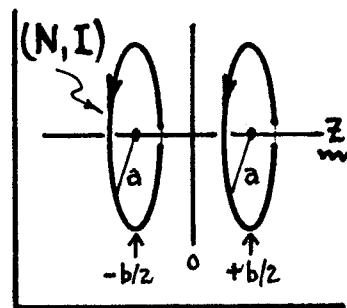


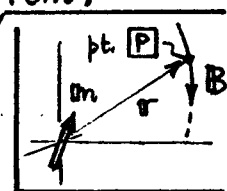
- 5) [15 pts]. A Helmholtz coil is a device to produce a relatively uniform B-field in a given volume. The coil consists of two identical circular loops, radius a with N turns carrying current I , sharing a common axis, and separated by distance b . Let the



Loop axis be the z -axis, and situate the loops at $z = \pm b/2$, as shown in sketch.

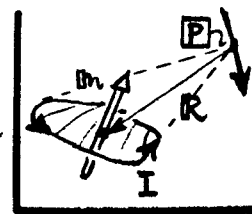
- (A) It is possible to choose b so that the axial field near the center of the coil (i.e. $z=0$) goes as: $B_z(z) = B_z(0)[1 - k(z/a)^4 + \dots]$. Find the condition on b which makes this so, and calculate the numerical constant k .
- (B) Find the radial field B_p close to the axis and near coil center for B_z of part (A).
- (C) Within a small volume at coil center, the total field $B = \sqrt{B_z^2 + B_p^2}$ is quite uniform. Picture this region as a cylinder, and find its dimensions and volume if -- inside the cylinder -- ΔB_z & ΔB_p are both $\leq \epsilon B_z(0)$, with $\epsilon \ll 1$. Calculate the coil radius a if $\epsilon = 1/10^4$ homogeneity is required over $\Delta z = 1 \text{ cm}$.

- 36) Consider a magnetic dipole field: $\mathbf{B} = \frac{1}{r^5} [3(\mathbf{m} \cdot \mathbf{r})\mathbf{r} - r^2 \mathbf{m}]$, at $r > 0$.



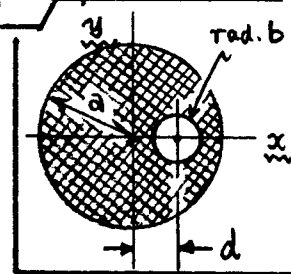
- (A) Show that $\mathbf{B} = -\nabla\psi$, where $\psi = \mathbf{m} \cdot \mathbf{r} / r^3$ is a scalar potential.

- (B) Suppose \mathbf{m} is generated by a current loop I as shown. Consistent with the dipole approximation ($R \gg$ loop size), show that the potential of part (A) is: $\psi = -(I/c)\Omega$, where Ω is the solid angle subtended by the loop at the field point P . Hence: $\mathbf{B} = (I/c)\nabla\Omega$, as quoted in Jkⁿ Prob.



- (5.1). HINT: $\Omega = \int \frac{1}{R^2} \hat{\mathbf{R}} \cdot d\mathbf{A}$, with the integral over the loop surface area.

- 37) [Jackson #5.3]. A cylindrical conductor of radius a has a hole of radius b bored \parallel to and at distance d from its axis ($d+b < a$). The current density is uniform and \parallel axis throughout the rest of the cylinder. If the conductor carries total current I , find the magnitude and direction of the B-field in the hole. You may assume $\nabla \cdot \mathbf{B} = 0$.



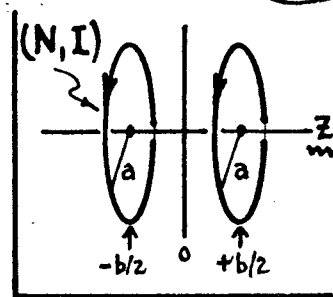
Φ 519 Solutions

S47

D [15 pts]. Analyse Helmholtz coil field for homogeneity.

(A) The double-loop configuration generates an axial field:

$$\rightarrow B_z(z) = \frac{2\pi NI}{ca} \left\{ [1+(\zeta-\beta)^2]^{-\frac{3}{2}} + [1+(\zeta+\beta)^2]^{-\frac{3}{2}} \right\}, \quad (1)$$



where $\zeta = z/a$ & $\beta = (b/2)/a$. Evidently $B_z(-z) = B_z(+z)$, so a Taylor expansion of $B_z(z)$ about $z=0$ will contain only even derivatives: $B_z(z) = B_z(0) + \frac{1}{2} z^2 B_z''(0) + \frac{1}{24} z^4 B_z^{IV}(0) + \dots$. The condition for a z^4 field is $B_z''(0) = 0$, i.e.

$$\rightarrow B_z''(0) = -\frac{3B_0}{a^2} \left\{ \frac{1-4(\zeta-\beta)^2}{[1+(\zeta-\beta)^2]^{7/2}} + \frac{1-4(\zeta+\beta)^2}{[1+(\zeta+\beta)^2]^{7/2}} \right\} \Big|_{z=0} = -\frac{6B_0}{a^2} \frac{1-4\beta^2}{(1+\beta^2)^{7/2}} = 0. \quad (2)$$

Here $B_0 = \frac{2\pi NI}{ca}$, and $B_z''(0) = 0$ if $\beta = b/2a = \frac{1}{2}$, i.e. if $\boxed{b=a}$. With this choice, the central field is $B_z(0) = (16/5^{3/2}) B_0$, and near $z=0$ we'll have

$$\rightarrow B_z(z) = B_z(0) [1 - k\zeta^4 + \dots], \text{ for } b=a, \text{ with } \zeta = z/a. \quad (3)$$

The const $k = (-) \frac{a^4}{24} B_z^{IV}(0) / B_z(0)$. After some arithmetic: $\boxed{k = 144/125}$.

(B) By the method used in Prob. 34: $\nabla \cdot \mathbf{B} = 0 \Rightarrow \frac{1}{\rho} \frac{\partial}{\partial \rho} (\rho B_\rho) \approx -(\partial B_z / \partial z) \Big|_{\rho=0}$, or:
 $B_\rho \approx -\frac{1}{2} \rho (\partial B_z / \partial z)_{\text{axis}}$, good to 1st order in ρ . Use Eq. (3) for $(\frac{\partial B_z}{\partial z})_{\text{axis}}$ to get:

$$\boxed{B_\rho \approx 2k B_z(0) \zeta^3 \rho / a}, \text{ for } \zeta = \frac{z}{a} \ll 1 \text{ \& } \rho \ll a. \quad (4)$$

(C) For a cylinder with ends at $\pm z$, the max. axial field change is: $\Delta B_z = k\zeta^4 B_z(0)$, so $\Delta B_z \leq \epsilon B_z(0)$ defines a cylinder of length: $L = 2a(\epsilon/k)^{1/4}$. From Eq. (4), the radial field change is max. at the ends also, and $\Delta B_\rho \leq \epsilon B_z(0)$ gives the cylinder radius: $\rho = \frac{a}{2}(\epsilon/k)^{1/4}$. So the "homogeneity" cylinder is...

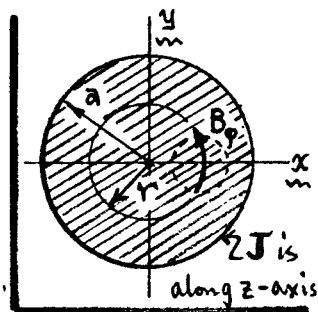
$$\boxed{\text{CYLINDER} \left\{ \begin{array}{l} \text{length: } L = 2a(\epsilon/k)^{1/4} \\ \text{radius: } \rho = \frac{1}{2}a(\epsilon/k)^{1/4} \end{array} \right\} \text{ volume: } V = \pi \rho^2 L = \frac{\pi}{2} a^3 (\epsilon/k)^{3/4}}. \quad (5)$$

If $\epsilon = 1/10^4$, $(\epsilon/k)^{1/4} = 0.09652$. And if $L \geq 1\text{cm}$, the coil radius must be $\underline{a \geq 5.18\text{ cm}}$. The cylinder $\rho = \frac{1}{4}L = 0.25\text{cm}$, and $V = 0.196\text{ cm}^3$.

7 B-field in a cylindrical hole in a cylindrical conductor.

1. First, consider the conductor without the hole (i.e. hole filled in).

At any $r \leq a$, the interior field will be in the ϕ (azimuthal) direction and of a size B_ϕ calculable by Ampere's Law, i.e.

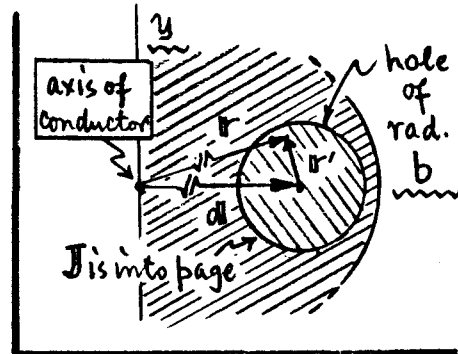


$$\rightarrow \nabla \times \mathbf{B} = \frac{4\pi}{c} \mathbf{J} \Rightarrow \oint_{\text{loop}} \mathbf{B} \cdot d\mathbf{l} = \frac{4\pi}{c} I(\text{enclosed});$$

So // circular loop of radius r } $B_\phi \cdot 2\pi r = \frac{4\pi}{c} J \cdot \pi r^2$, $\Rightarrow \mathbf{B} = \left(\frac{2\pi}{c} J \right) \hat{z} \times \mathbf{r}$. (1)

J is the const current density through the conductor, and \hat{z} a unit vector $\parallel \hat{z}$ -axis

2. Now, recreate the hole by adding an oppositely directed current density J to the hole region (hole of rad. b centered at $r=d$ on x -axis, etc.). Then the hole has net $J=0$, and the J_{opposite} generates a magnetic field of the form of Eq. (1), viz.



$$\mathbf{B}' = -\left(\frac{2\pi}{c} J \right) \hat{z} \times \mathbf{r}', \quad \mathbf{r}' \text{ from hole center.} \quad (2)$$

3. The net field in the hole is the superposition of \mathbf{B} [Eq. (1)] & \mathbf{B}' [Eq. (2)], i.e.

$$\rightarrow \mathbf{B}_{\text{hole}} = \mathbf{B} + \mathbf{B}' = \frac{2\pi J}{c} \hat{z} \times (\mathbf{r} - \mathbf{r}'). \quad (3)$$

But $\mathbf{r} - \mathbf{r}' = \mathbf{d} = \hat{x} d$, where d is the distance from conductor axis to hole axis. And $\hat{z} \times \hat{x} = \hat{y}$ lies along the y -axis, so that Eq. (3) becomes...

$$\rightarrow \mathbf{B}_{\text{hole}} = \left(\frac{2\pi J d}{c} \right) \hat{y}. \quad (4)$$

4. Finally (with hole present), conductor will be carrying current $I = J \cdot \pi(a^2 - b^2)$.

So the field in Eq. (4) is

$$\boxed{\mathbf{B}_{\text{hole}} = \left[\frac{2I d}{c} \left(\frac{1}{a^2 - b^2} \right) \right] \hat{y}}. \quad (5)$$

36) Achieve a representation of $\mathbf{B}_{\text{dipole}}$ via a scalar potential $\psi = \frac{\mathbf{m} \cdot \mathbf{r}}{r^3} = (-) \frac{I}{c} \Omega$.

(A) By the identity: $\nabla(\mathbf{a} \cdot \mathbf{b}) = (\mathbf{a} \cdot \nabla) \mathbf{b} + (\mathbf{b} \cdot \nabla) \mathbf{a} + \mathbf{a} \times (\nabla \times \mathbf{b}) + \mathbf{b} \times (\nabla \times \mathbf{a})$, with $\mathbf{a} = \mathbf{m} = \text{const}$, and $\mathbf{b} = \mathbf{r}/r^3$, we have...

$$\rightarrow \nabla(\mathbf{m} \cdot \mathbf{r}/r^3) = (\mathbf{m} \cdot \nabla) \frac{\mathbf{r}}{r^3} + 0 + \mathbf{m} \times (\nabla \times \frac{\mathbf{r}}{r^3}) + 0. \quad (1)$$

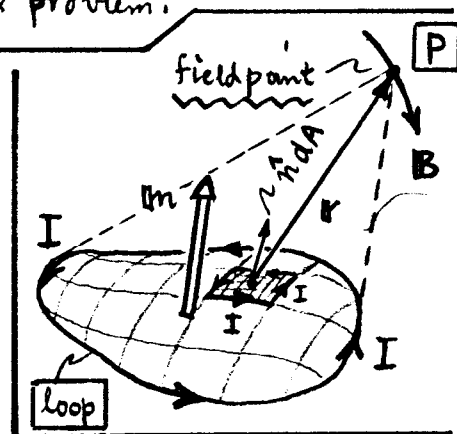
But: $\nabla \times (\mathbf{r}/r^3) = [\nabla(1/r^3)] \times \mathbf{r} + \frac{1}{r^3} (\nabla \times \mathbf{r}) = 0$, since $\nabla(1/r^3) = -\frac{3}{r^5} \mathbf{r}$ for $r \neq 0$, and the 1st term RHS vanishes because $\mathbf{r} \times \mathbf{r} = 0$. Then, $\psi = \mathbf{m} \cdot \mathbf{r}/r^3 \dots$

$$\rightarrow (-) \nabla \psi = -(\mathbf{m} \cdot \nabla) \frac{\mathbf{r}}{r^3} = \frac{1}{r^3} [3(\mathbf{m} \cdot \hat{\mathbf{r}}) \hat{\mathbf{r}} - \mathbf{m}] = \mathbf{B}_{\text{dipole}} \quad (2)$$

The details of $(\mathbf{m} \cdot \nabla) \mathbf{r}/r^3$ were worked out in class. Indeed the dipole field can be generated from a scalar potential: $\mathbf{B}_{\text{dipole}} = -\nabla \psi$, $\psi = (\mathbf{m} \cdot \mathbf{r})/r^3$.

This should not be astonishing, since $\nabla \times \mathbf{B} = 0$ for this problem.

(B) With \mathbf{m} generated by a loop current I , divide a surface through the loop into many small circuits by a mesh, as shown. Let each microloop carry current I ; then, because the currents cancel in the common branches of adjacent loops, the net effect of the microloops is the



same as that of the main loop carrying current I around its periphery only. Each microloop generates a moment: $d\mathbf{m} = \frac{I}{c} \hat{\mathbf{n}} dA$, where dA is the loop area and $\hat{\mathbf{n}}$ the local unit normal to the surface [cf. Jk^h Eg. (5.57)]. The total loop moment can then be represented as: $\mathbf{m} = \frac{I}{c} \int_{\text{loop}} \hat{\mathbf{n}} dA$.

At dipole distances, r changes negligibly as the vector $\mathbf{R} = (-) \mathbf{r}$, from pt. P to the loop ranges over the loop surface. The potential ψ of part (A) can be gotten from:

$$\rightarrow \psi = \int_{\text{loop}} (\mathbf{r} \cdot d\mathbf{m})/r^3 = -\frac{I}{c} \int_{\text{loop}} (\mathbf{R}/R^3) \cdot \hat{\mathbf{n}} dA = -\frac{I}{c} \Omega, \quad \Omega = \int_{\text{loop}} \frac{1}{R^2} \hat{\mathbf{R}} \cdot d\mathbf{A}. \quad (3)$$

Ω is the solid \angle subtended by the loop at the field point P . Then, as required, the field at P is: $\mathbf{B} = (I/c) \nabla \Omega$. Same result is quoted in Jk^h Prob. (5.1).