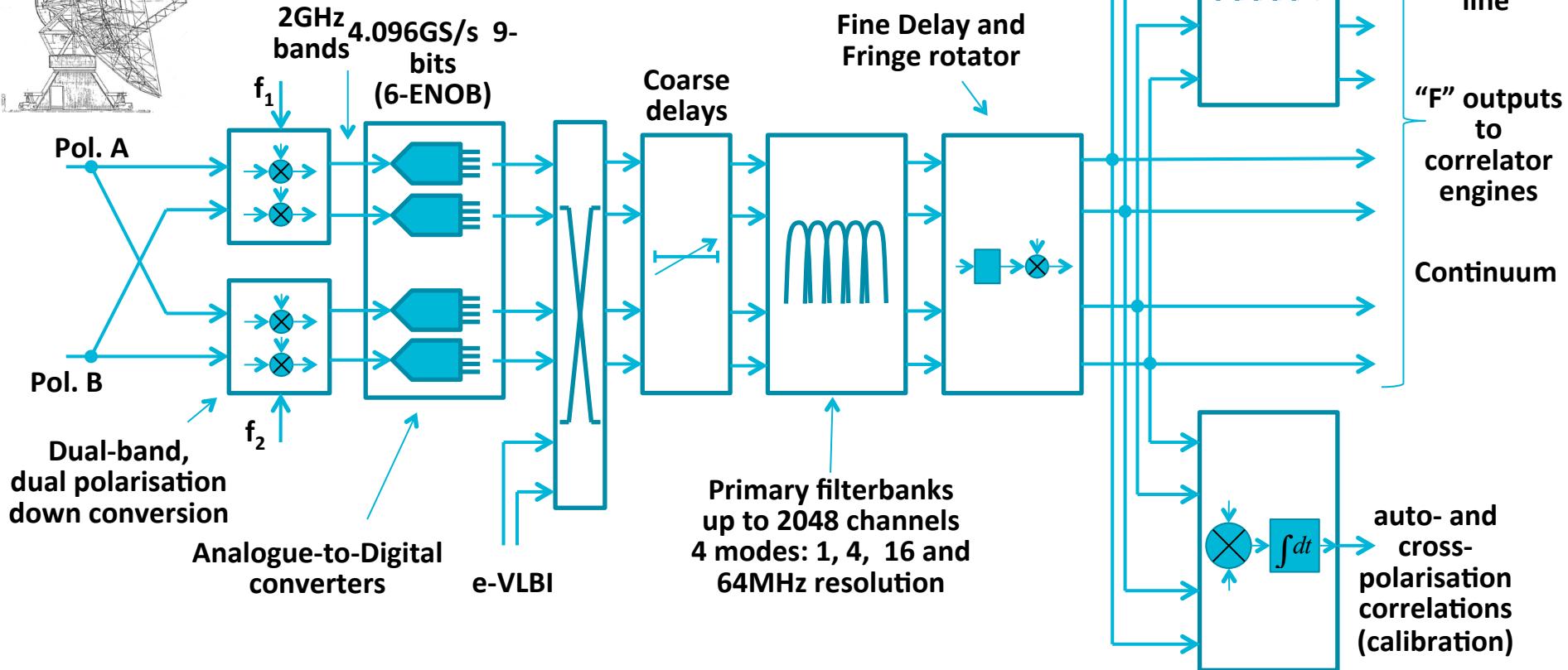


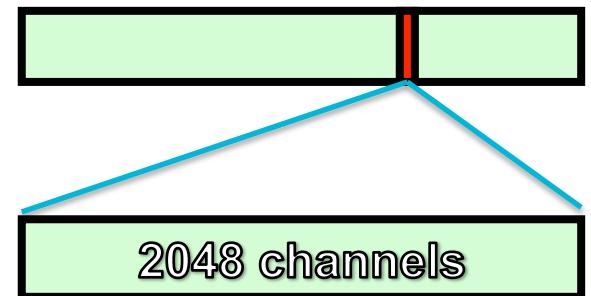
Image credit: John Tuthill



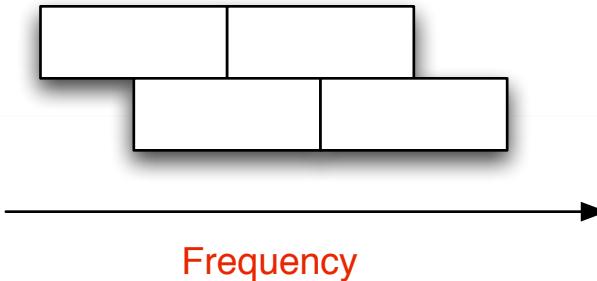
# Zoom modes: high spectral resolution



- Up to 16 “zoomed in” channels per each 2 GHz window (user selected)
- Positioned in steps equal to half of the wide-band spectral resolution (i.e. 0.5 MHz or 32 MHz)
- Each zoom window has 2048 spectral channels



Stitching zoom windows:



# CABB Correlator – correlation engines

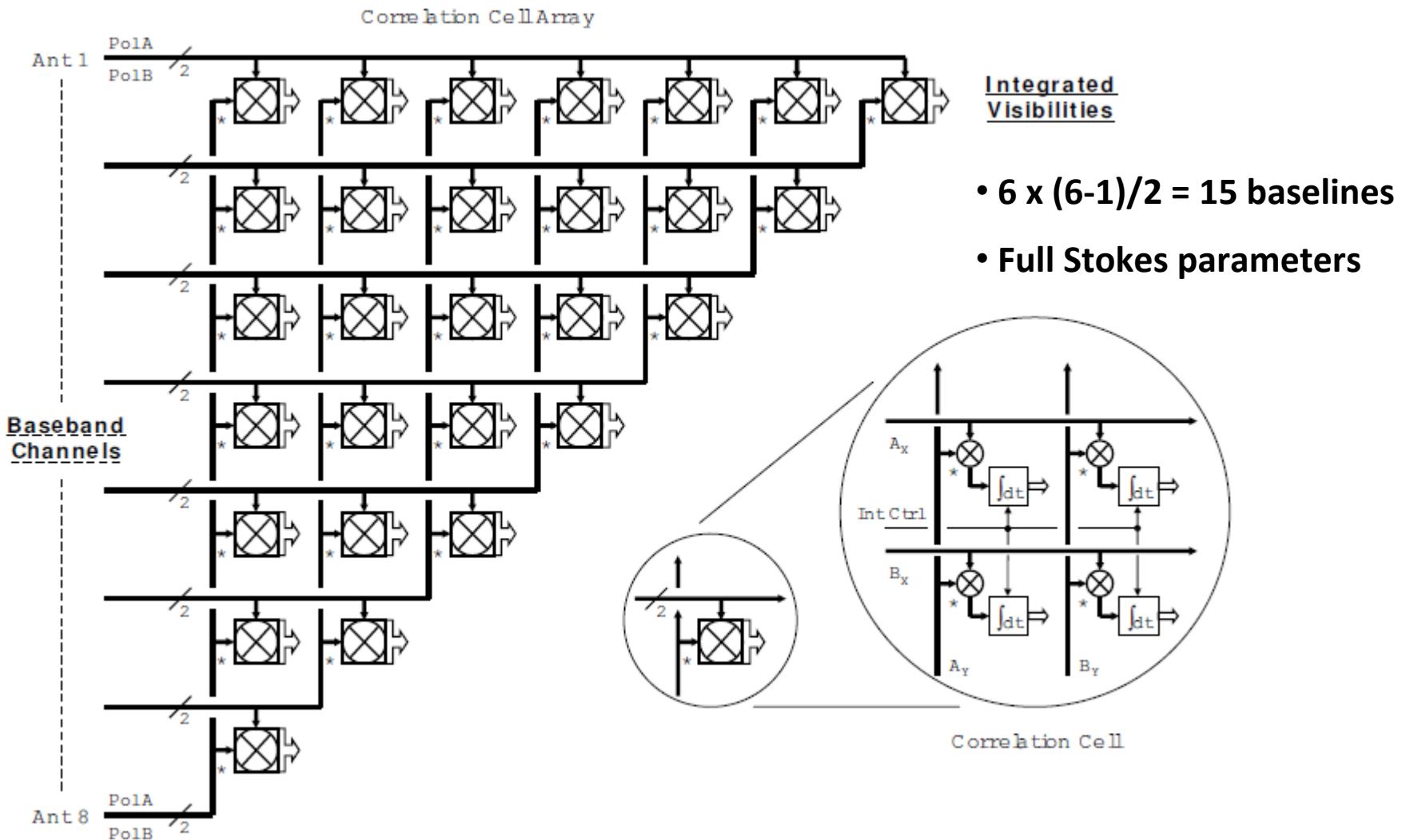


Image credit: John Tuthill

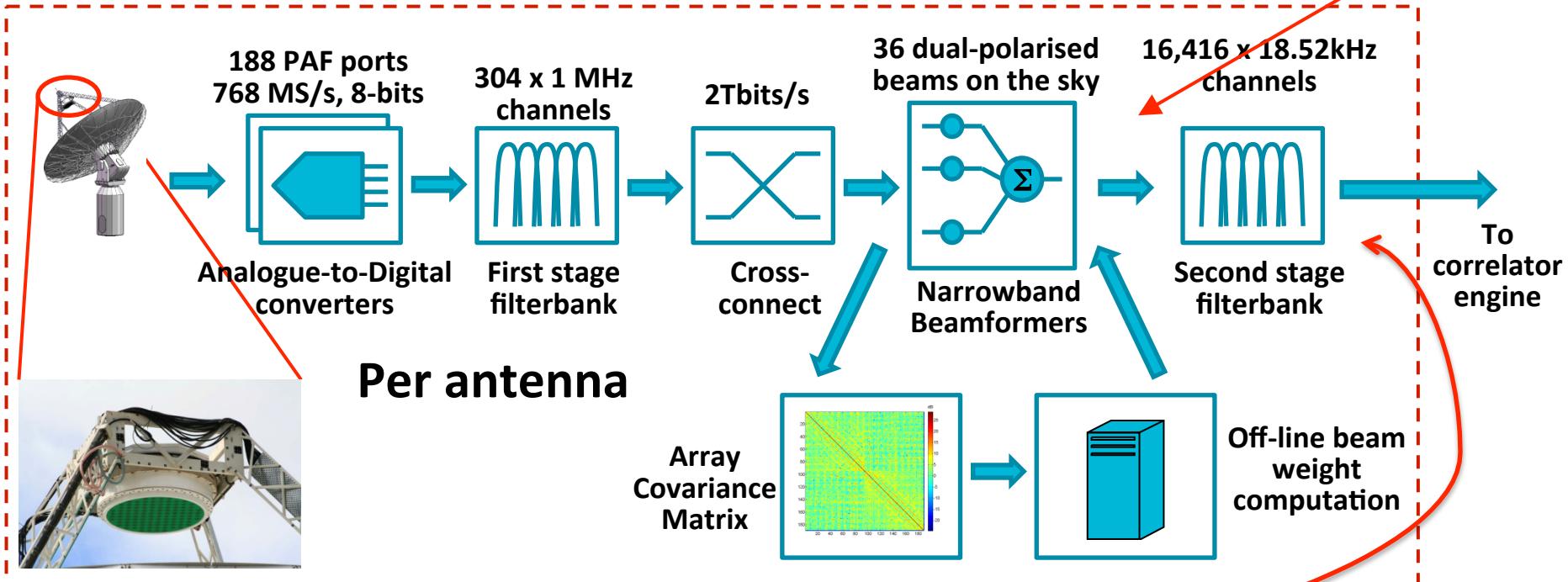
# Australian Square Kilometre Array Pathfinder

- Has beamformer before the correlator (needs channelisation)

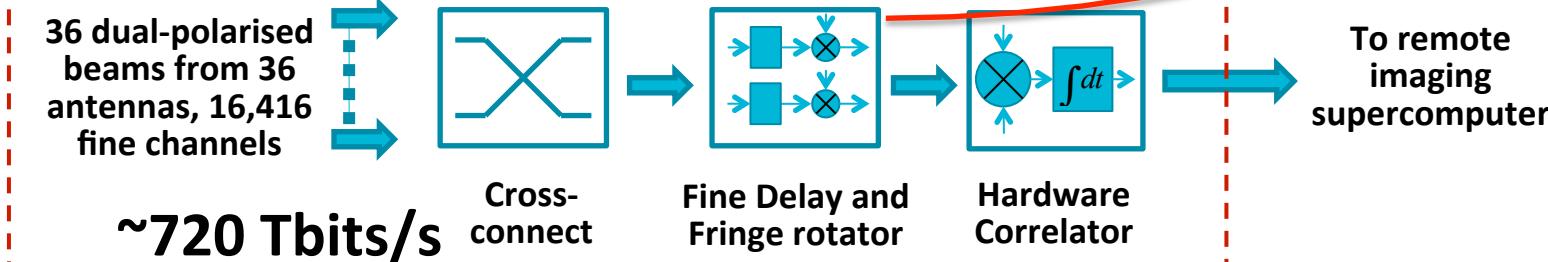


# ASKAP (BETA) digital back-end

Data throughput reduced by a factor of 2.7



Per antenna

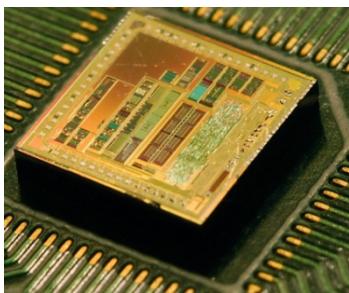


# Calculation Engines: so many choices...

Hard-wired logic

**ASIC's**

Application-Specific Integrated Circuit



- EVLA
- ALMA

- Less flexible
- Lower power/computation
- Higher initial development

Stored (programmed) logic

**FPGA's**

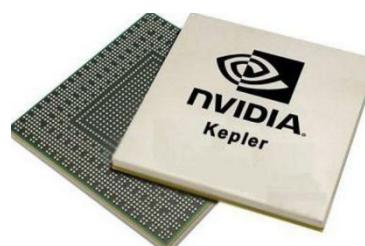
Field Programmable Gate Array



- CABB
- ASKAP

**GPU's**

Graphics Processing Unit



- MWA
- MeerKAT

**CPU's/DSP's**

Central Processing Unit/ Digital Signal Processor



- DiFX



- More flexible
- Higher power/computation
- Lower initial development



# Further Reading...

- Radio Astronomy:
  - H. C. Ko, "Coherence Theory in Radio-Astronomical Measurements," *IEEE Trans. Antennas & Propagation*, pp. 10-20, Vol. AP-15, No. 1, Jan. 1967.
  - G. B. Taylor, G. L. Carilli and R. A. Perley, *Synthesis Imaging in Radio Astronomy II*, Astron. Soc. Pac. Conf. Series, vol. 180, 2008.
- CABB
  - W. E. Wilson, et. al. "The Australia Telescope Compact Array Broadband Backend (CABB): Description & First Results," *Mon. Not. R. Astron. Soc.*, Feb. 2011
- ASKAP
  - D. R. DeBoer, et.al, "Australian SKA Pathfinder: A High-Dynamic Range Wide-Field of View Survey Telescope," *Proc. IEEE*, 2009.
- Filter Banks
  - R. E. Crochiere and L. R. Rabiner *Multirate Digital Signal Processing*, Prentice Hall, 1983.
  - F. J. Harris, *Multirate Signal Processing for Communication Systems*, Prentice Hall, 2008.
  - P. P. Vaidyanathan, *Multirate Systems And Filter Banks*, Prentice Hall, 1992.
- Beamforming
  - B. D. Van Veen and K. M. Buckley, "Beamforming: A Versatile Approach to Spatial Filtering," *IEEE ASSP Magazine*, April 1988

# Thank you

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