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Amplifier frequency response

Tutorial 6

- 1) At sufficiently high frequencies, the capacitive reactances of the coupling capacitors are very small according to the formula $X_C = \frac{1}{2\pi f C}$. That is why, the coupling capacitors do not have a significant effect on gain and they appear effectively as short circuits.
- 2) At lower frequencies, the reactance is greater, and it decreases as the frequency increases. At low frequencies, capacitively coupled amplifiers have less voltage gain than they have at higher frequencies. The reason is that at lower frequencies more signal voltage is dropped across C_1 and C_3 (coupling capacitors) because their reactances are higher. This higher signal voltage drop at lower frequencies reduces the voltage gain. Also, a phase shift is introduced by the coupling capacitors because C_1 forms a lead circuit with the R_{in} of the amplifier and C_3 forms a lead circuit with R_L in series with R_C or R_D . At lower frequencies, the reactance of bypass capacitor, C_2 , becomes significant and emitter is no longer at AC ground. The capacitive reactance X_{C_2} in parallel with R_E creates an impedance that reduces the gain.

$$3) \quad a) \quad f_c = \frac{1}{2\pi RC} = \frac{1}{2\pi \cdot (100 \Omega) \cdot (5 \mu F)} = 318.3 \text{ KHz.}$$

$$b) \quad f_c = \frac{1}{2\pi RC} = \frac{1}{2\pi \cdot (1 \text{ k}\Omega) \cdot (0.1 \mu F)} = 1591.5 \text{ KHz.}$$

$$4) \quad A_{p(dB)} = 10 \log \left(\frac{P_{out}}{P_{in}} \right) = 10 \log 10 = 10 \text{ dB}$$

$$5) \quad V_{in} = \frac{V_{out}}{A_v} = \frac{1.2 V_{rms}}{50} = 24 \text{ mV}_{rms}$$

$$A_{v(dB)} = 20 \log (A_v) = 20 \log 50 = 33.98 \text{ dB}$$

$$6) \quad a) \quad 10 \log \left(\frac{2 \text{ mV}}{1 \text{ mV}} \right) = 3.01 \text{ dBm}$$

$$b) \quad 10 \log \left(\frac{1 \text{ mV}}{1 \text{ mV}} \right) = 0 \text{ dBm}$$

$$c) \quad 10 \log \left(\frac{4 \text{ mV}}{1 \text{ mV}} \right) = 6.02 \text{ dBm}$$

$$d) \quad 10 \log \left(\frac{0.25 \text{ mV}}{1 \text{ mV}} \right) = -6.02 \text{ dBm}$$

$$7) \quad \theta = \tan^{-1} \left(\frac{X_{ce}}{R_{in}} \right) = \tan^{-1} \left(\frac{0.5 R_{in}}{R_{in}} \right) = \tan^{-1} (0.5) = 26.57 \text{ degree}$$

$$8) V_E = \left(\frac{R_2}{R_1 + R_2} \right) V_{CC} - 0.7V = \left(\frac{4.7k\Omega}{37.7k\Omega} \right) \cdot 20V - 0.7V = 1.79V$$

$$I_E = \frac{V_E}{R_E} = \frac{1.79V}{560\Omega} = 3.2mA$$

$$r'_e = \frac{25mV}{3.2mA} = 7.8\Omega$$

$$R_c = R_C \parallel R_L = \frac{2.2 \cdot 5.6}{7.8} = 1.58k\Omega$$

$$A_v = \frac{R_c}{r'_e} = \frac{1.58k\Omega}{7.8} = 202.56$$

$$C_{in(miller)} = C_{bc} \cdot (A_v + 1) = 814.24pF$$

$$C_{out(miller)} = C_{bc} \cdot \left(\frac{A_v + 1}{A_v} \right) = 4.02pF$$

9) ~~Diagram of a common emitter amplifier circuit~~

$$R_{in} = R_1 \parallel R_2 \parallel \beta \cdot r'_e = 910.9\Omega$$

$$f_{cl(input)} = \frac{1}{2\pi \cdot (R_s + R_{in}) \cdot C_1} = \frac{1}{2\pi \cdot (50 + 910.9)\Omega \cdot 0.1\mu F} = 1656.3Hz$$

$$f_{cl(output)} = \frac{1}{2\pi \cdot (R_c + R_L) \cdot C_3} = \frac{1}{2\pi \cdot (2.2 + 5.6)k\Omega \cdot 0.1\mu F} = 204.04Hz$$

$$10) R_{in(gate)} = \left| \frac{V_{GS}}{I_{GSS}} \right| = \left| \frac{-10V}{50 nA} \right| = 200 M\Omega$$

$$R_{in} = R_G \parallel R_{in(gate)} = 10 M\Omega \parallel 200 M\Omega = 9.52 M\Omega$$

$$f_{cl(input)} = \frac{1}{2\pi R_{in} \cdot C_1} = \frac{1}{2\pi \cdot (9.52 M\Omega) \cdot 0.005 \mu F} = 3.35 KHz$$

$$f_{cl(output)} = \frac{1}{2\pi \cdot (R_o + R_L) \cdot C_2} = \frac{1}{2\pi \cdot (560 \Omega + 10 K\Omega) \cdot 0.005 \mu F} = 3.016 KHz$$

$$11) V_B = \left(\frac{R_2}{R_1 + R_2} \right) \cdot V_{CC} = \left(\frac{4.7 K\Omega}{26.7 K\Omega} \right) \cdot 10V = 1.76V$$

$$V_E = V_B - 0.7V = 1.06V$$

$$I_E = \frac{V_E}{R_E} = \frac{1.06V}{470 \Omega} = 2.26 mA$$

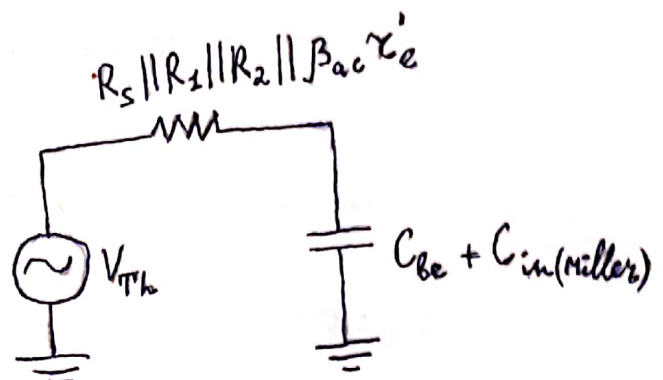
$$r'_e = \frac{25 mV}{I_E} = 11.1 \Omega$$

$$R_{in(tot)} = R_S \parallel R_1 \parallel R_2 \parallel \beta_{ac} \cdot r'_e = 378 \Omega$$

$$C_{in(miller)} = C_{bc} \cdot \left(\frac{R_C \parallel R_L}{r'_e} + 1 \right) = 240 pF$$

$$C_{in(tot)} = C_{in(miller)} + C_{be} = 240 pF + 20 pF = 260 pF$$

$$f_{cu(input)} = \frac{1}{2\pi \cdot R_{in(tot)} \cdot C_{in(tot)}} = \frac{1}{2\pi \cdot 378 \Omega \cdot 260 pF} = 1.62 MHz$$



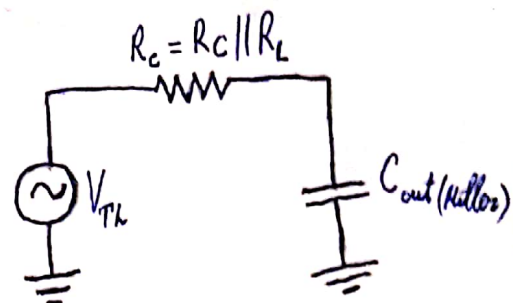
Equivalent high-frequency input RC circuit

$$A_{v(mid)} = \frac{R_c}{r'_e} = \frac{R_C \parallel R_L}{r'_e} = 99$$

$$C_{out(miller)} = C_{bc} \cdot \left(\frac{A_v + 1}{A_v} \right) = 2.42 pF$$

$$R_c = R_C \parallel R_L = 1.1 K\Omega$$

$$f_{cu(output)} = \frac{1}{2\pi \cdot R_c \cdot C_{out(miller)}} = 60 MHz$$



Equivalent high-frequency output RC circuit.