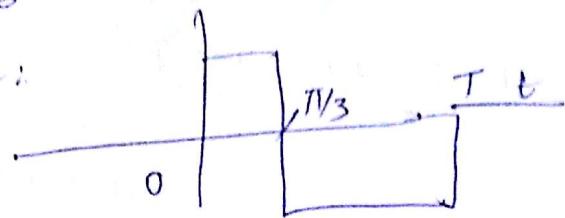


Let for $t = 0$

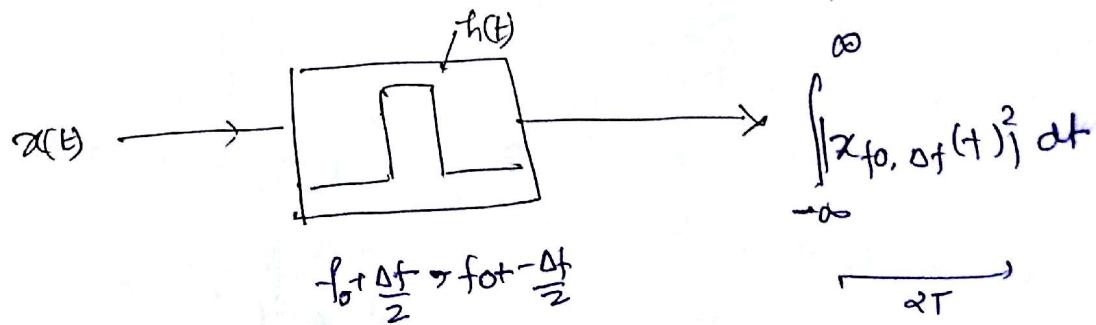
$$E[x(t)x^*(t-\tau)] \Rightarrow \sum_{-\infty}^{\infty} P |e^{-k\tau}|^2$$

Let say pulse ph:



Date 4-10-2016

$$\int_{-T}^{T} |x(t)|^2 dt = \int_{-\infty}^{\infty} |X(f)|^2 df$$



If $x(t)$ which is from $[-T, T]$ and passed through $h(t)$, i.e band pass

At the output $\int_{-\infty}^{\infty} |x_{f_0, \Delta f}(t)|^2 dt$, gives the gives energy with frequency component in $f_0 - \frac{\Delta f}{2}, f_0, f_0 + \frac{\Delta f}{2}$

And if we denote the $\int_{-\infty}^{\infty} |x_{f_0, \Delta f}(t)|^2 dt$ by a

then we will get average energy

$$\text{from parsel theorem } \int_{-\infty}^{\infty} x_{f_0, \Delta f}(t)^2 dt \Rightarrow \int_{f_0, \Delta f}^a (X(f))^2 df$$

$$\text{Now } x(t) \rightarrow \begin{array}{c} h(t) \\ \boxed{\text{ }} \end{array} \rightarrow \int_{-\infty}^{\infty} |x_{f_0, \Delta f}(t)|^2 dt = \int_{f_0 - \Delta f/2}^{f_0 + \Delta f/2} |X(f)|^2 df$$

These results are for only deterministic signals

so

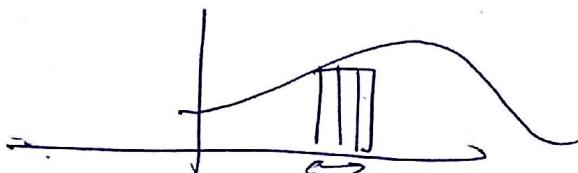
$$P\left(f_0 - \frac{\Delta f}{2}, f_0 + \frac{\Delta f}{2}\right) \Rightarrow \int_{f_0 - \frac{\Delta f}{2}}^{f_0 + \frac{\Delta f}{2}} |X(f)|^2 df / 2T$$

To find PSD divide it by Δf :

$$\Rightarrow \int_{f_0 - \frac{\Delta f}{2}}^{f_0 + \frac{\Delta f}{2}} \frac{|X(f)|^2 df}{2T \Delta f}$$

As deterministic signal $2T$ is fixed

$$\text{if } G(f) \triangleq |H(f)|$$



if we shrink the interval
 $f_0 - \frac{\Delta f}{2} \quad f_0 + \frac{\Delta f}{2}$
then $|H(f)|^2 \Rightarrow |H(f_0)|^2$
using same ~~same~~ point

or by Taylor Expansion

$$G(f) \approx G(f_0) + \Delta f G'(f)$$

So using result:

$$\int_{f_0 - \Delta f_2}^{f_0 + \Delta f_2} \frac{|X(f)|^2 df}{2T \Delta f} = \frac{|H(f_0)|^2 (f_0 + \Delta f_2 - f_0 + \Delta f_2)}{2T \Delta f} \approx \frac{|H(f_0)|^2}{2T}$$

~~Doubt~~
As T becomes larger $\frac{|H(f_0)|^2}{2T}$ will converge to a random value

Each time T' button is pressed $\geq N+1$ s.t. d sequences of $x(n)$
are generated

$$x(t) = \sum_{n=-N}^{n=N} x(n) p(t-nT)$$

$$x(n) = \sum_{k=-N}^{k=N} x(k) p[n-k]$$

$$x(t) = \sum_{n=-N}^{n=N} x(n) p(t-nT)$$

$$x(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt = \int_{-\infty}^{\infty} \left(\sum_{n=-N}^{n=N} x(n) p(t-nT) \right) e^{-j2\pi f t} dt = \sum_{n=-N}^{n=N} x(n) \int_{-\infty}^{\infty} p(t-nT) e^{-j2\pi f t} dt = \sum_{n=-N}^{n=N} x(n) p(f) e^{-j2\pi f (nT)}$$

$$\Rightarrow \int_{-\infty}^{\infty} \sum_{n=-N}^{n=N} \sum_{m=-N}^{m=N} |x(n)x^*(m)| p(t-nT)p^*(t-mT) e^{-j2\pi f (t-t)} dt_1 dt_2 = \sum_{n=-N}^{n=N} \sum_{m=-N}^{m=N} x(n)x^*(m) \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} p(t-nT) e^{-j2\pi f t_1} p^*(t-mT) e^{-j2\pi f t_2} dt_1 dt_2 = \sum_{n=-N}^{n=N} \sum_{m=-N}^{m=N} x(n)x^*(m) \left[\int_{-\infty}^{\infty} p(t-nT) e^{-j2\pi f t_1} dt_1 \right] \left[\int_{-\infty}^{\infty} p^*(t-mT) e^{-j2\pi f t_2} dt_2 \right] = \sum_{n=-N}^{n=N} \sum_{m=-N}^{m=N} x(n)x^*(m) p(f) e^{-j2\pi f nT} p^*(f) e^{j2\pi f mT}$$

$$\Rightarrow \sum_{n=-N}^{n=N} \sum_{m=-N}^{m=N} x(n)x^*(m) |p(f)| e^{-j2\pi f (n-m)T}$$

$$\Rightarrow \sum_{n=-N}^{n=N} \sum_{m=-N}^{m=N} a(n)b^*(m)c(n-m)$$

$$c(k+1)$$

$$N=M$$

$$c(k+1)$$

$$c(0) \sum_{k=-N}^{k=N} a(k)b[k]$$

$$c(1) \sum_{k=-N}^{k=N} a(k)b[k-1]$$

$$c(-1) \sum_{k=-N}^{k=N} a[k]b[-k]$$

$$c(N+1) = c[2N] x[n] x^*[n-1C]$$

$$\text{for } n \neq N+1: [c[i] \sum_{k=-N-i}^N x[k] x^*[k-1]]$$

$$\text{or } (6) \sum_{k=-N}^{k=N} x(k)x^*[k]$$

$$c[i] = \sum_{k=-N-i}^{N-i} x[k] x^*[k-1]$$

$$c(-i) = \sum_{k=-N+i}^{N+i} x[k] x^*[k+1]$$

$$\text{Using them in double summation sequence}$$

$$\sum_{m=-N}^{m=N} \sum_{n=-N}^{n=N} x[n] x^*[m] |p(f)|^2 e^{-j2\pi f (n-m)T}$$

$$= \frac{|p(f)|^2}{T} \sum_{i=1}^{2N} c[j2\pi f i T] \left(\sum_{k=-N+i}^{N+i} x[k] x^*[k-1] \right) / 2N$$

$$+ \frac{|p(f)|^2}{T} \sum_{i=2N}^1 e^{j2\pi f i T} \left(\sum_{k=-N}^{N-i} x[k] x^*[k+1] \right) / 2N$$

Continued..

Date: 5/10/2016

$$\frac{|X(f)|^2}{2\pi T} = \frac{|P(f)|^2}{T} \sum_{i=-N}^{N+1} e^{j2\pi f iT} \left(\sum_{k=-N+i}^{N-i} |x(k)|^2 + \sum_{i=1}^{2N} e^{-j2\pi f iT} \right)$$

$$= \left(\sum_{k=-N-i}^{N-i} x(k) x^*(k-i) \right) + \sum_{i=1}^{2N} e^{-j2\pi f iT} \left(\sum_{k=N+i}^{N+i} x(k) x^*(k+i) \right)$$

$$\frac{|X(f)|^2}{2\pi T} = \frac{|P(f)|^2}{T}$$

$$= \frac{|P(f)|^2}{T} \left[\sum_{k=-N}^N |x(k)|^2 + \sum_{i=1}^{2N} e^{-j2\pi f iT} \left(\sum_{k=-N+i}^{N-i} x(k) x^*(k-i) \right) \right]$$

$$\sum_{n=N}^N \sum_{m=-N}^N x(n) x^*(m) e^{-j2\pi f (n-m)T} + \sum_{i=1}^{2N} e^{-j2\pi f iT} \left(\sum_{k=N+i}^{N+i} x(k) x^*(k-i) \right)$$

for the integer part: $\frac{1}{2\pi T} \sum_{k=-N-i}^{N-i} x(k) x^*(k-i)$

$$= \sum_{k_1=m_1}^{k_2=m_2} x(k) x^*(k-p)$$

$\{x(n)\}_{n=-\infty}^{\infty}$
a discrete
time random
process

now take limit
 $m_2 \rightarrow \infty$
 $m_1 \rightarrow -\infty$
 $m_2 - m_1 \rightarrow \infty$

$$= \sum_{k_1=m_1}^{m_2} x(k) x^*(k-p)$$

$\sum_{k=m_2}^{k=m_1} x(k) x^*(k-p) \Rightarrow$ As each time, it will be same
as $m_2 - m_1 \rightarrow \infty$
No matter what sequence is there

$$\therefore \sum_{k=m_2}^{k=m_1} x(k) x^*(k-p) = R(p) \quad \text{--- (2)}$$

and $\{x(n)\}$ = {+A, -A}
with equal probability $\frac{1}{2}$ & $\frac{1}{2}$

To check for IDs

Let day outputs are generated multiple times from a machine
as 1st, 2nd, ..., nth, ..., mth $\xrightarrow{T_B}$

Trial #1	$x[n]$	$x[m]$
	+A	+A

Trial #	-A	+A
---------	----	----

Now say $P(x(n)=A, x(m)=A) = 1/9$

$$P(-A, +A) = 2/9$$

$$P(+A, -A) = 2/9$$

$$P(-A, -A) = 4/9$$

And let it be tested

$$P(x(n)=A) = 1/3$$

$$P(x(n) = -A) = 2/3$$

$$P(x(m)=A) = 1/3$$

$$P(x(m) = -A) = 2/3$$

No independence can be checked by $f_{xy}(x,y) = f_x(x)f_y(y)$

Definition $P(x(n)=a, x(m)=b) = P(x(n=a) \cap x(m=b))$

* Toss example

Now coming back to expression --- (2)

$$\text{And } \{x(n)\}_{n=0}^{\infty} = \{0, 1, 2, 1\} \text{ i.i.d}$$

Law of large numbers

$$\lim_{N \rightarrow \infty} \left(\frac{1}{N} \sum_{n=1}^N x[n] \right) \sim \text{i.i.d}$$

$$\boxed{\lim_{N \rightarrow \infty} \left(\frac{1}{N} \sum_{n=1}^N x[n] \right) \rightarrow E[x]}$$

It converges to truth probability 1.

Evaluating

$$\lim_{N \rightarrow \infty} \left(\frac{1}{N} \sum_{n=1}^N |x[n]|^2 \right) = E[|x|^2]$$

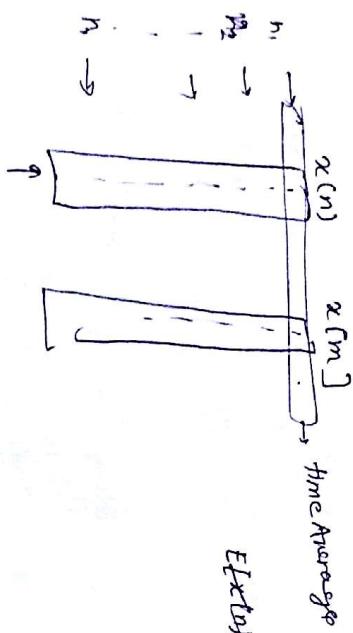
So

$$E[|x|^2] = \sum_{n=-\infty}^{\infty} |x[n]|^2$$

so

$$E[|x|^2] = \sum_{n=-\infty}^{\infty} |x[n]|^2 = \sum_{n=-\infty}^{\infty} \sum_{k=-\infty}^{\infty} x[k] x^*[k-n]$$

is a wss discrete-time sequence



Ensemble mean

Time Average

for a discrete time, random process

2. Ensemble means should be constant

And $E[\sin(\omega t)^2]$ would be independent of t or ω

Time averaged correlation $\Rightarrow \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{n=-N}^N x[n] x^*[n]$ long eqn

Ensime correlation $\rightarrow E[x(n)x^*(m)]$
For RSS - ergodic in mean, autocorrelation

$$\text{These results infer } \sum_{k_1, m_1}^m x[k] x^*[k-m_1] = R \delta_{m_1}$$

0

for $P \neq 0 \rightarrow$ terms will be zero

$$\text{Similarly } \sum_{k_1, m_1}^m e^{-j2\pi f_1 k_1} \sum_{k_2=-N}^{N+1} \frac{x[k] x^*[k-m_1]}{2N} = R \delta_{m_1}$$

so terms will go to

$$\sum_{k_1, m_1}^m e^{-j2\pi f_1 k_1} \sum_{k_2=-N}^{N+1} \frac{x[k] x^*[k-m_1]}{2N} = R \delta_{m_1}$$

$$e^{-j2\pi f_1 k_1} \sum_{k_2=-N}^{N+1} x[k] x^*[k-m_1]$$

$$R \delta_{m_1}$$

And for $N \rightarrow \infty$

$$E[x(t)] = \frac{1}{T} \left(\sum_{k_1=-\infty}^{\infty} R \delta_{m_1} e^{-j2\pi f_1 k_1} \right)$$

for the case where in
only terms, which is
non-zero

Question 1:

Let $\{b[n]\}_{n=-\infty}^{\infty}$ be a sequence of 1-bit information bits
 $(+1 \text{ and } -1)$ are equally likely
bit \rightarrow modulation symbol

$$\frac{b[n]}{b[n-1]} = \frac{b[n]}{s[n]} \quad s[n]$$

+1	+1	A + jA
+1	-1	A - jA
-1	+1	-A + jA
-1	-1	-A - jA

Complex waveform is represented by $x(t) = \sum_{n=-\infty}^{n=\infty} s[n] e^{jBt - n}$

Find the PSD of $x(t)$

\Rightarrow first to show $s[n]$ is an ergodic process or not

$$E[s(n)] = \cancel{s(n)} \quad R[s(n)R[s(t-n)]]$$

$$E[s(n)^2] = [A + jA] P[s(n)] = A + jA$$

$$\begin{aligned} & (2A - jA) [P[s(n)] = 2A - jA] \\ & + (A + jA) [P[s(n)] = -A + jA] \\ & + (A - jA) [P[s(n)] = -A - jA] \end{aligned}$$

$$E[s(n)] = A + jA \quad P[s(n)A + jA]$$

$$P[b[n]] = P[b[n-1] b[n]]$$

$$\xrightarrow{k=0} \frac{1}{4} [(A + jA) + jA - jA + A - jA] = \frac{(A - jA)}{4}$$

$$E[s(n)s(n-1)] =$$

$$\frac{1}{4} [(A + jA)^2 + jA \cdot jA + -jA \cdot jA + (A - jA)^2]$$

$$= \frac{1}{4} [A^2 + jA^2 + jA^2 + A^2]$$

$$E[s(n)s^*(n+1)] = E[s(n+1)s^*(n)]$$

$$E[s(n)]^2 = \frac{A^2}{2} [1 + 3j]$$

$$E[s(n)]^2 = \frac{A^2}{2} (1 - j3)$$

$$S_x(f) = \sum_k R_S(k) e^{jk2\pi f t} \frac{P(k)}{T}^3 = \frac{1}{T} \left(\sum_k R(k) e^{-jk2\pi f t} \right) \frac{P(k)}{T}^2$$

$$\xrightarrow{\text{PSD constant}} \frac{P(k)}{T} e^{-jk2\pi f t} \quad T = \frac{1}{B}$$

\Rightarrow RNN covariance matrix is symmetric also

$$\begin{aligned} &= y = x + N \\ &\quad \text{if } AC \text{ RNN sum} \\ &N = N^T + jN^A \\ &E[n n^A] = n^T \\ &|n| \leq 1 \\ &\text{for any } x \in \mathbb{R}^n \end{aligned}$$

$$\begin{aligned} \text{cov} \left(\begin{bmatrix} n^T \\ n^A \end{bmatrix} \right) & \quad \text{if complex } x^H A x \geq 0 \\ & \quad \text{for } \frac{P_D}{P_N} \text{ positive definite strictly} \end{aligned}$$

$$R_N^T = R_N \rightarrow \text{for symmetric case.}$$

$$\begin{aligned} \text{For } R_N = E \left\{ \begin{bmatrix} N^T - E(n^T) & [(N^T - E(n^T))(n^0 - E(n^0))] \\ [(N^T - E(n^T))(n^0 - E(n^0))] & N^0 - E(n^0) \end{bmatrix} \right\} \end{aligned}$$

To prove PSD

$$\begin{aligned} N^T R_N N^0 &\geq 0 \quad \text{As random noise due to } N \text{ here} \\ &E \left\{ \begin{bmatrix} x_1 & x_2 \end{bmatrix} \cdot \underbrace{\begin{bmatrix} N^T - E(n^T) & [(N^T - E(n^T))(n^0 - E(n^0))] \\ [(N^T - E(n^T))(n^0 - E(n^0))] & N^0 - E(n^0) \end{bmatrix}}_{\text{symmetric}} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \right\} \end{aligned}$$

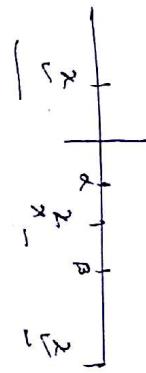
Let's take an example

$$\begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} x \\ y \end{bmatrix} \geq 0$$

$$(ax + c)y + (bx + d)x \geq 0$$

$$ax^2 + 2cx + b \geq 0$$

$$2x + \frac{\partial c}{\partial x}x + \frac{b}{a} \geq 0$$



can be interpreted as $(x-\alpha)(x-\beta) \geq 0$

Both α and β are complex conjugate

- $E \left\{ (n^T - E(n^T))^2 + x_2(n^0 - E(n^0))^2 \right\}$
- Above term of is positive definite ≥ 0

It has to ≤ 0

$$c^2 \leq ab$$

$$ab - c^2 \geq 0$$

19.10.2016

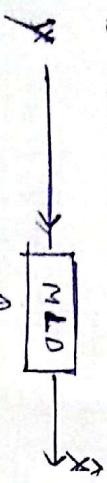
$\Rightarrow \frac{\partial R^2}{\partial x^2} \rightarrow$ Symmetric matrix condition

i) Diagonal elements $a \geq 0$ $b \geq 0$

ii) and Determinant ≤ 0

$$y = x + N$$

Need to find out optimal detector. but here equally likely so N is also suitable



Jointly Gaussian

$$(y^f, y^0)$$

$$\begin{aligned} P_1(y) &= P(x=+1) P(y|x=+1) \\ P_2(y) &= P(x=-1) P(y|x=-1) \end{aligned}$$

$$\begin{aligned} P(y/x=+1) &\stackrel{1}{\overbrace{(2\pi)^{|K|/2}}} e^{-\frac{1}{2} \left[(y^f-1)y^0 \right] \text{covariance}} \\ &\quad \cdot \int_{y^0}^{y^f-1} \left[\frac{1}{2} \right] \text{don't change} \end{aligned}$$

$$P(y/x=-1) = P(y = -1) =$$

$$\begin{aligned} \text{Now } & \frac{1}{e^{\frac{1}{2} \left[(y^f-1)y^0 \right] \left(\frac{Ry}{\mu_x} + (\mu_y - 1) \right)}} \\ & e^{-\frac{1}{2} \left[(y^f-1), y^0 \right] \left(\frac{Ry}{\mu_x} + (\mu_y - 1) \right)} \end{aligned}$$

$$\frac{x = +1 - ((y^f-1)^2 (Ry/\mu_x) + (\mu_y - 1))}{x = -1 - ((y^f-1)^2 (Ry/\mu_x) + (\mu_y - 1))}$$

$x < 0$ then $x = +1$
 $x > 0$ then $x = -1$

$$\begin{aligned} R_N &= \sigma^2 \begin{bmatrix} 1 & r \\ r & 1 \end{bmatrix} \\ R_N^{-1} &\Rightarrow \frac{1}{\sigma^2(1-r^2)} \begin{bmatrix} 1 & -r \\ -r & 2 \end{bmatrix} \\ &= [(y^f-1)y^0] \begin{bmatrix} 1 & -r \\ -r & 1 \end{bmatrix} \begin{bmatrix} y^f-1 \\ y^0 \end{bmatrix} \end{aligned}$$

$$\frac{\partial \ln P(y/x=+1)}{\partial y^f} = \frac{\partial}{\partial y^f} \left[(y^f-1)y^0 \right] R_N^{-1} \begin{bmatrix} y^f-1 \\ y^0 \end{bmatrix}$$

$$\begin{aligned} &= [(y^f-1) - ry^0] - r(y^f-1) + y^0 [-r(y^f-1) + y^0] \\ &= [(y^f-1) - ry^0] [(y^f-1) + y^0] - r[(y^f-1) - ry^0] + y^0 (-r(y^f-1) + y^0) \\ &= [(y^f-1) - ry^0]^2 + y^0 (-r(y^f-1) + y^0) \end{aligned}$$

$$\Rightarrow (y^f-1)^2 - 2r(y^f-1)y^0 + (y^0)^2$$

$$(y^f-1)^2 - 2r(y^f-1)y^0 + (y^0)^2 \geq (y^f-1)^2 - 2r((y^f-1)y^0 + (y^0)^2)$$

$$\begin{cases} x = +1 & (y^f-1)^2 \geq (y^f-1)y^0 + (y^0)^2 \\ x = -1 & (y^f-1)^2 \leq (y^f-1)y^0 + (y^0)^2 \\ x = +1 & (y^f-1)^2 > (y^f-1)y^0 + (y^0)^2 \end{cases}$$

$$x = \text{sign}(y^f - ry^0)$$

$\Rightarrow \rho_e$ expression

$$\Rightarrow \rho_e = P(X=+1)P(\text{Energy}/X=+1) + P(X=-1)P(\text{Energy}/X=-1)$$

$$\rho_e = P(N^f - rN^g < 0 | X=+1)$$

$$N^f = L + N^f$$

$$N^g = N^g$$

$$N^f - rN^g = 1 + (N^f - rN^g)$$

$$= P((N^f - rN^g) < -1 | X=+1)$$

Here N^f and N^g are independent of X so above expression

$$\text{surprise is } \rightarrow P(N^f - rN^g < -1)$$

As N^f and N^g are Gaussian R.V

$$\begin{aligned} E[N] &= 0 \\ E[\omega] &= E[\omega^2] = E[(N^f - rN^g)^2] \\ &= E[N^f]^2 + r^2 E[N^g]^2 + 2rE[N^f N^g] \end{aligned}$$

$$= r^2 + r^2 \sigma^2 - 2r\sigma^2 \cdot 2r$$

$$\Rightarrow \sigma^2(1+r^2) = 2r^2\sigma^2$$

$$= r^2(1+r^2)$$

$$\text{So PDF would be } -\frac{\omega^2}{2\sigma^2(1+r^2)} e^{-\omega^2/2\sigma^2} \approx e^{-x^2/2}$$

$$= \int_{-\infty}^{-t} \frac{1}{\sqrt{2\pi\sigma^2(1+r^2)}} e^{-x^2/2} dx$$

$$\rho_e \Rightarrow a \left(\frac{1}{\sqrt{r^2(1+r^2)}} \right)$$

$$\text{As } r \text{ is correlation coefficient} \\ \Rightarrow \frac{E[N^f N^g]}{\sqrt{E[N^f]^2 E[N^g]^2}} = r = \text{cor}(N^f, N^g)$$

$$\rho_e = a \left[\frac{1}{\sqrt{r^2 + r^2}} \right]$$

By subtracting, If a denotes variance of noise N^f and N^g As $1 + c \underline{N^f - rN^g} \rightarrow \underline{r=1}$ fully correlated

Minor 3rd part solution

$\frac{1+j}{1-j}$	$-1+j$	0	
$-1+j$	0	$-1-j$	

$$\rho_e \rightarrow x = (\chi^{(1)}, \chi^{(2)}, \chi^{(3)})$$

$$y = (y^{(1)}, y^{(2)}, y^{(3)})$$

$$d(x, y) = \|x - y\|$$

$$\sqrt{\frac{dy}{dx}}$$

Ans

$$x = [x^{(1)}, x^{(2)}, x^{(3)}]$$

of a column

$$\text{Energy} = \sqrt{(x^{(1)} - y^{(1)})^2 + (x^{(2)} - y^{(2)})^2 + (x^{(3)} - y^{(3)})^2}$$

$$\text{Ans} = \frac{1}{2} \times 4 + \frac{1}{2} \times 4 = \frac{9}{2}$$

$$\text{Ans} = \frac{9}{2}$$

Ans of information loss

$$\text{Here } \Omega\left(\frac{\sigma}{\sqrt{E}}\right)$$

\Rightarrow To find curve here we have to multiply code book by constant to vary Eb

$E_{\text{avg}} = C/\alpha^2$

$$C = C/\alpha^2 = 3/\alpha^2$$

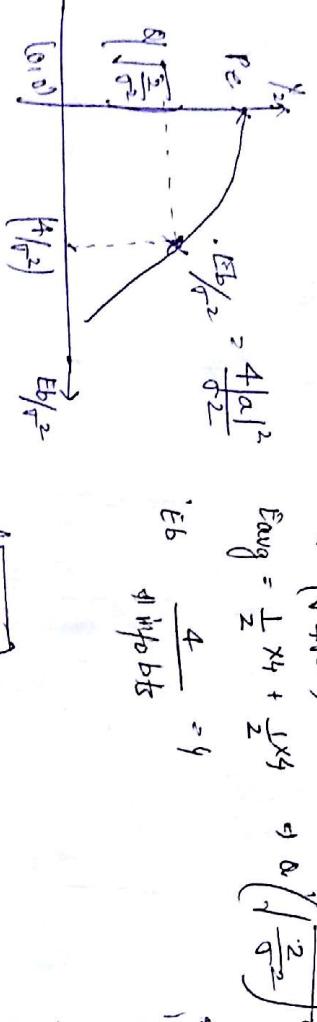
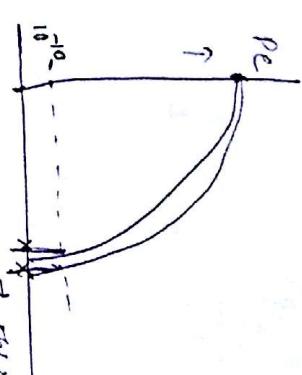
$$P_e^{(A)} \leq 3 \sqrt{\frac{10|\alpha|^2}{4\sigma^2}}$$

$$\frac{Eb}{\sigma^2} = \frac{3|\alpha|^2}{\sigma^2}$$

$$\frac{|\alpha|^2}{\sigma^2} = \frac{Eb}{3\sigma^2}$$

$$P_e^{(A)} \leq 3 \alpha \left(\sqrt{\frac{105}{4} \frac{Eb}{\sigma^2}} \right)$$

$\frac{Eb}{\sigma^2}$ requirement of B is more than A
So Code Book A is more efficient.



$$P_e \Rightarrow \Omega\left(\frac{d_{\min}}{2\sigma^2}\right)$$

$$\frac{\alpha^2}{\sigma^2} = \frac{Eb}{4\sigma^2}$$

$$\Omega\left(\frac{|\alpha|^2}{\sigma^2}\right)$$

$$P_e = \Omega\left(\sqrt{\frac{Eb}{2\sigma^2}}\right)$$

$$d_{\min} = \frac{1}{\sqrt{10}}$$

if we get

$$d_{\min}^2 = 10|\alpha|^2$$

$$P_e^{(A)} \leq 3 \alpha \left(\sqrt{\frac{d_{\min}^2}{4\sigma^2}} \right)$$

$$\frac{5}{6} \frac{Eb^A}{\sigma^2} = \frac{Eb^B}{2\sigma^2} \frac{Eb^A}{Eb^B} = \frac{5}{6}$$

Plotting both P_e on the same graph

$$P_e^A = P_e^B$$

$$10 \log_{10} \left(\frac{Eb^B}{Eb^A} \right)$$

$$= 10 \log_{10} \left(\frac{5}{3} \right) = 4.08 = 2.09 \text{ dB}$$

Scanned by CamScanner

Date: 14.10.2016

$\frac{1+j}{1-j}$	$-j+j$	0
$\frac{-1+j}{1-j}$	0	$-1-j$

④

for improving efficiency

keeping Eb same. dim same =

code B' will be more power efficient than code B if B's

curve is below B

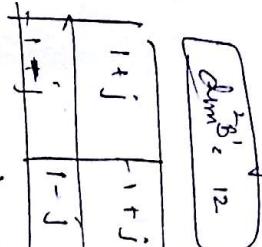
dim = 2

So dim increases by
a is changed to $-1-j$

a is changed to

$\frac{1+j}{1-j}$	$-1+j$	0
$\frac{-1+j}{1-j}$	0	$-1-j$

dim = 12



$$\text{dim} \geq \sqrt{4+4+2+2}$$

$\frac{1+j}{1-j}$	$-1+j$	0
$\frac{-1+j}{1-j}$	0	$-1-j$

dim = 16 ↓

↑ removed

+

n = 10000

n = -∞

n = +∞

n = 0

n = 1

n = 2

n = 3

n = 4

n = 5

n = 6

n = 7

n = 8

n = 9

n = 10

n = 11

n = 12

n = 13

n = 14

n = 15

b[n]

b[n-1]	b[n]	s[n]
+1	+1	$A(1+j)$
-1	+1	$A(-1+j)$
-1	-1	$A(-1-j)$

$$x(t) = \sum_{n=-\infty}^{\infty} s[n] \sin(Ct - n)$$

$$\begin{aligned} b_0 &= b_1 & -2 & -3 & -4 & -5 & -6 & -7 \\ b[n] &+1 & -1 & +1 & +1 & +1 & +1 & +1 \end{aligned}$$

$$\begin{aligned} n &= 1 & n &= 2 & n &= 3 \\ A(-1+j) & A(1-j) & A(1+j) \end{aligned}$$

bit rate = $\frac{10000}{1000/B}$
if 10000 bits
are taken
and 100 information bits
are transferred.

$$\begin{aligned} \text{bit rate} &= 1000 \\ &\approx B \text{ bits/sec} \end{aligned}$$

if 10000 bits
are taken
and 100 information bits
are transferred.

②

$$\text{Spectral efficiency} = \frac{B \text{ bits/sec}}{B \text{ Hz}}$$

$$= \frac{1 \text{ bits/sec-Hz}}{1 \text{ Hz}}$$

$$R_b = \frac{B}{T_s} \Rightarrow \frac{B}{T_s}$$

$$R_b = \frac{B}{T_s} = \frac{B}{\frac{T_s}{2}} = 2B$$

Properties of bit sequence & the symbol energy is

$$2^{N^2}$$

Also total energy $\rightarrow \frac{2^{N^2} \times 10000}{10001} \rightarrow$ total no. of sequences

$\frac{1}{10001} \rightarrow$ no. of information bits

Sequence can also be represented as

-1+j	0
-1+j	0
-1+j	0

Only symbols have to transmitted

Step 0

Now taking sequence in chunks of 10.

Average codeword length

$$E_{\text{avg}} = \frac{9 \times 2 A^2}{100}$$

$$= 0.2 A^2$$

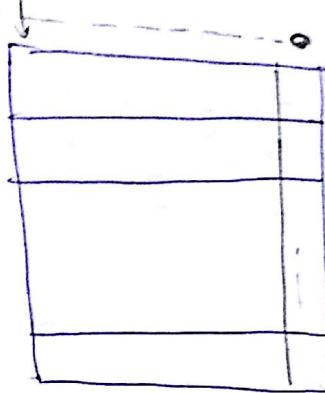
$$\begin{aligned} \text{Real part} &\rightarrow n^2 E[b(n)b(n-1)] + n^2 E[b(n-1)b(n-2)] \\ &\rightarrow n^2 E[b(n)b(n-1)] \xrightarrow{\text{as independent}} 0 \end{aligned}$$

$$= 0$$

$$= 0$$

$$\begin{aligned} \text{Imaginary part} &\rightarrow A^2 E[\{b(n-1)\}^2] - b(n)b(n-2) \\ &\rightarrow A^2 \cdot E[b(n-1)^2] \rightarrow 0 \\ &\Rightarrow j A^2 \end{aligned}$$

Step 1



If N is even, then N is divided in to four chunks

$$= \frac{N}{4} \times 2n^2 + \frac{N}{4} \times 2A^2 \rightarrow \frac{N}{4} \times 2n^2 + \frac{N}{4} \times 2A^2$$

$$\underline{N+1} \rightarrow \text{No. of info bits}$$

$$= \frac{Nn^2}{4}$$

$$E[S(n) S^*(n-k)] \rightarrow E[\beta(n)] \left\{ \begin{array}{l} A^2, k=0 \\ jA^2, k=1 \\ jA^2, k=2 \\ 0, k=3 \end{array} \right.$$

$$\begin{bmatrix} A^2 & 0 & 0 & 0 \\ jA^2 & A^2 & 0 & 0 \\ jA^2 & 0 & A^2 & 0 \\ 0 & 0 & 0 & A^2 \end{bmatrix}$$

$$E[S(n) S^*(n-1)]$$

$$\begin{aligned} \text{As } S[n] &\rightarrow [b(n) + jb(n-1)] \rightarrow b(n-1) - jb(n-2) \\ \times S^{*(n-1)} &\rightarrow A^2 \xrightarrow{\text{because } b(n-1)} \\ \Rightarrow S^{*(n-1)} & \end{aligned}$$

$$\rightarrow A^2 \left[b(n)b(n-1) - jb(n)b(n-2) - jb(n-1)^2 \right]$$

$$+ b(n-1)b(n-2)$$

$$\begin{aligned} \text{Real part} &\rightarrow n^2 E[b(n)b(n-1)] + n^2 E[b(n-1)b(n-2)] \\ &\rightarrow n^2 E[b(n)b(n-1)] \xrightarrow{\text{as independent}} 0 \end{aligned}$$

$$= 0 + 0$$

$$= 0$$

$$= 0$$

$$= 0$$

$$= 0$$

$$= 0$$

$$\text{PSD.} = \sum_{k=-\infty}^{\infty} R_S(k) e^{-j2\pi kfT}$$

$$R_S(f) = \frac{|P(f)|^2}{T} \sum_{k=-\infty}^{\infty}$$

Linear Modulation

$$\alpha(t) = \sum_{n=-\infty}^{\infty} x[n] p(t-nT)$$

$x[n]$ are iid zero mean var and $x[n] \in A$

R.F

$$S_x(f) = \frac{|P(f)|^2}{\pi} \sum_{k=-\infty}^{\infty} R_x[k] e^{-j2\pi kfT}$$

$$S_x(f) = \frac{|P(f)|^2}{\pi} R_x[0] \quad \text{as R.F}$$



$$\Rightarrow \int_{-\infty}^{w_b/2} S_x(f) df \quad \text{fraction power}$$

$$\int_{-\infty}^{\infty} S_x(f) df$$

≥ 0.99

To find out w_b such that above expression ≥ 0.99.

Given \Rightarrow $x[n] \sim \text{ggf. bandwith}$

$$\frac{w_b}{x(t)} \rightarrow \frac{1/T}{h(t)} \rightarrow y(t)$$

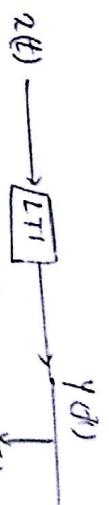
$$y(t) = x(t) * h(t)$$

$$\Rightarrow \left(\sum_{n=-\infty}^{\infty} x[n] p(t-nT) \right) * h(t)$$

$$x(t) > \delta(t)$$

$$y(t) = \left(\sum_{n=-\infty}^{\infty} x[n] p(t-nT) \right) * \delta(t)$$

$$y(t) = \left(\sum_{n=-\infty}^{\infty} x[n] p(t-nT) \right) * \delta(t)$$



$$x(t) \Rightarrow x(t) * q(t)$$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) q(t-\tau) d\tau$$

$$x(t) = \left[\sum_{n=-\infty}^{\infty} x[n] p(t-nT) \right] * q(t)$$

$$q(t) = p(t) * q(t)$$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) c(t-\tau) d\tau$$

$$z(t) = \sum_{n=-\infty}^{\infty} x[n] c(t-nT)$$

$$c(t) = p(t) * q(t)$$

$$y[m] = z(t=mT)$$

$$= x[m] = \sum_{n=-\infty}^{\infty} x[n] c(m-nT)$$

$$= \sum_{n=-\infty}^{\infty} x[n] c(t-mT)$$

$$y[m] = x(t=mT)$$

$$= \sum_{n=-\infty}^{\infty} x[n] c(t-mT)$$

$$= x[m] + x[m-1] c(t) + \dots$$

$\alpha \ln \int c(t) \rightarrow$ information part

$$\alpha \ln \int c(t) \rightarrow \text{intersymbol interference}$$

All the interference terms should be 0

$$c(t) = g(t) * p(t)$$

$$c(kT) = \begin{cases} 1, & k=0 \\ 0, & k \neq 0 \end{cases}$$

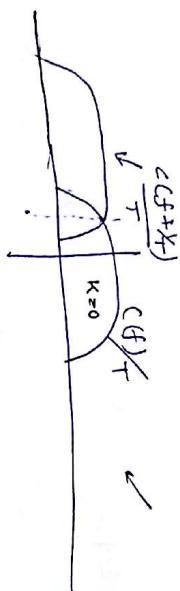
by sinc function

⇒ Above condition gets satisfied

In frequency domain

$$c(kT) = \begin{cases} 1, & k=0 \\ 0, & k \neq 0 \end{cases}$$

$$\frac{1}{T} \sum_{k=-\infty}^{\infty} e^{j(f + \frac{k}{T})} = 1$$



Taking $c(t) = \text{sinc}(t/T)$

$$c(t) \Rightarrow \begin{cases} T, & |f| < \frac{1}{2T} \\ 0, & |f| > \frac{1}{2T} \end{cases} \quad \sum_{k=-\infty}^{\infty} c(f + kT) = 1$$

