

03/12/23

Lecture-5

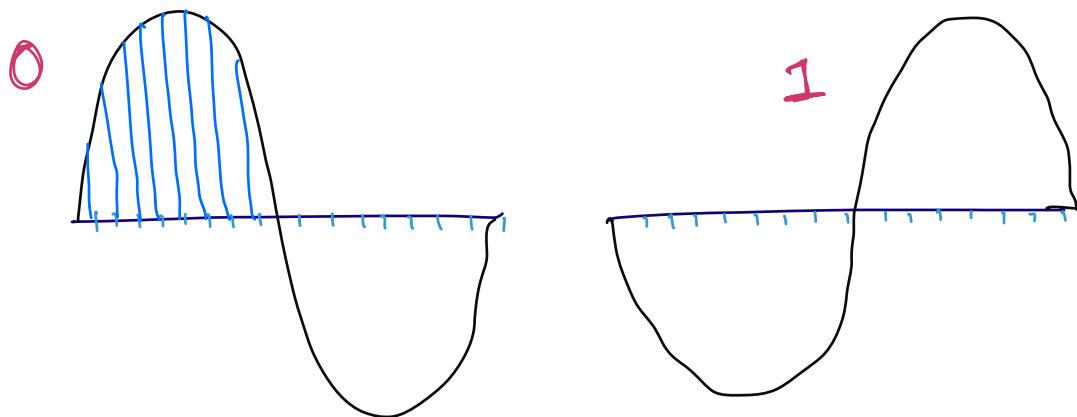
24/03/21

→ Continuation of discussion of Stochastic processes

We saw the definition of stochastic process can been characterised by the random variables at different time

Stamps. But it is unrealistic for $n \rightarrow \infty$

ex: BPSK signal (Binary phase shift key)



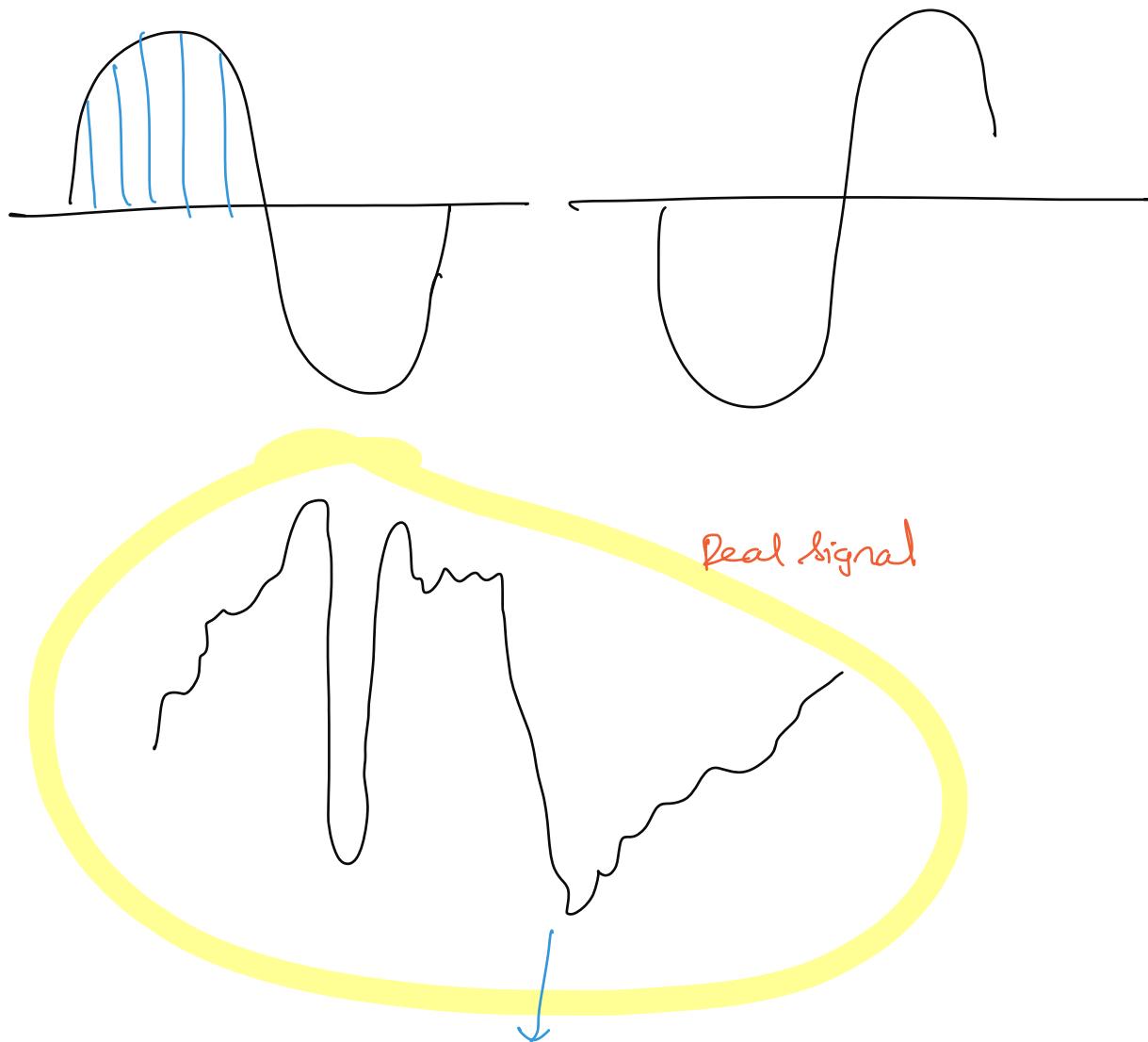
This signal consists of a sequence of waveforms and choose one waveform to transmit a One and another to transmit Zero

Usually in BPSK there are two stretches of sinusoids that can easily be distinguished.

∴ Our signal is usually a sequence of Time Windows each of it is a bit time and each time window

our profile cone Up-down (or) Down-Up
 • \cup '0' \cap '1'

→ The true signal that is present on the channel.
 If we are transmitting a radio signal using tristate-
 It will be very different because superimposed on this
 there will be Noise.



This can be a real zero for BPSK, it seems
 very corrupted but more or less the Up-Down trend

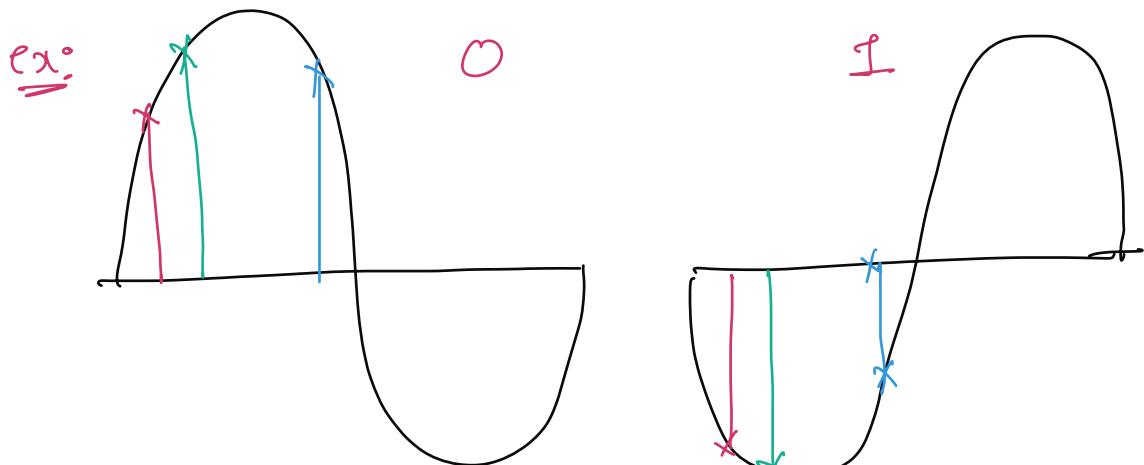
is followed.

But even the UP trend has a sharp Null in time.

→ So everything we have seen above in the realtime signal is a **stochastic process**, we never know what kind of Noise & at what Amplitude & phase would corrupt the original signal.

→ Everytime we transmit some information that we don't know from the beginning. ∴ It is a stochastic process for even for the information it carries

→ Let's focus on the second kind of Randomicity each one attached to a certain time stamp



• +ve value when we are transmitting 0

• -ve value when we are transmitting 1

∴ we don't know from the start what waveforms are transmitting so from the beginning there is a

probability distribution associating to that time instant.

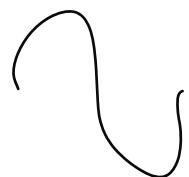
A RV may have all values +ve & -ve distribution

The same happens when we choose Green or Blue
time instants

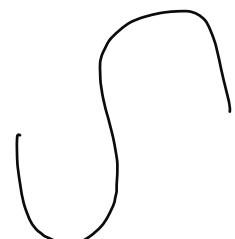
so RV at each time instant will be characterized by
a different PDF.

This is offcourse not enough PDF's of different
RV's will have some link.

*



$$\text{probability} = \frac{1}{2}$$



$$\text{probability} = \frac{1}{2}$$

→ We have to make all the different PDF's of the RV's
to agree to describe the common profile of the transmitted
signal

*

i.e. The PDF in all colored time instances agree
with each other.

- $\rightarrow f_{X(t')}$

- $\rightarrow f_{X(t'')}$

$$\rightarrow f_{X(t'')}$$

\therefore Once we know something about $f_{X(t')}$ then we can say something about the $f_{X(t'')}$

→ This means that we have bunch of PDF's but I have a lot of link b/w them to describe the waveform.

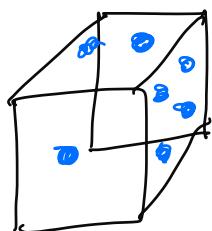
\therefore This waveform has been designed before hand so it is an artificial one, if we design it in a more complicated way then no. of links b/w PDF's will grow and it will be even more complicated.

* This makes the characterisation with TPDF very inefficient.

Q) What is the Alternative point of view ?

Ans Having completely different Approach.

→ consider the example of Rolling a Die



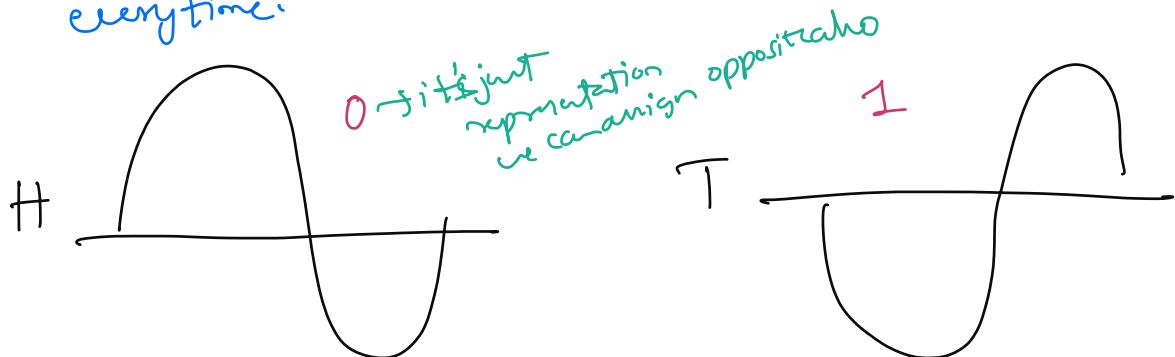
Theoretical
Mapping

Here
 $2 \in \mathbb{R}$

The number '2' has a probability of $\frac{1}{6}$
if the Die is a Fair one.

* When you can associate probabilities to Numbers
Why can't we associate probabilities to all waveforms.

It would be simple in our BPSK case, because
you can toss a coin and Toss it and get Head/Tail
(fair one)
everytime.



→ Here we are assigning probabilities to all waveforms.

* This New point of view will allow us to interpret quite well a lot of signals expressed with all the waveforms

BPSK is equivalent to Coin Tossing

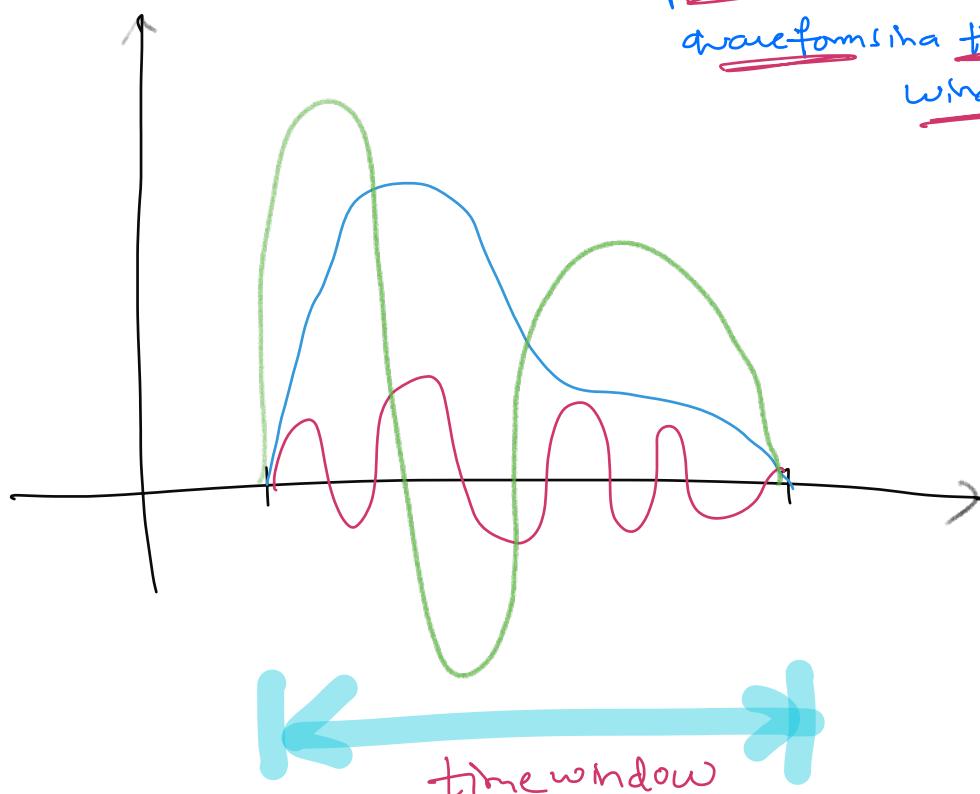
→ Therefore, our RV will not just give us random values but it will be whole waveforms in a Timewindow with different probabilities.

waveforms in a time window are Vector Space!

$$x(t) = \sum_j \alpha_j \phi_j(t)$$

scalars

fixed - non random
waveforms



- We have to attach probabilities to all the waveforms in a time window.

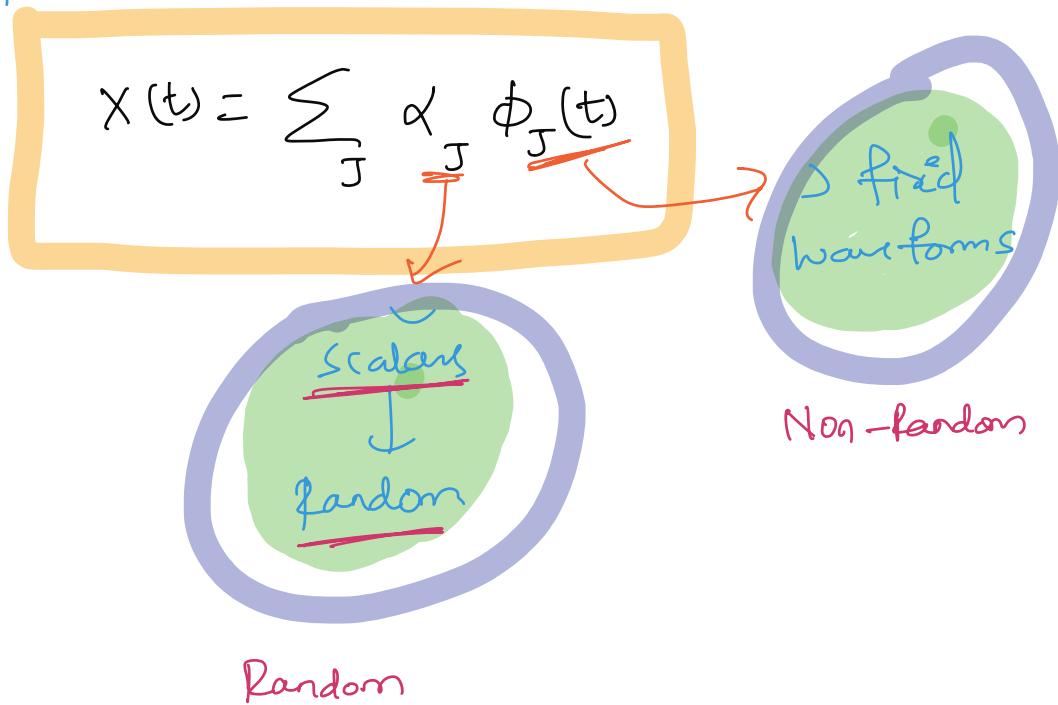
→ Q) Do we have a Basis for the Waveform Vector Space?

Any The answer to the question is No

Because, The Dimension of this Vector space
is Not Finite.

But in General, we can always take waveforms
that we are interested, adjust a pair of sinusoids
i.e. as in BPSK case sinusoid Up-down & Down-up
can still be produced by the linear combination of
some fundamental waveform that I assume to be
fixed.

So, Our Model would be set of waveforms that
are fixed.



→ This is a good way of attacking Random phenomena

to a waveform.

→ If we change α_j we are not only changing the value of F_j at one place but we change the value at all the places spanned by ϕ_j

Ex: 256-QAM modulation



$$d_0 \cos(2\pi f t) + d_1 \sin(2\pi f t)$$

QAM is able to transmit a single waveform upto 256 different symbols.

It does so carefully by choosing these 256 waveforms to which it attaches a probability of course. But these 256 waveforms one for every combination of 8 bits are defined in terms of cosine & sine on the same bandwidth.

∴ Given a bit frequency at Bit time that fix the frequency of sinusoid, what we are focusing is bit time 't' and choosing

Coefficients α_0, α_1 . So that the Linear Combination of two waveforms gives us the final waveform that I associate to each of the possible combination of the 8 bits.

→ So, we have to choose 256

pair of (α_0, α_1) to encode 8 bits

each of the waveforms will have a probability attached to them by means of RV's α_0, α_1 because ϕ_0, ϕ_1 are fixed.

→ If we accept this point of view all we have to do is to characterize separately the Random values of α and function ϕ_J .

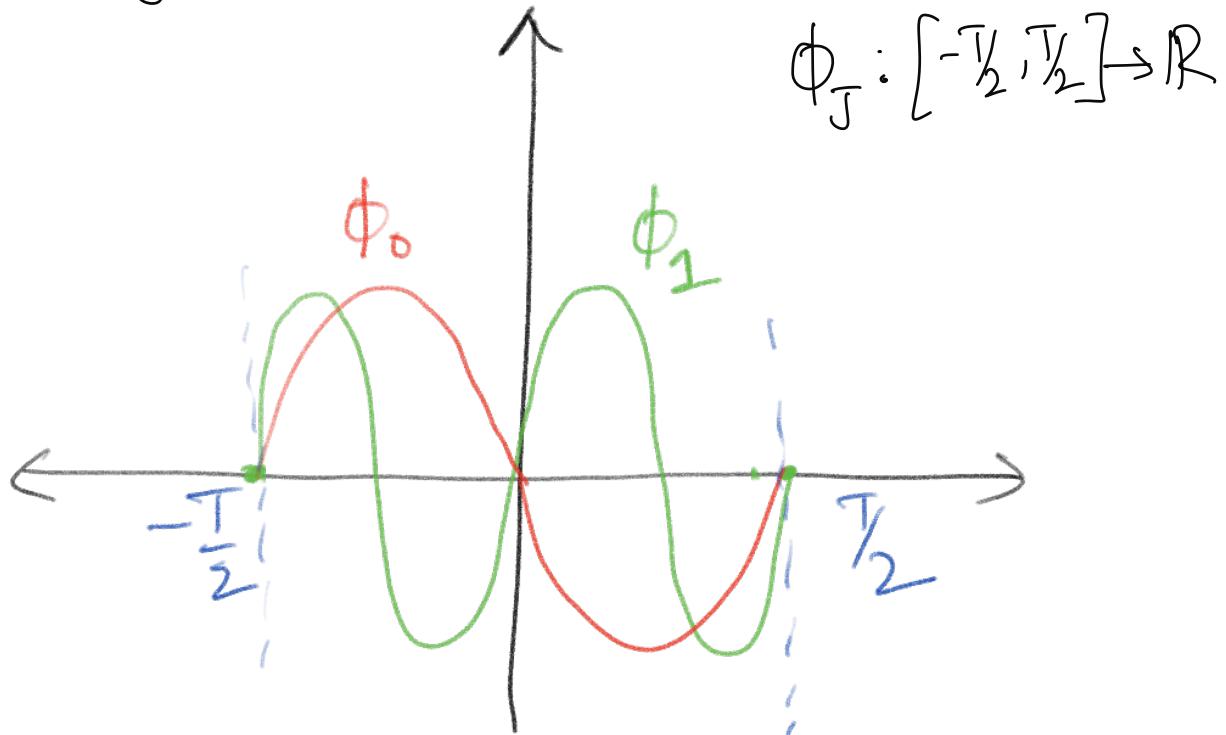
We should try and understand, how this ϕ_J should be made to be useful in the subsequent calculations.

Properties of ϕ_J :

- We would like ϕ_J to be

Orthogonal & possibly Orthonormal

ϕ_j are waveforms on a certain Timeinterval



→ $\alpha(t), \beta(t)$

Scalar product of α, β

$$\langle \alpha, \beta \rangle = \frac{T}{2} \int_{-\frac{T}{2}}^{\frac{T}{2}} \alpha^*(t) \beta(t) dt$$

Scalar product
of Two functions
defined over an
Interval.

* → Conjugation is need for the possibility
of α, β being complex values

→ Now that we have scalar product defined by the waveforms in time interval $[-T_2, T_2]$

→ We can say two waveforms are Orthogonal when the scalar product defined by them in a Time interval is zero

Ex: Two sinusoids with different frequencies

→ Beyond orthogonality we would also like them to be Ortho Normal. i.e their lengths are also equal to 1 apart from being orthogonal

Def: $\langle \phi_j, \phi_k \rangle = \begin{cases} 0 & j \neq k \\ 1 & j = k \end{cases}$

$$\langle \phi_j, \phi_j \rangle = \|\phi_j\|^2 = 1$$

$\uparrow L^2 \text{ Norm}$

→ Beautiful consequence of Ortho Normality.

known let
a waveform $X(t) = \sum_j \alpha_j \phi_j(t)$

- we want to know the coefficients of the waveform
i.e α_k for example

$$\langle \phi_k, x \rangle = \langle \phi_k, \sum_j \alpha_j \phi_j \rangle$$

Scalar product of
a single waveform ϕ_k with the Global waveform x

$$\Rightarrow \int_{T_1/2}^{T_2} \phi_k^*(t) \sum_j \alpha_j \phi_j(t) dt$$

$$\Rightarrow \sum_j \alpha_j \int_{T_1/2}^{T_2} \phi_k^*(t) \phi_j(t) dt$$

This integration has only two values

$$\langle \phi_k, \phi_j \rangle = \begin{cases} 0 & j \neq k \\ 1 & j = k \end{cases}$$

$$\boxed{\langle \phi_k, x \rangle = \alpha_k}$$

Note: → If we want to know what is the coefficient of ϕ_k to give a certain waveform $X(t)$ we may simply project our waveform $X(t)$

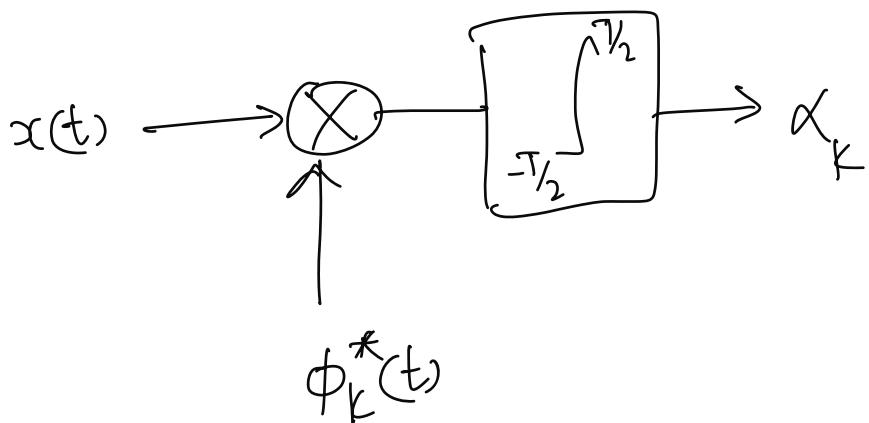
over the fixed waveform $\phi_k(t)$ that we use to express all our signals.

Fixed waveforms to have the above property should be orthogonal to each other

→ The operation we did above can be called

Correlator / projector

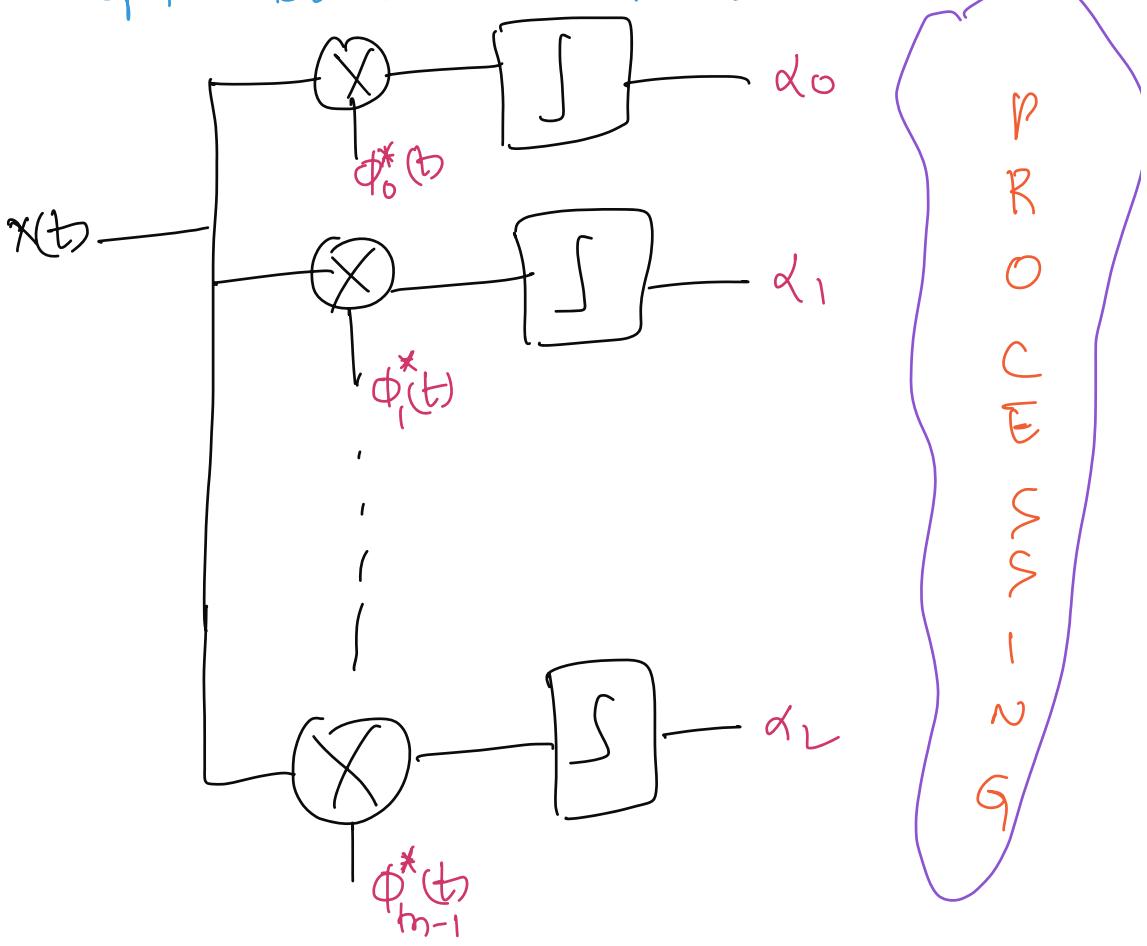
Correlator / projector is a block that performs the above Integral.



But our waveform $x(t)$ is characterized not only by one standard waveform but many other standard waveform from the Basis.

∴ we will have a lot of Projector in general to calculate the value of the coefficients

of the standard waveforms



(8)

→ From this situation on, we may ask ourselves beyond orthonormality is there anything we would like to have for our $\phi_j(t)$? which would make easier to model our process.

The idea is, Correlator/projector have a cost because we have to deploy specific $\phi_j(t)$ to calculate its coefficient.

Ideally we would like to employ least amount of projectors without losing any useful information.

In principle our basic criteria would be okay assuming we have deployed a correlator with respect to $\phi_0(t)$ and you want to decide whether it is sensible to deploy another correlator w.r.t. $\phi_1(t)$. One we have α_1 we must have some new piece of information if not so we would be satisfied by having α_0 w.r.t. $\phi_0(t)$.

- Q) How do we know a new piece of information has come in, when you know α_1 w.r.t. ϕ_0 ?

A) we can resort to the concept of prediction. Because starting from α_0 , if we know how to predict α_1 without deploying any correlator w.r.t. $\phi_1(t)$. It would be great.

We would deploy a correlator for one specific $\phi_j(t)$ if for sure we know that we can't predict α_j from α_0 .

In principle we would like to deploy .

correlators that give us Statistically Independent
 α_j (coefficients)

→ This is regrettable is too complicated to enforce
But we can revert to a limited form of inde-
pendence b/w both which is the property of
zero covariance.

→ We have seen earlier that, if we have two
random variables with zero covariance
it is not possible to predict one value from
another if we limit ourselves to linear
predictors.

→ Therefore, our limited criteria would be
 α_j could not be linearly predicted from
 α_k

$$\text{i.e. } \gamma_{\alpha_j \alpha_k} = 0 \quad \forall j \neq k$$

$$\gamma_{\alpha_j \alpha_j} \Rightarrow E \left[(\alpha_j - m_{\alpha_j})^* (\alpha_j - m_{\alpha_j}) \right]$$

assume

$$\alpha_j, \alpha_k \in \mathbb{R}$$

$$\rightarrow \alpha_j \Rightarrow \int_{-T/2}^{T/2} \phi_j(t) x(t) dt$$

$$m_{\alpha_j} = E[\alpha_j] = E \left[\int \phi_j(t) x(t) dt \right]$$

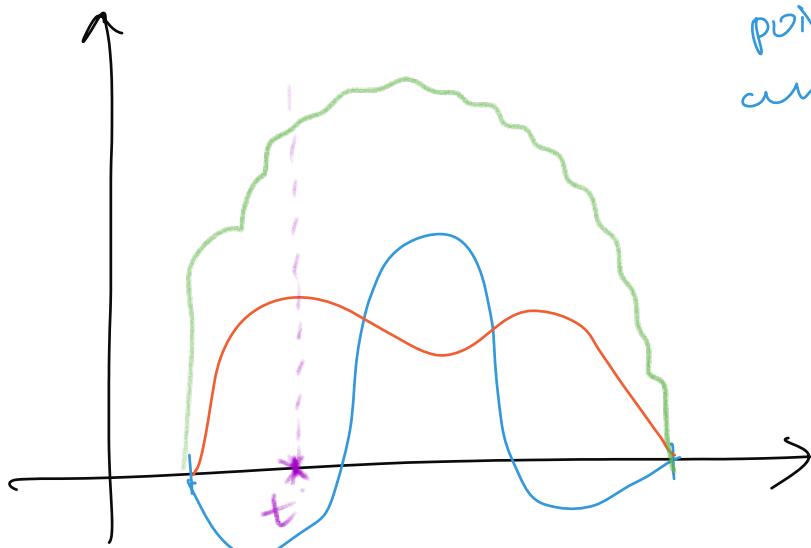
(Average)

$$= \int \underbrace{\phi_j(t)}_T E[x(t)] dt$$

since they are chosen to be fixed

$$\alpha_j = \int \phi_j(t) \underbrace{m_{x(t)}}_{\text{is the average of all points over } x(t) \text{ might occur at time instant } t} dt$$

is the average of all points over $x(t)$ might occur at time instant t



Similarly

$$\alpha_k = \int \phi_k(t) m_{x(t)} dt$$

Substituting α_J, α_K in the expression of covariance.

$$\begin{aligned}
\gamma_{\alpha_J \alpha_K} &= E \left[\left(\underbrace{\int \phi_J(t) x(t) dt - \int \phi_J(t) m_{x(t)} dt}_{\alpha_J} \right) \right. \\
&\quad \left. \left(\int \phi_K(t) x(t) dt - \int \phi_K(t) m_{x(t)} dt \right) \right] \\
&= E \left[\int \phi_J(t) (x(t) - m_{x(t)}) dt \right. \\
&\quad \left. \int \phi_K(t) (x(t) - m_{x(t)}) dt \right] \\
\Rightarrow E \left[\int \int \phi_J(t) \phi_K(s) \left(x(t) - m_{x(t)} \right) \right. \\
&\quad \left. \left(x(s) - m_{x(s)} \right) dt ds \right] \\
&\xrightarrow{\text{Here we have just exchanged the Notation.}} \\
&\int \int \phi_J(t) \phi_K(s) E \left[\left(x(t) - m_{x(t)} \right) \right. \\
&\quad \left. \left(x(s) - m_{x(s)} \right) dt ds \right]
\end{aligned}$$

$$\gamma_{\alpha_j \alpha_k} \Rightarrow \int \phi_j(t) \left(\int K_X(t, s) \phi_k(s) ds \right) dt$$

$\Psi(t)$

$K_X(t, s)$
 (Second order Covariance
function)

Here there is a Statistical characteristic of our process which is the 2nd order Covariance function of our process.

If we would like this Covariance

$$\gamma_{\alpha_j \alpha_k} = \begin{cases} 0 & \text{when } j \neq k \\ \sigma_{\alpha_j}^2 & \text{when } j = k \end{cases}$$

Variance

This would make it impossible to predict α_k if we know α_j by linear prediction.

→ The constraint of Zero Covariance is we would like to have when we want to design our $\phi_j(t)$ (set of basis functions)

i.e reward set of $\phi_j(t)$ should be
orthonormal
& covariance=0

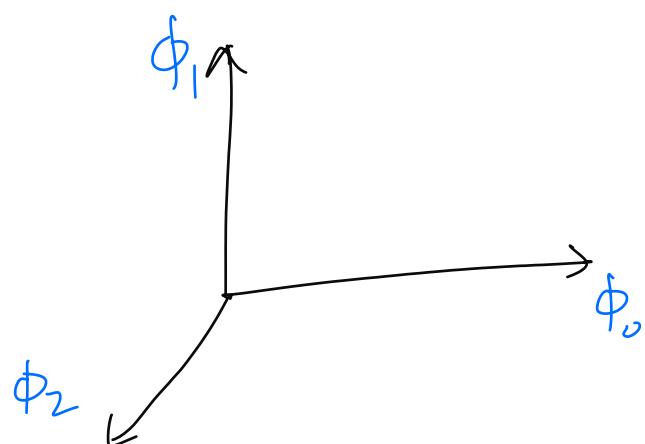
Example:

$$\gamma_{\alpha_J \alpha_K} = \int \phi_J(t) \left[\int k_x(s, t) \phi_K(s) ds \right] dt$$

$\underbrace{\qquad\qquad\qquad}_{\psi(t)}$

Let's say we have 3 ϕ

$\phi_0 \ \phi_1 \ \phi_2$ & $K=0$ i.e $\phi_K=\phi_0$



let's look at the expression of covariance above

$$\int \phi_j(t) \psi(t) dt = 0 \quad \left. \right\} \text{This is the expression of scalar product}$$

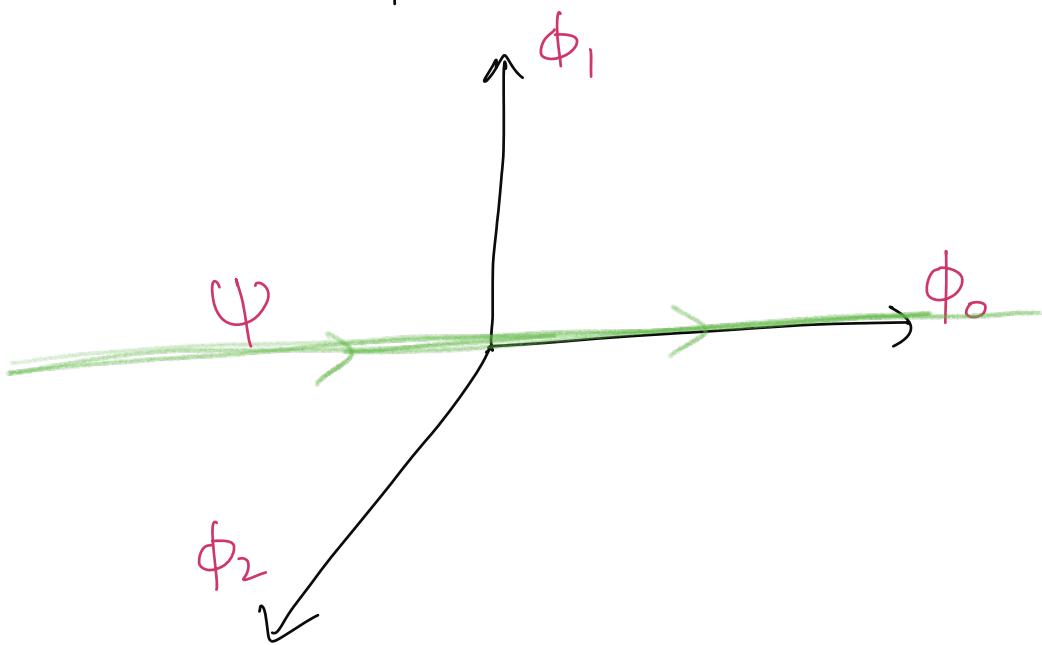
\Downarrow it includes ϕ_0

$$\langle \phi_j, \psi \rangle = 0$$

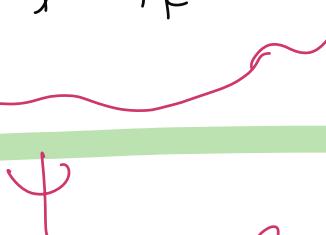
for $j=1, 2$

i.e. ϕ_0 should be orthogonal to ϕ_1 & ϕ_2

' ψ ' can be thought of a scaled version of ϕ_0



$$\Rightarrow \int k_x(t, s) \phi_k(s) ds = \lambda \phi_k(t)$$


 ↓
 Eigenvalue
 ↓
 Eigenfunction

ψ is a Multiple of ϕ_k , ϕ_k is the language of vectors and is equal to the Multiple of

EIGEN EQUATION

$$\int k_x(t, s) \phi_k(s) ds = \lambda \phi_k(t)$$

↑
 Eigenvalue
 ↓
 Eigenfunction

What we see here is a generalization

of the application of Matrix
to a vector

i.e. $AV = \lambda V \Rightarrow$ Eigen Equations

→ we can say $\phi_k(t)$ is an Eigenfunction
corresponding to λ is Eigen value

→ Therefore for our Design: Requirement

The projections α_J & α_K are not predictable by linear means by one and other

* We need to have the functions we are choosing ϕ_J & ϕ_K are also not only ORTHONORMAL but also be the EIGENFUNCTIONS of the Covariance function $k_X(t, s)$ of the process.

→ The problem is There is no guarantee that Eigen values & Eigen vectors exist for all k_X (Covariance function)

Note: The best ϕ_J^s are the Eigenfunctions of the Covariance function of the process. (i.e the Random process)

** → What we are gonna do is to look at a Fundamental theorem from Functional Analysis

But firstly look at the Covariance equation for $J=K$ which is equal to its Variance.

$$\gamma_{JJ} = \int \phi_J \int K_x(t,s) \phi_J(s) ds dt = \sigma_J^2$$

$$\Rightarrow \int \phi_J + \left[\int \frac{d}{dt} \phi_J(t) dt \right] = \sigma_J^2$$

Scalar

$$\Rightarrow \sqrt{\int \phi_J(t) \phi_J(t) dt} = \sigma_J$$

Scalar product of ϕ_J with itself

$$\Rightarrow \lambda \cdot I = \sigma_J^2$$

- The Eigen Function corresponding to ϕ_J is nothing but its Variance σ_J^2 .

$$\lambda_J = \sigma_J^2$$

\therefore once we get set of eigenfunctions of the covariance function then we will have

- the ① orthonormality
- ② zero covariance b/w the projections
- ③ Eigenvalues equal to the variance of the \mathbf{x}_j

\rightarrow But still we have to address existence of Eigenfunction

- ④ Do they exist? if so under what conditions?

$$K_x(t,s) = E \left[(x(t) - m_{x(t)})^* (x(s) - m_{x(s)}) \right]$$

our $K_x(t,s)$ is simply Hermitian

$$\text{i.e } K_x(t,s) = K_x^*(s,t)$$

\rightarrow If we go back to our linear algebra terms when a Matrix is Hermitian in Real domain it meant symmetric
The Eigen values & Eigen vectors are there

\rightarrow To check about the existence of Eigenfunction in functional space, we have to go to a theorem that is analogous to Spectral Decomposition Theorem.

Earlier during Spectral Decomposition Theorem we assume finite dimension.

* In this case it is weaker due to Two reasons

Additional Assumption instead of K_X being Hermitian

(i)

$$\iint |k_x(t,s)|^2 dt ds \leq +\infty$$

finite Energy Covariance

then for

$$\exists \phi_J \Rightarrow \boxed{\int k_x(t,s) \phi_J(s) ds = c_J \phi_J(t)}$$

(for some eigenfunctions)

But if

$\exists \psi$ such that \nexists

$$\int \phi_J^*(t) \psi(t) dt$$

Scalar product

$\psi(t)$ is orthogonal
to all the ϕ_J

i.e if the scalar product is zero then

$$\Rightarrow \psi \in (\text{ker } K_X)$$

i.e ψ transform K_X to zero

$$\Rightarrow \int k_x(t,s) \psi(s) ds = 0$$

~~$\int k_x(t,s) \psi(s) ds$~~

\therefore We can say some Eigenfunction might exist with Assumption-I

(ii)

Theorem about Eigenfunctions of

$k_x(t,s)$ (stronger)

if $k_x(t,s) = k_x^*(s,t)$

\Downarrow and

$$\iint |k_x(t,s)|^2 ds dt < +\infty$$

and

$k_x(s,t)$ is Strictly Positive Definite

Then ϕ_J and allows us to express any

process $x(t)$ s.t. $\int |x(t)|^2 dt < +\infty$



If $k_x(s, t)$ is strictly positive definite

Then Quadratic form

$$\forall g \neq 0 \quad \int \int k_x(t, s) g^*(t) g(s) dt ds > 0$$

In fact if $\exists \psi \quad s.t \quad \psi \perp \phi_J$
Then by the 1st condition

$$\int k_x(t, s) \psi(s) ds = 0$$

$$\Rightarrow \boxed{\int \psi^*(t) \int k_x(t, s) \psi(s) ds dt = 0}$$

We built a quadratic form above
by multiplying $\int \psi^*(t) dt$ to

$$\int k_x(t, s) \psi(s) ds = 0$$

$$\Rightarrow \int \int k_x(t, s) \psi^*(t) \psi(s) ds dt = 0$$

\Rightarrow But since $K_x(t,s)$ is positive definite
 $\Phi(t) = 0$ because nullvector is orthogonal to all the vectors

\rightarrow we are addressing:

- 1) every projector we lay down must be useful
- 2) Minimizing the No. of projectors, include only projectors that provide something new over

\rightarrow Even if the Theorem-2 provides everything we desire about Φ_J

Usually, the no. of Φ_J 's required to express something is infinite. and this is a Nonsense from an Engineering point of view because we cannot make any infinite Machine.

Therefore it would be beneficial for us to regard Φ_J 's which give us most of the information & discard others which only provide insignificant information.

Task: try to identify the MOST IMPORTANT

$$\phi_J^S$$

In general:

$$x(t) = \sum_{j=0}^n \alpha_j \phi_j^S(t)$$

↓ identify only useful
 ϕ_J^S

$$\mathcal{J} \subset \{0, 1, \dots\}$$

$$\hat{x}(t) = \sum_{j \in \mathcal{J}} \alpha_j \phi_j^S(t)$$

ex:
=

$$\mathcal{J} = \{0, 1, 2, 3\}$$

$$\hat{x}(t) = \sum_{j=0}^3 \alpha_j \phi_j^S(t)$$

$\hat{x}(t)$ would only be an estimate of $x(t)$

→ To know $\hat{x}(t)$ is good, The best way is do Error of Reconstruction.

$$\Sigma = E \left[\int |x(t) - \hat{x}(t)|^2 dt \right]$$

Since, we have waveforms the Error must be computed by an Integral.

$$= E \left[\int \left| \sum_{j=0}^{\infty} \alpha_j \phi_j(t) - \sum_{j \in J} \alpha_j \phi_j(t) \right|^2 dt \right]$$

$$\Sigma \Rightarrow E \left[\int \left| \sum_{j \notin J} \alpha_j \phi_j(t) \right|^2 \right]$$

$$\Rightarrow E \left[\sum_{j \notin J} \sum_{k \notin J} \alpha_j^* \alpha_k \phi_j^*(t) \phi_k(t) dt \right]$$

$$\Rightarrow E \left[\sum_{j \notin J} \sum_{k \notin J} \alpha_j^* \alpha_k \underbrace{\int \phi_j^*(t) \phi_k(t) dt}_{\text{brace}} \right]$$



$$J \neq k \rightarrow 0$$

$$J = k \rightarrow 1$$

$$\Rightarrow E \left[\sum_{J \notin J} \alpha_J^* \alpha_J \right]$$

$$\Rightarrow E \left[\sum_{J \notin J} (\alpha_J)^2 \right]$$

$$\varepsilon = \sum_{J \notin J} E[(\alpha_J)^2]$$

Exchanging Expectation with sum

~~*→~~ Usually projections are made zero-mean

Variance

$$\text{var} = E[(\alpha_J - \mu_{\alpha_J})^2] \quad ; \text{e} \quad \mu_{\alpha_J} = 0$$

$$\text{var} \Rightarrow E[(\alpha_J - \mu_{\alpha_J})^2] = E[(\alpha_J)^2]$$

$$\therefore \Sigma = \sum_{J \notin J} \sigma_J^2$$

recall, σ_J^2 are nothing but d_J

$$\Rightarrow \Sigma = \sum_{J \notin J} d_J$$

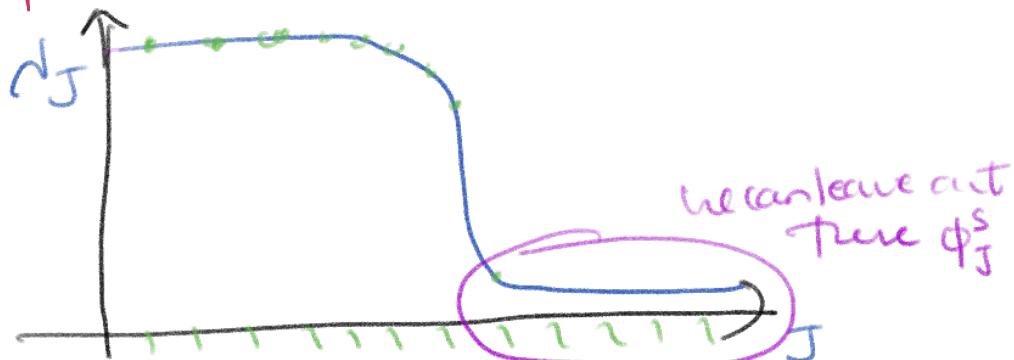
ERROR, is nothing but **Sum of Eigenvalues**

of ϕ_J that we are not taking.

Conclusion:

Avoid to take ϕ_J^s corresponding
to small d_J (Eigenvalues)

Empirical considerations



Assuming we have a strictly positive definite Covariance function.

→ The above consideration leads us to

Kahr vonen - Löve approximation

(as)

Compression

This is finishes part of the course where we are trying to characterize stochastic process



Stationarity, Ergodicity, Mixingness

All are related concepts. They are tough concepts but if you want to understand it from engineering application point of view and then develop the Math.

of all Engineering Courses

The aim is to build systems that process information, it may be Analog sys
Digital sys
Mp & MC

→ In the life time of a Device there are atleast 3 phases



- Design time, is the time which Engineers are mostly concerned.
- For sure design time should be as small as possible, it is disjoint from the working time.
- This is not true that the random signals we use during the Design time **sparks** may not be similar to what they might be subjected to the same in the real world during Working time

Regardless what random signals our system is subjected too it should work properly.



If the statistics that our System is subjected to in the '3' different time periods. its operation should remain good.

ex:

= But if it is a Mp which been designed for low power consumption.

* Then its power consumption would be larger in one of the above '3' regions. which is undesirable.

\therefore The statistical features ^{(inputs) of the process} which our System is designed to address are not constant then we would be in trouble.

* what prevents this from happening is an assumption that always we have in our processes which is STATIONARITY.

Stationarity:

It is much easier to define stationarity wrt to the Joint PDF characterization of stochastic process.