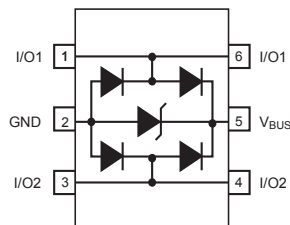
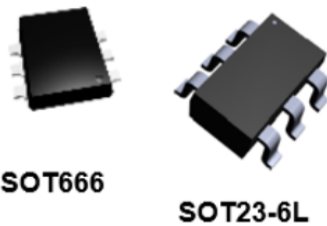


Very low capacitance ESD protection



Functional diagram (top view)

Features

- 2 data-line protection
- Protects VBUS
- Very low capacitance: 3.5 pF max.
- Very low leakage current: 150 nA max.
- SOT-666 and SOT23-6L packages
- RoHS compliant

Benefits

- Very low capacitance between lines to GND for optimized data integrity and speed
- Low PCB space consumption: 2.9 mm² max for SOT-666 and 9 mm² max for SOT23-6L
- Enhanced ESD protection: IEC 61000-4-2 level 4 compliance guaranteed at device level, hence greater immunity at system level
- ESD protection of VBUS
- High reliability offered by monolithic integration
- Low leakage current for longer operation of battery powered devices
- Fast response time
- Consistent D+ / D- signal balance:
 - Very low capacitance matching tolerance I/O to GND = 0.015 pF
 - Compliant with USB 2.0 requirements

Complies with the following standards:

- IEC 61000-4-2 level 4:
 - 15 kV (air discharge)
 - 8 kV (contact discharge)

Applications

- USB 2.0 ports up to 480 Mb/s (high speed)
- Compatible with USB 1.1 low and full speed
- Ethernet port: 10/100 Mb/s
- SIM card protection
- Video line protection
- Portable electronics

Description

The USBLC6-2SC6 and USBLC6-2P6 are monolithic application specific devices dedicated to ESD protection of high speed interfaces, such as USB 2.0, Ethernet links and video lines.

The very low line capacitance secures a high level of signal integrity without compromising in protecting sensitive chips against the most stringently characterized ESD strikes.

Product status link

[USBLC6-2](#)

1 Characteristics

Table 1. Absolute ratings ($T_{amb} = 25\text{ }^{\circ}\text{C}$)

Symbol	Parameter		Value	Unit
V_{PP}	Peak pulse voltage	IEC 61000-4-2 level 4 standard:		
		Air discharge	15	kV
		Contact discharge	15	
		MIL STD883G-Method 3015-7	25	
T_{stg}	Storage temperature range		-55 to +150	$^{\circ}\text{C}$
T_j	Operating junction temperature range		-40 to +150	$^{\circ}\text{C}$
T_L	Maximum lead temperature for soldering during 10 s at 5 mm		260	$^{\circ}\text{C}$

Table 2. Electrical characteristics ($T_{amb} = 25\text{ }^{\circ}\text{C}$)

Symbol	Parameter	Test conditions	Value			Unit
			Min.	Typ.	Max.	
I_{RM}	Leakage current	$V_{RM} = 5.25\text{ V}$		10	150	nA
V_{BR}	Breakdown voltage between V_{BUS} and GND	$I_R = 1\text{ mA}$	6			V
V_F	Forward voltage	$I_F = 10\text{ mA}$			1.1	V
V_{CL}	Clamping voltage	$I_{PP} = 1\text{ A}$, 8/20 μs Any I/O pin to GND			12	V
		$I_{PP} = 5\text{ A}$, 8/20 μs Any I/O pin to GND			17	
$C_{i/o-GND}$	Capacitance between I/O and GND	$V_R = 1.65\text{ V}$		2.5	3.5	pF
$\Delta C_{i/o-GND}$				0.015		
$C_{i/o-i/o}$	Capacitance between I/O	$V_R = 1.65\text{ V}$		1.2	1.7	pF
$\Delta C_{i/o-i/o}$				0.04		

1.1 Characteristics (curves)

Figure 1. Capacitance versus voltage (typical values)

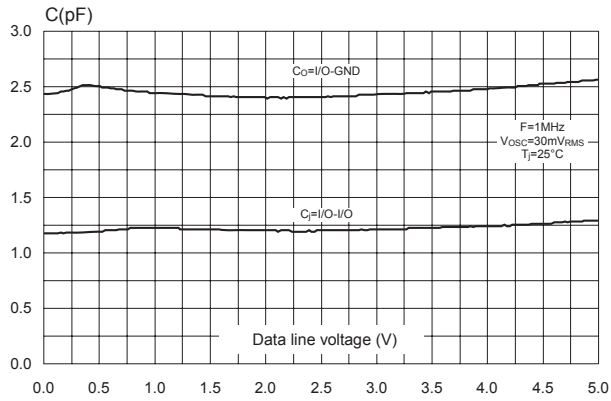


Figure 2. Line capacitance versus frequency (typical values)

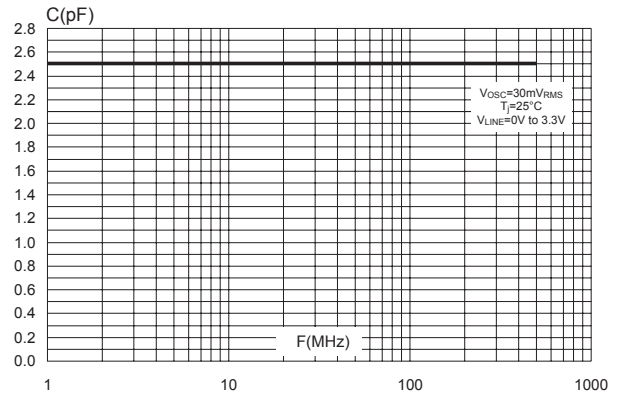


Figure 3. Relative variation of leakage current versus junction temperature (typical values)

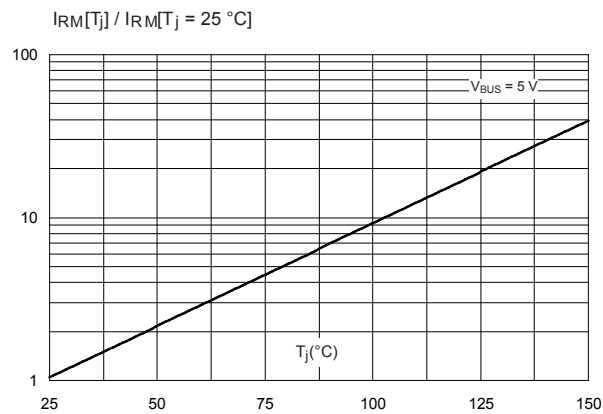
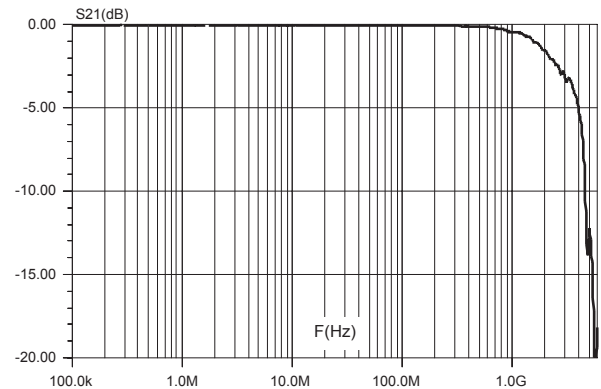


Figure 4. Frequency response



2 Technical information

2.1 Surge protection

The USBLC6-2 is particularly optimized to perform surge protection based on the rail to rail topology.

The clamping voltage V_{CL} can be calculated as follow:

$V_{CL+} = V_{TRANSIL} + V_F$ for positive surges

$V_{CL-} = -V_F$ for negative surges

with: $V_F = V_T + R_d \cdot I_p$

(V_F forward drop voltage) / (V_T forward drop threshold voltage)

and $V_{TRANSIL} = V_{BR} + R_{d_TRANSIL} \cdot I_p$

Calculation example

We assume that the value of the dynamic resistance of the clamping diode is typically:

$R_d = 0.5 \Omega$ and $V_T = 1.1 V$

We assume that the value of the dynamic resistance of the transil diode is typically:

$R_{d_TRANSIL} = 0.5 \Omega$ and $V_{BR} = 6.1 V$ For an IEC 61000-4-2 surge level 4 (Contact Discharge: $V_g = 8 kV$, $R_g = 330 \Omega$), $V_{BUS} = +5 V$, and if in first approximation, we assume that:

$I_p = V_g / R_g = 24 A$.

So, we find:

$V_{CL+} = +31.2 V$

$V_{CL-} = -13 V$

Note: The calculations do not take into account phenomena due to parasitic inductances.

2.2 Surge protection application example

If we consider that the connections from the pin V_{BUS} to V_{CC} , from I/O to data line and from GND to PCB GND plane are done by tracks of 10 mm long and 0.5 mm large, we assume that the parasitic inductances L_{VBUS} , $L_{I/O}$ and L_{GND} of these tracks are about 6 nH. So when an IEC 61000-4-2 surge occurs on data line, due to the rise time of this spike ($t_r = 1 ns$), the voltage V_{CL} has an extra value equal to $L_{I/O} \cdot dI/dt + L_{GND} \cdot dI/dt$.

The dI/dt is calculated as:

$dI/dt = I_p / t_r = 24 A/ns$

The overvoltage due to the parasitic inductances is:

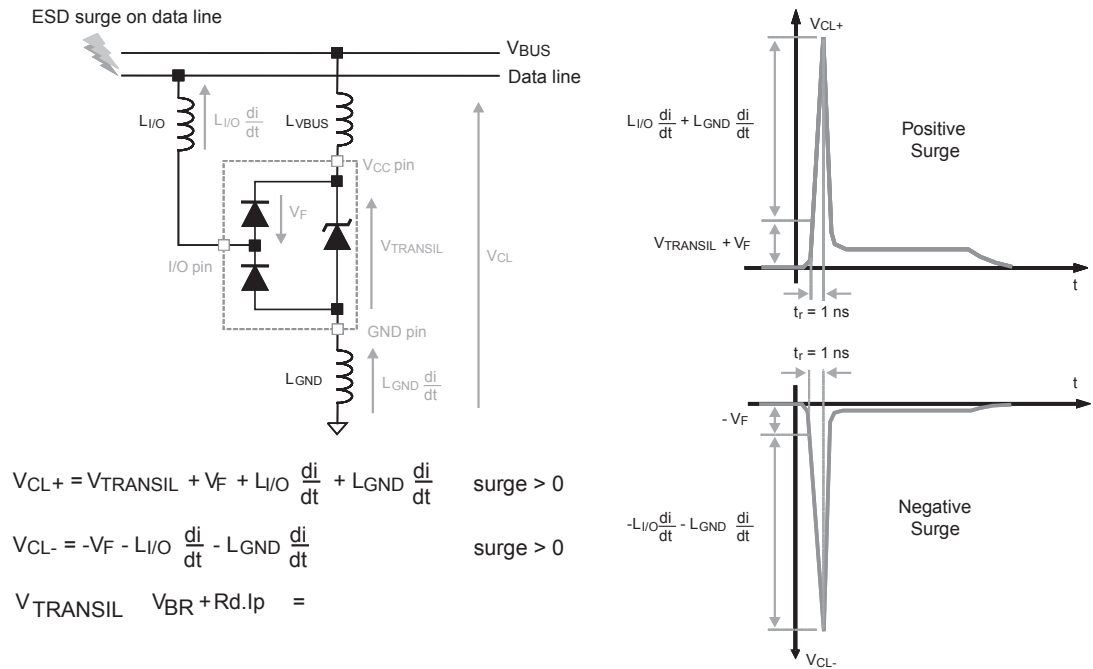
$L_{I/O} \cdot dI/dt = L_{GND} \cdot dI/dt = 6 nH \times 24 A/ns = 144 V$

By taking into account the effect of these parasitic inductances due to unsuitable layout, the clamping voltage will be:

$V_{CL+} = +31.2 + 144 + 144 = 319.2 V$

$V_{CL-} = -13.1 - 144 - 144 = -301.1 V$

We can significantly reduce this phenomena with simple layout optimization. It is for this reason that some recommendations have to be followed (see).

Figure 5. ESD behavior: parasitic phenomena due to unsuitable layout


2.3 How to ensure good ESD protection

While the USBLC6-2 provides high immunity to ESD surge, efficient protection depends on the layout of the board. In the same way, with the rail to rail topology, the track from data lines to I/O pins, from VCC to VBUS pin and from GND plane to GND pin must be as short as possible to avoid overvoltages due to parasitic phenomena (see Figure 6. ESD behavior: layout optimization and Figure 5. ESD behavior: parasitic phenomena due to unsuitable layout for layout consideration).

Figure 6. ESD behavior: layout optimization

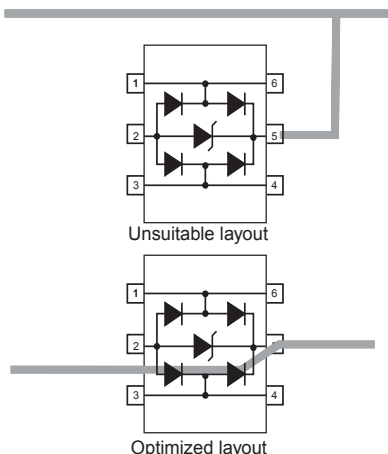


Figure 7. ESD behavior: measurement conditions

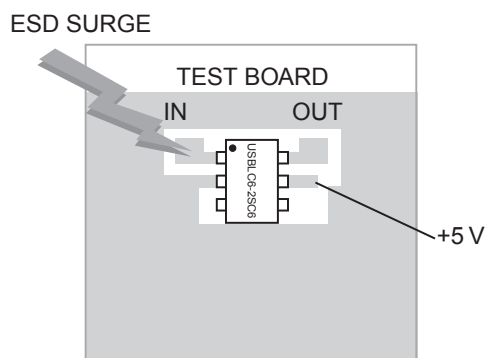


Figure 8. ESD response to IEC 61000-4-2 (+15 kV air discharge)

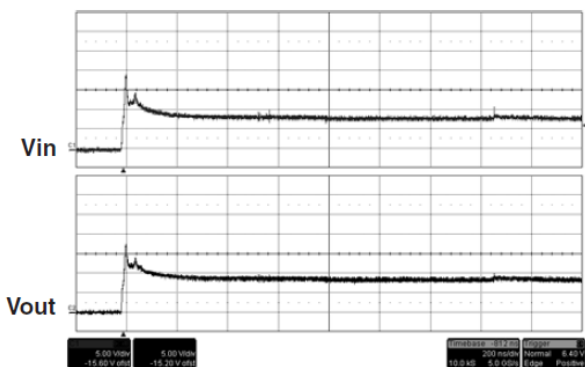
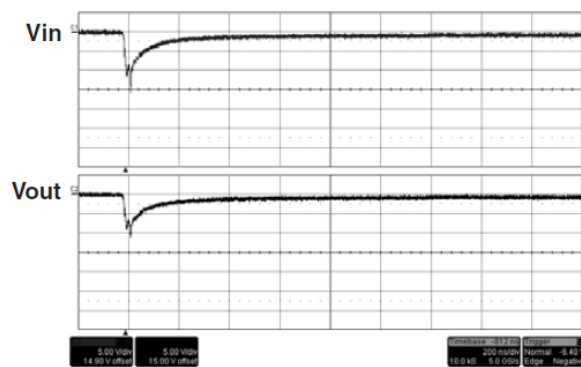


Figure 9. ESD response to IEC 61000-4-2 (-15 kV air discharge)



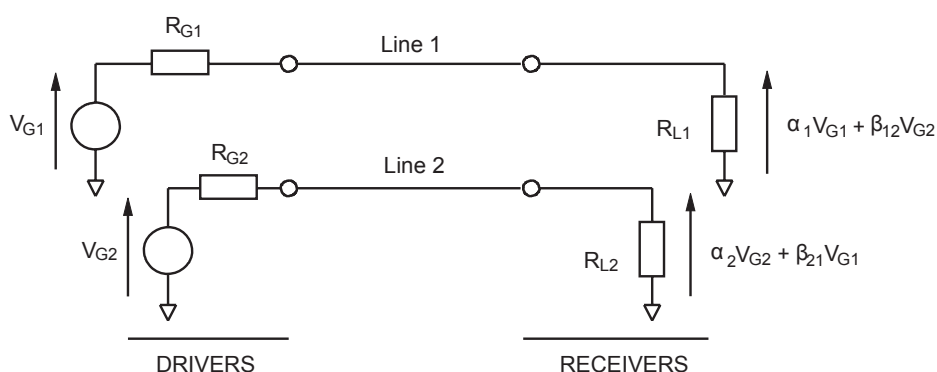
Note:

Important: A good precaution to take is to put the protection device as close as possible to the disturbance source (generally the connector).

2.4 Crosstalk behavior

2.4.1 Crosstalk phenomenon

Figure 10. Crosstalk phenomenon



The crosstalk phenomenon is due to the coupling between 2 lines. The coupling factor (β_{12} or β_{21}) increases when the gap across lines decreases, particularly in silicon dice. In the above example the expected signal on load R_{L2} is $\alpha_2 V_{G2}$, in fact the real voltage at this point has got an extra value $\beta_{21} V_{G1}$. This part of the V_{G1} signal represents the effect of the crosstalk phenomenon of the line 1 on the line 2. This phenomenon has to be taken into account when the drivers impose fast digital data or high frequency analog signals in the disturbing line. The perturbed line will be more affected if it works with low voltage signal or high load impedance (few k Ω).

Figure 11. Analog crosstalk measurements

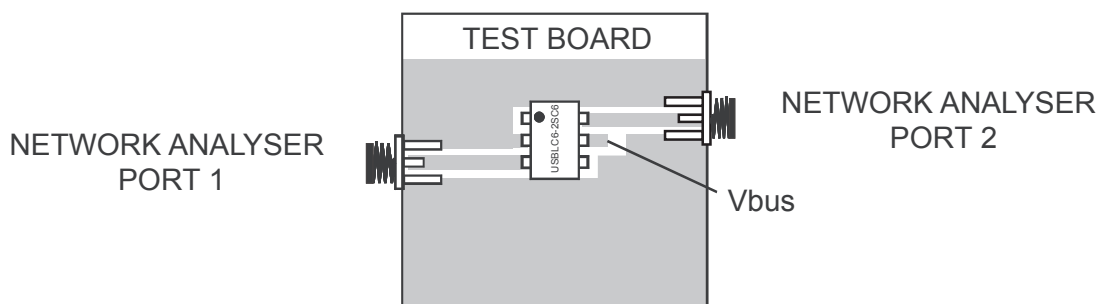
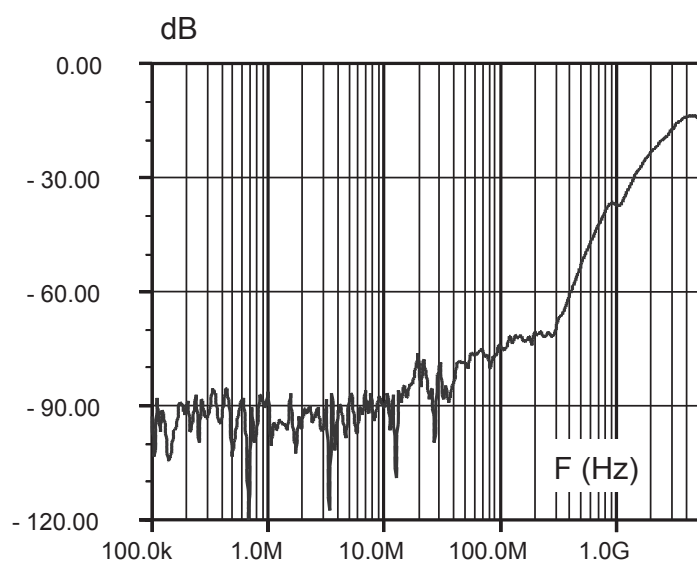


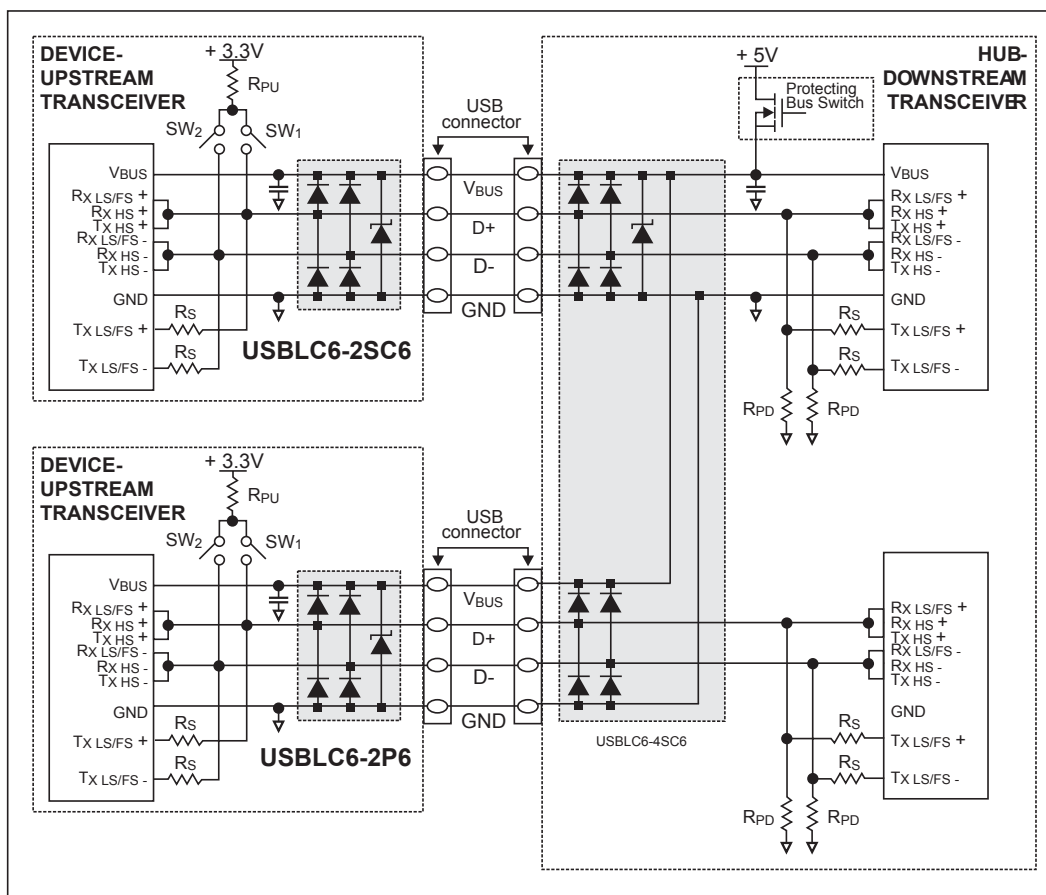
Figure 11. Analog crosstalk measurements shows the measurement circuit for the analog application. In usual frequency range of analog signals (up to 240 MHz) the effect on disturbed line is less than -55 dB (see Figure 12. Analog crosstalk results).

Figure 12. Analog crosstalk results


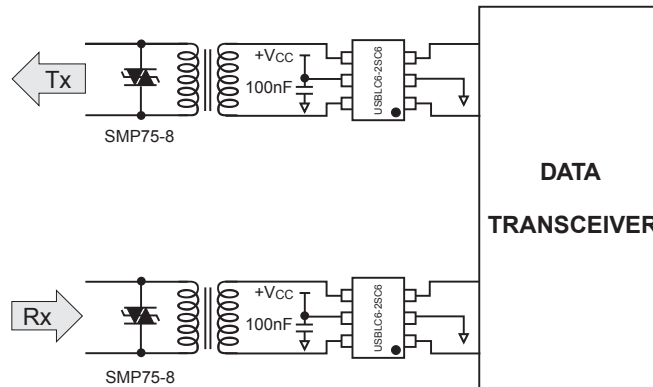
As the USBLC6-2 is designed to protect high speed data lines, it must ensure a good transmission of operating signals. The frequency response (Figure 4. Frequency response) gives attenuation information and shows that the USBLC6-2 is well suitable for data line transmission up to 480 Mbit/s while it works as a filter for undesirable signals like GSM (900 MHz) frequencies, for instance.

2.5 Application examples

Figure 13. USB 2.0 port application diagram using USBLC6-2

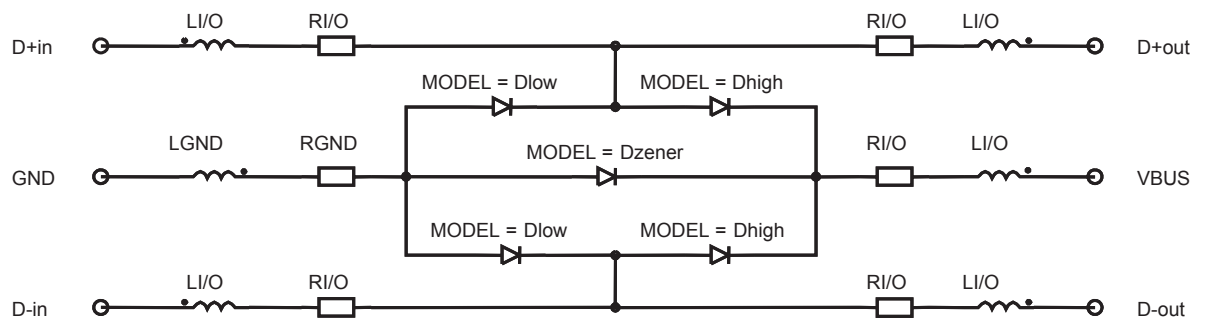


Mode	SW1	SW2
Low Speed LS	Open	Closed
Full Speed FS	Closed	Open
High Speed HS	Closed then open	Open

Figure 14. T1/E1/Ethernet protection


2.6 PSpice model

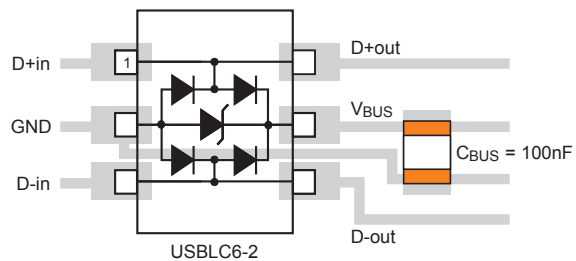
Figure 15. PSpice model shows the PSpice model of one USBLC6-2 cell. In this model, the diodes are defined by the PSpice parameters given in Figure 16. PSpice parameters.

Figure 15. PSpice model


Note: This simulation model is available only for an ambient temperature of 27 °C.

Figure 16. PSpice parameters

	Dlow	Dhigh	Dzener	LI/O	750p
BV	50	50	7.3	RI/O	110m
CJ0	0.9p	2.0p	40p	LGND	550p
IBV	1m	1m	1m	RGND	60m
M	0.3333	0.3333	0.3333		
RS	0.2	0.52	0.84		
VJ	0.6	0.6	0.6		
TT	0.1u	0.1u	0.1u		

Figure 17. USBLC6-2 PCB layout considerations


3 Package information

In order to meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

3.1 SOT23-6L package information

Figure 18. SOT23-6L package outline

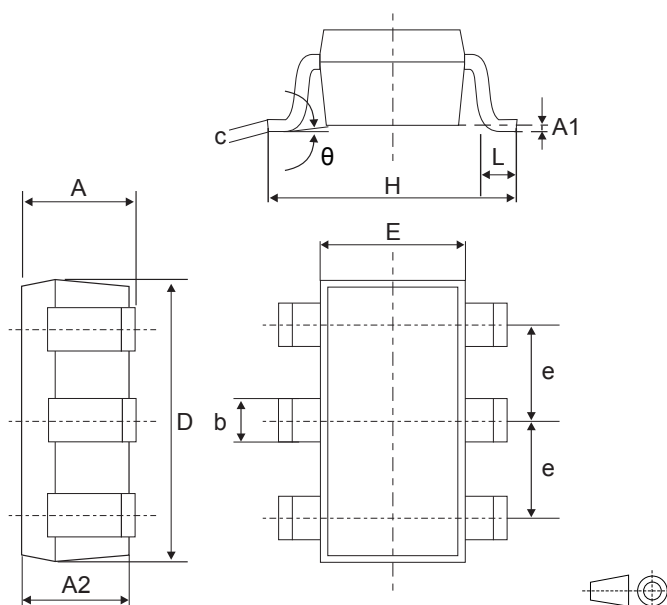
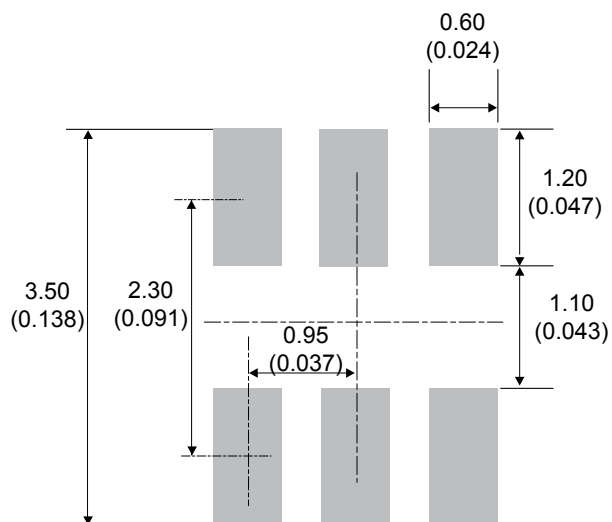


Table 3. SOT23-6L package mechanical data

Ref.	Dimensions					
	Millimeters			Inches ⁽¹⁾		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	0.9		1.45	0.0354		0.0571
A1	0		0.15	0		0.0059
A2	0.9		1.3	0.0354		0.0512
b	0.30		0.5	0.0118		0.0197
c	0.09		0.2	0.0035		0.0079
D	2.8		3.05	0.1102		0.1201
E	1.5		1.75	0.0591		0.0689
e		0.95			0.0374	
H	2.6		3	0.1024		0.1181
L	0.3		0.6	0.0118		0.0236
θ	0		10	0		0.3937

1. Value in inches are converted from mm and rounded to 4 decimal digits

Figure 19. Footprint recommendations, dimensions in mm (inches)



3.2 SOT-666 package information

Figure 20. SOT-666 package outline

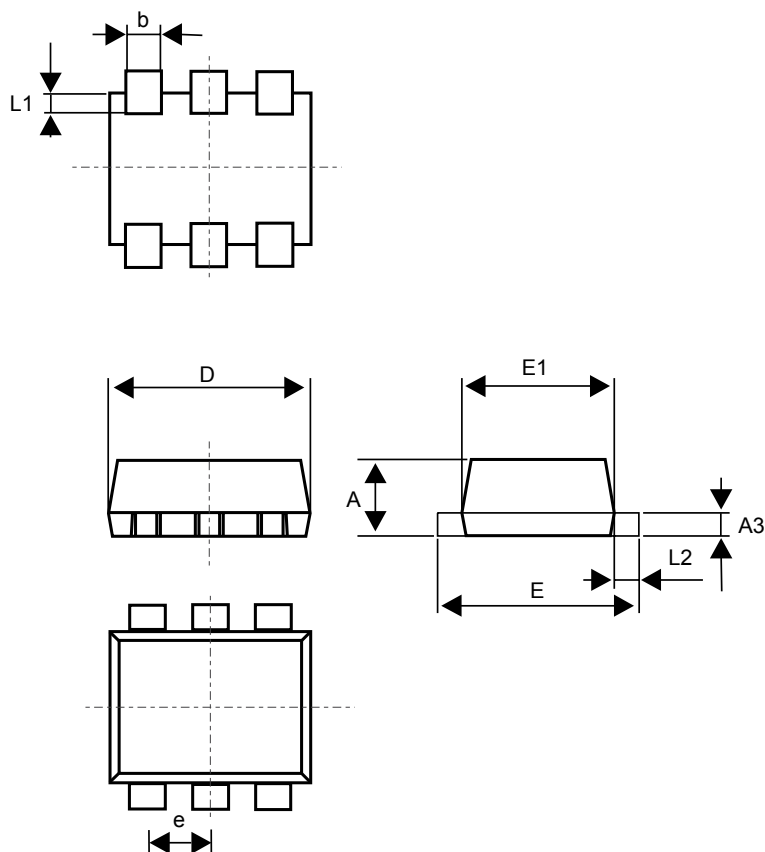
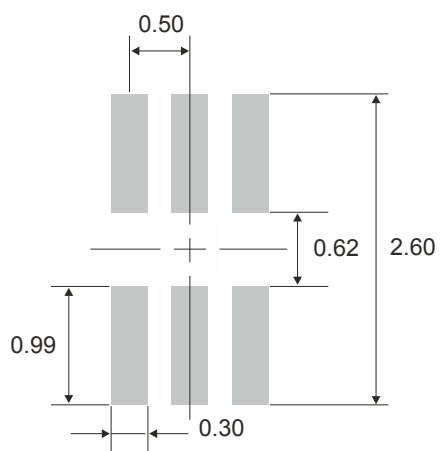


Table 4. SOT-666 package mechanical data

Ref.	Dimensions					
	Millimeters			Inches ⁽¹⁾		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	0.45		0.62	0.018		0.025
A3	0.08		0.18	0.003		0.007
b	0.17		0.34	0.007		0.013
D	1.50		1.70	0.059		0.067
E	1.50		1.70	0.059		0.067
E1	1.10		1.30	0.043		0.051
e		0.50			0.020	
L1		0.19			0.007	
L2	0.10		0.30	0.004		0.012

1. Value in inches are converted from mm and rounded to 4 decimal digits

Figure 21. Footprint recommendations, dimensions in mm



3.3 Packing information

Figure 22. Marking layout (refer to ordering information table for marking)

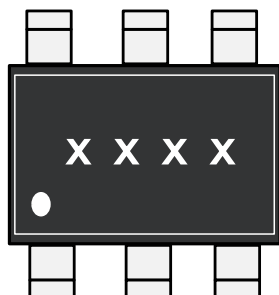
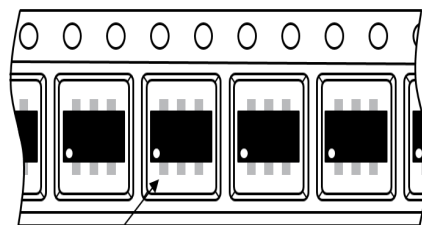


Figure 23. Package orientation in reel



Note: Pocket dimensions are not on scale
Pocket shape may vary depending on package

Figure 24. Tape and reel orientation

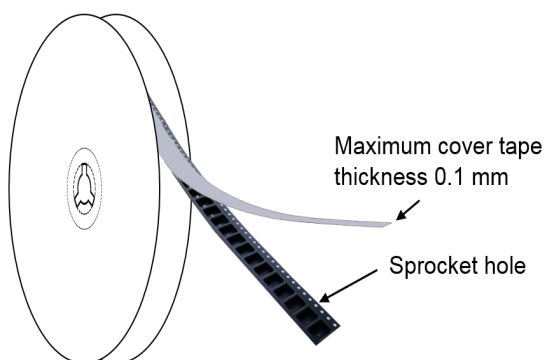


Figure 25. Reel dimensions (mm)

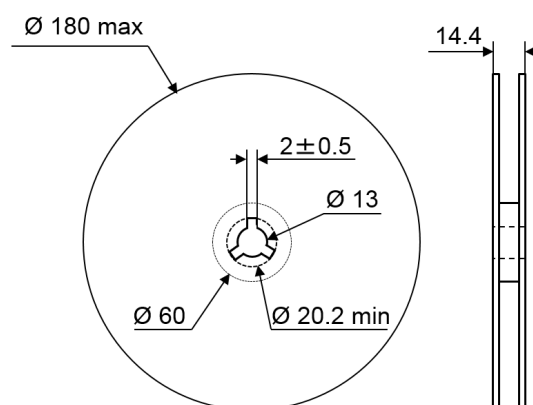


Figure 26. Inner box dimensions (mm)

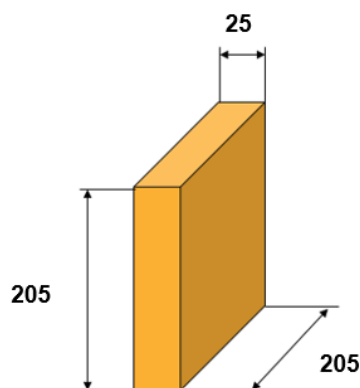
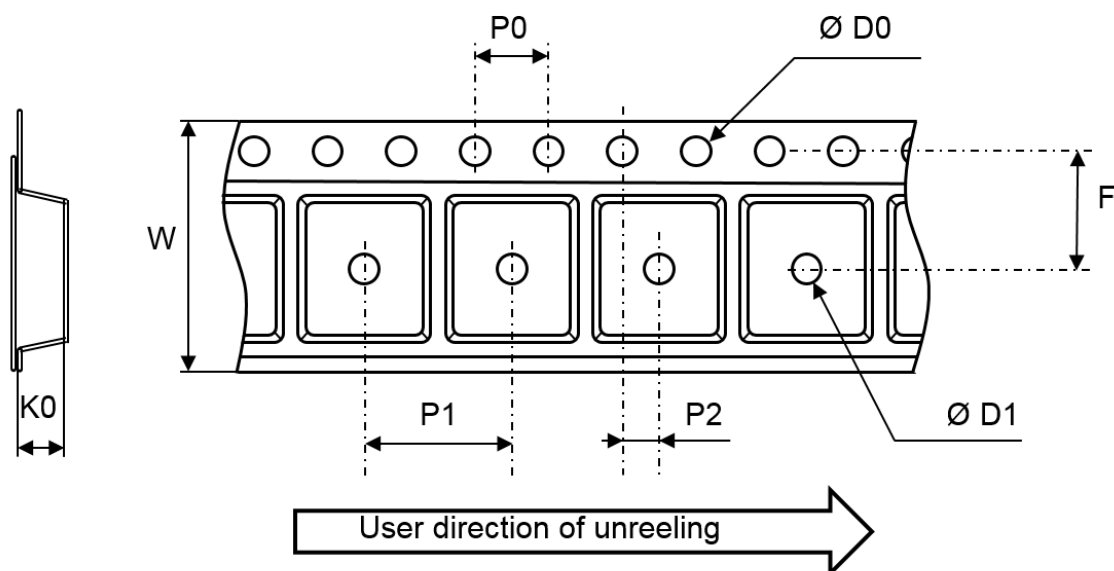


Figure 27. Tape and reel outline


Note: Pocket dimensions are not on scale
Pocket shape may vary depending on package

Table 5. Tape and reel mechanical data

Ref.	Dimensions		
	Millimeters		
	Min.	Typ.	Max.
P1	3.9	4	4.1
P0	3.9	4	4.1
D0	1.45	1.5	1.6
D1	1		
F	3.45	3.5	3.55
K0	1.3	1.4	1.6
P2	1.95	2	2.05
W	7.9	8	8.3

4 Ordering information

Figure 28. Ordering information scheme

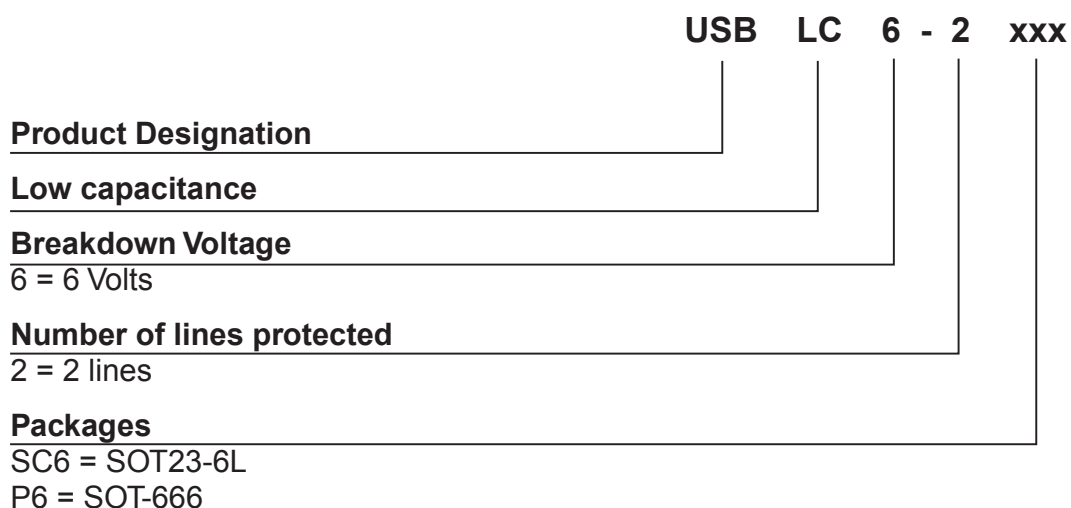


Table 6. Ordering information

Order code	Marking	Package	Weight	Base qty.	Delivery mode
USBLC6-2SC6 ⁽¹⁾	UL26	SOT23-6L	16.7 mg	3000	Tape and reel
USBLC6-2P6 ⁽¹⁾	F	SOT-666	2.9 mg	3000	Tape and reel

1. The marking code can be rotated by 90° to differentiate assembly location.

Revision history

Table 7. Document revision history

Date	Version	Changes
14-Mar-2005	1	Initial release.
07-Jun-2005	2	Format change to figure 3; no content changed.
20-Mar-2008	3	Added marking illustrations - Figures 21 and 23. Added ECOPACK statement. Updated operating junction temperature range in absolute ratings, page 2. Technical information section updated. Reformatted to current standards.
27-Jun-2011	4	Updated leakage current for VRM = 5.25 V as specified in USB standard. Updated marking illustrations Figure 21 and Figure 23.
24-Oct-2011	5	Updated legal statement.
16-Oct-2020	6	Minor text changes.
24-Dec-2021	7	Updated Section ■ Disclaimer Updated Section 1 Characteristics

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