

SERIAL COMMUNICATIONS

I²C and SPI digital communications are available. In both cases, the ADXL345 operates as a slave. I²C mode is enabled if the $\overline{\text{CS}}$ pin is tied high to $V_{\text{DD I/O}}$. The $\overline{\text{CS}}$ pin should always be tied high to $V_{\text{DD I/O}}$ or be driven by an external controller because there is no default mode if the $\overline{\text{CS}}$ pin is left unconnected. Therefore, not taking these precautions may result in an inability to communicate with the part. In SPI mode, the $\overline{\text{CS}}$ pin is controlled by the bus master. In both SPI and I²C modes of operation, data transmitted from the ADXL345 to the master device should be ignored during writes to the ADXL345.

SPI

For SPI, either 3- or 4-wire configuration is possible, as shown in the connection diagrams in Figure 34 and Figure 35. Clearing the SPI bit (Bit D6) in the DATA_FORMAT register (Address 0x31) selects 4-wire mode, whereas setting the SPI bit selects 3-wire mode. The maximum SPI clock speed is 5 MHz with 100 pF maximum loading, and the timing scheme follows clock polarity (CPOL) = 1 and clock phase (CPHA) = 1. If power is applied to the ADXL345 before the clock polarity and phase of the host processor are configured, the $\overline{\text{CS}}$ pin should be brought high before changing the clock polarity and phase. When using 3-wire SPI, it is recommended that the SDO pin be either pulled up to $V_{\text{DD I/O}}$ or pulled down to GND via a 10 k Ω resistor.

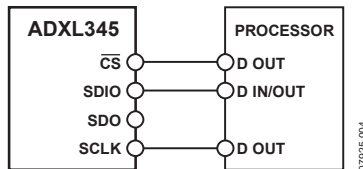


Figure 34. 3-Wire SPI Connection Diagram

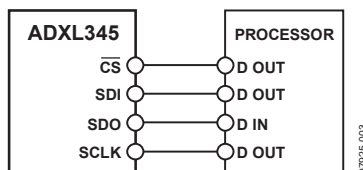


Figure 35. 4-Wire SPI Connection Diagram

$\overline{\text{CS}}$ is the serial port enable line and is controlled by the SPI master. This line must go low at the start of a transmission and high at the end of a transmission, as shown in Figure 37. SCLK is the serial port clock and is supplied by the SPI master. SCLK should idle high during a period of no transmission. SDI and SDO are the serial data input and output, respectively. Data is updated on the falling edge of SCLK and should be sampled on the rising edge of SCLK.

To read or write multiple bytes in a single transmission, the multiple-byte bit, located after the R/W bit in the first byte transfer (MB in Figure 37 to Figure 39), must be set. After the register addressing and the first byte of data, each subsequent set of clock pulses (eight clock pulses) causes the ADXL345 to point to the next register for a read or write. This shifting continues until the clock pulses cease and $\overline{\text{CS}}$ is deasserted. To perform reads or writes on different, nonsequential registers, $\overline{\text{CS}}$ must be deasserted between transmissions and the new register must be addressed separately.

The timing diagram for 3-wire SPI reads or writes is shown in Figure 39. The 4-wire equivalents for SPI writes and reads are shown in Figure 37 and Figure 38, respectively. For correct operation of the part, the logic thresholds and timing parameters in Table 9 and Table 10 must be met at all times.

Use of the 3200 Hz and 1600 Hz output data rates is only recommended with SPI communication rates greater than or equal to 2 MHz. The 800 Hz output data rate is recommended only for communication speeds greater than or equal to 400 kHz, and the remaining data rates scale proportionally. For example, the minimum recommended communication speed for a 200 Hz output data rate is 100 kHz. Operation at an output data rate above the recommended maximum may result in undesirable effects on the acceleration data, including missing samples or additional noise.

Preventing Bus Traffic Errors

The ADXL345 $\overline{\text{CS}}$ pin is used both for initiating SPI transactions, and for enabling I²C mode. When the ADXL345 is used on a SPI bus with multiple devices, its $\overline{\text{CS}}$ pin is held high while the master communicates with the other devices. There may be conditions where a SPI command transmitted to another device looks like a valid I²C command. In this case, the ADXL345 would interpret this as an attempt to communicate in I²C mode, and could interfere with other bus traffic. Unless bus traffic can be adequately controlled to assure such a condition never occurs, it is recommended to add a logic gate in front of the SDI pin as shown in Figure 36. This OR gate will hold the SDA line high when $\overline{\text{CS}}$ is high to prevent SPI bus traffic at the ADXL345 from appearing as an I²C start command.

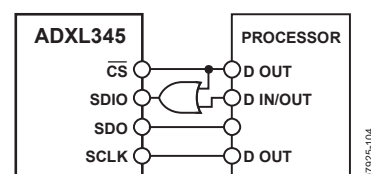


Figure 36. Recommended SPI Connection Diagram when Using Multiple SPI Devices on a Single Bus

REGISTER MAP

Table 19.

| Address | | Name | Type | Reset Value | Description |
|--------------|---------|----------------|------|-------------|---|
| Hex | Dec | | | | |
| 0x00 | 0 | DEVID | R | 11100101 | Device ID |
| 0x01 to 0x1C | 1 to 28 | Reserved | | | Reserved; do not access |
| 0x1D | 29 | THRESH_TAP | R/W | 00000000 | Tap threshold |
| 0x1E | 30 | OFSX | R/W | 00000000 | X-axis offset |
| 0x1F | 31 | OFSY | R/W | 00000000 | Y-axis offset |
| 0x20 | 32 | OFSZ | R/W | 00000000 | Z-axis offset |
| 0x21 | 33 | DUR | R/W | 00000000 | Tap duration |
| 0x22 | 34 | Latent | R/W | 00000000 | Tap latency |
| 0x23 | 35 | Window | R/W | 00000000 | Tap window |
| 0x24 | 36 | THRESH_ACT | R/W | 00000000 | Activity threshold |
| 0x25 | 37 | THRESH_INACT | R/W | 00000000 | Inactivity threshold |
| 0x26 | 38 | TIME_INACT | R/W | 00000000 | Inactivity time |
| 0x27 | 39 | ACT_INACT_CTL | R/W | 00000000 | Axis enable control for activity and inactivity detection |
| 0x28 | 40 | THRESH_FF | R/W | 00000000 | Free-fall threshold |
| 0x29 | 41 | TIME_FF | R/W | 00000000 | Free-fall time |
| 0x2A | 42 | TAP_AXES | R/W | 00000000 | Axis control for single tap/double tap |
| 0x2B | 43 | ACT_TAP_STATUS | R | 00000000 | Source of single tap/double tap |
| 0x2C | 44 | BW_RATE | R/W | 00001010 | Data rate and power mode control |
| 0x2D | 45 | POWER_CTL | R/W | 00000000 | Power-saving features control |
| 0x2E | 46 | INT_ENABLE | R/W | 00000000 | Interrupt enable control |
| 0x2F | 47 | INT_MAP | R/W | 00000000 | Interrupt mapping control |
| 0x30 | 48 | INT_SOURCE | R | 00000010 | Source of interrupts |
| 0x31 | 49 | DATA_FORMAT | R/W | 00000000 | Data format control |
| 0x32 | 50 | DATA0 | R | 00000000 | X-Axis Data 0 |
| 0x33 | 51 | DATA1 | R | 00000000 | X-Axis Data 1 |
| 0x34 | 52 | DATA0 | R | 00000000 | Y-Axis Data 0 |
| 0x35 | 53 | DATA1 | R | 00000000 | Y-Axis Data 1 |
| 0x36 | 54 | DATA0 | R | 00000000 | Z-Axis Data 0 |
| 0x37 | 55 | DATA1 | R | 00000000 | Z-Axis Data 1 |
| 0x38 | 56 | FIFO_CTL | R/W | 00000000 | FIFO control |
| 0x39 | 57 | FIFO_STATUS | R | 00000000 | FIFO status |

DATA FORMATTING OF UPPER DATA RATES

Formatting of output data at the 3200 Hz and 1600 Hz output data rates changes depending on the mode of operation (full-resolution or fixed 10-bit) and the selected output range.

When using the 3200 Hz or 1600 Hz output data rates in full-resolution or ± 2 g, 10-bit operation, the LSB of the output data-word is always 0. When data is right justified, this corresponds to Bit D0 of the DATAx0 register, as shown in Figure 49. When data is left justified and the part is operating in ± 2 g, 10-bit mode, the LSB of the output data-word is Bit D6 of the DATAx0 register. In full-resolution operation when data is left justified, the location of the LSB changes according to the selected output range.

For a range of ± 2 g, the LSB is Bit D6 of the DATAx0 register; for ± 4 g, Bit D5 of the DATAx0 register; for ± 8 g, Bit D4 of the DATAx0 register; and for ± 16 g, Bit D3 of the DATAx0 register. This is shown in Figure 50.

The use of 3200 Hz and 1600 Hz output data rates for fixed 10-bit operation in the ± 4 g, ± 8 g, and ± 16 g output ranges provides an LSB that is valid and that changes according to the applied acceleration. Therefore, in these modes of operation, Bit D0 is not always 0 when output data is right justified and Bit D6 is not always 0 when output data is left justified. Operation at any data rate of 800 Hz or lower also provides a valid LSB in all ranges and modes that changes according to the applied acceleration.

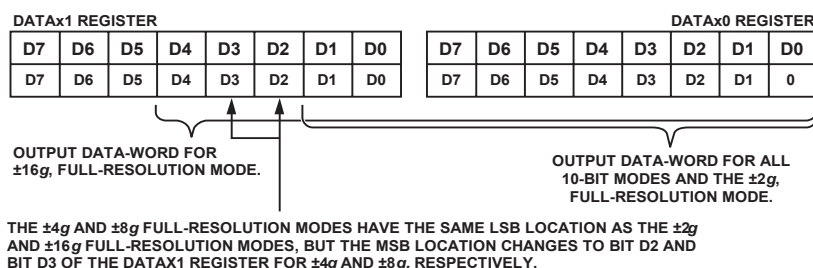


Figure 49. Data Formatting of Full-Resolution and ± 2 g, 10-Bit Modes of Operation When Output Data Is Right Justified

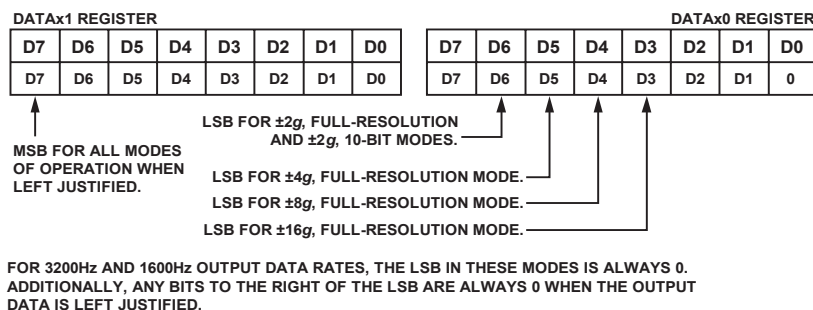


Figure 50. Data Formatting of Full-Resolution and ± 2 g, 10-Bit Modes of Operation When Output Data Is Left Justified