

# CET 141: Day 6

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# Objectives (learning points)

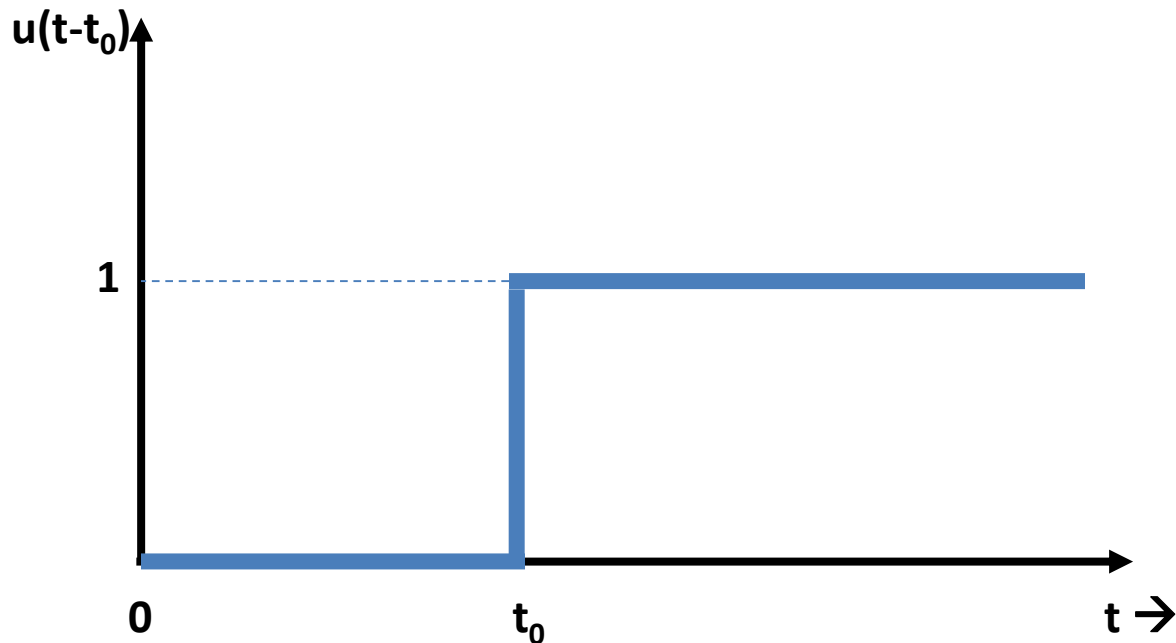
1. Characteristics of RL and RC circuits when there is an on-off switch
  - Analyze when the switch becomes on  $\leftrightarrow$  off
  - Define a step function to simulate the on-off switch behavior mathematically
2. Definition of “Response” with several concepts
  - i.e., complete, natural, forced....

# Agenda

1. The unit step function
2. Initial conditions of switched circuits
3. First-order (1<sup>st</sup> order) circuits
4. Stability of the 1<sup>st</sup> order circuit
5. Time constant  $\tau$
6. transient R. and steady-state R. for 1<sup>st</sup>ord. circuits
7. Simple RC, RL circuit analysis (1<sup>st</sup> ord.)
8. Examples of RL, RC circuits

# 1. The unit step function

➔ To simulate changes of signal status easily (switching)



$$u(t - t_0) = \begin{cases} 0 & t < t_0 \\ 1 & t > t_0 \end{cases}$$

Normally we define  $t_0=0$

## 2. Initial conditions of switched circuits

- Initial condition was given at  $t_0$
- A switch was **on at  $t_0$**
- A status **before** switching ( $t < t_0$ ): steady-state

	Cap response	Ind response
$i(t_0)$	Discontinuous $i_c(t_0^+) \neq i_c(t_0^-)$	Continuous $i_L(t_0^+) = i_L(t_0^-)$ $v_L(t) = L \frac{d}{dt} i_L(t)$
$v(t_0)$	Continuous $v_c(t_0^+) = v_c(t_0^-)$ $i_c(t) = C \frac{d}{dt} v_c(t)$	Discontinuous $v_L(t_0^+) \neq v_L(t_0^-)$

Must be differentiable

- At DC ( $f=0$ ,  $t=\infty$ ): also steady-state
  - An inductor acts as like a short-circuit (connected wire)
  - A cap acts as like a open-circuit (disconnected wire)

$$v_L(t) = L \frac{d}{dt} i_L(t)$$

If  $i_L(t)$  = a Constant (DC)  
 $\rightarrow V_L(t)=0 \rightarrow$  S.C.

$$i_c(t) = C \frac{d}{dt} v_c(t)$$

If  $v_c(t)$  = a Constant (DC)  
 $\rightarrow i_c(t)=0 \rightarrow$  O.C.

### 3. 1<sup>st</sup> order circuits

- Def: Only one capacitor or one inductor circuits
  - Not an absolute number of C or L
  - Simplification by Thevenin or Norton
  - Therefore, kinds of components in the circuit defines RC or RL circuits
- Responses: **Circuit outputs**



Complete response = Natural R. + Forced R.



Homogeneous  
sol.  
(when  $f=0$ )



Particular sol.

# Recap: 1<sup>st</sup> order diff eq. solution w/ a constant f(t)

$$x' + \frac{x}{\tau} = K$$

$$x' = \frac{d}{dt}x \quad \begin{array}{l} x \text{ is a function of time} \rightarrow x(t) \\ \tau \text{ is a time constant} \end{array}$$

$$\frac{dx}{dt} = \frac{K\tau - x}{\tau} \quad \longleftrightarrow \quad \frac{dx}{x - K\tau} = -\frac{dt}{\tau}$$

$$\int \frac{dx}{x - K\tau} = \int -\frac{dt}{\tau} \quad \longrightarrow \quad \ln(x - K\tau) = -\frac{t}{\tau} + \text{Const.}$$

$$e^{-t/\tau + \text{Const.}} = x - K\tau$$

$$x \text{ is a function of time} \rightarrow x(t)$$

$$x(t) = K\tau + A \cdot e^{-t/\tau}$$

$$A = e^{\text{Const.}} \rightarrow \text{a constant}$$

Forced R.  
Particular sol

Natural R.  
Homogeneous sol

## 4. Stability of the 1<sup>st</sup> order circuit

(by investigating **the natural response**)

Homogeneous portion of the answer

$$A \cdot e^{-t/\tau}$$

As  $t \rightarrow \infty$ , if  $\tau > 0$ , this term converges to 0

$\rightarrow$  before  $t < \infty$ , this term makes circuits unstable

$A \cdot e^{-t/\tau}$	When $\tau > 0$ , $R \neq 0$	When $\tau < 0$
$R_{Th} > 0$ is required to make 1 <sup>st</sup> order circuit stable	$t \rightarrow 0$ Stable (the natural R disappears $\rightarrow$ only forced R remains)	Unstable (without $t \rightarrow 0$ ) (the natural R grows $\rightarrow$ relatively small forced R.)

## <Summary of stability>

		t=0	t=τ	t=∞
<b>x(t)</b> 1 <sup>st</sup> PDE	F.R.: $K\tau$	0%	63.2%	100%
	N.R.: $Ae^{-t/\tau}$	100%	36.7%	0%

Unstable  Stable

Then what is the  $\tau$  in the circuit responses?

## 5. The time constant $\tau$

- Def: a special time which represents the speed of a particular system
- Properties:
  - Can respond to **CHANGE**: typically a factor of  $e^{-1}$  (i.e.,  $1 - \frac{1}{e} = 0.6321 \dots$ )
  - $5\tau$  is regarded as its maximum **CHANGE**

- Kinds of  $\tau$

- RC time constant: for RC circuit

- $\tau_{RC} = RC$  (where  $R = R_{th}$ )
- The RC unit:  $\Omega F = \frac{V}{A} \frac{Col}{V} = \frac{sec}{Col} Col = sec$
- Time required to charge a cap (through R) by 63.2% or discharge the cap by 36.8%

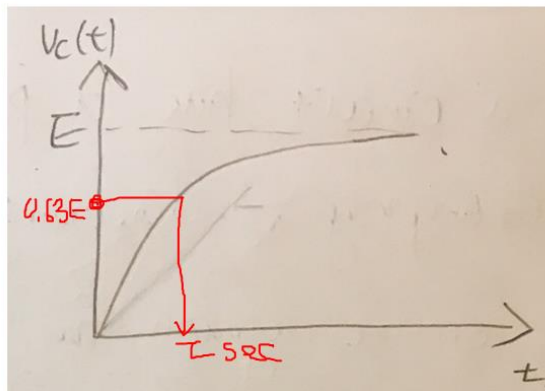
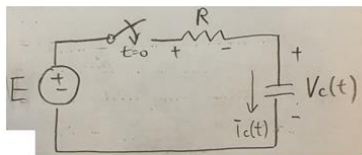
When  $t = \tau = RC$

$t = \tau$  a special case to define the time constant

$$v_c(t) = E(1 - e^{-t/RC})$$

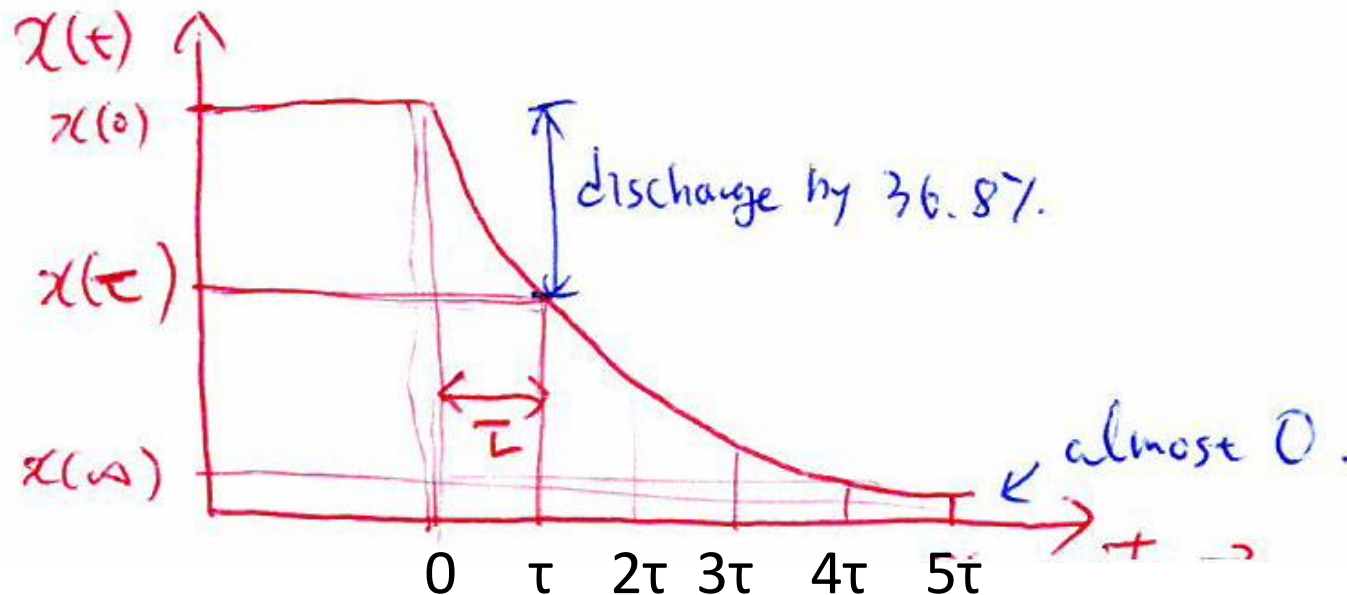
$$v_c(RC) = E(1 - e^{-RC/RC}) = E(1 - e^{-1}) = 0.63E$$

0.3678....



– RL time constant: for RL circuit

- $\tau_{RL} = L/R$  (where  $R = R_{th}$ )
- The  $L/R$  unit:  $\frac{H}{\Omega} = \frac{V \cdot sec}{A} * \frac{A}{V} = sec$
- Time required to energize an Ind (through R) by 63.2% or de-energize the Ind by 36.8%
- After  $5\tau \rightarrow$  considered as 0%





# So far, we've talked about..

1. The unit step function
2. Initial conditions of switched circuits
3. First-order circuits
4. Stability of the 1<sup>st</sup> order circuit
5. Time constant  $\tau$

Going back to our original business,  
let's talk about the circuit again.

6. 1<sup>st</sup>ord. circuit: transient R. and steady-state R.
7. Simple RC, RL circuit analysis
8. Examples of RL, RC circuits

In terms of  
PDE solutions

$$\begin{aligned}\text{Complete response} &= \text{Natural R.} + \text{Forced R.} \\ &= \underset{f(t)=0}{\text{Homog. S.}} + \underset{f(t)=\text{something}}{\text{Partic. S.}}\end{aligned}$$

$$= \underline{\text{Transient R.}} + \underline{\text{Steady-state R.}}$$

Dies out eventually  
when  $t \rightarrow \infty$

Dominant response  
when  $t \rightarrow \infty$

When we input signals for example

- A sinusoidal function or
- A constant (DC) signal

To circuits, we have

Complete response = Transient R. + Steady-state R

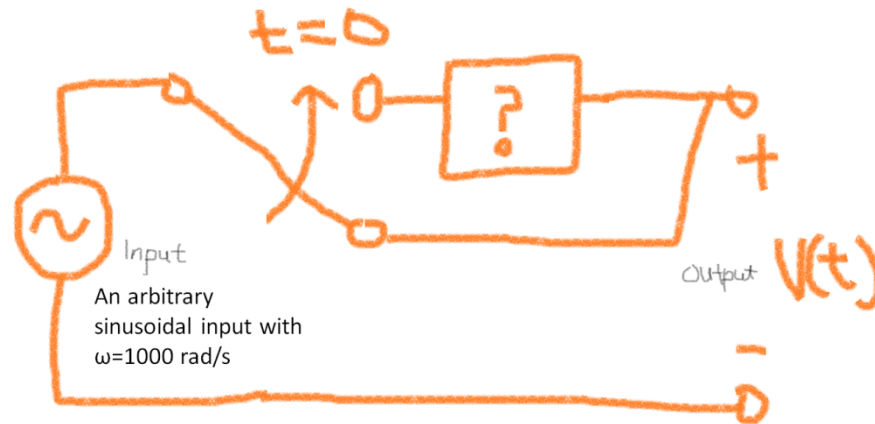
- Existing momentarily due to changing in the circuit (i.e.,  $0 \rightarrow 1$  or  $1 \rightarrow 0$ )
- Becomes 0 when  $t \rightarrow \infty$

- Stable response
- i.e., before switching /after switching

Note that when the circuit is 1<sup>st</sup> order, usually,  
Transient.R= Natural.R, and Steady-state.R=Forced. R, but not always....

## 6. The 1<sup>st</sup> order circuit: T. R. and SS R.

Complete R. = Transient R. + Steady-state R

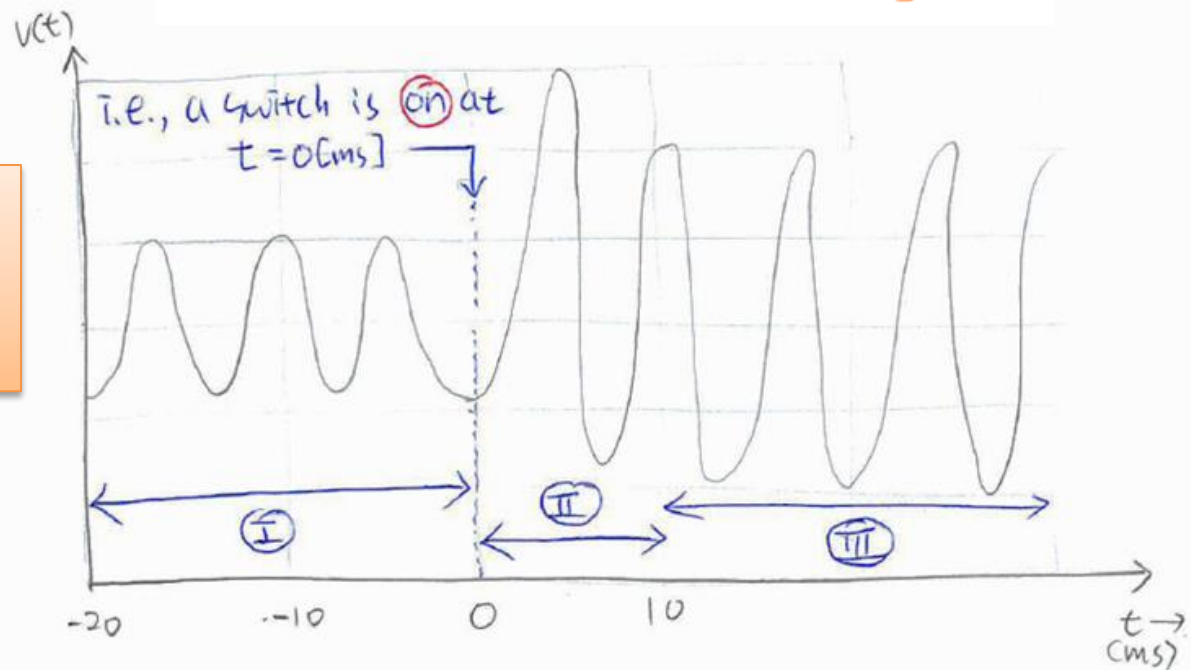


*Before switching*

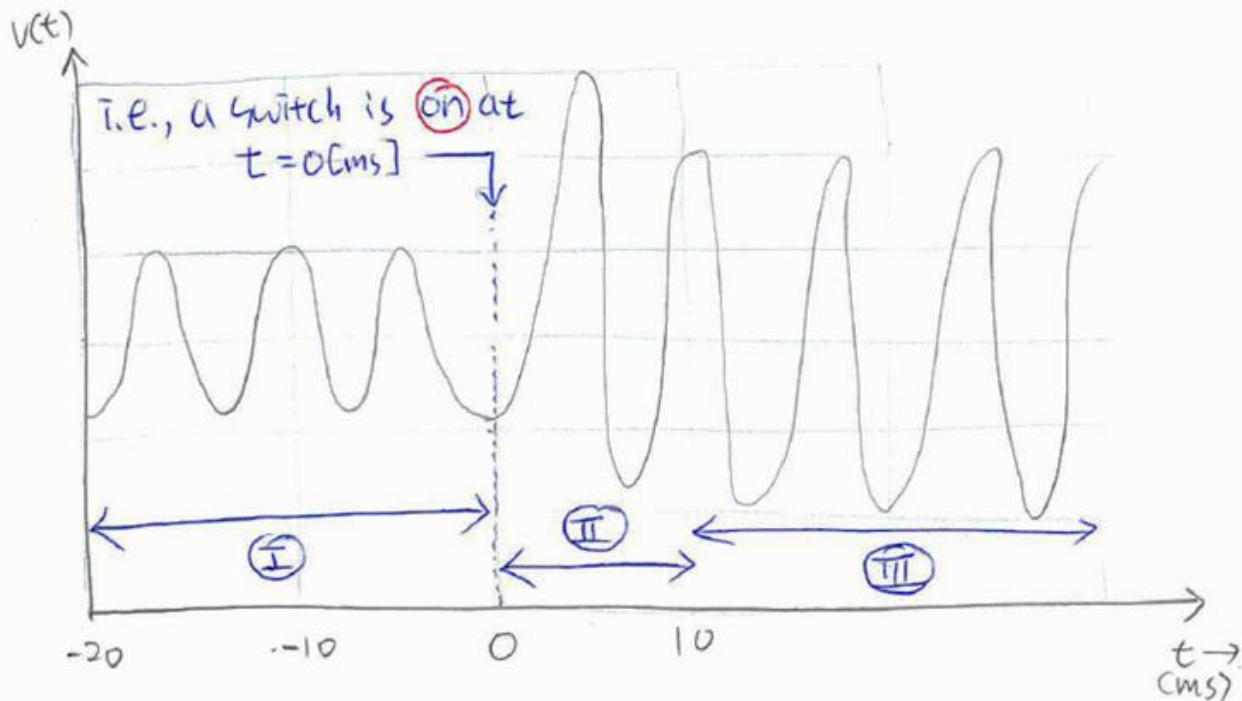
$$v(t) = A \cos(1000t + \theta)$$

at switchin ( $t = 0$ )

$$v(0) = A \cos \theta$$



- I. A steady state when the switch is opened  
(not yet closed)
- II. A transient state right after the switch is closed
  - A momentarily status due to changing its status
- III. A steady state when the switch is closed

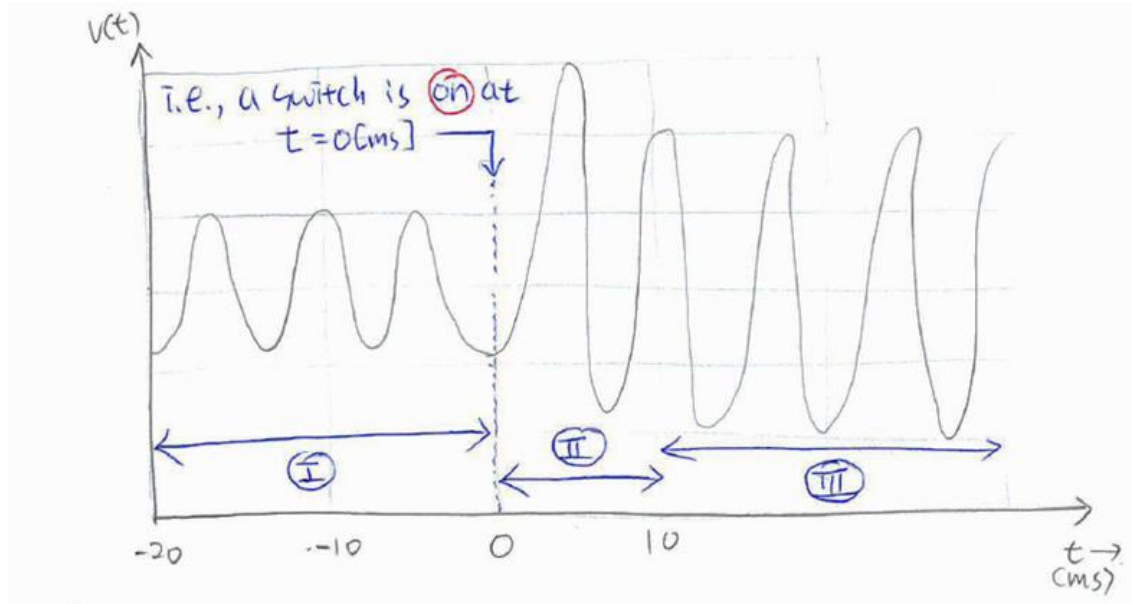


# Partic. Sol. forms for the 1<sup>st</sup> order PDE

$$X' + aX = f(t) \rightarrow \text{Where } X=X(t) \text{ a function of time}$$

$f(t)$	$X_p(t)$ guess
A constant	A constant
$A \cdot t$	$B \cdot t + C$
$A \cdot e^{\omega t}$	$B \cdot e^{\omega t}$
Sinusoidal with $\omega$	Sinusoidal with $\omega$ diff amplitude and phase ( $\varphi$ )

Where A, B, and C are constants

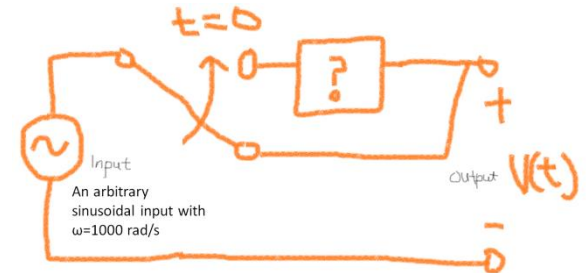


*Before switching*

$$v(t) = A \cos(1000t + \theta)$$

at switchin ( $t = 0$ )

$$v(0) = A \cos \theta$$



- 1<sup>st</sup> order PDE i.e:  $Z * \cos(1000t + \vartheta) = X * v(t) + Y * \frac{d}{dt} v(t)$
- The complete sol for  $v(t)$

$$v(t) = K_1 \cos(1000t + \varphi) + K_2 e^{-t/\tau} [\text{V}]$$

- Particul. sol
- Steady-state R.
- Dominant R. (stable) when  $t \rightarrow \infty$

- Homogen. sol
- Transient R
- Becomes 0 when  $t \rightarrow \infty$

# 7. Simple RC, RL circuits' analysis

- Two properties to be used

1. When in steady state,  $t < 0$  ( $t < t_0$ ) or  $t \rightarrow \infty$

- A Capacitor

- Open circuit ( $i=0$ )
- Use  $v_c(0^-) = v_c(0^+)$  to get  $v_c(0)$   
or  $v_c(t_0^-) = v_c(t_0^+)$  to get  $v_c(t_0)$

To define  $i_c(0)$ , it has to be differentiable

$$i_c(t) = C \frac{d}{dt} v_c(t)$$

- An inductor

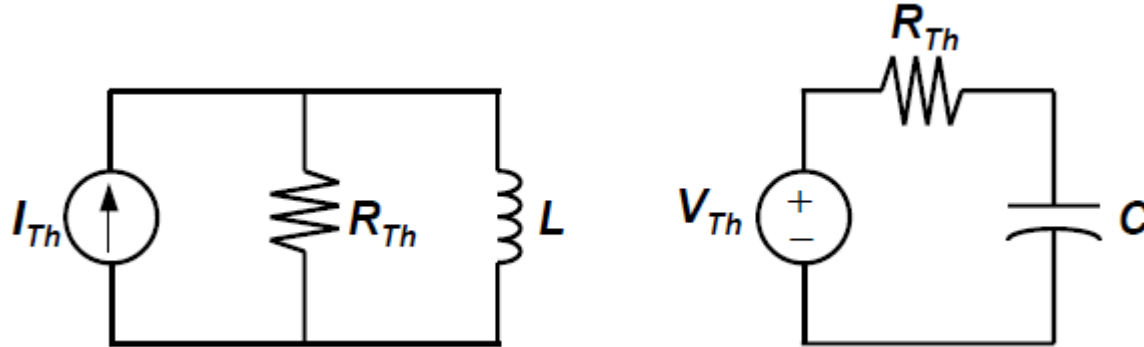
- Short circuit ( $v=0$ )
- Use  $i_L(0^-) = i_L(0^+)$  to get  $i_L(0)$   
or  $i_L(t_0^-) = i_L(t_0^+)$  to get  $i_L(t_0)$

- In steady state, an inductor behaves like a short circuit
- In steady state, a capacitor behaves like an open circuit



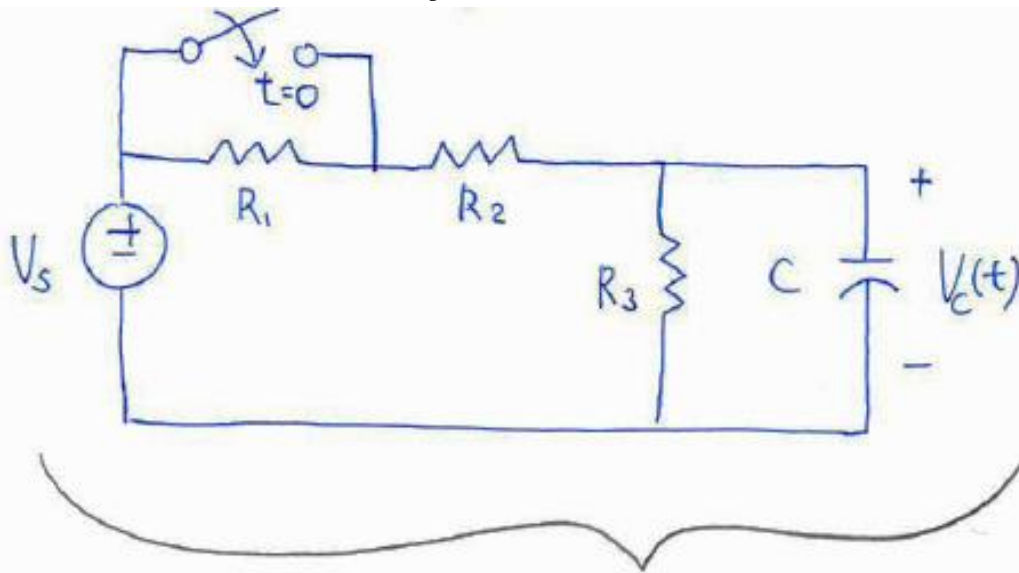
## 2. When $t > 0$ or $t > t_0$

- RC circuits: Thevenin equivalent  $\rightarrow$  KVL or node methods
- RL circuits: Norton equivalent  $\rightarrow$  KCL or node methods

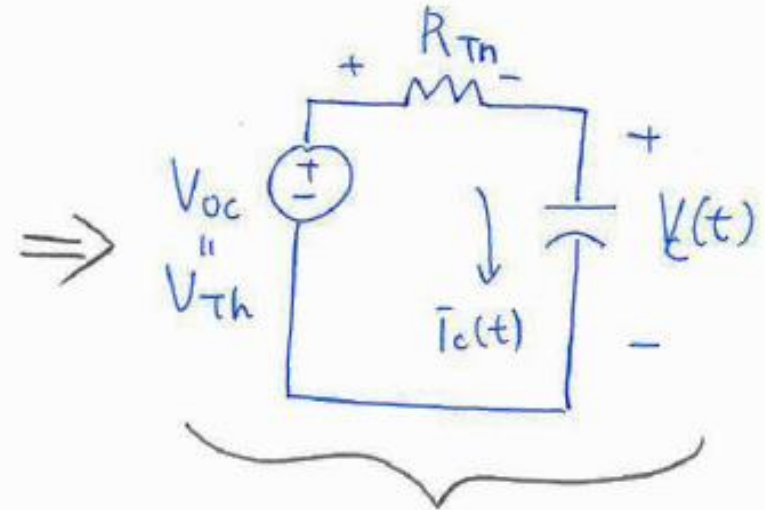


- Then, solve for 1<sup>st</sup> order PDE
  - Usually 2 unknowns (i.e.,  $K_1$  and  $K_2$ )
  - Need 2 boundary conditions including an initial condition ( $t=0$ )
    - Usually conditions at  $t=0$  and  $t \rightarrow \infty$

- RC analysis



Any RC circuit can be  
reduced to

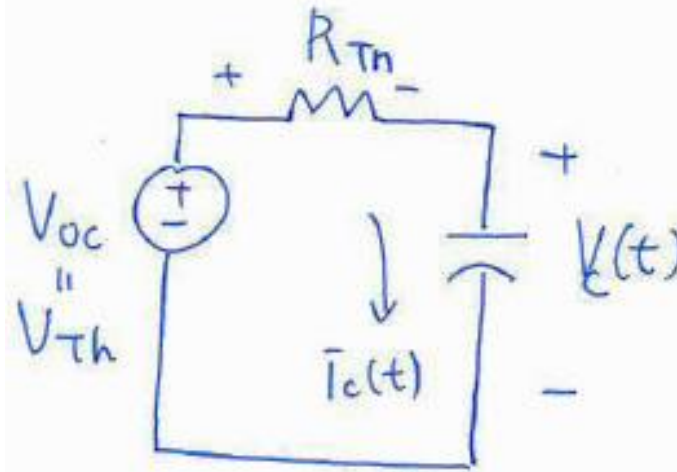


$\Rightarrow$  Thevenin eqv.

When  $t = 0^+$  the  
switch is closed

$$\left\{ \begin{aligned} V_{Th} &= V_{oc} = \frac{R_3}{R_2 + R_3} V_s \\ R_{Th} &= R_2 || R_3 = \frac{R_2 \cdot R_3}{R_2 + R_3} \end{aligned} \right.$$

- RC analysis (cont..)
  - KVL on Thevenin



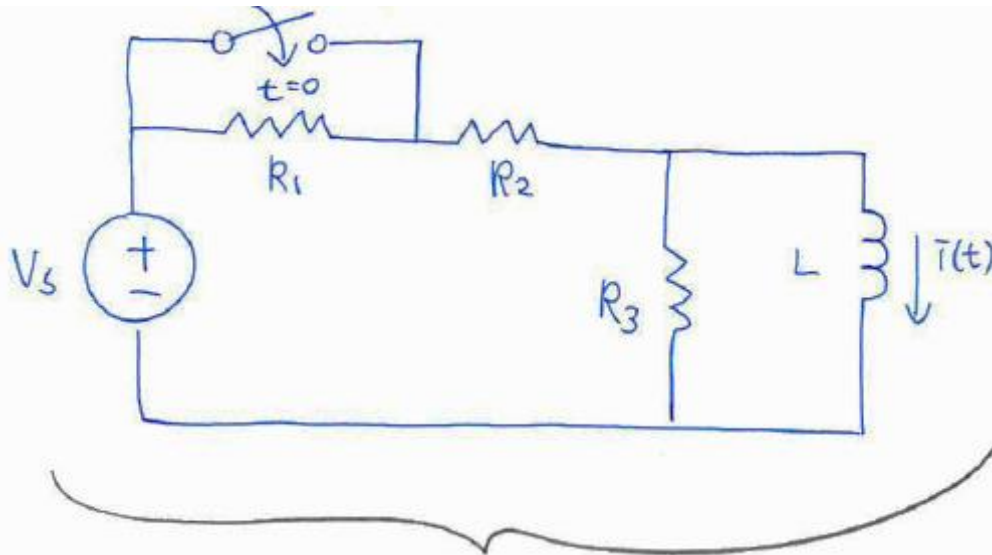
$$(Gain) = (Drop)$$

$$V_{oc} = R_{Th} \bar{I}_c(t) + V_c(t)$$

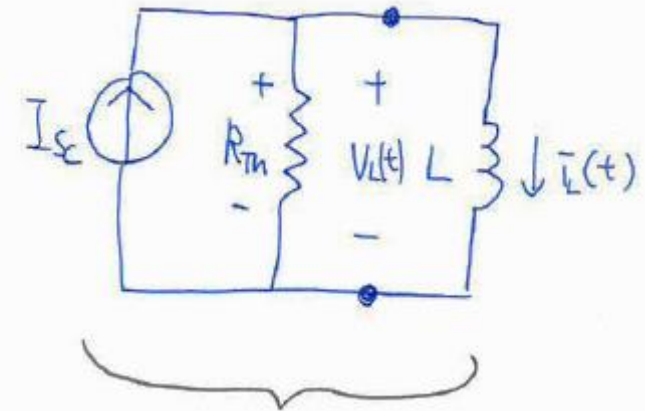
$$= R_{Th} \left( C \cdot \frac{d}{dt} V_c(t) \right) + V_c(t)$$

$$= \underbrace{R_{Th} \left( C \cdot \frac{d}{dt} V_c(t) \right)}_{\tau} + V_c(t)$$

- RL analysis

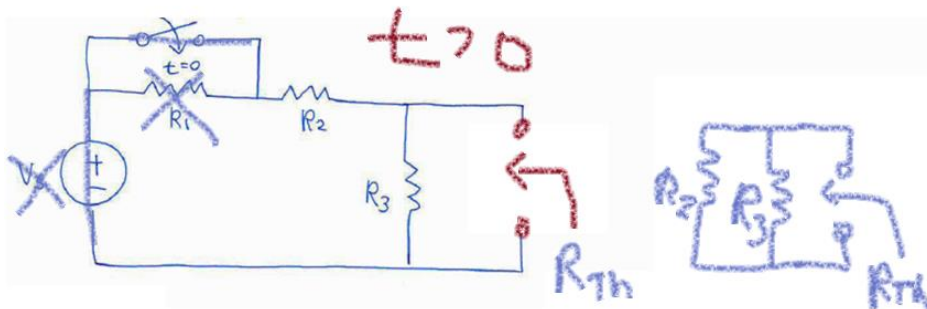


any RL circuits can be reduced to



Norton eqv.

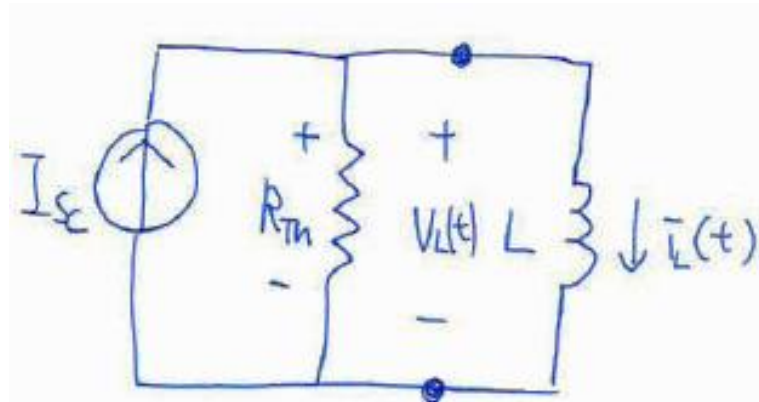
When  $t = 0^+$  the switch is closed



$$I_{SC} = \frac{V_{OC}}{R_{Th}}$$

$$R_{Th} = R_2 || R_3 = \frac{R_2 \cdot R_3}{R_2 + R_3}$$

- RL analysis (cont..)
  - KCL on Norton



$$I_{in} = I_{out}$$

$$I_{sc} = \frac{V_L(t)}{R_{Th}} + I_L(t) \Rightarrow \frac{1}{R_{Th}} \left( L \cdot \frac{d}{dt} I_L(t) \right) + I_L(t)$$

$$\Rightarrow \underbrace{\frac{1}{R_{Th}} \left( L \frac{d}{dt} I_L(t) \right)}_{\tau} + I_L(t)$$

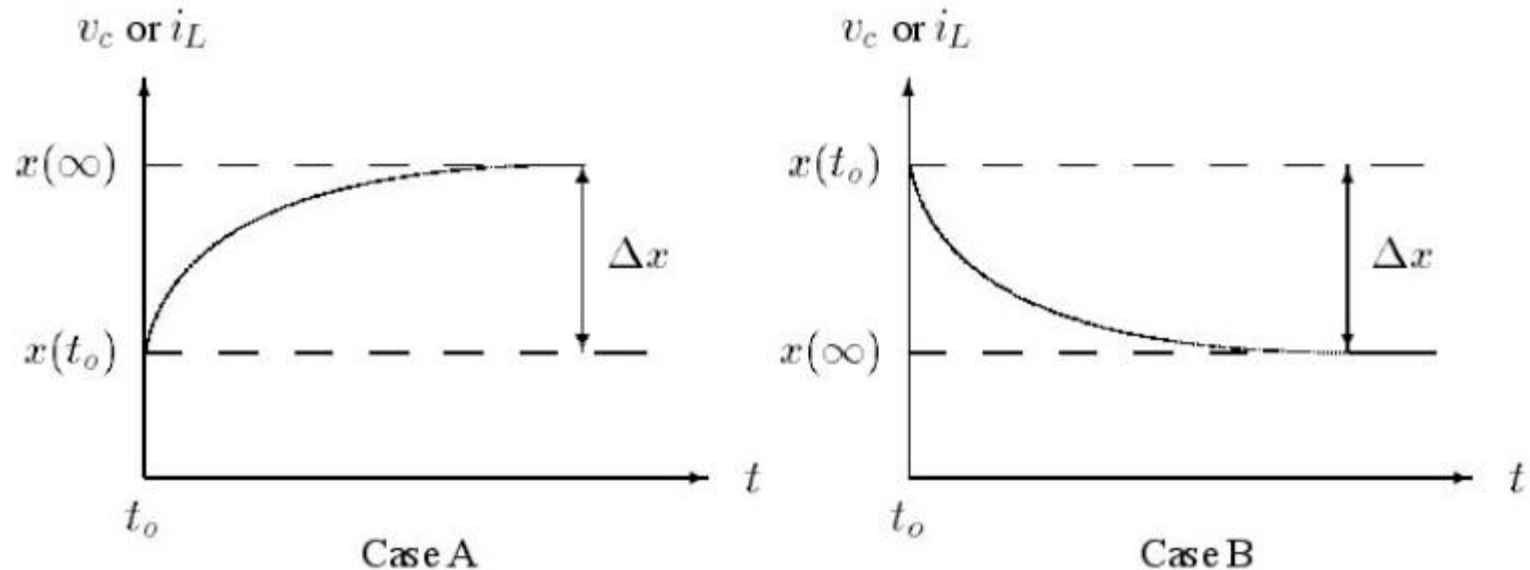
- Based on previous RC and RL analysis, we can generalize the 1<sup>st</sup> order PDE form when each circuit is reduced to Thevenin or Norton equivalent

$$V_{oc} = \underbrace{R_{Th}}_{\tau} \left( C \frac{d}{dt} V_c(t) \right) + V_c(t)$$

$$I_{sc} = \underbrace{\frac{1}{R_{Th}}}_{\tau} \left( L \frac{d}{dt} I_L(t) \right) + I_L(t)$$

$$\tau \frac{d}{dt} v_{or} i(t) + v_{or} i(t) = V_{oc} \text{ or } I_{sc}$$

# More in general ...



Case A: 
$$x(t) = x(t_o) + \Delta x \left(1 - e^{-(t-t_o)/\tau}\right)$$

Case B: 
$$x(t) = x(\infty) + \Delta x e^{-(t-t_o)/\tau}$$

- Previous examples:

?