

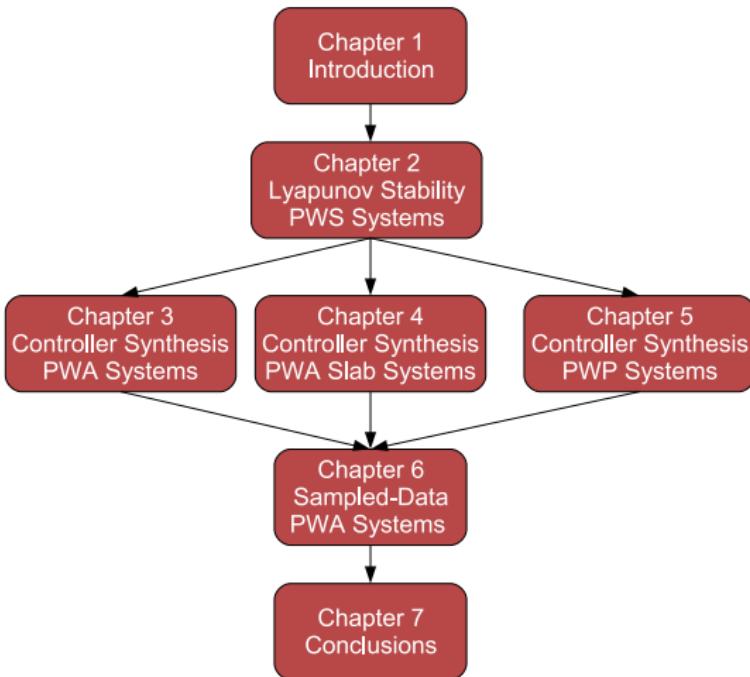
Stability Analysis and Controller Synthesis for a Class of Piecewise Smooth Systems

The Oral Examination
for the Degree of Doctor of Philosophy
Behzad Samadi

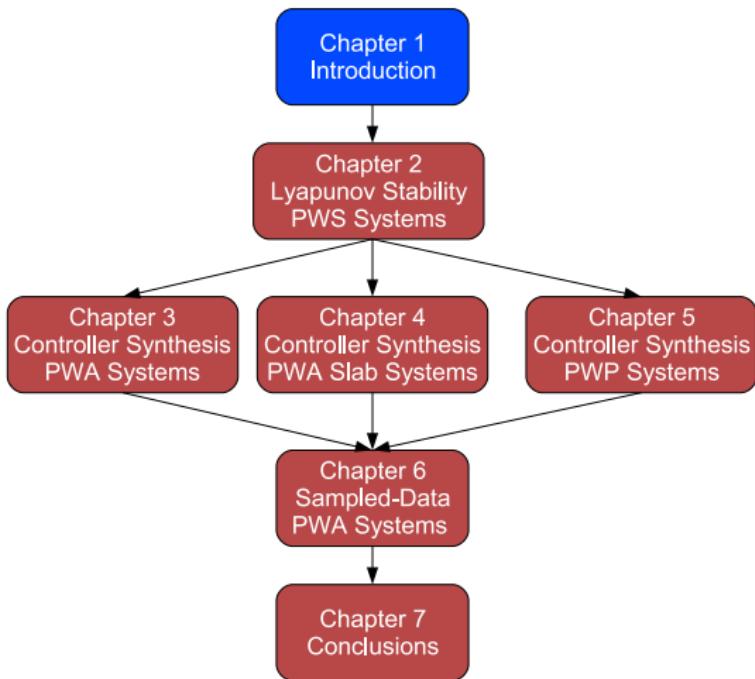
Department of Mechanical and Industrial Engineering
Concordia University

18 April 2008
Montreal, Quebec
Canada

Outline



Introduction



Practical Motivation



©Quanser

Memoryless Nonlinearities



Saturation



Dead Zone



Coulomb &
Viscous Friction

Theoretical Motivation

- Richard Murray, California Institute of Technology: “

Rank	Top Ten Research Problems in Nonlinear Control
10	Building representative experiments for evaluating controllers
9	Convincing industry to invest in new nonlinear methodologies
8	Recognizing the difference between regulation and tracking
7	Exploiting special structure to analyze and design controllers
6	Integrating good linear techniques into nonlinear methodologies
5	Recognizing the difference between performance and operability
4	Finding nonlinear normal systems for control
3	Global robust stabilization and local robust performance
2	Magnitude and rate saturation
1	Writing numerical software for implementing nonlinear theory

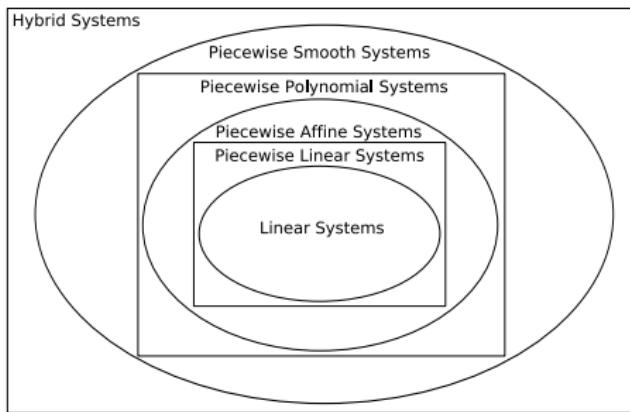
- ...This is more or less a way for me to think online, so I wouldn't take any of this too seriously.”

Special Structure

- Piecewise smooth system:

$$\dot{x} = f(x) + g(x)u$$

where $f(x)$ and $g(x)$ are piecewise continuous and bounded functions of x .



Objective

To develop a **computational tool** to design controllers for **piecewise smooth systems** using **convex optimization** techniques.

Objective

To develop a **computational tool** to design controllers for **piecewise smooth systems** using **convex optimization** techniques.

- Why convex optimization?
 - There are numerically efficient tools to solve convex optimization problems.
 - Linear Matrix Inequalities (**LMI**)
 - Sum of Squares (**SOS**) programming

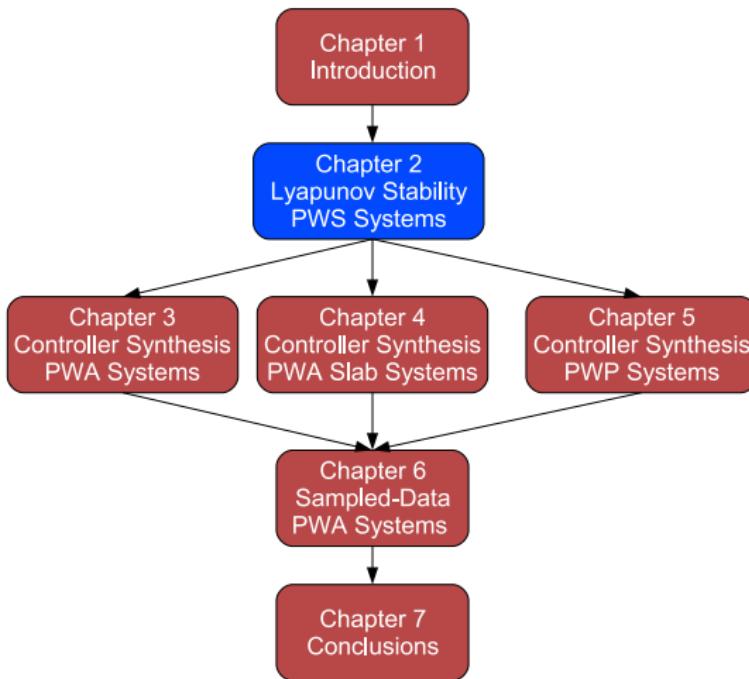
Literature

- Hassibi and Boyd (1998) - Quadratic stabilization and control of piecewise linear systems - **Limited to piecewise linear controllers** for PWA slab systems
- Johansson and Rantzer (2000) - Piecewise linear quadratic optimal control - **No guarantee for stability**
- Feng (2002) - Controller design and analysis of uncertain piecewise linear systems - **All local subsystems should be stable**
- Rodrigues and How (2003) - Observer-based control of piecewise affine systems - **Bilinear matrix inequality**
- Rodrigues and Boyd (2005) - Piecewise affine state feedback for piecewise affine slab systems using convex optimization - Stability analysis and synthesis using **parametrized linear matrix inequalities**

Major Contributions

- ① To propose a two-step controller synthesis method for a class of uncertain nonlinear systems described by PWA differential inclusions.
- ② To introduce for the first time a duality-based interpretation of PWA systems. This enables controller synthesis for PWA slab systems to be formulated as a convex optimization problem.
- ③ To propose a nonsmooth backstepping controller synthesis for PWP systems.
- ④ To propose a time-delay approach to stability analysis of sampled-data PWA systems.

Lyapunov Stability for Piecewise Smooth Systems



Lyapunov Stability for Piecewise Smooth Systems

Theorem (2.1)

For nonlinear system $\dot{x}(t) = f(x(t))$, if there exists a continuous function $V(x)$ such that

$$V(x^*) = 0$$

$$V(x) > 0 \text{ for all } x \neq x^* \text{ in } \mathcal{X}$$

$$t_1 \leq t_2 \Rightarrow V(x(t_1)) \geq V(x(t_2))$$

then $x = x^*$ is a stable equilibrium point. Moreover if there exists a continuous function $W(x)$ such that

$$W(x^*) = 0$$

$$W(x) > 0 \text{ for all } x \neq x^* \text{ in } \mathcal{X}$$

$$t_1 \leq t_2 \Rightarrow V(x(t_1)) \geq V(x(t_2)) + \int_{t_1}^{t_2} W(x(\tau)) d\tau$$

and

$$\|x\| \rightarrow \infty \Rightarrow V(x) \rightarrow \infty$$

then all trajectories in \mathcal{X} asymptotically converge to $x = x^*$.

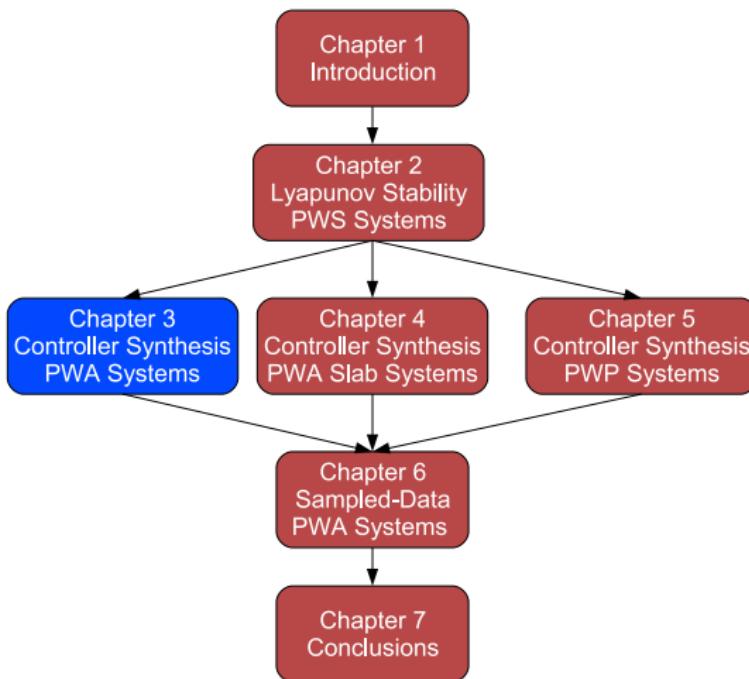
Lyapunov Stability for Piecewise Smooth Systems

Why nonsmooth analysis?

- Discontinuous vector fields
- Piecewise smooth Lyapunov functions

Lyapunov Function	Vector Field	
	Continuous	Discontinuous
Smooth		
Piecewise Smooth		

Extension of local linear controllers to global PWA controllers for uncertain nonlinear systems



Extension of linear controllers to PWA controllers

Objectives:

- **Global robust stabilization and local robust performance**
- Integrating good **linear techniques** into **nonlinear methodologies**

PWA Differential Inclusions

- Consider the following **uncertain** nonlinear system

$$\dot{x} = f(x) + g(x)u$$

Let

$$\dot{x} \in \mathbf{Conv}\{\sigma_1(x, u), \dots, \sigma_K(x, u)\}$$

where

$$\sigma_\kappa(x, u) = A_{i\kappa}x + a_{i\kappa} + B_{i\kappa}u, x \in \overline{\mathcal{R}}_i,$$

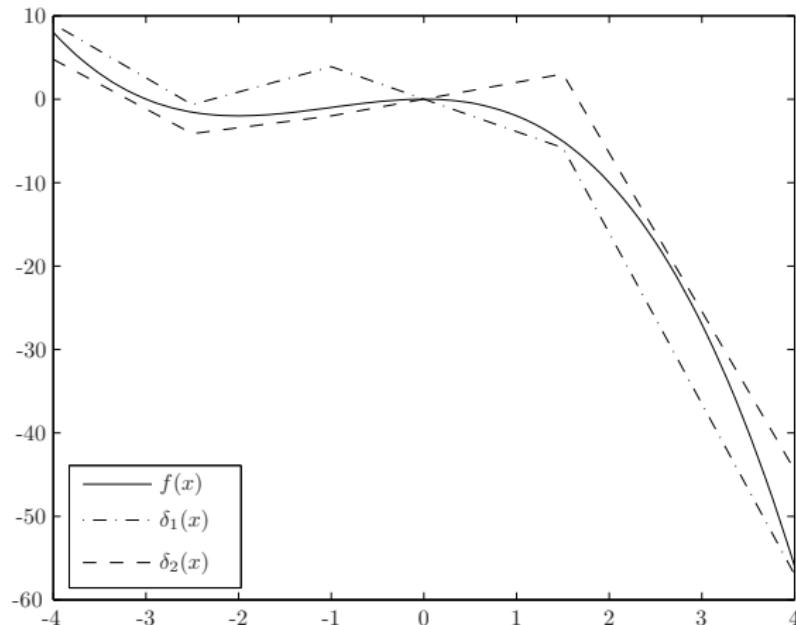
with

$$\mathcal{R}_i = \{x | E_i x + e_i \succ 0\}, \text{ for } i = 1, \dots, M$$

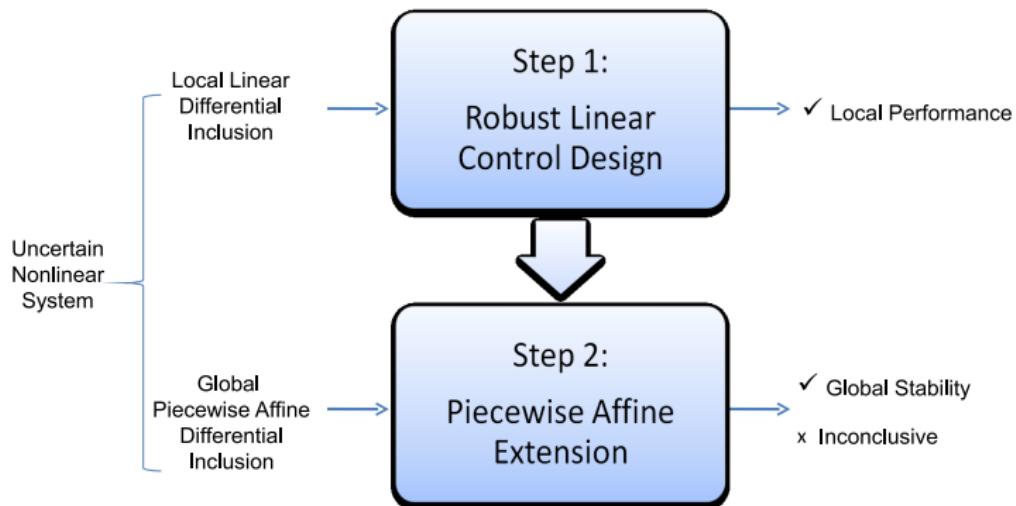
\succ represents an elementwise inequality.

PWA Differential Inclusions

PWA bounding envelope for a nonlinear function



Extension of linear controllers to PWA controllers



Extension of linear controllers to PWA controllers

Theorem (3.2)

Let there exist matrices $\bar{P}_i = \bar{P}_i^T$, \bar{K}_i , Z_i , \bar{Z}_i , Λ_{ik} and $\bar{\Lambda}_{ik}$ that verify the conditions for all $i = 1, \dots, M$, $\kappa = 1, \dots, \mathcal{K}$ and for a given decay rate $\alpha > 0$, desired equilibrium point x^* , linear controller gain \bar{K}_{i^*} and $\epsilon > 0$, then **all trajectories of the nonlinear system in \mathcal{X} asymptotically converge to $x = x^*$.**

Extension of linear controllers to PWA controllers

Theorem (3.2)

Let there exist matrices $\bar{P}_i = \bar{P}_i^T$, \bar{K}_i , Z_i , \bar{Z}_i , Λ_{ik} and $\bar{\Lambda}_{ik}$ that verify the conditions for all $i = 1, \dots, M$, $\kappa = 1, \dots, \mathcal{K}$ and for a given decay rate $\alpha > 0$, desired equilibrium point x^* , linear controller gain \bar{K}_{i^*} and $\epsilon > 0$, then **all trajectories of the nonlinear system in \mathcal{X} asymptotically converge to $x = x^*$.**

- If the conditions are feasible, the resulting PWA controller provides global robust stability and local robust performance.

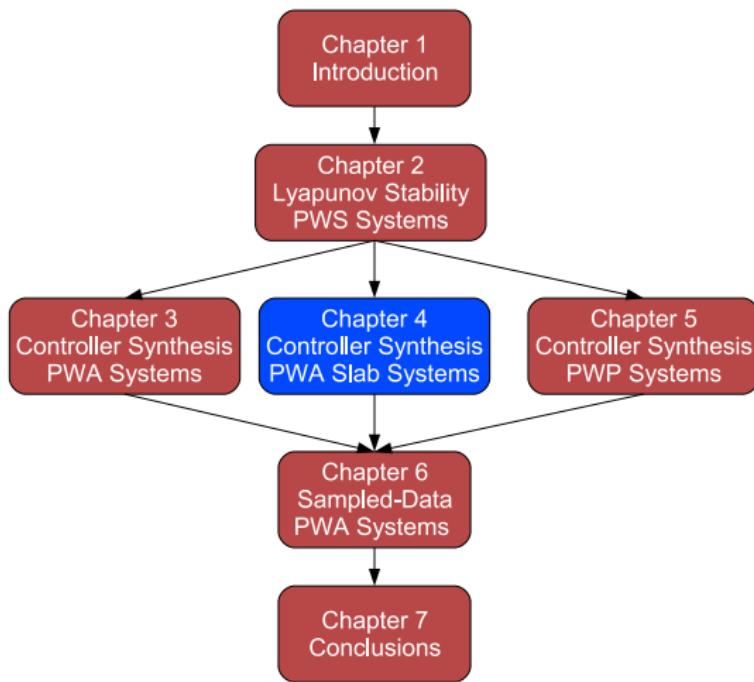
Extension of linear controllers to PWA controllers

Theorem (3.2)

Let there exist matrices $\bar{P}_i = \bar{P}_i^T$, \bar{K}_i , Z_i , \bar{Z}_i , Λ_{ik} and $\bar{\Lambda}_{ik}$ that verify the conditions for all $i = 1, \dots, M$, $\kappa = 1, \dots, \mathcal{K}$ and for a given decay rate $\alpha > 0$, desired equilibrium point x^* , linear controller gain \bar{K}_{i^*} and $\epsilon > 0$, then **all trajectories of the nonlinear system in \mathcal{X} asymptotically converge to $x = x^*$.**

- If the conditions are feasible, the resulting PWA controller provides global robust stability and local robust performance.
- The synthesis problem includes a set of Bilinear Matrix Inequalities (BMI). In general, it is not convex.

Controller synthesis for PWA slab differential inclusions: a duality-based convex optimization approach



Controller synthesis for PWA slab differential inclusions: a duality-based convex optimization approach

Objective:

- To formulate controller synthesis for **stability** and **L_2 gain performance** of piecewise affine **slab** differential inclusions as a set of **LMIs**.

L_2 Gain Analysis for PWA Slab Differential Inclusions

- PWA slab differential inclusion:

$$\dot{x} \in \text{Conv}\{A_{i\kappa}x + a_{i\kappa} + B_{w_{i\kappa}}w, \kappa = 1, 2\}, \quad (x, w) \in \mathcal{R}_i^{\mathcal{X} \times \mathcal{W}}$$

$$y \in \text{Conv}\{C_{i\kappa}x + c_{i\kappa} + D_{w_{i\kappa}}w, \kappa = 1, 2\}$$

$$\mathcal{R}_i^{\mathcal{X} \times \mathcal{W}} = \{(x, w) \mid \|L_i x + l_i + M_i w\| < 1\}$$

- Parameter set:

$$\Phi = \left\{ \begin{bmatrix} A_{i\kappa_1} & a_{i\kappa_1} & B_{w_{i\kappa_1}} \\ L_i & l_i & M_i \\ C_{i\kappa_2} & c_{i\kappa_2} & D_{w_{i\kappa_2}} \end{bmatrix} \middle| i = 1, \dots, M, \kappa_1 = 1, 2, \kappa_2 = 1, 2 \right\}$$

A duality-based convex optimization approach

Dual Parameter Set

$$\Phi^T = \left\{ \begin{bmatrix} A_{i\kappa_1}^T & L_i^T & C_{i\kappa_2}^T \\ a_{i\kappa_1}^T & l_i & c_{i\kappa_2}^T \\ B_{w_{i\kappa_1}}^T & M_i^T & D_{w_{i\kappa_2}}^T \end{bmatrix} \mid i = 1, \dots, M, \kappa_1 = 1, 2, \kappa_2 = 1, 2 \right\}$$

LMI Conditions for L₂ Gain Analysis

Parameter Set

$$P > 0,$$

$$\begin{bmatrix} A_{i\kappa_1}^T P + PA_{i\kappa_1} + C_{i\kappa_2}^T C_{i\kappa_2} & * \\ B_{w_{i\kappa_1}}^T P + D_{w_{i\kappa_2}}^T C_{i\kappa_2} & -\gamma^2 I + D_{w_{i\kappa_2}}^T D_{w_{i\kappa_2}} \end{bmatrix} < 0$$

for $i \in \mathcal{I}(0, 0)$, $\kappa_1 = 1, 2$ and $\kappa_2 = 1, 2$ and $\lambda_{i\kappa_1\kappa_2} < 0$

$$\begin{bmatrix} \begin{pmatrix} A_{i\kappa_1}^T P + PA_{i\kappa_1} \\ + C_{i\kappa_2}^T C_{i\kappa_2} \\ + \lambda_{i\kappa_1\kappa_2} L_i^T L_i \end{pmatrix} & * & * \\ a_{i\kappa_1}^T P + c_{i\kappa_2}^T C_{i\kappa_2} + \lambda_{i\kappa_1\kappa_2} I_i L_i & \lambda_{i\kappa_1\kappa_2} (\beta_i^2 - 1) + c_{i\kappa_2}^T c_{i\kappa_2} & * \\ \begin{pmatrix} B_{w_{i\kappa_1}}^T P \\ + D_{w_{i\kappa_2}}^T C_{i\kappa_2} \\ + \lambda_{i\kappa_1\kappa_2} M_i^T L_i \end{pmatrix} & D_{w_{i\kappa_2}}^T c_{i\kappa_2} + \lambda_{i\kappa_1\kappa_2} I_i M_i^T & \begin{pmatrix} -\gamma^2 I \\ + D_{w_{i\kappa_2}}^T D_{w_{i\kappa_2}} \\ + \lambda_{i\kappa_1\kappa_2} M_i^T M_i \end{pmatrix} \end{bmatrix} < 0$$

for $i \notin \mathcal{I}(0, 0)$, $\kappa_1 = 1, 2$ and $\kappa_2 = 1, 2$

LMI Conditions for L₂ Gain Analysis

Dual Parameter Set

$$Q > 0,$$

$$\begin{bmatrix} A_{i\kappa_1} Q + QA_{i\kappa_1}^T + B_{W_{i\kappa_1}} B_{W_{i\kappa_1}}^T & * \\ C_{i\kappa_2} Q + D_{W_{i\kappa_2}} B_{W_{i\kappa_1}}^T & -\gamma^2 I + D_{W_{i\kappa_2}} D_{W_{i\kappa_2}}^T \end{bmatrix} < 0$$

for $i \in \mathcal{I}(0, 0)$, $\kappa_1 = 1, 2$ and $\kappa_2 = 1, 2$ and $\mu_{i\kappa_1\kappa_2} < 0$

$$\begin{bmatrix} \begin{pmatrix} A_{i\kappa_1} Q + QA_{i\kappa_1}^T \\ +B_{W_{i\kappa_1}} B_{W_{i\kappa_1}}^T \\ +\mu_{i\kappa_1\kappa_2} a_{i\kappa_1} a_{i\kappa_1}^T \end{pmatrix} & * & * \\ L_i Q + M_i B_{W_{i\kappa_1}}^T + \mu_{i\kappa_1\kappa_2} l_i a_{i\kappa_1}^T & \mu_{i\kappa_1\kappa_2} (\ell_i^2 - 1) + M_i M_i^T & * \\ \begin{pmatrix} C_{i\kappa_2} Q + D_{W_{i\kappa_2}} B_{W_{i\kappa_1}}^T \\ +\mu_{i\kappa_1\kappa_2} c_{i\kappa_2} c_{i\kappa_2}^T \end{pmatrix} & D_{W_{i\kappa_2}} M_i^T + \mu_{i\kappa_1\kappa_2} l_i c_{i\kappa_2} & \begin{pmatrix} -\gamma^2 I \\ +D_{W_{i\kappa_2}} D_{W_{i\kappa_2}}^T \\ +\mu_{i\kappa_1\kappa_2} c_{i\kappa_2} c_{i\kappa_2}^T \end{pmatrix} \end{bmatrix} < 0$$

for $i \notin \mathcal{I}(0, 0)$, $\kappa_1 = 1, 2$ and $\kappa_2 = 1, 2$

PWA L_2 gain controller synthesis

- The goal is to limit the L_2 gain from w to y using the following PWA controller:

$$u = K_i x + \mathbf{k}_i, \quad x \in \mathcal{R}_i$$

- New variables:

$$Y_i = K_i Q$$

$$Z_i = \mu_i k_i$$

$$W_i = \mu_i k_i k_i^T$$

- Problem:** W_i is not a linear function of the unknown parameters μ_i , Y_i and Z_i .

A duality-based convex optimization approach

Proposed solutions:

- *Convex relaxation:* Since $W_i = \mu_i k_i k_i^T \leq 0$, if the synthesis inequalities are satisfied with $W_i = 0$, they are satisfied with any $W_i \leq 0$. Therefore, the synthesis problem can be made convex by omitting W_i .

A duality-based convex optimization approach

Proposed solutions:

- *Convex relaxation:* Since $W_i = \mu_i k_i k_i^T \leq 0$, if the synthesis inequalities are satisfied with $W_i = 0$, they are satisfied with any $W_i \leq 0$. Therefore, the synthesis problem can be made convex by omitting W_i .
- *Rank minimization:* Note that $W_i = \mu_i k_i k_i^T \leq 0$ is the solution of the following rank minimization problem:

$$\begin{aligned} & \min \mathbf{Rank} X_i \\ \text{s.t. } & X_i = \begin{bmatrix} W_i & Z_i \\ Z_i^T & \mu_i \end{bmatrix} \leq 0 \end{aligned}$$

Rank minimization is also not a convex problem. However, trace minimization works practically well as a heuristic solution

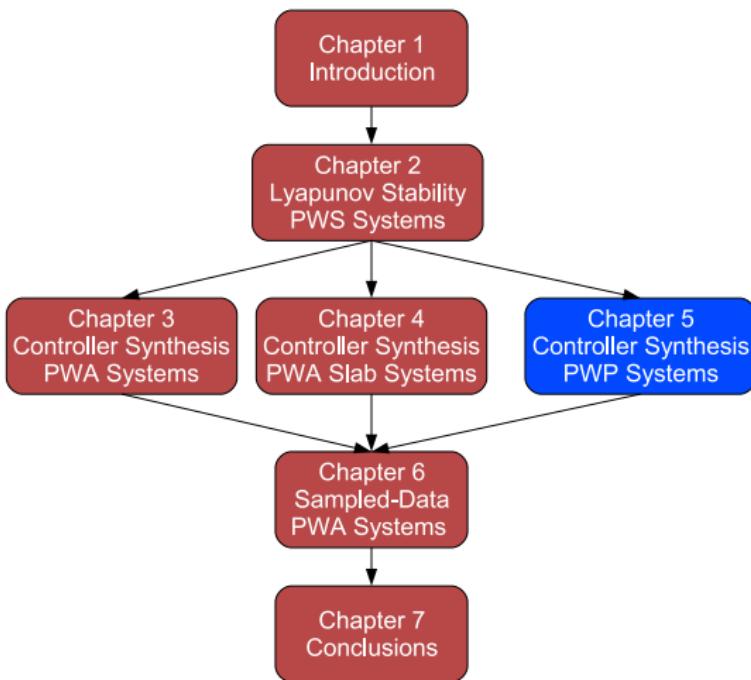
$$\min \mathbf{Trace} X_i, \text{ s.t. } X_i = \begin{bmatrix} W_i & Z_i \\ Z_i^T & \mu_i \end{bmatrix} \leq 0$$

A duality-based convex optimization approach

The following problems for PWA slab differential inclusions with PWA outputs were formulated as a set of LMI s :

- Stability analysis (Propositions 4.1 and 4.2)
- L_2 gain analysis (Propositions 4.3 and 4.4)
- Stabilization using PWA controllers (Propositions 4.5 and 4.6)
- PWA L_2 gain controller synthesis (Propositions 4.7 and 4.8)

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach



Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Objective:

- To formulate controller synthesis for a class of piecewise polynomial systems as a **Sum of Squares (SOS)** programming.

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Objective:

- To formulate controller synthesis for a class of piecewise polynomial systems as a **Sum of Squares (SOS)** programming.
- SOSTOOLS, a MATLAB toolbox that handles the general SOS programming, was developed by S. Prajna, A. Papachristodoulou and P. Parrilo.

$$p(x) = \sum_{i=1}^m f_i^2(x) \geq 0$$

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

- PWP system in *strict feedback from*

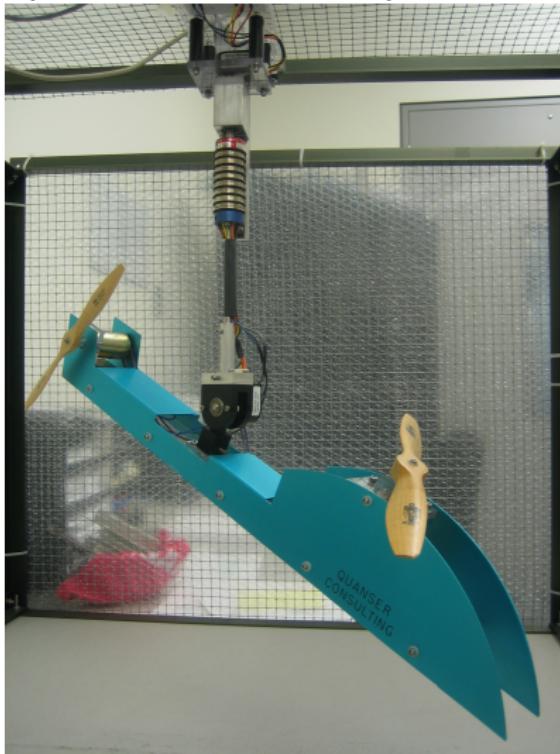
$$\left\{ \begin{array}{ll} \dot{x}_1 = f_{1i_1}(x_1) + g_{1i_1}(x_1)x_2, & \text{for } x_1 \in \mathcal{P}_{1i_1} \\ \dot{x}_2 = f_{2i_2}(x_1, x_2) + g_{2i_2}(x_1, x_2)x_3, & \text{for } \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \in \mathcal{P}_{2i_2} \\ \vdots & \\ \dot{x}_k = f_{ki_k}(x_1, x_2, \dots, x_k) + g_{ki_k}(x_1, x_2, \dots, x_k)u, & \text{for } \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_k \end{bmatrix} \in \mathcal{P}_{ki_k} \end{array} \right.$$

where

$$\mathcal{P}_{ji} = \left\{ \begin{bmatrix} x_1 \\ \vdots \\ x_j \end{bmatrix} \mid E_{ji}(x_1, \dots, x_j) \succ 0 \right\}$$

Sampled-Data PWA Systems: A Time-Delay Approach

Motivation: Toycopter, a 2 DOF helicopter model



Sampled-Data PWA Systems: A Time-Delay Approach

Example:

- Pitch model of the experimental helicopter:

$$\dot{x}_1 = x_2$$

$$\begin{aligned}\dot{x}_2 = & \frac{1}{I_{yy}}(-m_{heli}l_{cgx}g\cos(x_1) - m_{heli}l_{cgz}g\sin(x_1) - F_{kM}\operatorname{sgn}(x_2) \\ & - F_{vM}x_2 + u)\end{aligned}$$

where x_1 is the pitch angle and x_2 is the pitch rate.

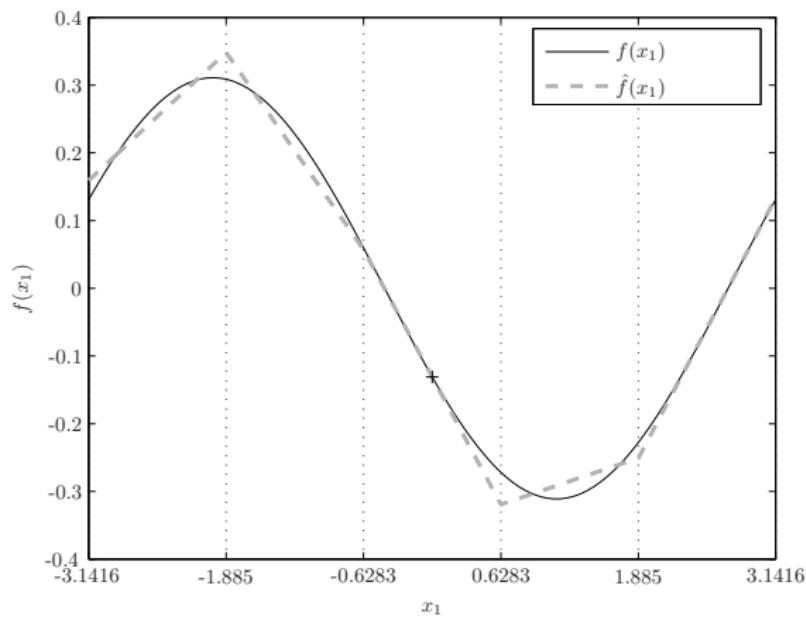
- Nonlinear part:

$$f(x_1) = -m_{heli}l_{cgx}g\cos(x_1) - m_{heli}l_{cgz}g\sin(x_1)$$

- PWA part:

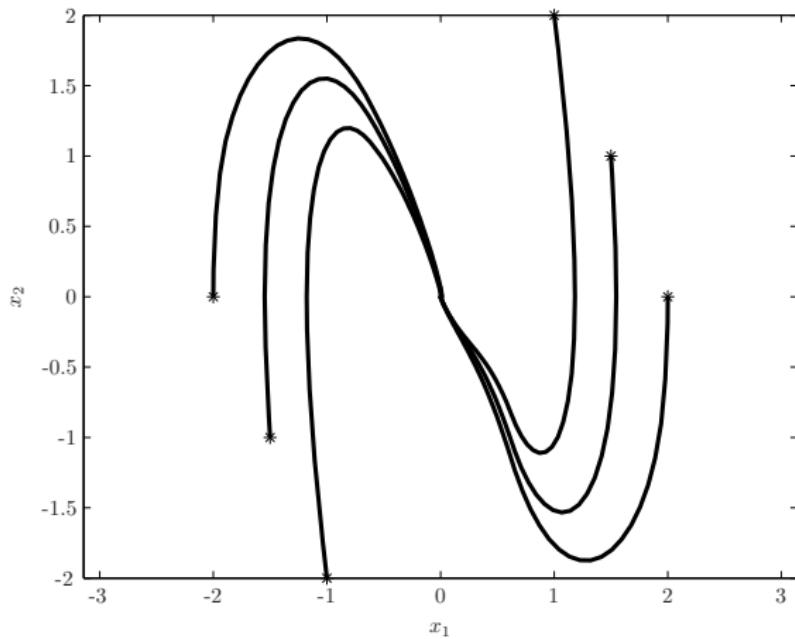
$$f(x_2) = -F_{kM}\operatorname{sgn}(x_2)$$

Sampled-Data PWA Systems: A Time-Delay Approach



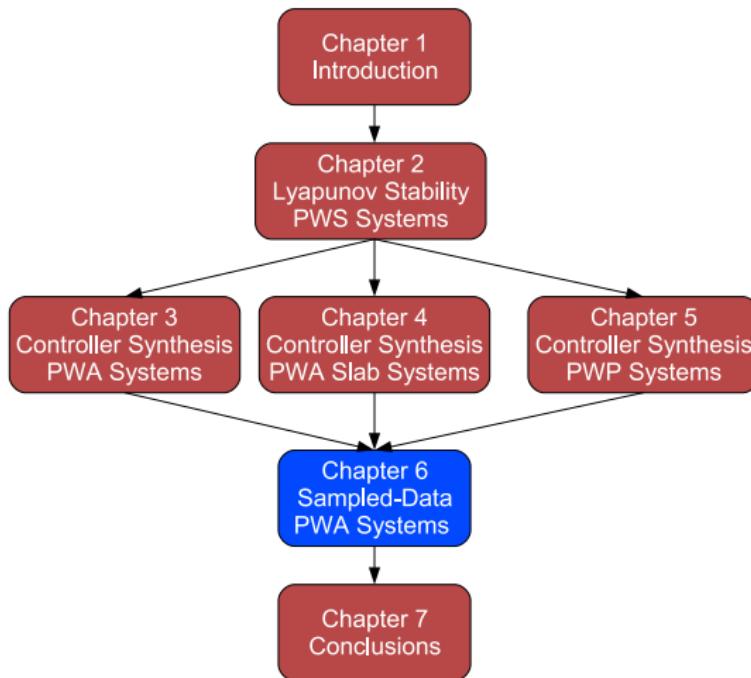
PWA approximation - Helicopter model

Sampled-Data PWA Systems: A Time-Delay Approach



Continuous time PWA controller

Sampled-Data PWA Systems: A Time-Delay Approach



Sampled-Data PWA Systems: A Time-Delay Approach

- Sampled-data PWA controller

$$u(t) = K_i x(t_k) + k_i, \quad x(t_k) \in \mathcal{R}_i$$

Sampled-Data PWA Systems: A Time-Delay Approach

- The closed-loop system can be rewritten as

$$\dot{x}(t) = A_i x(t) + a_i + B_i(K_i x(t_k) + k_i) + B_i w,$$

for $x(t) \in \mathcal{R}_i$ and $x(t_k) \in \mathcal{R}_j$ where

$$w(t) = (K_j - K_i)x(t_k) + (k_j - k_i), \quad x(t) \in \mathcal{R}_i, \quad x(t_k) \in \mathcal{R}_j$$

The input $w(t)$ is a result of the fact that $x(t)$ and $x(t_k)$ are not necessarily in the same region.

Sampled-Data PWA Systems: A Time-Delay Approach

Theorem (6.1)

For the sampled-data PWA system, assume there exist symmetric positive matrices P, R, X and matrices N_i for $i = 1, \dots, M$ such that the conditions are satisfied and let there be constants Δ_K and Δ_k such that

$$\|w\| \leq \Delta_K \|x(t_k)\| + \Delta_k$$

Then, all the trajectories of the sampled-data PWA system in \mathcal{X} converge to the following invariant set

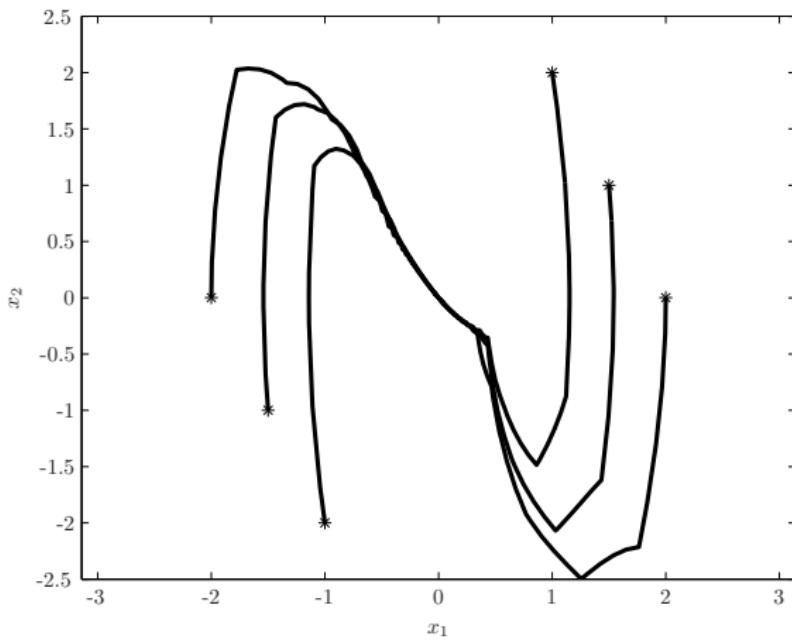
$$\Omega = \{x_s \mid V(x_s, \rho) \leq \sigma_a \mu_\theta^2 + \sigma_b\}$$

Sampled-Data PWA Systems: A Time-Delay Approach

Solving an optimization problem to maximize τ_M subject to the constraints of the main theorem and $\eta > \gamma > 1$ leads to

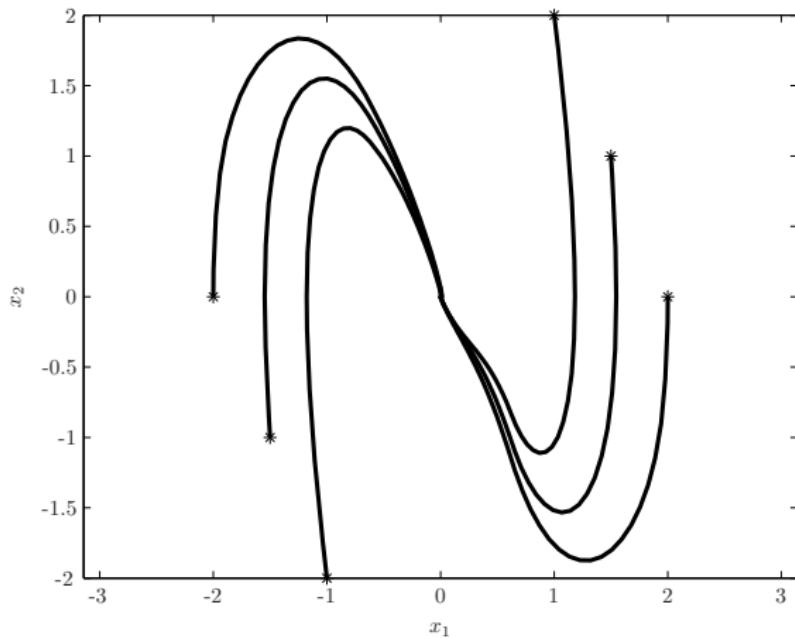
$$\tau_M^* = 0.2193$$

Sampled-Data PWA Systems: A Time-Delay Approach



Sampled data PWA controller for $T_s = 0.2193$

Sampled-Data PWA Systems: A Time-Delay Approach



Continuous time PWA controller

Breaking News

- The day before yesterday, I almost had a heart attack when...

Breaking News

- The day before yesterday, I almost had a heart attack when...
- I found a **LITTLE MINUS SIGN MISTAKE** in the proof of Theorem 6.1.

Breaking News

- The day before yesterday, I almost had a heart attack when...
- I found a **LITTLE MINUS SIGN MISTAKE** in the proof of Theorem 6.1.
- **Good news** is that fixing the mistake slightly changes the LMI conditions for PWA systems and that does not affect the numerical results much. For example for the helicopter model, τ_M was 0.2193. Now, it is 0.2175.

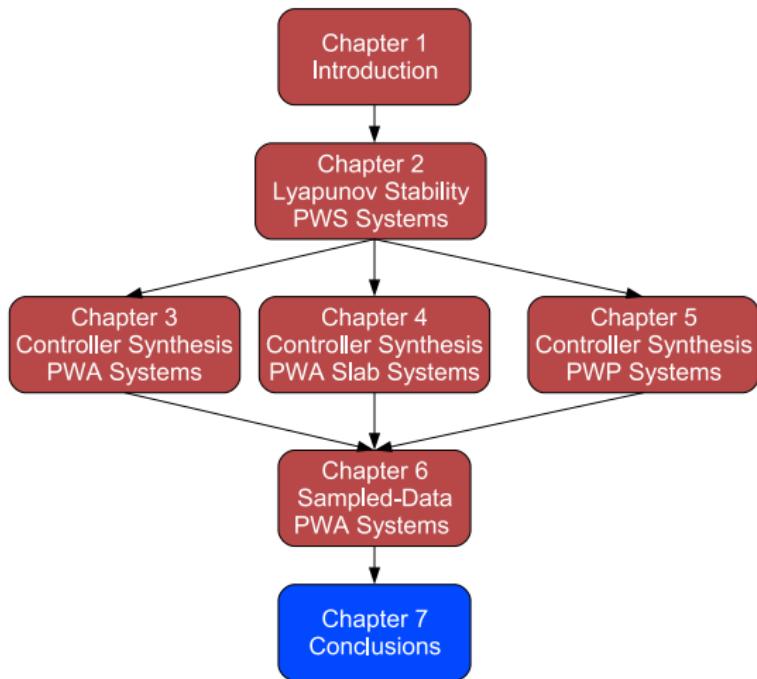
Breaking News

- The day before yesterday, I almost had a heart attack when...
- I found a **LITTLE MINUS SIGN MISTAKE** in the proof of Theorem 6.1.
- **Good news** is that fixing the mistake slightly changes the LMI conditions for PWA systems and that does not affect the numerical results much. For example for the helicopter model, τ_M was 0.2193. Now, it is 0.2175.
- **Bad news** is that for linear systems I get exactly the LMI conditions of [82]. Therefore, the result for linear systems and the linear example are no longer valid.

Breaking News

- The day before yesterday, I almost had a heart attack when...
- I found a **LITTLE MINUS SIGN MISTAKE** in the proof of Theorem 6.1.
- **Good news** is that fixing the mistake slightly changes the LMI conditions for PWA systems and that does not affect the numerical results much. For example for the helicopter model, τ_M was 0.2193. Now, it is 0.2175.
- **Bad news** is that for linear systems I get exactly the LMI conditions of [82]. Therefore, the result for linear systems and the linear example are no longer valid.
- **Good news** is that I can still keep my friendship with my undergrad classmate who is the first author of [82].

Conclusions



Summary of Major Contributions

- ① To propose a two-step controller synthesis method for a class of uncertain nonlinear systems described by PWA differential inclusions.
- ② To introduce for the first time a duality-based interpretation of PWA systems. This enables controller synthesis for PWA slab systems to be formulated as a convex optimization problem.
- ③ To propose a nonsmooth backstepping controller synthesis for PWP systems.
- ④ To propose a time-delay approach to stability analysis of sampled-data PWA systems.

Publications

- ① B. Samadi and L. Rodrigues, "Extension of local linear controllers to global piecewise affine controllers for uncertain nonlinear systems," accepted for publication in the *International Journal of Systems Science*.
- ② B. Samadi and L. Rodrigues, "Controller synthesis for piecewise affine slab differential inclusions: a duality-based convex optimization approach," under second revision for publication in *Automatica*.
- ③ B. Samadi and L. Rodrigues, "Backstepping Controller Synthesis for Piecewise Polynomial Systems: A Sum of Squares Approach," in preparation
- ④ B. Samadi and L. Rodrigues, "Sampled-Data Piecewise Affine Systems: A Time-Delay Approach," to be submitted.

Publications

- ① B. Samadi and L. Rodrigues, "Backstepping Controller Synthesis for Piecewise Polynomial Systems: A Sum of Squares Approach," submitted to *the 46th Conference on Decision and Control*, cancun, Mexico, Dec. 2008.
- ② B. Samadi and L. Rodrigues, "Sampled-Data Piecewise Affine Slab Systems: A Time-Delay Approach," in *Proc. of the American Control Conference*, Seattle, WA, Jun. 2008.
- ③ B. Samadi and L. Rodrigues, "Controller synthesis for piecewise affine slab differential inclusions: a duality-based convex optimization approach," in *Proc. of the 46th Conference on Decision and Control*, New Orleans, LA, Dec. 2007.
- ④ B. Samadi and L. Rodrigues, "Backstepping Controller Synthesis for Piecewise Affine Systems: A Sum of Squares Approach," in *Proc. of the IEEE International Conference on Systems, Man, and Cybernetics (SMC 2007)*, Montreal, Oct. 2007.
- ⑤ B. Samadi and L. Rodrigues, "Extension of a local linear controller to a stabilizing semi-global piecewise-affine controller," *7th Portuguese Conference on Automatic Control*, Lisbon, Portugal, Sep. 2006.

Questions



Open Problems

- Can general PWP/PWA controller synthesis be converted to a convex problem?
- What is the dual of a PWA system?

Special Structure

Type	$f(x)$	$g(x)$
Piecewise Smooth	Piecewise Continuous	Piecewise Continuous
Piecewise Polynomial	Piecewise Polynomial	Piecewise Polynomial
Piecewise Affine	Piecewise Affine	Piecewise Constant
Piecewise Linear	Piecewise Linear	Piecewise Constant
Linear	Linear	Constant

Convex Optimization

- “In fact the great watershed in optimization is not between **linearity** and **nonlinearity**, but **convexity** and **nonconvexity**.” (Rockafellar, SIAM review, 1993)
- The hard part is to find out if a problem can be formulated as a convex optimization problem.

Extension of linear controllers to PWA controllers

- Piecewise Quadratic Lyapunov function

$$V(x) = x^T P_i x + 2q_i^T x + r_i, \text{ for } x \in \overline{\mathcal{R}}_i$$

Extension of linear controllers to PWA controllers

- Conditions on the PWA controller:

$$\bar{K}_i = \bar{K}_{i^*}, \text{ if } x^* \in \overline{\mathcal{R}}_i$$

$$(\bar{A}_{i\kappa} + \bar{B}_{i\kappa}\bar{K}_i)\bar{x}^* = 0, \text{ if } x^* \in \overline{\mathcal{R}}_i$$

$$(\bar{A}_{i\kappa} + \bar{B}_{i\kappa}\bar{K}_i)\bar{F}_{ij} = (\bar{A}_{j\kappa} + \bar{B}_{j\kappa}\bar{K}_j)\bar{F}_{ij}, \text{ if } \overline{\mathcal{R}}_i \cap \overline{\mathcal{R}}_j \neq \emptyset$$

Extension of linear controllers to PWA controllers

- Conditions on the PWA controller:

$$\bar{K}_i = \bar{K}_{i^*}, \text{ if } x^* \in \overline{\mathcal{R}}_i$$

$$(\bar{A}_{i\kappa} + \bar{B}_{i\kappa}\bar{K}_i)\bar{x}^* = 0, \text{ if } x^* \in \overline{\mathcal{R}}_i$$

$$(\bar{A}_{i\kappa} + \bar{B}_{i\kappa}\bar{K}_i)\bar{F}_{ij} = (\bar{A}_{j\kappa} + \bar{B}_{j\kappa}\bar{K}_j)\bar{F}_{ij}, \text{ if } \overline{\mathcal{R}}_i \cap \overline{\mathcal{R}}_j \neq \emptyset$$

- Continuity of the Lyapunov function:

$$\bar{F}_{ij}^T(\bar{P}_i - \bar{P}_j)\bar{F}_{ij} = 0, \text{ if } \overline{\mathcal{R}}_i \cap \overline{\mathcal{R}}_j \neq \emptyset$$

Extension of linear controllers to PWA controllers

- Conditions on the PWA controller:

$$\bar{K}_i = \bar{K}_{i^*}, \text{ if } x^* \in \overline{\mathcal{R}}_i$$

$$(\bar{A}_{ik} + \bar{B}_{ik}\bar{K}_i)\bar{x}^* = 0, \text{ if } x^* \in \overline{\mathcal{R}}_i$$

$$(\bar{A}_{ik} + \bar{B}_{ik}\bar{K}_i)\bar{F}_{ij} = (\bar{A}_{jk} + \bar{B}_{jk}\bar{K}_j)\bar{F}_{ij}, \text{ if } \overline{\mathcal{R}}_i \cap \overline{\mathcal{R}}_j \neq \emptyset$$

- Continuity of the Lyapunov function:

$$\bar{F}_{ij}^T(\bar{P}_i - \bar{P}_j)\bar{F}_{ij} = 0, \text{ if } \overline{\mathcal{R}}_i \cap \overline{\mathcal{R}}_j \neq \emptyset$$

- Positive definiteness of the Lyapunov function:

$$\bar{P}_i\bar{x}^* = 0, \text{ if } x^* \in \overline{\mathcal{R}}_i$$

$$P_i > \epsilon I, \text{ if } x^* \in \overline{\mathcal{R}}_i, E_i x^* + e_i \neq 0$$

$$\begin{cases} Z_i \in \mathbb{R}^{n \times n}, Z_i \succeq 0 \\ P_i - E_i^T Z_i E_i > \epsilon I \end{cases}, \text{ if } x^* \in \overline{\mathcal{R}}_i, E_i x^* + e_i = 0$$

$$\begin{cases} \bar{Z}_i \in \mathbb{R}^{(n+1) \times (n+1)}, \bar{Z}_i \succeq 0 \\ \bar{P}_i - \bar{E}_i^T \bar{Z}_i \bar{E}_i > \epsilon \bar{I} \end{cases}, \text{ if } x^* \notin \overline{\mathcal{R}}_i$$

Extension of linear controllers to PWA controllers

- Monotonicity of the Lyapunov function:

for i such that $x^* \in \overline{\mathcal{R}}_i$, $E_i x^* + e_i \neq 0$,

$$P_i(A_{ik} + B_{ik}K_i) + (A_{ik} + B_{ik}K_i)^T P_i < -\alpha P_i$$

for i such that $x^* \in \overline{\mathcal{R}}_i$, $E_i x^* + e_i = 0$,

$$\begin{cases} \Lambda_{ik} \in \mathbb{R}^{n \times n}, \quad \Lambda_{ik} \succeq 0 \\ P_i(A_{ik} + B_{ik}K_i) + (A_{ik} + B_{ik}K_i)^T P_i + E_i^T \Lambda_{ik} E_i < -\alpha P_i \end{cases}$$

for i such that $x^* \notin \overline{\mathcal{R}}_i$,

$$\begin{cases} \bar{\Lambda}_{ik} \in \mathbb{R}^{(n+1) \times (n+1)}, \quad \bar{\Lambda}_{ik} \succeq 0 \\ \bar{P}_i(\bar{A}_{ik} + \bar{B}_{ik}\bar{K}_i) + (\bar{A}_{ik} + \bar{B}_{ik}\bar{K}_i)^T \bar{P}_i + \bar{E}_i^T \bar{\Lambda}_{ik} \bar{E}_i < -\alpha \bar{P}_i \end{cases}$$

Extension of linear controllers to PWA controllers

Consider the following second order system

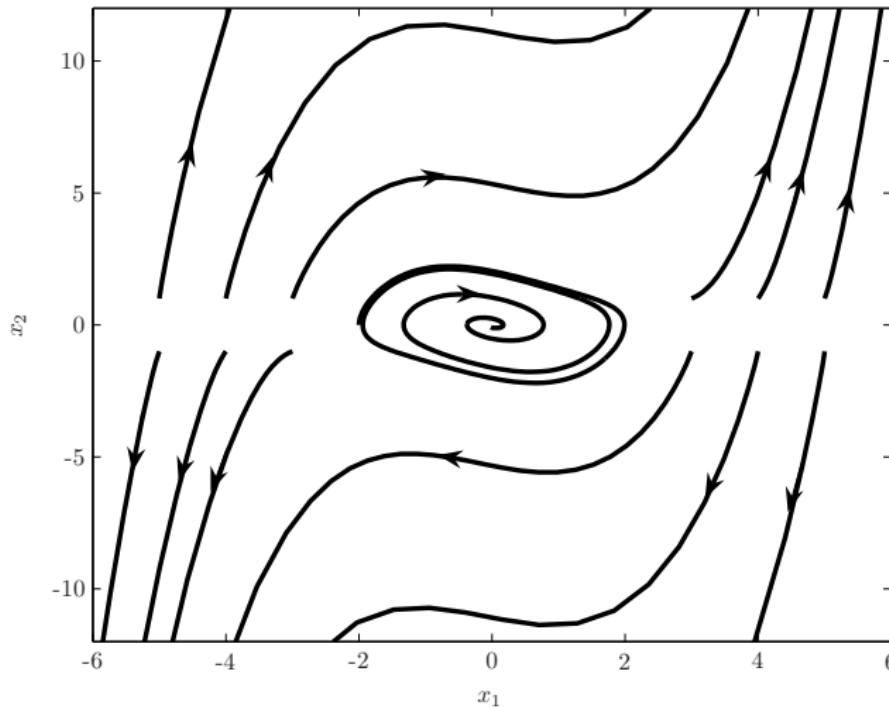
$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -x_1 + 0.5x_2 - 0.5x_1^2 x_2 + u$$

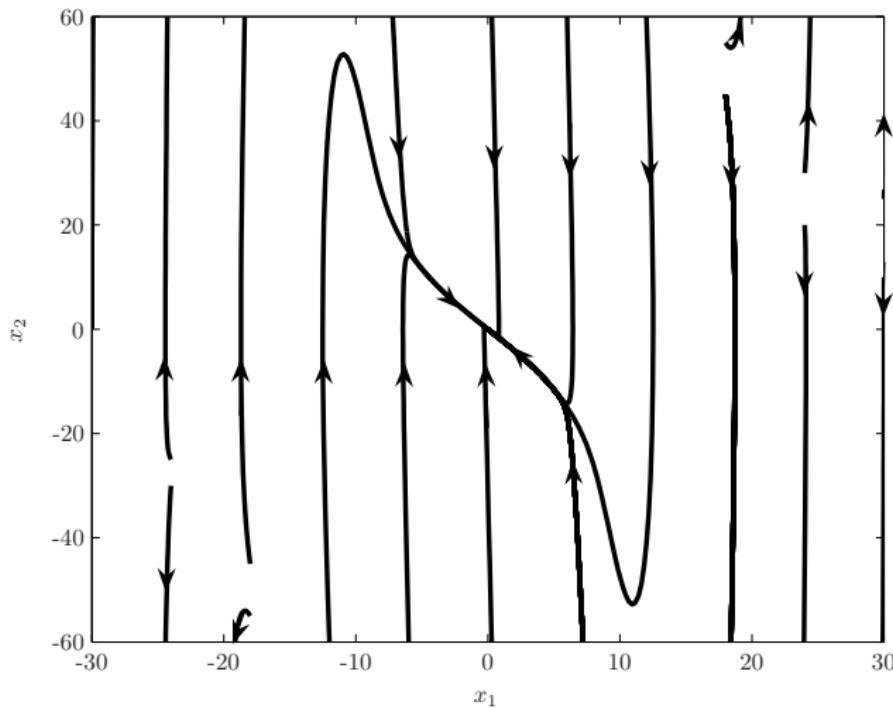
with the following domain:

$$\mathcal{X} = \left\{ \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \mid -30 < x_1 < 30, -60 < x_2 < 60 \right\}$$

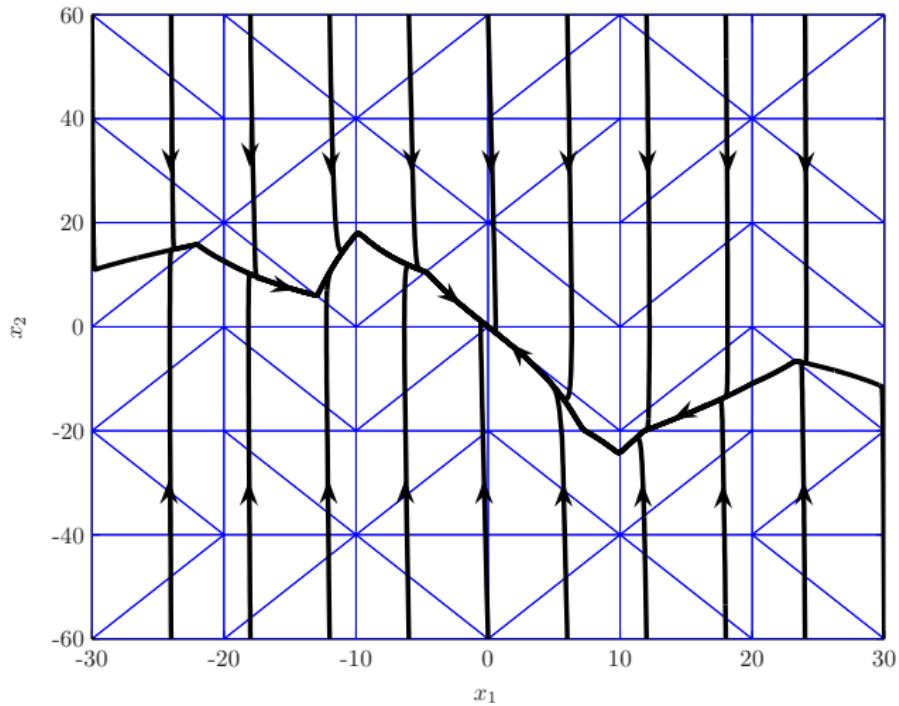
Extension of linear controllers to PWA controllers



Extension of linear controllers to PWA controllers



Extension of linear controllers to PWA controllers



Extension of linear controllers to PWA controllers

- *The main limitation:* The controller synthesis is formulated as a set of Bilinear Matrix Inequalities (BMI).
- *Linear systems analogy:* Consider the linear system $\dot{x} = Ax + Bu$ with a candidate Lyapunov function $V(x) = x^T Px$ and a state feedback controller $u = Kx$. Sufficient conditions for the stability of this system are:
 - Positivity: $V(x) = x^T Px > 0$ for $x \neq 0$
 - Monotonicity: $\dot{V}(x) = \dot{x}^T Px + x^T P\dot{x} < 0$

$$P > 0$$

$$(A^T + K^T B^T)P + P(A + BK) < 0$$

- BMIs are nonconvex problems in general.

Extension of linear controllers to PWA controllers

Contributions:

- To propose a two-step controller synthesis method for a class of uncertain nonlinear systems described by PWA differential inclusions.
 - The proposed method has two objectives: global stability and local performance. It thus enables to use well known techniques in linear control design for local stability and performance while delivering a global PWA controller that is guaranteed to stabilize the nonlinear system.
 - Differential inclusions are considered. Therefore the controller is robust in the sense that it can stabilize any piecewise smooth nonlinear system bounded by the differential inclusion.
 - Stability is guaranteed even if sliding modes exist.

A duality-based convex optimization approach

Linear systems analogy:

- Consider the dual system

$$\dot{z} = (A + BK)^T z$$

with a candidate Lyapunov function

$$V(x) = x^T Q x$$

- Sufficient conditions for the stability of this system are:
 - Positive definiteness: $Q > 0$
 - Monotonicity: $(A + BK)Q + Q(A^T + K^T B^T) < 0$

Define $Y = KQ$.

- Positive definiteness: $Q > 0$
- Monotonicity: $AQ + BY + QA^T + Y^T B^T < 0$

Then $K = YQ^{-1}$.

A duality-based convex optimization approach

- PWA slab system:

$$\dot{x} = A_i x + a_i, \quad x \in \mathcal{R}_i$$

$$\mathcal{R}_i = \{x \mid \|L_i x + l_i\| < 1\}$$

- Literature
 - Hassibi and Boyd (1998)
 - Rodrigues and Boyd (2005)

A duality-based convex optimization approach

- Sufficient conditions for stability

$$P > 0,$$

$$A_i^T P + PA_i + \alpha P < 0, \text{ for } 0 \in \overline{\mathcal{R}}_i,$$

$$\left\{ \begin{array}{l} \lambda_i < 0, \\ \begin{bmatrix} A_i^T P + PA_i + \alpha P + \lambda_i L_i^T L_i & Pa_i + \lambda_i l_i L_i^T \\ a_i^T P + \lambda_i l_i L_i & \lambda_i (\beta_i^2 - 1) \end{bmatrix} < 0, \text{ for } 0 \notin \overline{\mathcal{R}}_i \end{array} \right.$$

- Hassibi and Boyd (1998)

A duality-based convex optimization approach

- Sufficient conditions for stability

$$Q > 0,$$

$$A_i Q + Q A_i^T + \alpha Q < 0, \quad \text{for } 0 \in \overline{\mathcal{R}}_i,$$

$$\left\{ \begin{array}{ll} \mu_i < 0, \\ \begin{bmatrix} A_i Q + Q A_i^T + \alpha Q + \mu_i a_i a_i^T & Q L_i^T + \mu_i l_i a_i \\ L_i Q + \mu_i l_i a_i^T & \mu_i (\beta_i^2 - 1) \end{bmatrix} < 0, \quad \text{for } 0 \notin \overline{\mathcal{R}}_i \end{array} \right.$$

- Hassibi and Boyd (1998)

A duality-based convex optimization approach

- Parameter set:

$$\Omega = \left\{ \begin{bmatrix} A_i & a_i \\ L_i & l_i \end{bmatrix} \mid i = 1, \dots, M \right\}$$

- Dual parameter set

$$\Omega^T = \left\{ \begin{bmatrix} A_i^T & L_i^T \\ a_i^T & l_i \end{bmatrix} \mid i = 1, \dots, M \right\}$$

A duality-based convex optimization approach

- PWA slab system:

$$\dot{x} = A_i x + a_i + B_{w_i} w,$$

$$x \in \mathcal{R}_i = \{x \mid \|L_i x + l_i\| < 1\},$$

$$y = C_i x + D_{w_i} w,$$

- Parameter set:

$$\Phi = \left\{ \begin{bmatrix} A_i & a_i & B_{w_i} \\ L_i & l_i & 0 \\ C_i & 0 & D_{w_i} \end{bmatrix} \middle| i = 1, \dots, M \right\}$$

- Hassibi and Boyd (1998)

A duality-based convex optimization approach

Summary:

- Introducing PWA slab differential inclusions
- Introducing the dual parameter set
- Extending the L_2 gain analysis and synthesis to PWA slab differential inclusions with PWA outputs
- Extending the definition of the regions of a PWA slab differential inclusion
- Proposing two methods to formulate the PWA controller synthesis for PWA slab differential inclusions as a convex problem

Sum of Squares Programming

A sum of squares program is a convex optimization program of the following form:

$$\text{Minimize} \quad \sum_{j=1}^J w_j \alpha_j$$

$$\text{subject to} \quad f_{i,0} + \sum_{j=1}^J \alpha_j f_{i,j}(x) \text{ is SOS, for } i = 1, \dots, I$$

where the α_j 's are the scalar real decision variables, the w_j 's are some given real numbers, and the $f_{i,j}$ are some given multivariate polynomials.

Sum of Squares Programming

A sum of squares program is a convex optimization program of the following form:

$$\text{Minimize} \quad \sum_{j=1}^J w_j \alpha_j$$

$$\text{subject to} \quad f_{i,0} + \sum_{j=1}^J \alpha_j f_{i,j}(x) \text{ is SOS, for } i = 1, \dots, I$$

where the α_j 's are the scalar real decision variables, the w_j 's are some given real numbers, and the $f_{i,j}$ are some given multivariate polynomials.

- SOSTOOLS, a MATLAB toolbox that handles the general SOS programming, was developed by S. Prajna, A. Papachristodoulou and P. Parrilo.

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

- Consider the following PWP system:

$$\begin{aligned}\dot{x} &= f_i(x) + g_i(x)z, \quad x \in \mathcal{P}_i \\ \dot{z} &= u\end{aligned}$$

where

$$\mathcal{P}_i = \{x | E_i(x) \succ 0\}$$

where $E_i(x) \in \mathbb{R}^{p_i}$ is a vector polynomial function of x and \succ represents an elementwise inequality.

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Backstepping as a Lyapunov function construction method:

- Consider $\dot{x} = f_i(x) + g_i(x)z$, $x \in \mathcal{P}_i$

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Backstepping as a Lyapunov function construction method:

- Consider $\dot{x} = f_i(x) + g_i(x)z$, $x \in \mathcal{P}_i$
- Assume that there exists a polynomial control $z = \gamma(x)$ and $V(x)$ is an SOS Lyapunov function for the closed loop system verifying

$$\begin{cases} V(x) - \lambda(x) \text{ is SOS} \\ -\nabla V(x)^T (f_i(x) + g_i(x)\gamma(x)) - \Gamma_i(x)^T E_i(x) - \alpha V(x) \text{ is SOS} \end{cases}$$

for $i = 1, \dots, M$ and any $\alpha > 0$, where $\lambda(x)$ is a positive definite SOS polynomial, $\Gamma_i(x)$ is an SOS vector function

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Backstepping as a Lyapunov function construction method:

- Consider $\dot{x} = f_i(x) + g_i(x)z$, $x \in \mathcal{P}_i$
- Assume that there exists a polynomial control $z = \gamma(x)$ and $V(x)$ is an SOS Lyapunov function for the closed loop system verifying

$$\begin{cases} V(x) - \lambda(x) \text{ is SOS} \\ -\nabla V(x)^T (f_i(x) + g_i(x)\gamma(x)) - \Gamma_i(x)^T E_i(x) - \alpha V(x) \text{ is SOS} \end{cases}$$

for $i = 1, \dots, M$ and any $\alpha > 0$, where $\lambda(x)$ is a positive definite SOS polynomial, $\Gamma_i(x)$ is an SOS vector function

- Consider now the following candidate Lyapunov function

$$V_\gamma(x, z) = V(x) + \frac{1}{2}(z - \gamma(x))^T(z - \gamma(x))$$

Note that $V_\gamma(x, z)$ is a positive definite function.

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

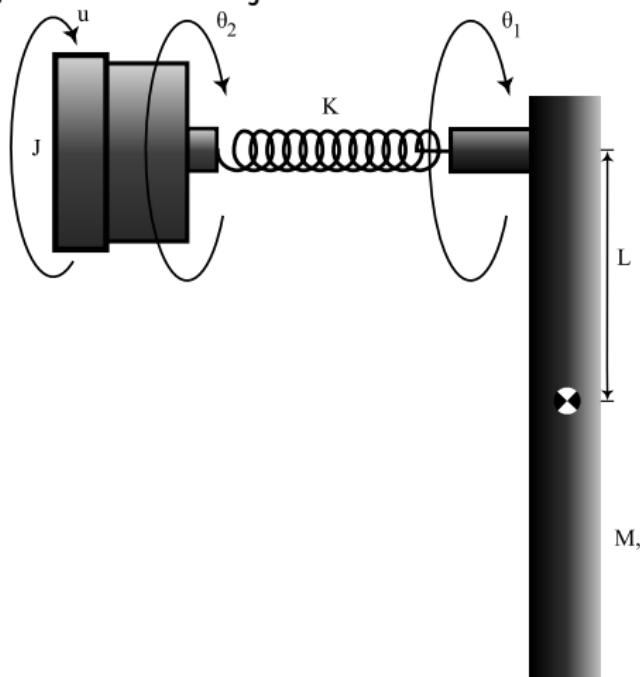
- *Controller synthesis:* The synthesis problem can be formulated as the following SOS program.

$$\begin{aligned} \text{Find } \quad & u = \gamma_2(x, z) \text{ and } \Gamma(x, z) \\ \text{s.t. } \quad & -\nabla_x V_\gamma(x, z)^\top (f_i(x) + g_i(x)z) \\ & -\nabla_z V_\gamma(x, z)^\top u \\ & -\Gamma(x, z)^\top E_i(x, z) - \alpha V_\gamma(x, z) \text{ is SOS,} \\ & \Gamma(x, z) \text{ is SOS} \\ & \gamma_2(0, 0) = 0 \end{aligned}$$

where $i_2 = 1, \dots, M_2$ and $\gamma_2(x_1, x_2)$ is a polynomial function of x_1 and x_2 .

Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Example: Single link flexible joint robot:



Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Example: Single link flexible joint robot:

$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -\frac{MgL}{I} \sin(x_1) - \frac{K}{I}(x_1 - x_3)$$

$$\dot{x}_3 = x_4$$

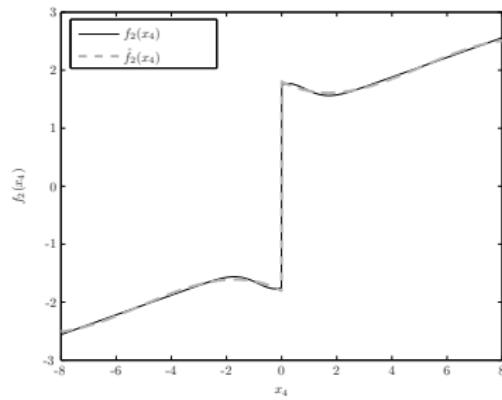
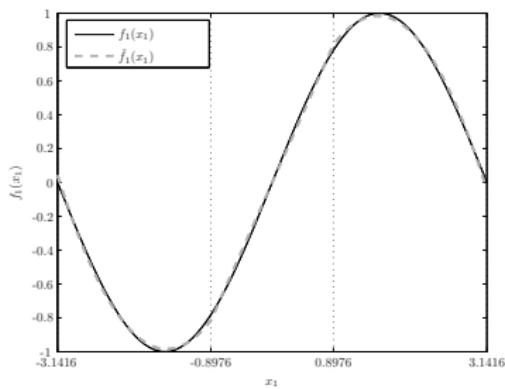
$$\dot{x}_4 = -\frac{f_2(x_4)}{J} + \frac{K}{J}(x_1 - x_3) + \frac{1}{J}u$$

where $x_1 = \theta_1$, $x_2 = \dot{\theta}_1$, $x_3 = \theta_2$ and $x_4 = \dot{\theta}_2$. u is the motor torque and $f_2(x_4)$ denotes the motor friction which is described by

$$f_2(x_4) = b_m x_4 + \text{sgn}(x_4) \left(F_{cm} + (F_{sm} - F_{cm}) \exp\left(-\frac{x_4^2}{c_m^2}\right) \right)$$

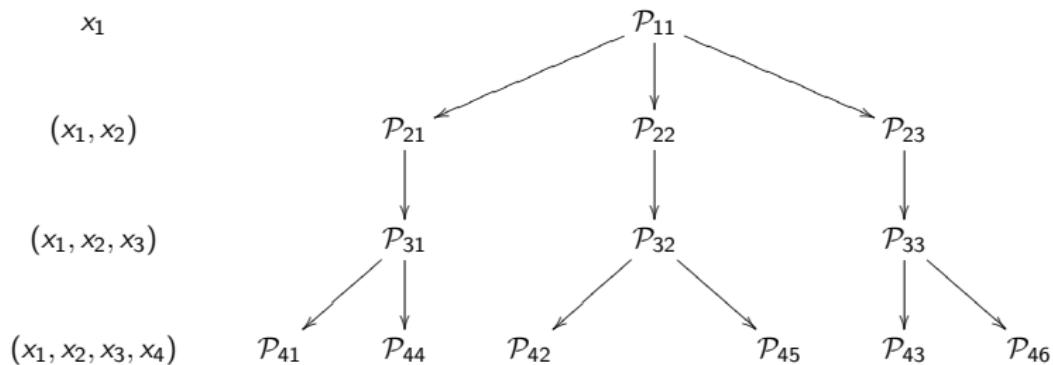
Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

PWP approximation:



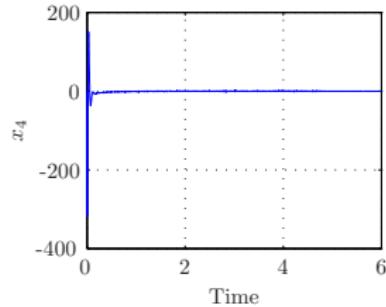
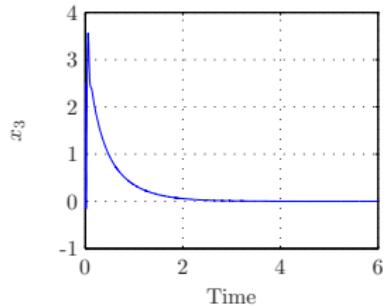
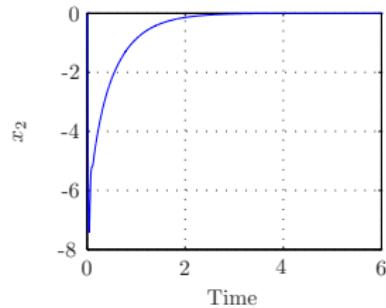
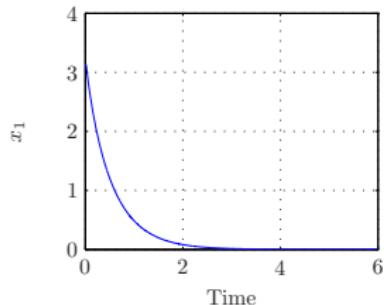
Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Typical structure of the regions of a PWP system in strict feedback form



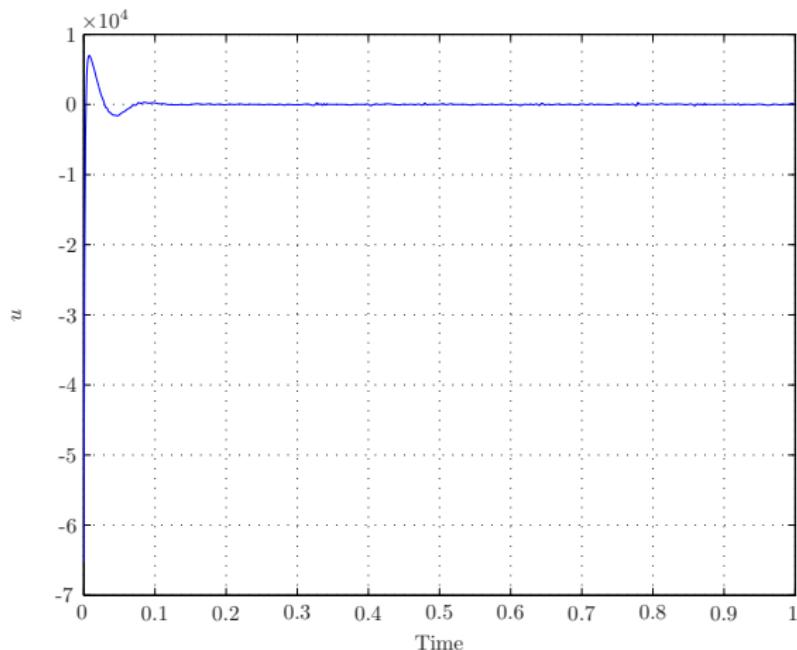
Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

State variables of the nonlinear model - PWA controller:



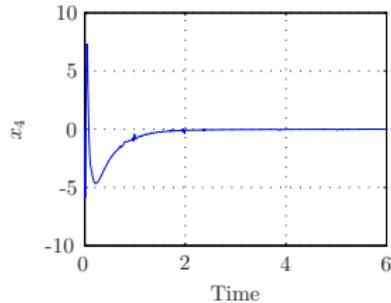
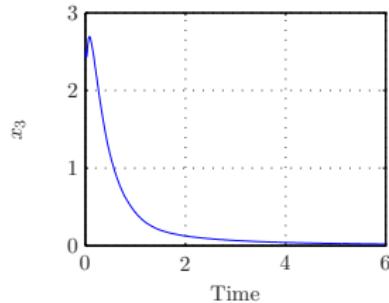
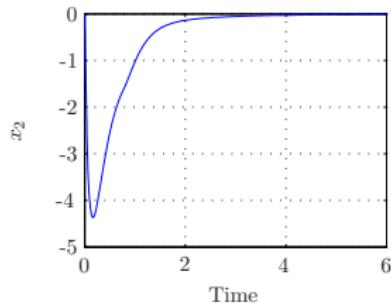
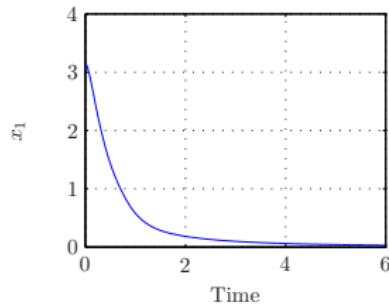
Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Control input - PWA controller:



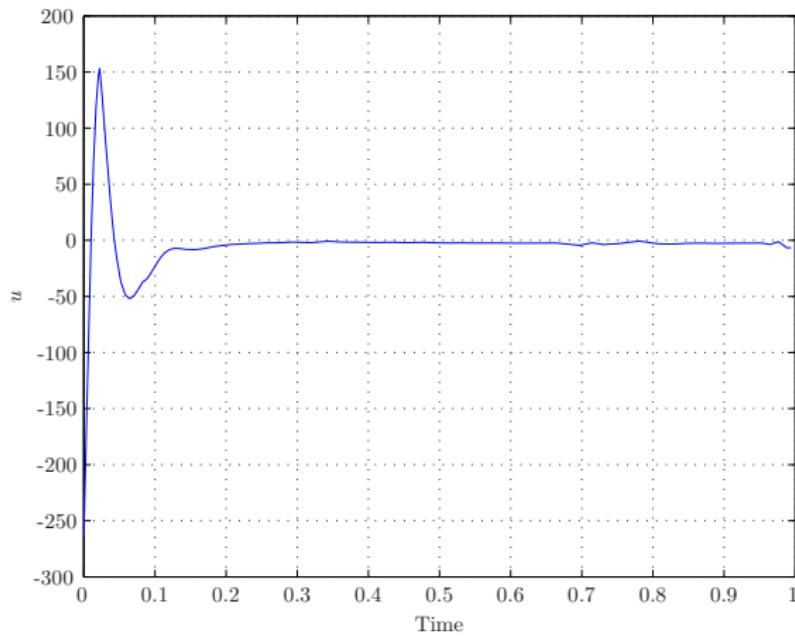
Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

State variables of the nonlinear model - PWP controller:



Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Control input - PWP controller:



Backstepping Controller Synthesis for PWP Systems: A Sum of Squares Approach

Summary of the contributions:

- Introducing PWP systems in *strict feedback form*
- Formulating backstepping controller synthesis for PWP systems as a convex optimization problem
 - Polynomial Lyapunov functions for PWP systems with discontinuous vector fields
 - PWP Lyapunov functions for PWP systems with continuous vector fields
 - Numerical tools such as SOSTOOLS and Yalmip/SeDuMi

Sampled-Data PWA Systems: A Time-Delay Approach

- PWA system

$$\dot{x} = A_i x + a_i + B_i u, \quad \text{for } x \in \mathcal{R}_i$$

with the region \mathcal{R}_i defined as

$$\mathcal{R}_i = \{x | E_i x + e_i \succ 0\},$$

Sampled-Data PWA Systems: A Time-Delay Approach

- PWA system

$$\dot{x} = A_i x + a_i + B_i u, \quad \text{for } x \in \mathcal{R}_i$$

with the region \mathcal{R}_i defined as

$$\mathcal{R}_i = \{x | E_i x + e_i \succ 0\},$$

- Continuous-time PWA controller

$$u(t) = K_i x(t) + k_i, \quad x(t) \in \mathcal{R}_i$$

Sampled-Data PWA Systems: A Time-Delay Approach

- Lyapunov-Krasovskii functional:

$$V(x_s, \rho) := V_1(x) + V_2(x_s, \rho) + V_3(x_s, \rho)$$

where

$$x_s(t) := \begin{bmatrix} x(t) \\ x(t_k) \end{bmatrix}, \quad t_k \leq t < t_{k+1}$$

$$V_1(x) := x^T P x$$

$$V_2(x_s, \rho) := \int_{-\tau_M}^0 \int_{t+r}^t \dot{x}^T(s) R \dot{x}(s) ds dr$$

$$V_3(x_s, \rho) := (\tau_M - \rho)(x(t) - x(t_k))^T X (x(t) - x(t_k))$$

and P , R and X are positive definite matrices.

Sampled-Data PWA Systems: A Time-Delay Approach

- for all i such that $0 \in \overline{\mathcal{R}}_i$,

$$\begin{bmatrix} \Psi_i + \tau_M M_{1i} & \\ B_i^T [P \ 0] + \tau_M B_i^T X [I \ -I] & \begin{bmatrix} P \\ 0 \end{bmatrix} B_i + \tau_M \begin{bmatrix} I \\ -I \end{bmatrix} X B_i \\ & -\gamma I \end{bmatrix} < 0$$

$$\begin{bmatrix} \Psi_i + \tau_M M_{2i} & \\ \tau_M B_i^T R [A_i \ B_i K_i] + B_i^T [P \ 0] & \begin{bmatrix} A_i^T \\ K_i^T B_i^T \end{bmatrix} R B_i + \begin{bmatrix} P \\ 0 \end{bmatrix} B_i \quad \tau_M N_i \\ \tau_M N_i^T & \begin{bmatrix} \tau_M B_i^T R B_i - \gamma I \\ 0 \\ -\frac{\tau_M}{2} R \end{bmatrix} \end{bmatrix}$$

Sampled-Data PWA Systems: A Time-Delay Approach

- for all i such that $0 \notin \bar{\mathcal{R}}_i$, $\bar{\Lambda}_i \succ 0$,

$$\begin{bmatrix} \bar{\Psi}_i + \tau_M \bar{M}_{1i} & \begin{bmatrix} P \\ 0 \\ 0 \end{bmatrix} B_i + \tau_M \begin{bmatrix} I \\ -I \\ 0 \end{bmatrix} X B_i \\ B_i^T \begin{bmatrix} P & 0 & 0 \end{bmatrix} + \tau_M B_i^T X \begin{bmatrix} I & -I & 0 \end{bmatrix} & -\gamma I \end{bmatrix} < 0$$

$$\begin{bmatrix} \bar{\Psi}_i + \tau_M \bar{M}_{2i} & \tau_M \begin{bmatrix} A_i^T \\ K_i^T B_i^T + a_i^T \\ P \\ 0 \\ 0 \end{bmatrix} R B_i & \tau_M \begin{bmatrix} N_i \\ 0 \end{bmatrix} \\ \hline \tau_M B_i^T R \begin{bmatrix} A_i & B_i K_i & B_i k_i + a_i \end{bmatrix} + B_i^T \begin{bmatrix} P & 0 & 0 \end{bmatrix} & \tau_M B_i^T R B_i - \gamma I & 0 \\ \hline \tau_M \begin{bmatrix} N_i^T & 0 \end{bmatrix} & 0 & -\frac{\tau_M}{2} R \end{bmatrix} < 0$$

Sampled-Data PWA Systems: A Time-Delay Approach

Summary of the contributions

- Formulating stability analysis of sampled-data PWA systems as a convex optimization problem

Active Suspension System

Example: Active suspension with a nonlinear damper

